#### Outline of Lecture

- Continuous-Frequency Spectrum (Fourier Transform)
  - Fourier Transform
- 2 Fourier Transform Properties
- Fourier Transform of Common Signals
  - Dirac- $\delta$  Function
  - DC Signal
  - Complex Exponential Signals
  - Cosine Functions
  - Sine Functions
  - Arbitrary Periodic Signals
  - Dirac Comb Function
- Fourier Transforms of Common Signals (Table)

## Continuous-Frequency Spectrum (Fourier Transform)

#### 1. Fourier Transform

- In the last lecture, we have shown that the discrete-frequency spectrum of a periodic signal,  $x_p(t)$  is given by its complex exponential Fourier series coefficient,  $c_k$ .
- ▶ The Fourier series expansion of an aperiodic signal does not exist. Instead we make use of another mathematical tool called Fourier transform to derive the spectrum of an aperiodic signal in the continuous-frequency domain, f.
- $\blacktriangleright$  The continuous-frequency spectrum of an aperiodic signal x(t) is given by its Fourier transform, X(f) :

$$\left.\begin{array}{l} \text{forward} \\ \text{Fourier transform} \\ \text{t-domain to f-domain} \end{array}\right\} X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt \tag{1}$$

$$\left.\begin{array}{l} \text{inverse} \\ \text{Fourier transform} \\ \text{f-domain to t-domain} \end{array}\right\}x(t)=\int_{-\infty}^{\infty}X(f)e^{j2\pi ft}df \tag{2}$$

lacktriangledown x(t) and X(f) are Fourier transform pairs, denoted by  $x(t) \hookrightarrow X(f)$ 

### Example 1

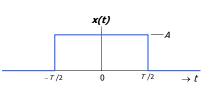
Spectrum of 
$$x(t) = Arect\left(\frac{t}{T}\right)$$

$$x(t) = Arect\left(\frac{t}{T}\right)$$

$$= \begin{cases} A & -\frac{T}{2} \le t < \frac{T}{2} \\ 0 & elsewhere \end{cases}$$

The Fourier transform of x(t) is :

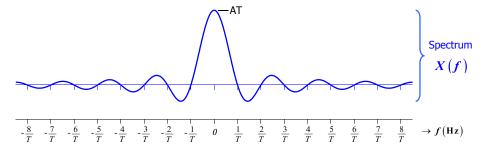
$$\begin{split} X(f) &= \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt \\ &= \int_{-0.5T}^{0.5T} A e^{-j2\pi f t} dt \\ &= \begin{cases} AT \frac{\sin(\pi f T)}{\pi f T} & f \neq 0 \\ AT & f = 0 \end{cases} \\ &= AT \text{sinc}(fT) \end{split}$$



$$\underbrace{\left[A\mathrm{rect}\left(\frac{t}{T}\right) \,\leftrightarrow\, AT\mathrm{sinc}(fT)\right]}$$

 $Fourier\ transform\ pair$ 

Since X(f) is real, the spectrum can be plotted in one single plot :



X(f) can also be written in terms of its magnitude and phase :

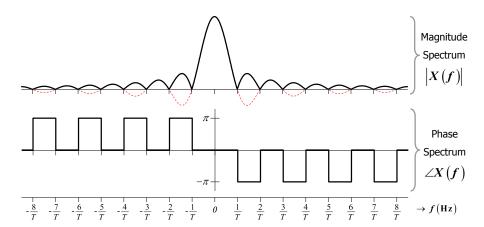
$$X(f) = AT \mathrm{sinc}(fT) = |X(f)| e^{j\angle X(f)} \text{ where } \left\{ \begin{array}{l} |X(f)| = AT |\mathrm{sinc}(fT)| \\ \angle X(f) = \left\{ \begin{array}{l} 0 & X(f) > 0 \\ \pm \pi & X(f) < 0 \end{array} \right. \end{array} \right.$$

