

EE2023/TEE2023 TUTORIAL 2 (SOLUTIONS)**Solution to Q.1**

Description of $x(t)$:

- $x(t)$ is a REAL & EVEN function of t \therefore Spectrum is REAL and SYMMETRIC
- $x(t)$ has an average (or DC) value of 2 \therefore Zero-frequency component has value 2
- $x(t)$ is APERIODIC $\therefore \{\pi, \pi^2, \pi^3\} \dots$ has no common factor
- $x(t)$ is a POWER SIGNAL \therefore $\begin{cases} \text{Spectrum is defined only at discrete} \\ \text{frequency points (sum of sinusoids)} \end{cases}$

Since $x(t)$ is non-periodic, it does not have a Fourier series expansion.

Solution to Q.2

- (a) The fundamental frequency of $x(t) = 6\sin(12\pi t) + 4\exp\left(j\left(8\pi t + \frac{\pi}{4}\right)\right) + 2$ is $\begin{cases} f_p = \text{HCF}\{6, 4\} = 2 \\ T_p = 0.5 \end{cases}$.

Re-write $x(t)$ as a sum of complex exponentials:

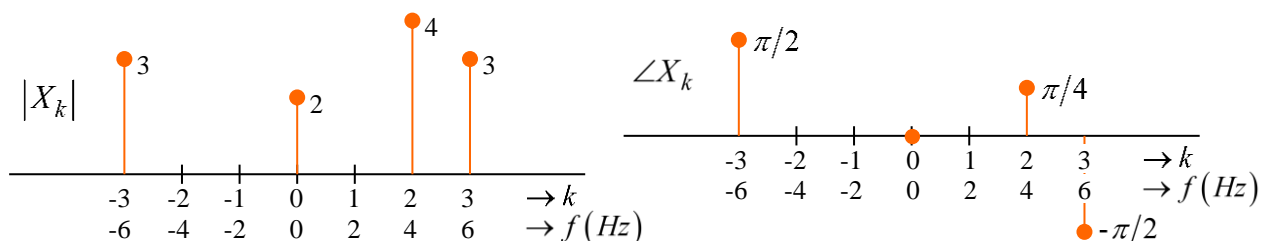
$$\begin{aligned} x(t) &= \frac{6}{j2} [\exp(j12\pi t) - \exp(-j12\pi t)] + 4\exp(j\pi/4)\exp(j8\pi t) + 2 \\ &= j3\exp(-j12\pi t) + 2 + 4\exp(j\pi/4)\exp(j8\pi t) - j3\exp(j12\pi t) \end{aligned} \quad (1)$$

Express $x(t)$ as a complex exponential Fourier series:

$$\begin{aligned} x(t) &= \sum_{k=-\infty}^{\infty} X_k \exp\left(j2\pi \frac{k}{T_p} t\right) = \sum_{k=-\infty}^{\infty} X_k \exp(j4\pi k t) \\ &= \begin{pmatrix} \dots + X_{-3} \exp(-j12\pi t) + X_{-2} \exp(-j8\pi t) + X_{-1} \exp(-j4\pi t) \\ + X_0 \\ + X_1 \exp(j4\pi t) + X_2 \exp(j8\pi t) + X_3 \exp(j12\pi t) + \dots \end{pmatrix} \end{aligned} \quad (2)$$

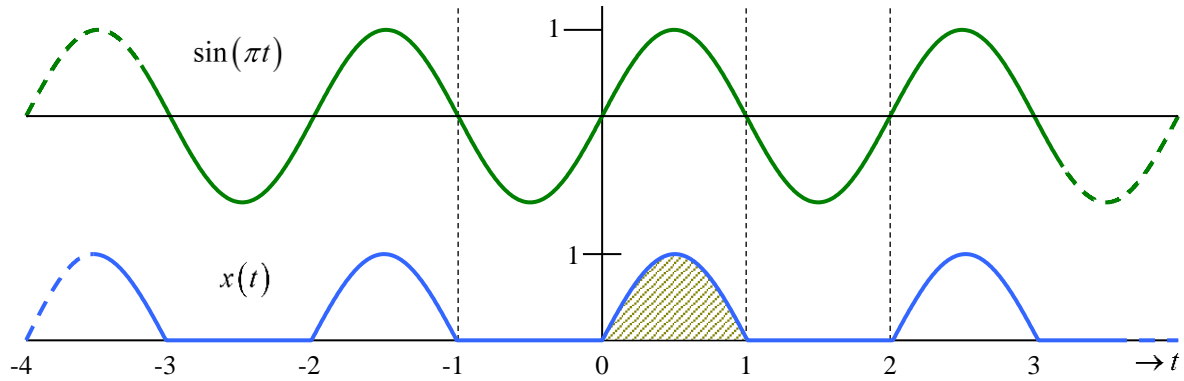
Comparing coefficients of complex exponential terms in (1) and (2), we conclude that:

