## **ELEC 341**

# **MATLAB Project**

Submitted to Prof. Jane Wang

Apr 10, 2020

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Note that all MATLAB code is attached to the end of this report.

Problems are in Modern Control Systems 12<sup>th</sup> edition--just realized right before the deadline-The numbers given in Problem 2, 5, 10 may be slightly different in 12<sup>th</sup>&13<sup>th</sup> versions.

### **Problem 1. CP7.10**

a)

Characteristic Equation:

$$det[sI - A] = s^3 + (2 + k)s^2 + 5s + 1 = 0$$

b)

The Routh Array:

$$s^3$$
 1
 5

  $s^2$ 
 $2+k$ 
 1

  $s^1$ 
 $5-\frac{1}{2+k}$ 
 0

  $s^0$ 
 1
 0

To make the system stable:

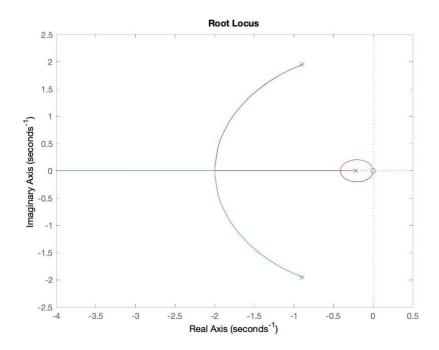
$$2+k>0$$
,  $5-\frac{1}{2+k}>0$ 

Thus,

$$k > -\frac{9}{5}$$

c)

**Root Locus:** 



For k > 0, the system is always stable since the plot is on the left-hand side of jw-axis.

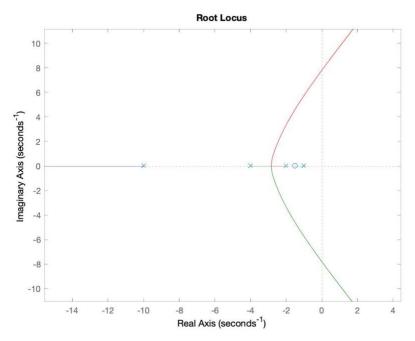
### Problem 2. DP7.12

a)

Characteristic Equation:

 $1 + K * \frac{s + 1.5}{(s + 1)(s + 2)(s + 4)(s + 10)} = 0$ 

**Root Locus:** 



When K = 100, roots are:

$$s = -11.38, -1.45, -2.09 \pm j3.10$$

When K = 300, roots are:

$$s = -12.94, -1.48, -1.29 \pm j5.10$$

When K = 600, roots are:

$$s = -14.44, -1.49, -0.53 \pm j6.72$$

b)

When K=100,

$$T_s = \frac{4}{\zeta \omega_n} = \frac{4}{2.09} \cong 1.91 \, s$$

$$\omega_n = \sqrt{2.09^2 + 3.1^2} = 3.74$$

$$\zeta = \frac{2.09}{3.74} = 0.56$$

Thus,

$$P.O.\% = 13\%$$

Steady-state error:

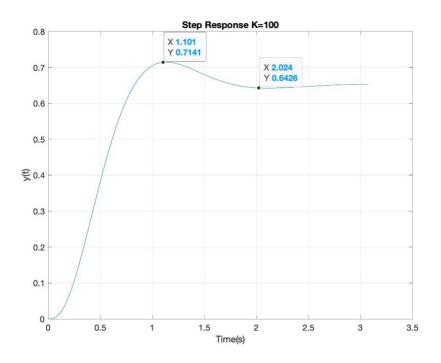
$$e_{ss} = \lim_{s \to 0} s * \frac{1}{s} * T(s) = \frac{\frac{1.5}{80} * 100}{1 + \frac{1.5}{80} * 100} = 0.65$$

We then apply the same calculation to K=300, 600.

c)

$$T(s) = \frac{K * (s + 1.5)}{(s + 1)(s + 2)(s + 4)(s + 10) + K * (s + 1.5)}$$
$$= \frac{Ks + 1.5K}{s^4 + 17s^3 + 84s^2 + (148 + K)s + 80 + 1.5K}$$