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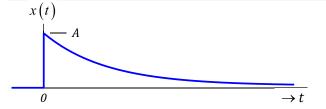
where the corresponding plots are :



#### Example 2

Spectrum of an exponentially decaying pulse :  $x(t) = Ae^{-\alpha t}u(t)$ 

$$x(t) = Ae^{-\alpha t}u(t) = \left\{ egin{array}{ll} Ae^{-\alpha t} & t \geq 0 \\ 0 & t < 0 \end{array} \right.$$
 ... assume that  $\alpha > 0$ 

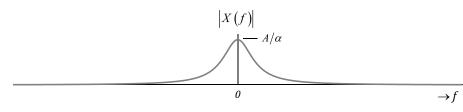


The Fourier transform of x(t) is as follows :

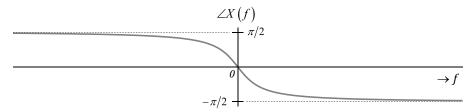
$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft}dt = \int_{0}^{\infty} Ae^{-\alpha t}e^{-j2\pi ft}dt$$

$$= \int_{0}^{\infty} Ae^{-(\alpha+j2\pi f)t}dt = A\left[\frac{e^{-(\alpha+j2\pi f)t}}{-(\alpha+j2\pi f)}\right]_{0}^{\infty} = \frac{A}{\alpha+j2\pi f}$$

 $\mbox{Magnitude spectrum}: \ |X(f)| = \sqrt{X(f)X^*(f)} = \frac{A}{\sqrt{\alpha^2 + 4\pi^2 f^2}}$ 



Phase spectrum : 
$$\angle X(f) = \tan^{-1}\left(\frac{Im[X(f)]}{Re[X(f)]}\right) = -\tan^{-1}\left(\frac{2\pi f}{\alpha}\right)$$



Spectrum of X(f) cannot be combined into one plot as X(f) is complex.

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2. Fourier Transform Properties

$$\text{Let } \left\{ \begin{array}{ll} X(f) = \mathcal{F}\left\{x(t)\right\} & \text{denote Fourier transform of } x(t) \\ x(t) \leftrightarrow X(f) & \text{denote Fourier transform pair} \end{array} \right.$$

## Property A (Linearity)

$$\alpha x_1(t) + \beta x_2(t) \leftrightarrow \alpha X_1(f) + \beta X_2(f)$$

#### Example 3

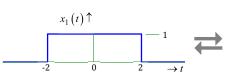
Find the Fourier transform of  $Y(f) = 0.5x_1(t) + 1.5x_2(t)$ .

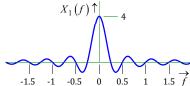
#### <u>Answer</u>: Applying linearity property:

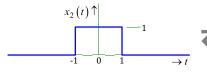
$$\begin{array}{lcl} Y(f) & = & 0.5X_1(f) + 1.5X_2(f) \\ & = & 0.5\left[4\operatorname{sinc}(4f)\right] + 1.5\left[2\operatorname{sinc}(2f)\right] \\ & = & 2\operatorname{sinc}(4f) + 3\operatorname{sinc}(2f) \end{array}$$

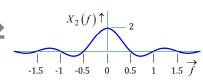
#### Time domain

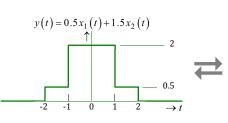
#### Frequency domain

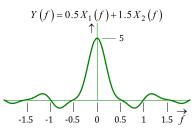












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## Property B (Time Scaling)

$$x(\beta t) \leftrightarrow \frac{1}{|\beta|} X\left(\frac{f}{\beta}\right)$$

#### Example 4

$$\mathit{Given}\left[x(t) = \mathit{rect}\left(t\right)\right] \leftrightarrow \left[X(f) = \mathit{sinc}\left(f\right)\right]$$