$$X_{-3} = j3, \quad X_0 = 2, \quad X_2 = 4\exp\left(j\frac{\pi}{4}\right), \quad X_3 = -j3 \quad \text{and} \quad [X_k = 0; k \neq 0, 2, \pm 3].$$



Remarks: If a periodic signal is given as a sum of sinusoids, then its Fourier series coefficients can be evaluated using the above method without the need to perform any integration.

(b) $x(t) = \frac{1}{2}(|\sin(\pi t)| + \sin(\pi t))$: Half-wave rectification of $\sin(\pi t)$.



Period of $x(t)$: $T = 2$; Fundamental frequency = 0.5

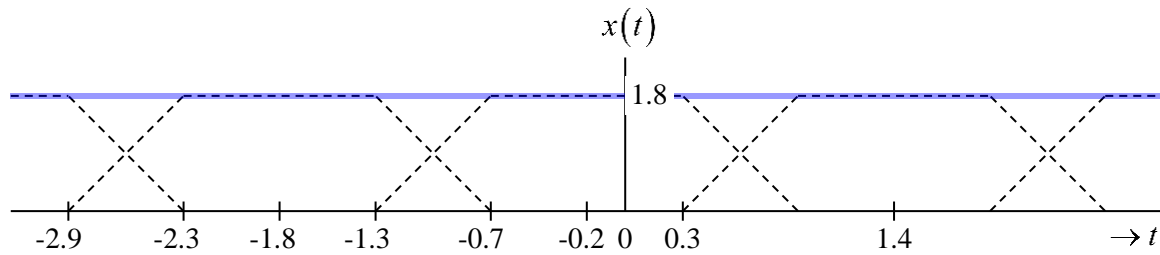
Frequency of 4th harmonic = $4 \times 0.5 = 2$ Hz

The Fourier series coefficients, X_k , can also be calculated as follows, though this is not required in the question:

$$\begin{aligned}
 X_k &= \frac{1}{T} \int_0^T x(t) \exp(-j2\pi kt/T) dt = \frac{1}{2} \int_0^2 x(t) \exp(-j\pi kt) dt \\
 &= \frac{1}{2} \int_0^1 \sin(\pi t) \exp(-j\pi kt) dt = \frac{1}{2} \int_0^1 \frac{1}{j2} [\exp(j\pi t) - \exp(-j\pi t)] \exp(-j\pi kt) dt \\
 &= \frac{1}{j4} \int_0^1 \exp(-j\pi(k-1)t) - \exp(-j\pi(k+1)t) dt \\
 &= \frac{1}{j4} \left[\frac{\exp(-j\pi(k-1)t)}{-j\pi(k-1)} - \frac{\exp(-j\pi(k+1)t)}{-j\pi(k+1)} \right]_0^1 \\
 &= \frac{1}{j4} \left[\exp(-j\pi k) \left(\frac{\exp(j\pi)}{-j\pi(k-1)} - \frac{\exp(-j\pi)}{-j\pi(k+1)} \right) - \left(\frac{1}{-j\pi(k-1)} - \frac{1}{-j\pi(k+1)} \right) \right] \\
 &= \frac{1}{j4} \left[(-1)^k \left(\frac{1}{j\pi(k-1)} - \frac{1}{j\pi(k+1)} \right) + \left(\frac{1}{j\pi(k-1)} - \frac{1}{j\pi(k+1)} \right) \right] \\
 &= \frac{(-1)^k}{4\pi} \left[-\frac{1}{(k-1)} + \frac{1}{(k+1)} \right] + \frac{1}{4\pi} \left[-\frac{1}{(k-1)} + \frac{1}{(k+1)} \right] \\
 &= \frac{(-1)^k}{4\pi} \left[-\frac{(k+1)}{(k^2-1)} + \frac{(k-1)}{(k^2+1)} \right] + \frac{1}{4\pi} \left[-\frac{(k+1)}{(k^2-1)} + \frac{(k-1)}{(k^2+1)} \right] \\
 &= \frac{(-1)^k}{4\pi} \left[\frac{2}{(1-k^2)} \right] + \frac{1}{4\pi} \left[\frac{2}{(1-k^2)} \right] \\
 &= \begin{cases} \frac{1}{\pi(1-k^2)}; & k = \text{even} \\ 0; & k = \text{odd} \end{cases} \\
 &= \frac{1+(-1)^k}{2\pi(1-k^2)}
 \end{aligned}$$

