

EE2023 TUTORIAL 1 (SOLUTIONS)

Solution to Q.1

Write z in polar form:

$$z = x + jy = |z| \exp(j\angle z).$$

Since adding integer multiples of 2π to $\angle z$ does not affect the value of z , we may also express z as

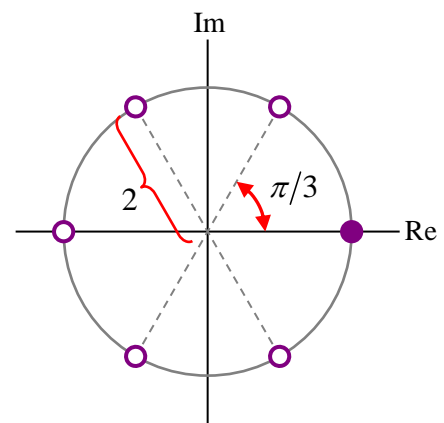
$$z = |z| \exp(j(\angle z + 2k\pi))$$

where k is an integer. This leads to

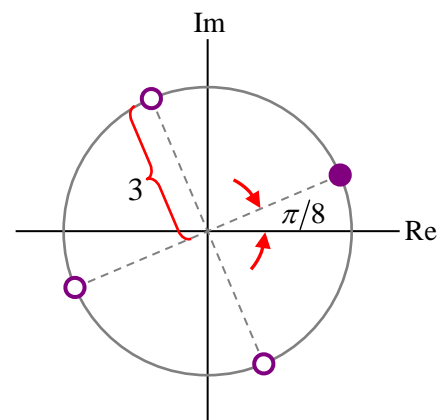
$$\sqrt[N]{z} = |z|^{1/N} \exp\left(j\left(\frac{\angle z}{N} + \frac{2k\pi}{N}\right)\right); \quad k = 0, 1, \dots, N-1,$$

which yields the N distinct values of $\sqrt[N]{z}$.

$$\sqrt[6]{64} : \left\{ \begin{array}{l} z = 64 \rightarrow \begin{cases} |z| = 64 \\ \angle z = 0 \end{cases} \\ \sqrt[6]{64} = |z|^{1/N} \exp\left(j\left(\frac{\angle z}{N} + \frac{2k\pi}{N}\right)\right) \Big|_{z=64, N=6} \\ = 2 \exp\left(j\left(\frac{k\pi}{3}\right)\right); \quad k = 0, 1, \dots, 5 \\ = \begin{cases} (2), 2 \exp\left(j\left(\frac{\pi}{3}\right)\right); 2 \exp\left(j\left(\frac{2\pi}{3}\right)\right); \\ (-2), 2 \exp\left(j\left(\frac{4\pi}{3}\right)\right); 2 \exp\left(j\left(\frac{5\pi}{3}\right)\right) \end{cases} \end{array} \right.$$



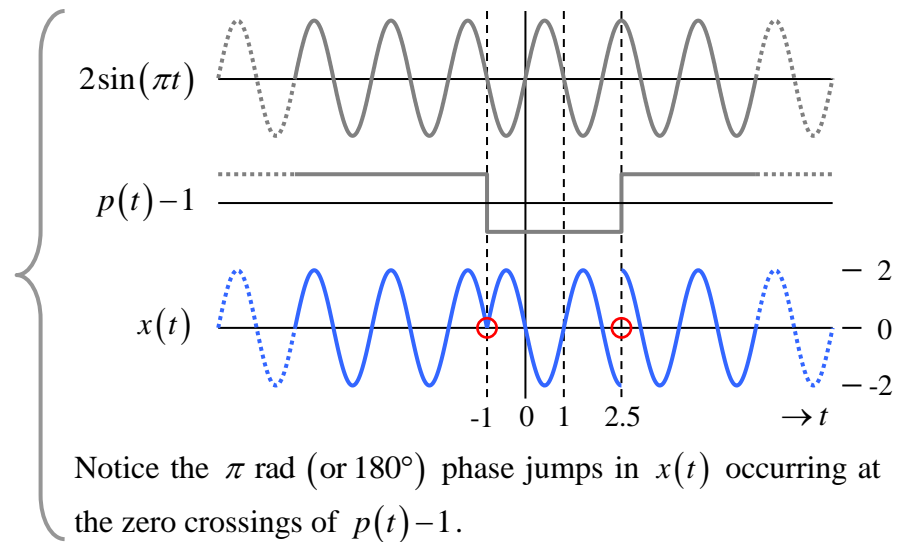
$$\sqrt[4]{j81} : \left\{ \begin{array}{l} z = j81 \rightarrow \begin{cases} |z| = 81 \\ \angle z = \frac{\pi}{2} \end{cases} \\ \sqrt[4]{j81} = |z|^{1/N} \exp\left(j\left(\frac{\angle z}{N} + \frac{2k\pi}{N}\right)\right) \Big|_{z=81, N=4} \\ = 3 \exp\left(j\left(\frac{\pi}{8} + \frac{k\pi}{2}\right)\right); \quad k = 0, 1, \dots, 3 \\ = \begin{cases} 3 \exp\left(j\left(\frac{\pi}{8}\right)\right), 3 \exp\left(j\left(\frac{5\pi}{8}\right)\right), \\ 3 \exp\left(j\left(\frac{9\pi}{8}\right)\right), 3 \exp\left(j\left(\frac{13\pi}{8}\right)\right) \end{cases} \end{array} \right.$$



