Lecture #13

Boolean Algebra

NUS | School of
National University | Computing
of Singapore

Lecture #13: Boolean Algebra

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    Sum-of-Minterms and Product-of-Maxterms

1. Digital Circuits (1/2)

- Two voltage levels
  - High/true/1/asserted
  - Low/false/0/deasserted

High

Low

Signals in digital circuit   *Signals in analog circuit*

- Advantages of digital circuits over analog circuits
  - More reliable (simpler circuits, less noise-prone)
  - Specified accuracy (determinable)
  - Abstraction can be applied using simple mathematical model – Boolean Algebra
  - Ease design, analysis and simplification of digital circuit – Digital Logic Design

1. Digital Circuits (2/2)

- Combinational: no memory, output depends solely on the input
  - Gates
  - Decoders, multiplexers
  - Adders, multipliers

- Sequential: with memory, output depends on both input and current state
  - Counters, registers
  - Memories

2. Boolean Algebra

- Boolean values:
  - True (T or 1)
  - False (F or 0)

- Connectives
  - Conjunction (AND)
    - $A \cdot B$; $A \wedge B$
  - Disjunction (OR)
    - $A + B$; $A \vee B$
  - Negation (NOT)
    - A'; $\overline{A}$; $\neg A$;

In CS2100, we use the symbols 1 for true, 0 for false, · for AND, + for OR, and ' for negation (you may use the accent bar). Please follow.

- Truth tables
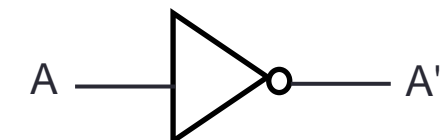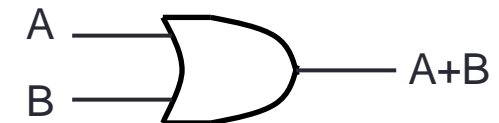
A	B	A · B
0	0	0
0	1	0
1	0	0
1	1	1

A	B	A + B
0	0	0
0	1	1
1	0	1
1	1	1

A	A'
0	1
1	0

- Logic gates

A
B ———— A·B

A
B ———— A+B

A ———— A'
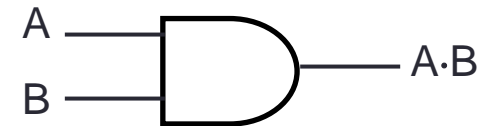
2. Boolean Algebra: AND

- Do write the AND operator · (instead of omitting it)
  - Example: Write ***a·b*** instead of ***ab***
  - Why? Writing ***ab*** could mean that it is a 2-bit value.

3. Truth Table

- Provide a listing of every possible combination of inputs and its corresponding outputs.
  - Inputs are usually listed in binary sequence.

- Example
  - Truth table with 3 inputs x, y, z and 2 outputs (y + z) and (x · (y + z)).

x	y	z	y + z	x · (y + z)
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	1	0
1	0	0	0	0
1	0	1	1	1
1	1	0	1	1
1	1	1	1	1

3. Proof using Truth Table

- Prove: $x \cdot (y + z) = (x \cdot y) + (x \cdot z)$
  - Construct truth table for LHS and RHS

x	y	z	y + z	$x \cdot (y + z)$	$x \cdot y$	$x \cdot z$	$(x \cdot y) + (x \cdot z)$
0	0	0	0	**0**	0	0	**0**
0	0	1	1	**0**	0	0	**0**
0	1	0	1	**0**	0	0	**0**
0	1	1	1	**0**	0	0	**0**
1	0	0	0	**0**	0	0	**0**
1	0	1	1	**1**	0	1	**1**
1	1	0	1	**1**	1	0	**1**
1	1	1	1	**1**	1	1	**1**

  - Check that column for LHS = column for RHS

- DLD page 59 Quick Review Questions Question 3-1.