Solution to Q.3

Graphically, we observe that $x(t) = \sum_{n=-\infty}^{\infty} 2p(t - 1.6n) = 1.8$.



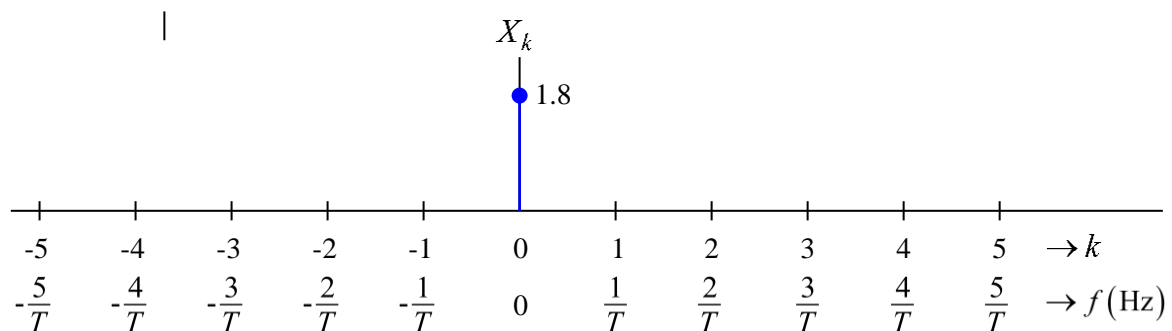
By Deduction:

- $x(t)$ has a *zero-frequency* component of value 1.8, which implies that $X_0 = 1.8$.
- $x(t)$ has no *non-zero frequency* components, which implies that $X_k = 0$; $k \neq 0$.

By Derivation:

Since $x(t)$ is a constant (or a DC signal), it may be treated as a periodic signal of arbitrary period T , where $0 < T < \infty$. Its Fourier series coefficients can thus be computed as

$$\begin{aligned}
 X_k &= \frac{1}{T} \int_{-T/2}^{T/2} 1.8 \exp\left(-j2\pi \frac{k}{T} t\right) dt \\
 &= \frac{1.8}{T} \left[\frac{\exp(-j2\pi kt/T)}{-j2\pi k/T} \right]_{-T/2}^{T/2} \\
 &= \frac{1.8}{T} \left[\frac{\exp(-j\pi k)}{-j2\pi k/T} - \frac{\exp(j\pi k)}{-j2\pi k/T} \right] \\
 &= 1.8 \frac{\sin(\pi k)}{\pi k} \\
 &= 1.8 \operatorname{sinc}(k) \\
 &= \begin{cases} 1.8; & k = 0 \\ 0; & k \neq 0 \end{cases}
 \end{aligned}$$



Solution to Q.4

This is the continuous frequency spectrum of a periodic signal. Recall that for a periodic signal, $x_p(t)$, we have following representations :

Fourier series (discrete frequency spectrum) : $x_p(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_p t}$

Fourier transform (continuous frequency spectrum) : $X_p(f) = \sum_{k=-\infty}^{\infty} c_k \delta(f - k f_p)$

In both representations, c_k are the Fourier series coefficients.