Find the Fourier 
$$\begin{cases} y_1(t) = x(0.5t) & \cdots \text{ expansion in time domain} \\ y_2(t) = x(2t) & \cdots \text{ compression in time domain} \end{cases}$$

#### Answer: Applying time scaling property:

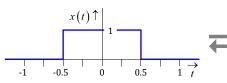
$$Y_1(f) = 2X(2f) = 2\operatorname{sinc}(2f)$$
 ... compression in frequency domain  $Y_2(f) = 0.5X(0.5f) = 0.5\operatorname{sinc}(0.5f)$  ... expansion in frequency domain

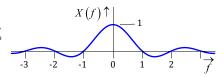
- Expansion in time domain leads to compression in frequency domain.
- Compression in time domain leads to expansion in frequency domain.
- Hence time-spread is inversely proportional to frequency-spread.

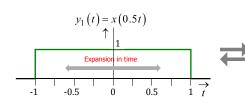
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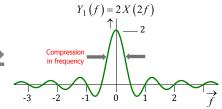
#### Time domain

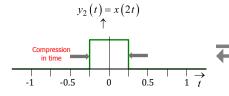
#### Frequency domain

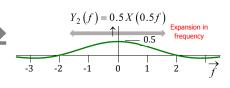












## Property C (Duality)

$$X(t) \leftrightarrow x(-f) \text{ or } X(-t) \leftrightarrow x(f)$$

#### Example 5

$$\mathit{Given}\left[x(t) = \tfrac{1}{2}\mathit{tri}\left(\tfrac{t}{2}\right)\right] \leftrightarrow \left[X(f) = \mathit{sinc}^2(2f)\right]$$

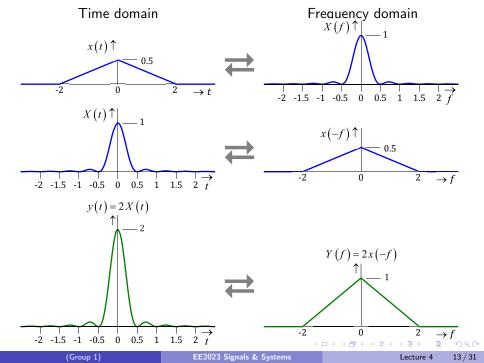
Find the Fourier transform of  $y(t) = 2 \operatorname{sinc}^2(2t)$ .

Answer: Note that y(t) = 2X(t). Applying duality property:

$$\underbrace{\left[\frac{1}{2}\operatorname{tri}\left(\frac{t}{2}\right) \leftrightarrow \widetilde{\operatorname{sinc}^2(2f)}\right]}_{\text{Given}} \overset{by \, duality}{\underset{property}{\longrightarrow}} \left[\underbrace{\frac{X(t)}{\operatorname{sinc}^2(2t)} \leftrightarrow \frac{1}{2}\operatorname{tri}\left(\frac{-f}{2}\right)}_{X(t)}\right]$$

Hence  $Y(f) = \mathcal{F}\left\{2X(t)\right\} = \mathcal{F}\left\{2\operatorname{sinc}^2(2t)\right\} = \operatorname{tri}\left(\frac{f}{2}\right)$ Note that the negative sign in  $\operatorname{tri}(.)$  is dropped because it is an even function i.e.  $\operatorname{tri}(-f) = \operatorname{tri}(f)$ .

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## Property D (Time Shifting)

$$x(t-t_0) \leftrightarrow X(f)e^{-j2\pi f t_0}$$

#### Example 6

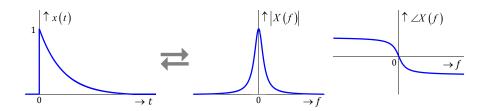
Given 
$$\left[x(t)=e^{-t}u(t)\right]\leftrightarrow \left[X(f)=\frac{1}{1+j2\pi f}\right]$$

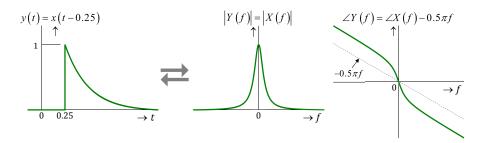
Find the Fourier transform of y(t) = x(t - 0.25).