4. Precedence of Operators

- ## Precedence from highest to lowest
  - Not (')
  - And (·)
  - Or (+)

  Note the difference with CS1231/CS1231S. Here in CS2100, AND has higher precedence than OR.

- ## Examples:
  - A · B + C = (A · B) + C

    Hence, A · B + C is <u>not</u> ambiguous in CS2100.
  - X + Y' = X + (Y')
  - P + Q' · R = P + ((Q') · R)

- ## Use parenthesis to overwrite precedence. Examples:
  - A · (B + C)   [Without parenthesis, it means A·B+C or (A·B)+C]
  - (P + Q)' · R   [Without parenthesis, it means P+Q'·R or P+(Q'·R)]

5. Laws of Boolean Algebra

Identity laws	
$A + 0 = 0 + A = A$	$A \cdot 1 = 1 \cdot A = A$
Inverse/complement laws	
$A + A' = A' + A = 1$	$A \cdot A' = A' \cdot A = 0$
Commutative laws	
$A + B = B + A$	$A \cdot B = B \cdot A$
Associative laws *	
$A + (B + C) = (A + B) + C$	$A \cdot (B \cdot C) = (A \cdot B) \cdot C$
Distributive laws	
$A \cdot (B + C) = (A \cdot B) + (A \cdot C)$	$A + (B \cdot C) = (A + B) \cdot (A + C)$

* Due to the associative laws, A + B + C is unambiguous. It may be evaluated as A + (B + C) or (A + B) + C. Likewise for A·B·C.

6. Duality

- If the AND/OR operators and identity elements 0/1 in a Boolean equation are interchanged, it remains valid.

- Example:
  - The dual equation of $a+(b \cdot c)=(a+b) \cdot (a+c)$ is $a \cdot (b+c)=(a \cdot b)+(a \cdot c)$.

- Duality gives free theorems – "two for the price of one", as a Boolean equation is logically equivalent to its dual. So, you prove one theorem and the other comes for free!

- Examples:
  - If $(x+y+z)' = x' \cdot y' \cdot z'$ is valid, then its dual $(x \cdot y \cdot z)' = x'+y'+z'$ is also valid.
  - If $x+1 = 1$ is valid, then its dual $x \cdot 0 = 0$ is also valid.

Do not confuse duality with negation!

7. Theorems

Idempotency	
X + X = X	X · X = X
One element / Zero element	
X + 1 = 1 + X = 1	X · 0 = 0 · X = 0
Involution	
(X')' = X	
Absorption 1	
X + X·Y = X	X·(X + Y) = X
Absorption 2	
X + X'·Y = X + Y	X·(X' + Y) = X·Y
DeMorgans' (can be generalised to more than 2 variables)	
(X + Y)' = X' · Y'	(X · Y)' = X' + Y'
Consensus	
X·Y + X'·Z + Y·Z = X·Y + X'·Z	(X+Y)·(X'+Z)·(Y+Z) = (X+Y)·(X'+Z)

7. Proving a Theorem

- Theorems can be proved using truth table, or by algebraic manipulation using other theorems/laws.

- Example: Prove absorption theorem X + X·Y = X

> X + X·Y   = X·1 + X·Y (by identity law)
>
> = X·(1+Y) (by distributivity)
>
> = X·1 (by one element law)
> = X (by identity law)

- By the principle of duality, we may also cite (<u>without proof</u>) that X·(X+Y) = X.

8. Boolean Functions

- Examples of Boolean functions (logic equations):

  $F1(x,y,z) = x \cdot y \cdot z'$

  $F2(x,y,z) = x + y' \cdot z$

  $F3(x,y,z) = x' \cdot y' \cdot z + x' \cdot y \cdot z + x \cdot y'$

  $F4(x,y,z) = x \cdot y' + x' \cdot z$

x	y	z	F1	F2	F3	F4
0	0	0	0	0	0	0
0	0	1	0	1	1	1
0	1	0	0	0	0	0
0	1	1	0	0	1	1
1	0	0	0	1	1	1
1	0	1	0	1	1	1
1	1	0	1	1	0	0
1	1	1	0	1	0	0

From the truth table, F3 = F4.