Given the spectrum in the figure, we recognize that $X(f)$ is the continuous frequency spectrum and hence $X(f)$ can be written as :

$$X(f) = 2e^{j0.25\pi}\delta(f + 32) + 8e^{j0.5\pi}\delta(f + 24) + 5\delta(f + 16) + 5\delta(f - 16) + 8e^{-j0.5\pi}\delta(f - 24) + 2e^{-j0.25\pi}\delta(f - 32)$$

Recall that by applying the frequency shifting property of the Fourier Transform, you get

$$\mathcal{F}\{e^{j2\pi f_0 t}\} = \delta(f - f_0) \text{ since } \mathcal{F}\{1\} = \delta(f).$$

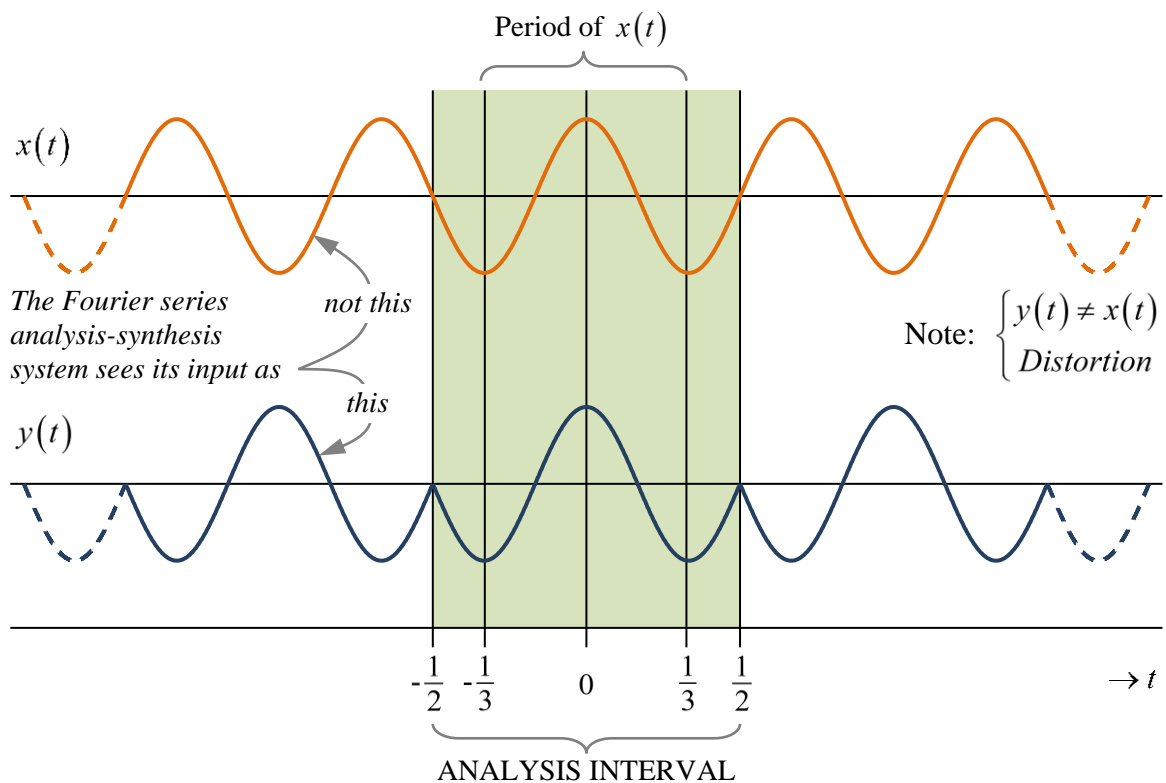
Hence applying this property to $X(f)$, we get

$$\begin{aligned} x(t) &= 2e^{j0.25\pi}e^{-j2\pi 32t} + 8e^{j0.5\pi}e^{-j2\pi 24t} + 5e^{-j2\pi 16t} + 5e^{j2\pi 16t} + 8e^{-j0.5\pi}e^{j2\pi 24t} + 2e^{-j0.25\pi}e^{j2\pi 32t} \\ &= 4\cos(64\pi t - 0.25\pi) + 16\cos(48\pi t - 0.5\pi) + 10\cos(32\pi t) \\ &= 4\cos(64\pi t - 0.25\pi) + 16\sin(48\pi t) + 10\cos(32\pi t) \end{aligned}$$

Hence $x(t)$ consists of only real sinusoids.