#### Answer: Applying the time-shifting property:

$$\begin{split} Y(f) &= \mathcal{F}\left\{x(t-0.25)\right\} = X(f)e^{-j2\pi f(0.25)} \\ |Y(f)| &= |X(f)|.\underbrace{\left|e^{-j2\pi f(0.25)}\right|}_{1} = |X(f)| = \frac{1}{\sqrt{1+4\pi^2 f^2}} \\ \angle Y(f) &= \angle X(f) + \angle e^{-j2\pi f(0.25)} \end{split}$$

$$\angle Y(f) = \angle X(f) + \angle e^{-j2\pi f(0.25)}$$
  
=  $\angle X(f) - 0.5\pi f$   
=  $-\tan^{-1}(2\pi f) - 0.5\pi f$ 





Notice that the time shifted function y(t) has the same magnitude spectrum as x(t) but the phase spectrum is shifted by  $-2\pi f t_0 = -0.5\pi f$ .

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## Property E (Frequency Shifting (Modulation))

$$x(t)e^{j2\pi f_0t} \leftrightarrow X(f-f_0)$$

#### Example 7

$$\mathit{Given}\left[x(t) = \frac{1}{\sqrt{2\pi}}e^{-\frac{t^2}{2}}\right] \leftrightarrow \left[X(f) = e^{-2\pi^2 f^2}\right]$$

Find the Fourier transform of 
$$y(t) = x(t)e^{j2\pi 5t}$$
 
$$= \underbrace{x(t)\cos(2\pi 5t)}_{\text{Re}[y(t)]} + \underbrace{jx(t)\sin(2\pi 5t)}_{\text{Im}[y(t)]}$$

Answer: Applying the frequency-shifting property:

$$Y(f) = \mathcal{F} \{x(t)e^{j2\pi 5t}\}$$

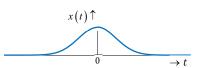
$$= X(f-5)$$

$$= e^{-2\pi^2(f-5)^2}$$

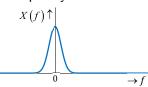
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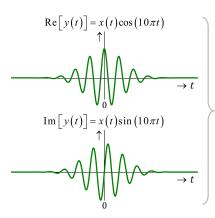
#### Time domain

#### Frequency domain

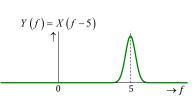












## Property F (Differentiation in the Time Domain)

$$\frac{d}{dt}x(t) \leftrightarrow j2\pi f X(f)$$

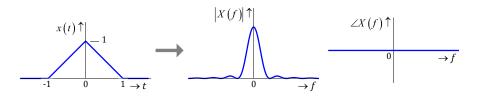
#### Example 8

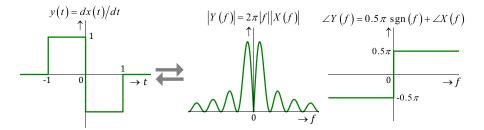
Given 
$$[x(t) = tri(t)] \leftrightarrow [X(f) = sinc^2(f)]$$
  
Find the Fourier transform of  $y(t) = \frac{dx(t)}{dt}$ .