Can you prove F3 = F4 by using Boolean Algebra?

9. Complement Functions

- Given a Boolean function F, the complement of F, denoted as F', is obtained by <u>interchanging 1 with 0</u> in the function's output values.

- Example: F1 = x·y·z'

- What is F1' ?

  - F1' = (x·y·z')'
       = x' + y' + (z')' (DeMorgan's)
       = x' + y' + z   (Involution)

x	y	z	F1	F1'
0	0	0	0	1
0	0	1	0	1
0	1	0	0	1
0	1	1	0	1
1	0	0	0	1
1	0	1	0	1
1	1	0	1	0
1	1	1	0	1

10. Standard Forms (1/2)

- Certain types of Boolean expressions lead to circuits that are desirable from an implementation viewpoint.

- Two standard forms:
  - Sum-of-Products (SOP)
  - Product-of-Sums (POS)

- Literals
  - A Boolean variable on its own or in its complemented form
  - Examples: (1) x, (2) x', (3) y, (4) y'

- Product term
  - A single literal or a logical product (AND) of several literals
  - Examples: (1) x, (2) x·y·z', (3) A'·B, (4) A·B, (5) d·g'·v·w

10. Standard Forms (2/2)

- Sum term
  - A single literal or a logical sum (OR) of several literals
  - Examples: (1) x, (2) x+y+z', (3) A'+B, (4) A+B, (5) c+d+h'+j

- Sum-of-Products (SOP) expression
  - A product term or a logical sum (OR) of several product terms
  - Examples: (1) x, (2) x + y·z', (3) x·y' + x'·y·z, (4) A·B + A'·B',
    (5) A + B'·C + A·C' + C·D

- Product-of-Sums (POS) expression
  - A sum term or a logical product (AND) of several sum terms
  - Examples: (1) x, (2) x·(y+z'), (3) (x+y')·(x'+y+z),
    (4) (A+B)·(A'+B'), (5) (A+B+C)·D'·(B'+D+E')

- Every Boolean expression can be expressed in SOP or POS form.
  - DLD page 59 Quick Review Questions Questions 3-2 to 3-5.

Quiz Time!

> **SOP** expr: A product term or a logical sum (OR) of several product terms.
> **POS** expr: A sum term or a logical product (AND) of several sum terms.

- Put the right ticks in the following table.

	Expression	*SOP?*	*POS?*
(1)	X'·Y + X·Y' + X·Y·Z	✓	✗
(2)	(X+Y')·(X'+Y)·(X'+Z')	✗	✓
(3)	X' + Y + Z	✓	✓
(4)	X·(W' + Y·Z)	✗	✗
(5)	X·Y·Z'	✓	✓
(6)	W·X'·Y + V·(X·Z + W')	✗	✗

11. Minterms and Maxterms (1/2)

- A minterm of *n* variables is a <u>product term</u> that contains *n* literals from all the variables.

  - Example: On 2 variables x and y, the minterms are: x'·y', x'·y, x·y' and x·y

- A maxterm of *n* variables is a <u>sum term</u> that contains *n* literals from all the variables.

  - Example: On 2 variables x and y, the maxterms are: x'+y', x'+y, x+y' and x+y

- In general, with *n* variables we have up to 2^n minterms and 2^n maxterms.

11. Minterms and Maxterms (2/2)

- The minterms and maxterms on 2 variables are denoted by m0 to m3 and M0 to M3 respectively.

x	y	Minterms		Maxterms	
		Term	Notation	Term	Notation
0	0	x'·y'	m0	x+y	M0
0	1	x'·y	m1	x+y'	M1
1	0	x·y'	m2	x'+y	M2
1	1	x·y	m3	x'+y'	M3

- Important fact: Each minterm is the <u>complement</u> of its corresponding maxterm. Likwise, each maxterm is the complement of its corresponding minterm.