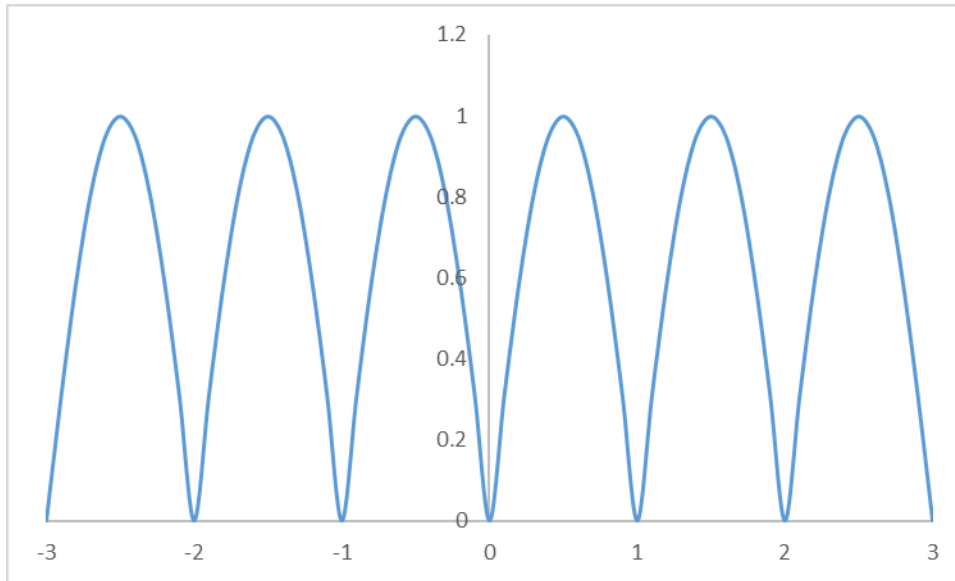
Solution to Q.5

- (a) The computation of the Fourier series coefficients, c_k , assumes that $x(t)$ has a period of 1 over the interval $[-0.5, 0.5]$.
- (b) The synthesis of $y(t)$ uses c_k as Fourier series coefficients to synthesize a periodic signal of period equal to 1.
- (c) The first graph illustrates the signal $x(t) = \cos(3\pi t)$, which has a period of $2/3$ seconds. However, since the analysis and synthesis systems are based on a periodic signal with period of 1 second, then the portion of $x(t)$ of duration 1 second is analysed and synthesized to give the output shown in the lower graph.



Supplementary Questions (Solutions)

S1 The signal $x(t) = |\sin(\pi t)|$ is as follows:



Hence it has a period of 1 second.

$$c_0 = \int_0^1 \sin(\pi t) dt = \frac{1}{\pi} [\cos(\pi t)]_0^1 = -\frac{2}{\pi}$$

$$\begin{aligned} c_k &= \int_0^1 \sin(\pi t) e^{-j2\pi kt} dt \\ &= \int_0^1 \frac{1}{2j} [e^{j\pi t} - e^{-j\pi t}] e^{-j2\pi kt} dt \\ &= \frac{1}{2j} \int_0^1 [e^{j\pi t(1-2k)} - e^{-j\pi t(1+2k)}] dt \\ &= \frac{1}{2j} \left[\frac{e^{j\pi t(1-2k)}}{j\pi(1-2k)} + \frac{e^{-j\pi t(1+2k)}}{-j\pi(1+2k)} \right]_0^1 \\ &= -\frac{1}{2\pi} \left[\frac{e^{j\pi(1-2k)} - 1}{(1-2k)} + \frac{e^{-j\pi(1+2k)} - 1}{(1+2k)} \right] \\ &= -\frac{1}{2\pi} \left[\frac{-1-1}{(1-2k)} + \frac{-1-1}{(1+2k)} \right] \\ &= -\frac{1}{2\pi} \left[\frac{-2}{(1-2k)} + \frac{-2}{(1+2k)} \right] \\ &= -\frac{1}{2\pi} \left[\frac{-2(1+2k) - 2(1-2k)}{1-4k^2} \right] \\ &= -\frac{1}{2\pi} \cdot \frac{4}{4k^2 - 1} \end{aligned}$$

$$\text{Hence, we have: } x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi kt} = -\frac{2}{\pi} \sum_{k=-\infty}^{\infty} \frac{1}{4k^2 - 1} e^{j2\pi kt}$$