Solution to Q.2

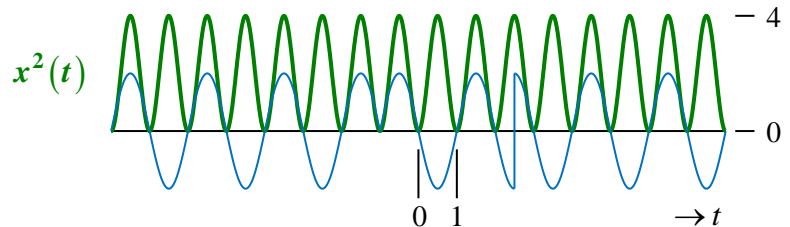
(a) $p(t) = 2 - 2\text{rect}\left(\frac{t-0.75}{3.5}\right)$

(b) By inspection, $x(t)$ is not periodic.



(c)

$$\begin{aligned} x^2(t) &= 4 \sin^2(\pi t) \underbrace{(p(t)-1)^2}_1 \\ &= 4 \sin^2(\pi t) \\ &= 2(1 - \cos(2\pi t)) \end{aligned}$$



Note that $x^2(t)$ is periodic with a period of $T = 1$.

Total energy:
$$E = \int_{-\infty}^{\infty} x^2(t) dt = \sum_{n=-\infty}^{\infty} \underbrace{\int_{nT}^{(n+1)T} x^2(t) dt}_{\substack{\text{over one period} \\ \text{thus independent} \\ \text{of } n}} = \left(\underbrace{\int_0^T x^2(t) dt}_{\text{finite}} \right) \sum_{n=-\infty}^{\infty} 1 = \infty$$

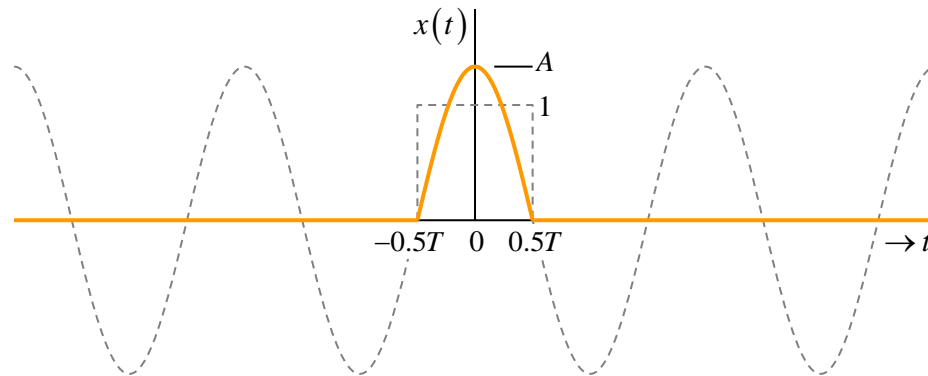
Average Power:
$$P = \frac{1}{T} \int_{-T/2}^{T/2} x^2(t) dt = \int_{-0.5}^{0.5} 2(1 - \cos(2\pi t)) dt = 2$$

 $x^2(t)$ is periodic. \therefore
 P can be obtained
 by averaging over
 one period.