  - Example: m2 = x·y'
    m2' = (x·y')' = x' + (y')' = x' + y = M2

Quiz Time Again!

- Ability to convert minterms and maxterms from its Boolean expression to its notation (and vice versa) is important.
- Test yourself with the following quiz, assuming that you are given a Boolean function on 4 variables A, B, C, D.

Minterm

	Boolean expression	*Minterm notation*
(1)	A'·B'·C·D	m3
(2)	**A·B'·C·D'**	m10
(3)	**A·B'·C·D**	m11
(4)	A·B·C·D'	**m14**
(5)	A·B'·C'·D	**m9**

Maxterm

	Boolean expression	*Maxterm notation*
(1)	A+B+C'+D'	M3
(2)	**A'+B'+C+D'**	M13
(3)	**A+B+C+D**	M0
(4)	A+B+C'+D	**M2**
(5)	A'+B+C+D'	**M9**

12. Canonical Forms

- Canonical/normal form: a unique form of representation.
  - Sum-of-minterms = Canonical sum-of-products
  - Product-of-maxterms = Canonical product-of-sums

12.1 Sum-of-Minterms

- Given a truth table, example:

- Obtain sum-of-minterms expression by gathering the minterms of the function (where output is 1).

x	y	z	F1	F2	F3
0	0	0	0	0	0
0	0	1	0	1	1
0	1	0	0	0	0
0	1	1	0	0	1
1	0	0	0	1	1
1	0	1	0	1	1
1	1	0	1	1	0
1	1	1	0	1	0

F1 = x·y·z' = m6

F2 = x'·y'·z + x·y'·z' + x·y'·z + x·y·z' + x·y·z
  = m1 + m4 + m5 + m6 + m7 = Σm(1,4,5,6,7) or Σm(1,4 – 7)

F3 = x'·y'·z + x'·y·z + x·y'·z' + x·y'·z
  = m1 + m3 + m4 + m5 = Σm(1,3,4,5) or Σm(1,3 – 5)

12.2 Product-of-Maxterms

- Given a truth table, example:

- Obtain product-of-maxterms expression by gathering the maxterms of the function (where output is 0).

x	y	z	F1	F2	F3
0	0	0	0	⓪	0
0	0	1	0	1	1
0	1	0	0	⓪	0
0	1	1	0	⓪	1
1	0	0	0	1	1
1	0	1	0	1	1
1	1	0	1	1	0
1	1	1	0	1	0

$F2 = (x+y+z) \cdot (x+y'+z) \cdot (x+y'+z')$
$\quad = M0 \cdot M2 \cdot M3 = \Pi M(0,2,3)$

$F3 = (x+y+z) \cdot (x+y'+z) \cdot (x'+y'+z) \cdot (x'+y'+z')$
$\quad = M0 \cdot M2 \cdot M6 \cdot M7 = \Pi M(0,2,6,7)$

12.3 Conversion of Standard Forms

- We can convert between sum-of-minterms and product-of-maxterms easily

- Example: F2 = Σm(1,4,5,6,7) = ΠM(0,2,3)

- Why? See F2' in truth table.

- F2' = m0 + m2 + m3
  Therefore,
  F2 = (m0 + m2 + m3)'
      = m0' · m2' · m3' (by DeMorgan's)
      = M0 · M2 · M3   (as mx' = Mx)

x	y	z	F2	F2'
0	0	0	0	1
0	0	1	1	0
0	1	0	0	1
0	1	1	0	1
1	0	0	1	0
1	0	1	1	0
1	1	0	1	0
1	1	1	1	0

- Read up DLD section 3.4, pg 57 – 58.
- Quick Review Questions: pg 60 – 61, Q3-6 to 3-13.

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