<u>Answer</u>: Applying the differentiation in time domain property:

$$\begin{split} Y(f) &= \mathcal{F}\left\{\frac{dx(t)}{dt}\right\} = j2\pi f X(f) \\ |Y(f)| &= 2\pi |f| |X(f)| = 2\pi |f| \mathrm{sinc}^2(f) \\ \angle Y(f) &= \angle j2\pi f + \angle X(f) \\ &= 0.5\pi \mathrm{sgn}\left(f\right) + \underbrace{\angle X(f)}_{0} \\ &= 0.5\pi \mathrm{sgn}\left(f\right) \end{split} \qquad \qquad \begin{aligned} & \mathrm{sgn}(f) = \left\{\begin{array}{cc} +1 & f > 0 \\ -1 & f < 0 \end{array}\right. \\ \angle j2\pi f &= \angle jf \\ &= \left\{\begin{array}{cc} 0.5\pi & f > 0 \\ -0.5\pi & f < 0 \end{array}\right. \\ &= 0.5\pi \mathrm{sgn}\left(f\right) \end{aligned}$$

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Property G (Integration in the Time Domain)

$$\int_{-\infty}^t x(\tau)d\tau \leftrightarrow \frac{1}{j2\pi f}X(f) + \frac{1}{2}X(0)\delta(f)$$

where  $\delta(f)$  is the unit impulse and  $X(0) = \int_{-\infty}^{\infty} x(t)dt$ .

#### Example 9

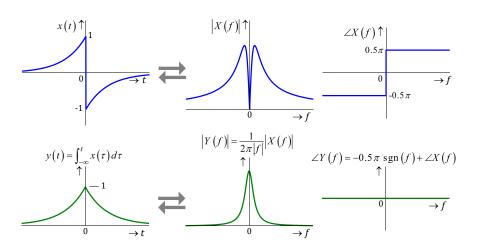
$$\left[x(t)=e^t u(-t)-e^{-t} u(t)\right] \leftrightarrow X(f) = \frac{j4\pi f}{1+4\pi^2 f^2} \rightarrow \begin{array}{c} |X(f)|=\frac{4\pi |f|}{1+4\pi^2 f^2} \\ \angle X(f)=\frac{1}{2}\pi \mathrm{sgn}(f) \end{array}$$

Find the Fourier transform of  $y(t) = \int_{-\tau}^{\tau} x(\tau)d\tau$ .

### Answer: Applying the integration in the time domain property:

$$\begin{split} Y(f) &= \mathcal{F}\left\{\int_{-\infty}^t x(\tau)d\tau\right\} = \frac{1}{j2\pi f}X(f) + \frac{1}{2}\underbrace{X(0)}_0\delta(f) = \frac{2}{1+4\pi^2f^2}\\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f) = 0 \text{ because } Y(f) > 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \text{ because } Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi^2f^2}, \quad \angle Y(f)| = 0 \; \forall \; f \in \mathbb{R} \\ |Y(f)| &= \frac{2}{1+4\pi$$

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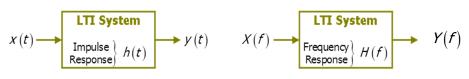
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### Property H (Convolution in Time equiv. to Multiplication in Freq.)

$$\underbrace{\int_{-\infty}^{\infty} x_1(\tau) x_2(t-\tau) d\tau}_{x_1(t) * x_2(t)} \leftrightarrow \underbrace{X_1(\mathbf{f}).X_2(f)}_{\textit{multiplication in } \textit{f}}$$

#### Note that

- \* denotes the convolution operator
- $w(t) * v(t) = \int_{-\infty}^{\infty} w(\tau)v(t-\tau)d\tau = \int_{-\infty}^{\infty} v(\tau)w(t-\tau)d\tau$
- This property is used extensively in the study of linear time-invariant systems.



## TIME-DOMAIN convolution

$$y(t) = x(t) * h(t)$$

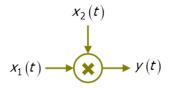
# FREQUENCY-DOMAIN multiplication

$$Y(f) = H(f)X(f)$$

### Property I (Multiplication in Time equiv. to Convolution in Freq.)

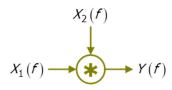
$$\underbrace{x_1(t).x_2(t)}_{\text{multiplication in time}} \leftrightarrow \underbrace{\int_{-\infty}^{\infty} X_1(\tau) X_2(f-\tau) d\tau}_{X_1(f)*X_2(f)}$$

This property is used extensively in radio communication systems.