Conclusion: $x(t)$ is an aperiodic power signal.

Solution to Q.3

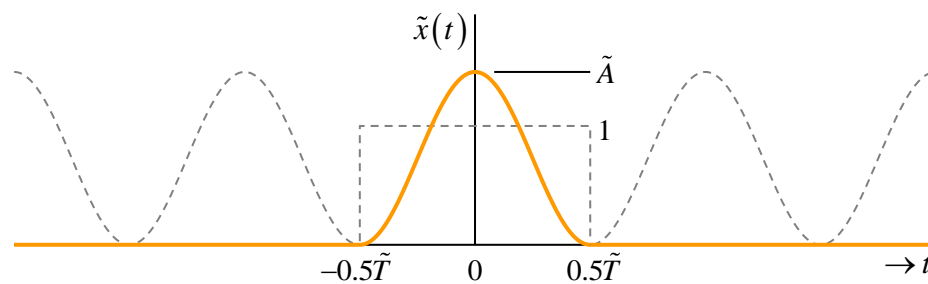
Half-cosine pulse: $x(t) = A \cos\left(\frac{\pi t}{T}\right) \text{rect}\left(\frac{t}{T}\right)$



$$x^2(t) = \frac{A^2}{2} \left[1 + \cos\left(\frac{2\pi t}{T}\right) \right] \text{rect}\left(\frac{t}{T}\right)$$

Energy: $E = \frac{A^2}{2} \int_{-0.5T}^{0.5T} \underbrace{1 + \cos\left(\frac{2\pi t}{T}\right)}_{\substack{\int \text{over one} \\ \text{period} = 0}} dt = \frac{1}{2} A^2 T$

Raised-cosine pulse: $\tilde{x}(t) = \frac{\tilde{A}}{2} \left(1 + \cos\left(\frac{2\pi t}{\tilde{T}}\right) \right) \text{rect}\left(\frac{t}{\tilde{T}}\right)$



$$\tilde{x}^2(t) = \frac{\tilde{A}^2}{4} \left[\frac{3}{2} + 2 \cos\left(\frac{2\pi t}{\tilde{T}}\right) + \frac{1}{2} \cos\left(\frac{4\pi t}{\tilde{T}}\right) \right] \text{rect}\left(\frac{t}{\tilde{T}}\right)$$

Energy: $\tilde{E} = \frac{\tilde{A}^2}{4} \int_{-0.5\tilde{T}}^{0.5\tilde{T}} \underbrace{\frac{3}{2} + 2 \cos\left(\frac{2\pi t}{\tilde{T}}\right)}_{\substack{\int \text{over one} \\ \text{period} = 0}} + \underbrace{\frac{1}{2} \cos\left(\frac{4\pi t}{\tilde{T}}\right)}_{\substack{\int \text{over two} \\ \text{periods} = 0}} dt = \frac{3}{8} \tilde{A}^2 \tilde{T}$

Both $x(t)$ and $\tilde{x}(t)$ will have the same energy if $A^2 T = \frac{3}{4} \tilde{A}^2 \tilde{T}$.

Solution to Q.4

(a) Let m , n and k be positive integers. Based on the definition of a periodic signal, we have

$$\begin{aligned}x_1(t) &= x_1(t + T_1) = x_1(t + mT_1) \\x_2(t) &= x_2(t + T_2) = x_2(t + nT_2)\end{aligned}$$

Hence,

$$x_0(t) = x_1(t) + x_2(t) = x_1(t + mT_1) + x_2(t + nT_2).$$

If $mT_1 = nT_2 = kT_0$, then

$$x_0(t) = x_1(t + kT_0) + x_2(t + kT_0) = x_0(t + kT_0)$$

which shows that $x_0(t)$ is periodic with a period of T_0 and a fundamental frequency of $f_0 = \frac{1}{T_0}$.