When K=100,

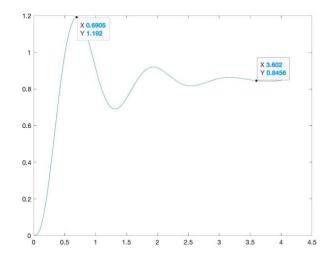


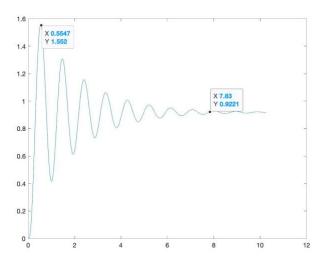
From the graph:

$$O.P.\% \cong 11.12\%$$
,  $T_s \cong 2 s$ 

The results are pretty close to the predicted values from part b.

Here are also the plots for K=300, 600. Again, the O.P. and Ts are similar to what we obtain from part b.



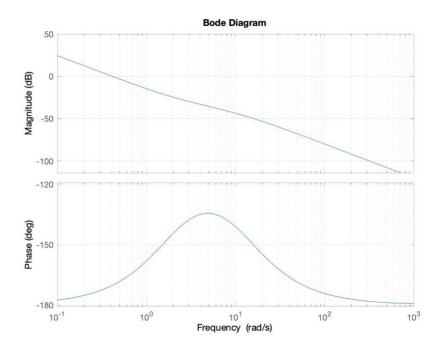


# Problem 3. DP8.1

a)

$$G_c(s) * G(s) = \frac{s+2}{s^2(s+12)}$$

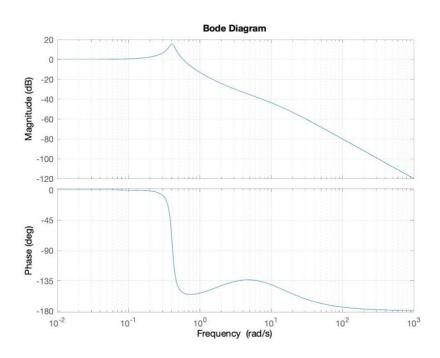
**Bode Plots:** 



b)

$$T(s) = \frac{s+2}{s^3 + 12s^2 + s + 2}$$

Bode Plots:

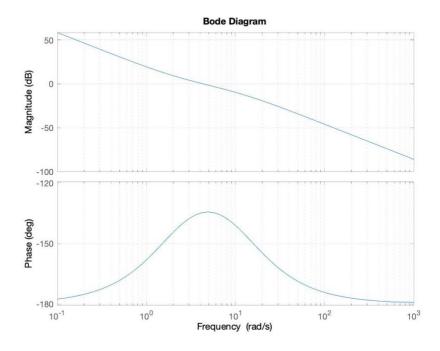


c)

K=50:

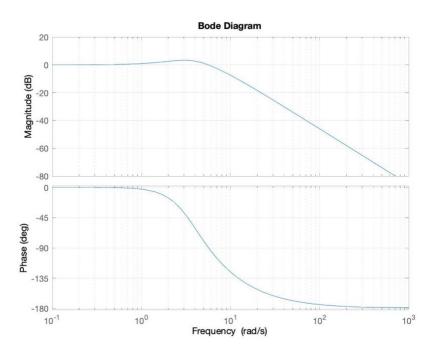
$$G_c(s) * G(s) = \frac{50s + 100}{s^2(s + 12)}$$

**Bode Plots:** 



$$T(s) = \frac{50s + 100}{s^3 + 12s^2 + 50s + 100}$$

**Bode Plots:** 



d)

So that:

 $M_p \le 2$ 

Due to the fact that the system has to be stable:

$$14 \le K \le 350$$

We choose:

$$K=350$$
,  $\omega_B=29 \ rad/sec$ 

e)