# TIME-DOMAIN multiplication

$$y(t) = x_1(t).x_2(t)$$



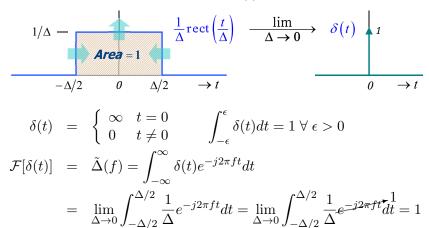
# FREQUENCY-DOMAIN convolution

$$Y(f) = X_1(f) * X_2(f)$$

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## Fourier Transform of Common Signals

1. The unit impulse or Dirac- $\delta$  function,  $\delta(t)$  defined in Lecture 2.



Therefore  $\mathcal{F}[\delta(t)] = 1$ .

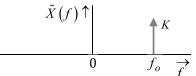
2. The DC signal  $x_{dc}(t) = K$ .

$$\mathcal{F}[K\delta(t)] = K \xrightarrow{\text{duality}} \mathcal{F}[K] = K\delta(f) \qquad X_{dc}(f) \uparrow \\ \text{Therefore} \\ X_{dc}(f) = \mathcal{F}[x_{dc}(t) = K] = K\delta(f) \qquad 0 \qquad f_o \xrightarrow{f}$$

3. Complex exponential signal

$$\tilde{x}(t) = Ke^{j2\pi f_0 t}$$

Therefore



$$\underbrace{\tilde{X}(f) = \mathcal{F}[\tilde{x}(t)] = X_{dc}(f - f_0)}_{= K\delta(f - f_0)} = K\delta(f - f_0)$$

frequency-shifting property

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4. Cosine :  $x_c(t) = K \cos(2\pi f_0 t)$ .

$$x_c(t) = K \cos(2\pi f_0 t)$$

$$= \frac{K}{2} \left[ e^{j2\pi f_0 t} + e^{-j2\pi f_0 t} \right]$$

$$X_c(f) = \frac{K}{2} \left[ \delta(f - f_0) + \delta(f + f_0) \right]$$

$$X_c(f) = \frac{K}{2} \left[ \delta(f - f_0) + \delta(f + f_0) \right]$$

$$\begin{array}{c|c}
X_c(f) \uparrow \\
\hline
K/2 & & & \downarrow \\
-f_o & 0 & f_o \xrightarrow{f}
\end{array}$$

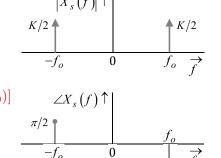
5. Sine : 
$$x_s(t) = K \sin(2\pi f_0 t)$$
.

$$x_s(t) = K \sin(2\pi f_0 t)$$
$$= \frac{K}{2i} \left[ e^{j2\pi f_0 t} - e^{-j2\pi f_0 t} \right]$$

$$X_{s}(f) = \frac{K}{2j} [\delta(f - f_{0}) - \delta(f + f_{0})]$$

$$= \frac{K}{2} e^{-j\frac{\pi}{2}} \delta(f - f_{0})$$

$$+ \frac{K}{2} e^{j\frac{\pi}{2}} \delta(f + f_{0})$$



(Group 1)

6. Arbitrary Periodic Signal,  $x_p(t)$ , period  $T_p$ 

$$\underbrace{x_p(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi f_p t}}_{\text{Fourier series expansion}} \underbrace{c_k = \frac{1}{T_p} \int_{-\frac{T_p}{2}}^{\frac{T_p}{2}} x_p(t) e^{-j2\pi f_p t} dt}_{\text{Fourier series coefficients}}$$

applying Fourier transform to  $x_n(t)$ ,

$$X_p(f) = \mathcal{F}[x_p(t)] = \mathcal{F}\left[\sum_{k=-\infty}^{\infty} c_k e^{j2\pi f_p t}\right] = \sum_{k=-\infty}^{\infty} c_k \mathcal{F}\left[e^{j2\pi f_p t}\right]$$