Under this condition, (i.e. $mT_1 = nT_2 = kT_0$),

$$\left. \begin{aligned} \frac{1}{T_1} &= m \frac{1}{kT_0} \\ \frac{1}{T_2} &= n \frac{1}{kT_0} \end{aligned} \right\} \text{ implying that } \begin{cases} \dots\dots \frac{1}{kT_0} \text{ are common factors of } \left\{ \frac{1}{T_1}, \frac{1}{T_2} \right\} \dots\dots \\ \text{and} \\ \frac{1}{T_0} \text{ is the highest common factor (HCF) of } \left\{ \frac{1}{T_1}, \frac{1}{T_2} \right\} \\ \text{or} \\ f_0 \text{ is the highest common factor (HCF) of } \{f_1, f_2\} \end{cases}$$

Conclusion: For $x_0(t)$ to be periodic, $\{f_1, f_2\}$ must have a HCF. In turn, this HCF is the fundamental frequency of $x_0(t)$

(b) i. $x(t) = \cos(3.2t) + \sin(1.6t) + \exp(j2.8t) \dots \begin{cases} \cos(3.2t) & \text{has a frequency of } 3.2 \text{ rad/s} \\ \sin(1.6t) & \text{has a frequency of } 1.6 \text{ rad/s} \\ \exp(j2.8t) & \text{has a frequency of } 2.8 \text{ rad/s} \end{cases}$

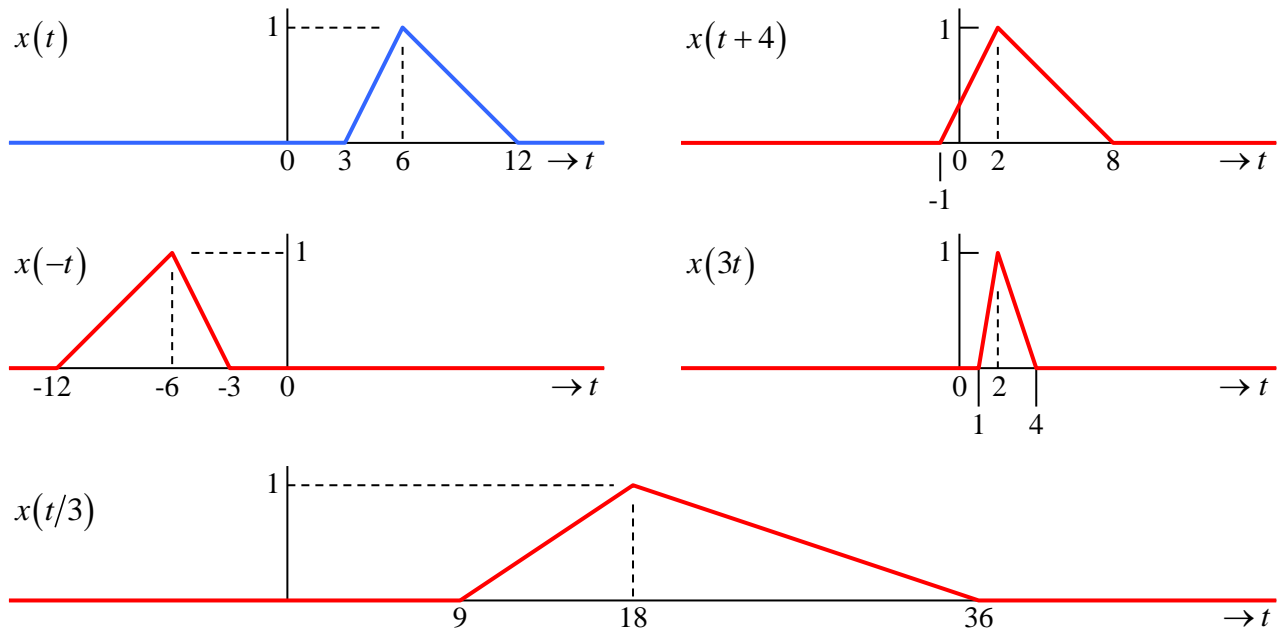
Highest common factor (HCF) of $\{3.2, 1.6, 2.8\}$ exists and is equal to 0.4. Thus, $x(t)$ is periodic and has a fundamental frequency of 0.4 rad/s (or $0.2/\pi$ Hz) and a fundamental period of 5π s.