Steady-state Error for a ramp input:

$$e_{ss} = \lim_{s \to 0} s * R(s) * T(s) = 0$$

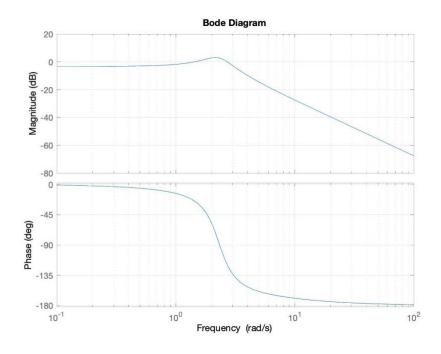
## Problem 4. DP8.3

$$T(s) = \frac{K(s+5)}{s^3 + 7s^2 + 12s + 10 + 5K}$$

From the question (20lgM=3), we get

$$K = 4.2$$

**Body Plot:** 



We can tell from the graph that the system bandwidth is:

Steady-state error:

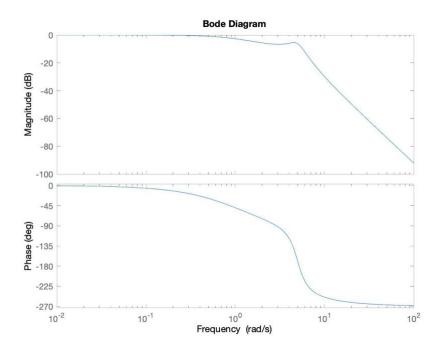
$$e_{ss} = \lim_{s \to 0} s * E(s) = \lim_{s \to 0} (1 - T(s)) = 0.32$$

## Problem 5. CP8.6

Open-loop system:

$$L_s = \frac{25}{s^3 + 3s^2 + 27s + 25}$$

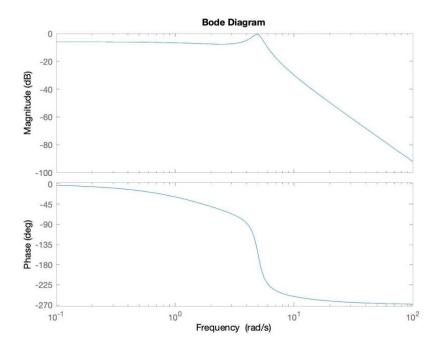
Bode Plot:



Closed-loop system:

$$L_s = \frac{25}{s^3 + 3s^2 + 27s + 50}$$

**Bode Plot:** 



## Problem 6. DP10.1

$$G(s) = \frac{20}{s(s+2)}$$

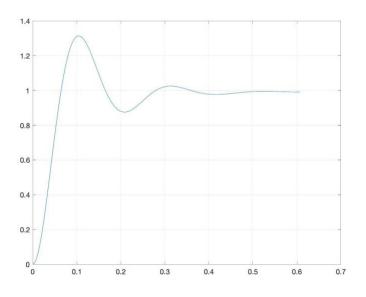
The possible lead and lag compensators can be:

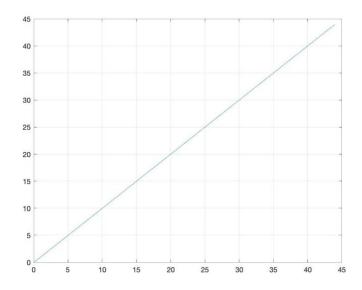
$$G_{lead}(s) = \frac{50(s+1)}{s+20}, \qquad G_{lag}(s) = \frac{(s+0.1)}{s+0.022}$$

Thus,

$$L_s = \frac{1000(s+1)(s+0.1)}{s(s+2)(s+20)(s+0.022)}$$

Step and ramp responses are shown as follows:



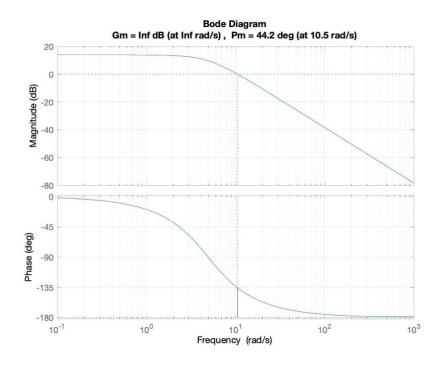


## Problem 7. CP10.2

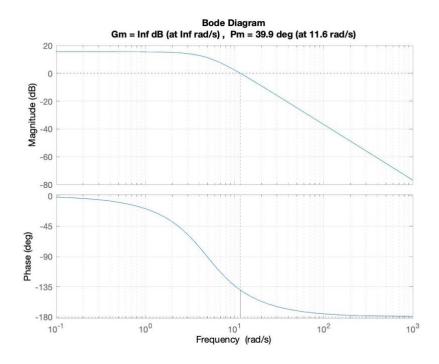
Open-loop system:

 $L_s = \frac{24.2}{s^2 + 8s + 24.2}$ 

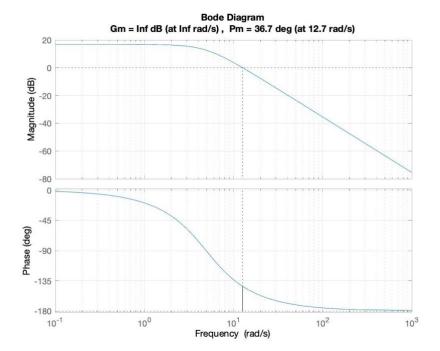
Bode Plot (K=5):



Bode Plot (K=6):



Bode Plot (K=7):



By iterating different values of K and checking phase margin in MATLAB, we conclude that:

$$K = 6$$

K=6 gives us a phase margin of about 40 degrees.

### Problem 8. CP10.4

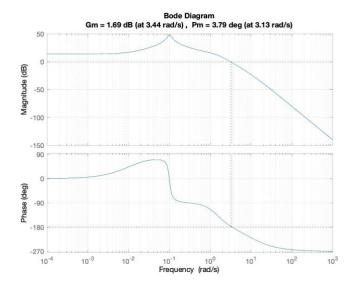
a)

In order to have 0.P.% < 10%, we have:

$$\zeta > 0.6$$
, phase margin  $> 60^{\circ}$ 

Using MATLAB, we found out the phase margin of the uncompensated system is  $3.79^{\circ}$  which is far less than the expected value. Meanwhile,  $\alpha$  is found out to be 10.8 here.

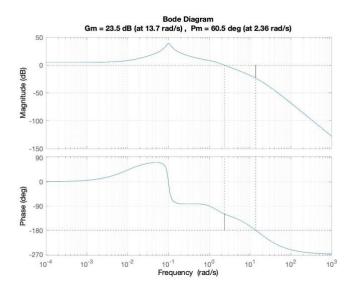
Bode Plot:



From the graph above,  $G(j\omega)$  is  $0 \ dB$  at about  $2 \ rad/sec$ . We can also find out K converges to 4, and  $p=\alpha z=21.7$ . So that we can design the phase lead compensator to be as follows:

$$G_c(s) = \frac{4(s+2)}{s+22}$$

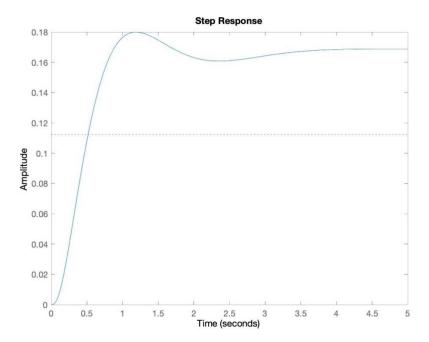
New Bode Plot:



This system now satisfies the requirement.

# b)

In MATLAB, we can draw the step response for the input of  $60^{\circ}/sec$ :



# Problem 9. CP11.3

Given:

$$A = \begin{bmatrix} 0 & 1 \\ -1 & 2 \end{bmatrix}, \qquad B = \begin{bmatrix} 1 \\ 1 \end{bmatrix}, \qquad C = \begin{bmatrix} 1 & -1 \end{bmatrix}, \qquad D = \begin{bmatrix} 0 \end{bmatrix},$$
 
$$s_1 = -1, \qquad s_2 = -2$$

The gain matrix of the closed-loop system is easily calculated in MATLAB:

$$K = [0.5 \ 0.5]$$

## **Problem 10. DP11.7**

1)

$$A - BK - LC = \begin{bmatrix} -L_1 & 1 & 0 \\ -L_2 & 0 & 1 \\ -2 - K_1 - L_3 & -5 - K_2 & -10 - K_3 \end{bmatrix}$$