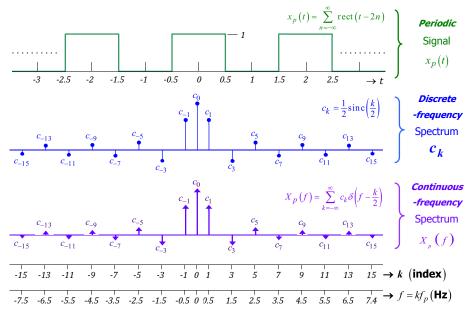
Substituting  $\mathcal{F}\left[e^{j2\pi f_pt}\right]=\delta(f-\mathbf{k}f_p)$ , we get :

$$X_p(f) = \sum_{k = -\infty}^{\infty} c_k \delta(f - kf_p)$$
 (3)

Important Note The Fourier transform  $X_p(f)$  of any periodic signal  $x_p(t)$  can be obtained by first computing the Fourier series coefficient  $c_k$  and substituting into (3).

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Illustrating using the square wave you have seen before :  $T_p=2, f_p=0.5\ \mathrm{Hz}$ 



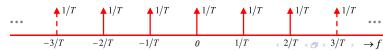
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7. An interesting periodic signal which we will encounter often is the Dirac Comb, defined as  $s_T(t) = \sum_n \delta(t-nT)$  where T is the fundamental period.

The discrete-frequency spectrum  $c_k$  [Fourier series coeff of  $s_T(t)$ ]

The continuous-freq. spectrum  $S_T(f)$  [Fourier transform of  $s_T(t)$ ]

$$S_T(f) = \sum_k c_k \delta(f - \frac{k}{T}) = \frac{1}{T} \sum_k \delta(f - \frac{k}{T})$$



## Fourier Transforms of Common Signals

Signal	x(t)	X(f)
Constant (DC)	K	$K\delta(f)$
Impulse	$\delta(t)$	1
Step function	u(t)	$\frac{1}{2}\left[\delta(f) + \frac{1}{j\pi f}\right]$
Sign or Signum	$\operatorname{sgn}(t)$	$\frac{1}{j\pi f}$
Rectangular	$\operatorname{rect}\left(rac{t}{T} ight)$	Tsinc $(Tf)$
Triangular	$tri\left(rac{t}{T} ight)$	$T$ sinc $^2$ $(Tf)$
Cardinal sine	$sinc\left(rac{t}{T} ight)$	Trect $(Tf)$
Complex exponential	$e^{j2\pi f_0 t}$	$\delta(f-f_0)$
Cosine	$\cos(2\pi f_0 t)$	$0.5[\delta(f - f_0) + \delta(f + f_0)]$
Sine	$\sin(2\pi f_0 t)$	$-0.5j[\delta(f-f_0)-\delta(f+f_0)]$
Gaussian	$e^{-\left(\frac{t^2}{\alpha^2}\right)}$	$\alpha\sqrt{0.5}e^{-\alpha^2\pi^2f^2}$
Comb	$\sum_{k} \delta(t - kT)$	$\frac{1}{T}\sum_{k}\delta\left(f-\frac{k}{T}\right)$

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#### Exercise 1

The waveforms of 4 pulses,  $x_1(t)$ ,  $x_2(t)$ ,  $x_3(t)$  and  $x_4(t)$  are shown below. Derive the spectrum  $X_1(f)$ ,  $X_2(f)$ ,  $X_3(f)$  and  $X_4(f)$  and express the spectra of  $X_2(f)$ ,  $X_3(f)$  and  $X_4(f)$  in terms of  $X_1(f)$ .