REMARKS: Although $x(t)$ is periodic with a fundamental frequency of 0.4 rad/s, it does not contain the fundamental frequency component itself.

(b) ii. $x(t) = \cos(4t) + \sin(\pi t) \dots \begin{cases} \cos(4t) & \text{has a frequency of } 4 \text{ rad/s} \\ \sin(\pi t) & \text{has a frequency of } \pi \text{ rad/s} \end{cases}$

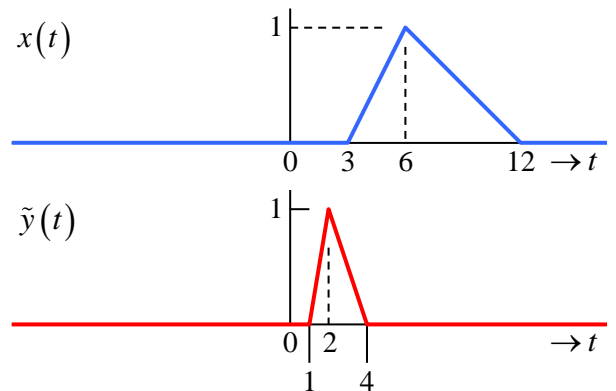
Highest common factor (HCF) of $\{4, \pi\}$ does not exist. Thus, $x(t)$ is not periodic.

REMARKS: Summing sinusoids does not necessarily lead to a periodic signal unless the frequencies of the sinusoids are harmonics of a common fundamental frequency.

Solution to Q.5**(a)****(b)** We observe that $y(t)$ is a time-scaled, -reversed and -shifted version of $x(t)$.

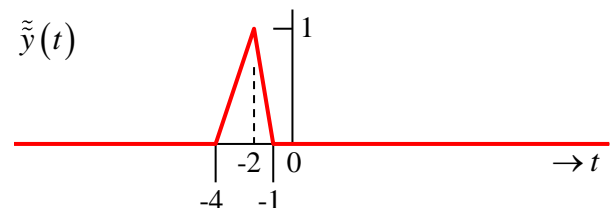
For problems of this nature, we should start with time-scaling first since it involves linear warping of the time axis. If we were to start with time-shifting and/or time-reversal, we may have to redo them after time-scaling. However, this sequence of operation need not be followed if we are sketching the signal from the mathematical expression.

Comparing $x(t)$ and $y(t)$, we note that $y(t)$ involves time-scaling (or contraction) of $x(t)$ by a factor of 3.



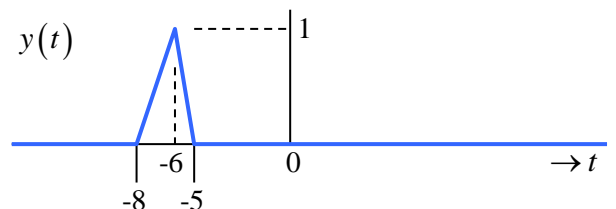
Time-scaling of $x(t)$: $\tilde{y}(t) = x(3t)$

Time-reversal of $\tilde{y}(t)$: $\tilde{\tilde{y}}(t) = \tilde{y}(-t) = x(-3t)$



Time shifting of $\tilde{\tilde{y}}(t)$: $\begin{cases} y(t) = \tilde{\tilde{y}}(t+4) \\ = x(-3(t+4)) \end{cases}$

$$\therefore y(t) = x(-3(t+4))$$



Solution to Q.6

Let

$$\delta(t) = \lim_{\Delta \rightarrow 0} x(t) \text{ where } x(t) = \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta}\right).$$

Here, we have implicitly assumed that $\Delta > 0$.

Due to symmetry, we have $\delta(\beta t) = \delta(|\beta|t)$.