We have:

$$c_0 = a_0 = \frac{2}{\pi}$$

$$A_k = |c_k| = \frac{2}{\pi} \frac{1}{4k^2 - 1}$$

$$\angle c_k = \theta_k = -\pi$$

$$\begin{aligned} x(t) &= a_0 + 2 \sum A_k \cos(2\pi kt + \theta_k) \\ &= \frac{2}{\pi} + 2 \sum_{k=1}^{\infty} \frac{2}{\pi} \frac{1}{4k^2 - 1} \cos(2\pi kt - \pi) \\ &= \frac{2}{\pi} - \sum_{k=1}^{\infty} \frac{4}{\pi} \frac{1}{4k^2 - 1} \cos(2\pi kt - \pi) \end{aligned}$$

S2 Given $x(t) = t^2$; $-\pi < t < \pi$ and $x(t + 2\pi) = x(t)$, we have a periodic signal with period of 2π .

$$c_0 = \frac{1}{2\pi} \int_{-\pi}^{\pi} t^2 dt = \frac{1}{2\pi} \left[\frac{1}{3} t^3 \right]_{-\pi}^{\pi} = \frac{1}{6\pi} [\pi^3 + \pi^3] = \frac{\pi^2}{3}$$

$$c_k = \frac{1}{2\pi} \int_{-\pi}^{\pi} t^2 e^{-j2\pi \left(\frac{k}{2\pi}\right)t} dt = \frac{1}{2\pi} \int_{-\pi}^{\pi} t^2 e^{-jkt} dt$$

Consider $\int t^2 e^{-jkt} dt$, we have:

$$\begin{aligned} \int t^2 e^{-jkt} dt &= t^2 \left(\frac{e^{-jkt}}{-jk} \right) - \int \frac{e^{-jkt}}{jk} \cdot 2t \cdot dt \\ &= \frac{jt^2}{k} e^{-jkt} - \frac{2j}{k} \int t e^{-jkt} dt \\ &= \frac{jt^2}{k} e^{-jkt} - \frac{2j}{k} \left[t \frac{e^{-jkt}}{-jk} - \int \frac{e^{-jkt}}{-jk} dt \right] \\ &= \frac{jt^2}{k} e^{-jkt} - \frac{2j}{k} \left[-\frac{te^{-jkt}}{jk} + \frac{1}{jk} \frac{e^{-jkt}}{-jk} \right] \\ &= \frac{jt^2}{k} e^{-jkt} + \frac{2}{k^2} t e^{-jkt} - \frac{2j}{k^3} e^{-jkt} \end{aligned}$$

Hence:

$$\begin{aligned} c_k &= \frac{1}{2\pi} \left[\frac{jt^2}{k} e^{-jkt} + \frac{2}{k^2} t e^{-jkt} - \frac{2j}{k^3} e^{-jkt} \right]_{-\pi}^{\pi} \\ &= \frac{1}{2\pi} \left[\frac{j\pi^2}{k} e^{-j\pi k} + \frac{2\pi}{k^2} e^{-j\pi k} - \frac{2j}{k^3} e^{-j\pi k} - \frac{j\pi^2}{k} e^{j\pi k} + \frac{2\pi}{k^2} e^{j\pi k} - \frac{2j}{k^3} e^{j\pi k} \right] \\ &= \frac{1}{2\pi} \left[\frac{2\pi}{k^2} e^{-j\pi k} + \frac{2\pi}{k^2} e^{j\pi k} \right] = \frac{1}{2\pi} \left[\frac{4\pi}{k^2} (-1)^k \right] = \frac{2}{k^2} (-1)^k \end{aligned}$$

Hence we have: $x(t) = \frac{\pi^2}{3} + \sum_k \frac{2}{k^2} (-1)^k e^{jkt}$

$$\text{S3} \quad x(t) = c_0 + \sum c_k \cos\left(2\pi \frac{k}{T_0} t - \theta_k\right)$$

$$\text{Hence: } X_0 = c_0 \text{ and } \left|X_k\right| = \frac{1}{2} c_k \text{ and } \angle X_k = \tan(-\theta_k) = -\tan(\theta_k)$$