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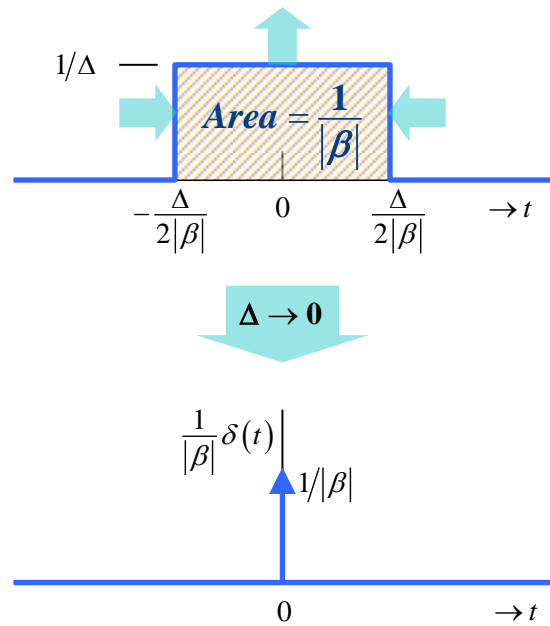
$$= \lim_{\Delta \rightarrow 0} x(|\beta|t) = \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta/|\beta|}\right)$$

But
$$\left(\begin{aligned} \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta/|\beta|}\right) &= \begin{cases} \infty; & t = 0 \\ 0; & t \neq 0 \end{cases} \\ \lim_{\Delta \rightarrow 0} \int_{-\infty}^{\infty} \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta/|\beta|}\right) dt &= \frac{1}{|\beta|} \end{aligned} \right).$$

Hence,
$$\delta(\beta t) = \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta/|\beta|}\right) = \frac{1}{|\beta|} \delta(t).$$

Domain-scaling of $\delta(\cdot)$ is often encountered in transforming a spectrum containing $\delta(\cdot)$ between cyclic-frequency (f) and radian frequency (ω) domain:

$$\left[\delta(\omega) = \delta(2\pi f) = \frac{1}{2\pi} \delta(f) \right] \text{ or } \left[\delta(f) = \delta\left(\frac{\omega}{2\pi}\right) = 2\pi \delta(\omega) \right].$$



Supplementary Questions (Solutions)

S1(a) Given that integration of unit step function, $u(t)$, is a ramp, i.e. $t.u(t)$, then $x(t)$ is made up of:

$$\begin{aligned} x(t) &= \frac{1}{2} [t.u(t)].u(2-t) \\ &= \left[\int_{-\infty}^{\infty} \frac{1}{2} u(\tau) d\tau \right] u(2-t) \end{aligned}$$

S1(b) The signal $x(t)$ is observed to be made up of various $u(t)$ functions that are shifted in time and/or reversed in time. Hence:

$$x(t) = u(t+1) + 2u(t) - u(t-1) - u(t-2) - u(t-3)$$

S2(a) Given: $x(t) = \cos(2t + 0.25\pi)$

$x(t)$ is periodic with an angular frequency of 2 rads/s.

Hence, its frequency is $\frac{2}{2\pi} = \frac{1}{\pi}$ and period of π .

$$\begin{aligned} P &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} |x(t)|^2 dt \\ &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \cos^2(2t + 0.25\pi) dt \\ &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \frac{1}{2} [1 + \cos(4t + 0.5\pi)] dt \\ &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \frac{1}{2} dt + \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \frac{1}{2} \cos(4t + 0.5\pi) dt \\ &= \frac{1}{2\pi} [t]_{-\pi/2}^{\pi/2} \\ &= \frac{1}{2\pi} \left[\frac{\pi}{2} + \frac{\pi}{2} \right] \\ &= \frac{1}{2} \end{aligned}$$

Note that $\int_{-\pi/2}^{\pi/2} \frac{1}{2} \cos(4t + 0.5\pi) dt = 0$.

S2(b)

$$\begin{aligned}
P &= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} |x(t)|^2 dt \\
&= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \left| \frac{1}{2} [1 + \cos(2t)] \right|^2 dt \\
&= \frac{1}{\pi} \int_{-\pi/2}^{\pi/2} \frac{1}{4} [1 + \cos^2(2t)] dt \\
&= \frac{1}{4\pi} \int_{-\pi/2}^{\pi/2} dt + \frac{1}{4\pi} \int_{-\pi/2}^{\pi/2} \cos^2(2t) dt \\
&= \frac{1}{4\pi} [t]_{-\pi/2}^{\pi/2} + \frac{1}{4\pi} \int_{-\pi/2}^{\pi/2} \frac{1}{2} [1 + \cos(4t)] dt \\
&= \frac{1}{4} + \frac{1}{8\pi} \int_{-\pi/2}^{\pi/2} dt + \frac{1}{8\pi} \int_{-\pi/2}^{\pi/2} \cos(4t) dt \\
&= \frac{1}{4} + \frac{1}{8\pi} [t]_{-\pi/2}^{\pi/2} \\
&= \frac{1}{4} + \frac{1}{8} \\
&= \frac{3}{8}
\end{aligned}$$

Note that $\frac{1}{8\pi} \int_{-\pi/2}^{\pi/2} \cos(4t) dt = 0$.

S2(c) $x(t) = \cos(2\pi t)u(t)$ is not a periodic signal since $x(t) = 0$ for $t < 0$.

S2(d) $x(t) = e^{j\pi t}$; $f = 0.5$; $T = 2$; and $x(t)$ is periodic.

$$\begin{aligned}
P &= \frac{1}{2} \int_{-1}^1 |e^{j\pi t}|^2 dt \\
&= \frac{1}{2} \int_{-1}^1 1 dt \\
&= \frac{1}{2} [t]_{-1}^1 \\
&= 1
\end{aligned}$$

$$S3(a) \quad \int_{-\infty}^t \cos(\tau)u(\tau)d\tau = \int_0^t \cos(\tau)d\tau = \sin(t)u(t)$$

Note that since $\cos(\tau)u(\tau)$ is zero for negative time, then the integration will yield $\sin(t)$ for positive time only.

$$S3(b) \quad \int_{-\infty}^t \cos(\tau)\delta(\tau)d\tau = \int_{-\infty}^t \cos(0)\delta(\tau)d\tau = \int_{-\infty}^t 1.\delta(\tau)d\tau = u(t)$$

$$S3(c) \quad \int_{-\infty}^{\infty} \cos(t)u(t-1)dt = \int_1^{\infty} \cos(t)dt = \infty$$

Note that integrating $\cos(t)$ from 1 to infinity will result in zero since the positive areas will cancel out the negative areas.

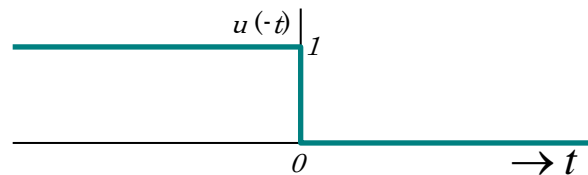
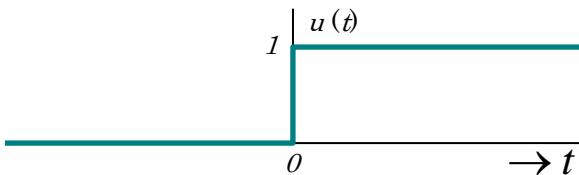
S3(d)

$$\begin{aligned} \int_0^{2\pi} t \cdot \sin\left(\frac{t}{2}\right) \delta(\pi - t) dt &= \int_0^{2\pi} t \cdot \sin\left(\frac{t}{2}\right) \delta(-(t - \pi)) dt \\ &= \int_0^{2\pi} \pi \sin\left(\frac{\pi}{2}\right) \delta(-(t - \pi)) dt \\ &= \pi \int_0^{2\pi} \delta(-(t - \pi)) dt \\ &= \pi \end{aligned}$$

Note that $\int_0^{2\pi} \delta(-(t - \pi)) dt = 1$, as it is the area within $\delta(-(t - \pi))$

$$S4 \quad x(t) = x_e(t) + x_o(t) \quad ; \quad x_e(t) = \frac{1}{2}[x(t) + x(-t)] \quad ; \quad x_o(t) = \frac{1}{2}[x(t) - x(-t)] \quad ; \quad \text{Given: } x(t) = u(t)$$

$$x_e(t) = \frac{1}{2}[u(t) + u(-t)] = \begin{cases} 1, & t = 0 \\ 0.5 & t \neq 0 \end{cases}$$



$$x_o(t) = \frac{1}{2}[u(t) - u(-t)] = \begin{cases} -0.5, & t < 0 \\ 0.5, & t \geq 0 \end{cases} = \frac{1}{2} \text{sgn}(t)$$