Here we choose:

$$K_1 = -2$$
,  $K_2 = -5$ ,  $K_3 = -10$ 

Thus,

$$\det(\lambda I - (A - BK - LC)) = 0$$

Plugging in, we have:

$$\lambda^3 + L_1 * \lambda^2 + L_2 * \lambda + L_3 = 0$$

According to R-H table, to make the system stable:

$$\begin{cases} L_2 > \frac{L_3}{L_1} \\ L_3 > 0 \end{cases}$$

Since  $\omega_B \ge 10 \ dB$ , we can also choose:

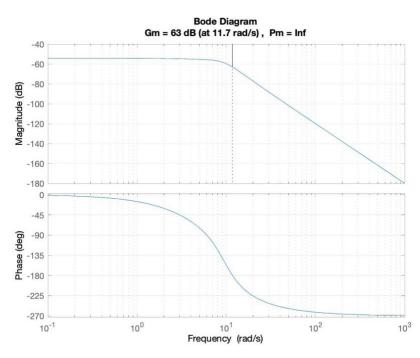
$$\omega_n = 8 \, rad/sec$$

Thus,

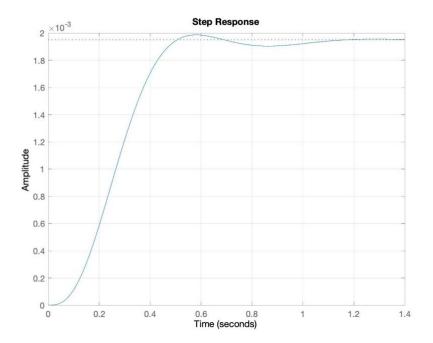
$$L = \begin{bmatrix} 14\\138\\512 \end{bmatrix}$$

In MATLAB:

Bode Plot w/ gain margin greater than 6 dB:



#### Step Response:



Steady-state error meets the requirement.

## **Problem 11**

1)

$$Q(s) = 1 + \frac{s + \alpha}{s^3 + (1 + \alpha)s^2 + (\alpha - 1)s + (1 - \alpha)} = 0$$

$$s^{3} + (1 + \alpha)s^{2} + (\alpha - 1)s + (1 - \alpha) + s + \alpha = 0$$

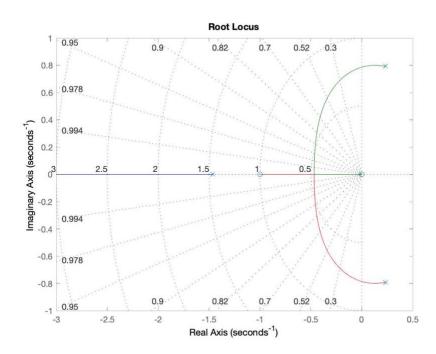
$$s^3 + s^2 + 1 + \alpha(s^2 + s) = 0$$

$$1 + \frac{\alpha(s^2 + s)}{s^3 + s^2 + 1} = 0$$

Thus,

$$L_s = \frac{(s^2 + s)}{s^3 + s^2 + 1}$$

**Root Locus:** 



2)

$$Q(s) = s^3 + (1 + \alpha)s^2 + \alpha s + 1 = 0$$

The Routh Array:

$$\begin{array}{c|cccc} s^3 & 1 & \alpha \\ s^2 & 1+\alpha & 1 \\ s^1 & \alpha - \frac{1}{1+\alpha} & 0 \\ s^0 & 1 & 0 \end{array}$$

To make the system stable:

$$1 + \alpha > 0, \qquad \alpha - \frac{1}{1 + \alpha} > 0$$

Thus,

$$\alpha > 0.61803$$

$$E(s) = \frac{R(s)}{1 + L_s} = \frac{\frac{1}{s}}{1 + \frac{s + \alpha}{s^3 + (1 + \alpha)s^2 + (\alpha - 1)s + (1 - \alpha)}}$$

$$e_{ss} = \lim_{s \to 0} s * E(s) = \frac{1}{1 + \frac{\alpha}{1 - \alpha}} \le 10\%$$

$$\alpha \ge 0.9$$

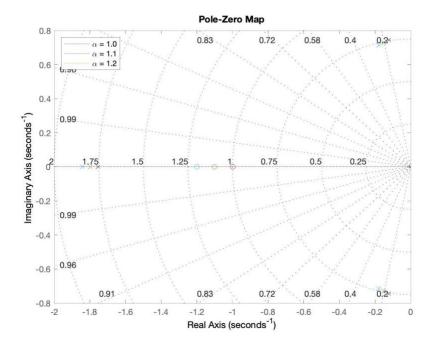
4)

$$T(s) = \frac{s + \alpha}{s^3 + (1 + \alpha)s^2 + \alpha s + 1}$$

To satisfy the requirement in part 3, we choose:

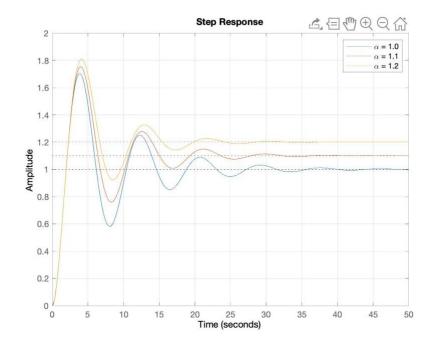
$$\alpha = 1.0,$$
 1.1, 1.2

The roots are plotted as follows:



For a specific  $\alpha$ , we have one real zero, one real pole, and two complex poles. This matches the root locus plot we obtained in part 1.

#### Step response is printed as:



 $\alpha=1.2$  gives us the best result since the settling time would be less than the other two, and the same is the steady-state error.

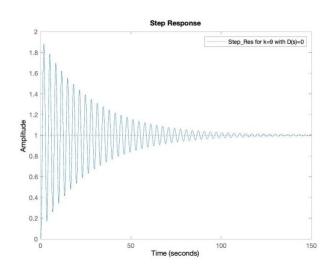
## **Problem 12**

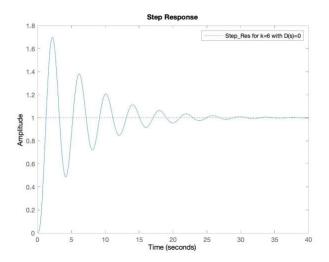
# 1.a)

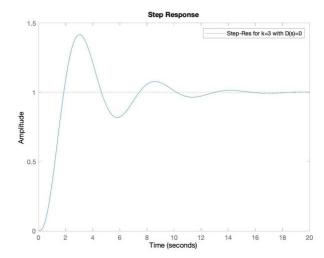
$$L_s = G_c * G = \frac{2K}{s(s+1)(s+4)}$$

$$T_D(s) = \frac{\frac{2}{s(s+1)(s+4)}}{1 + \frac{2K}{s(s+1)(s+4)}} = \frac{2}{s(s+1)(s+4) + 2K}$$

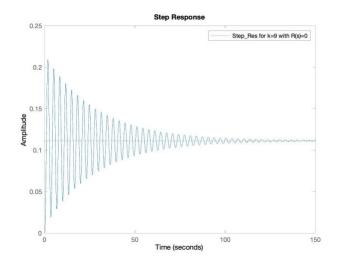
Input Step Response:

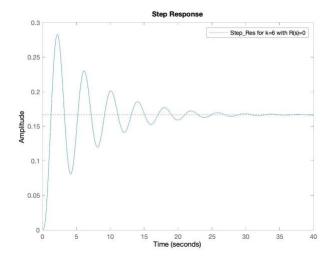


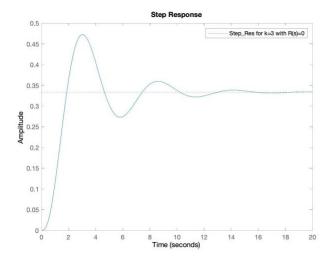




#### Noise Step Response:







#### Steady-state Error for a step input:

$$e_{ss} = \lim_{s \to 0} s * \frac{1}{s} * T_D(s) = \frac{1}{K}$$

1.b)

$$T_R(s) = \frac{2K}{s(s+1)(s+4) + 2K} = \frac{2K}{s^3 + 5s^2 + 4s + 2K}$$

The Routh Array:

$$s^3$$
 1
 4

  $s^2$ 
 5
  $2K$ 
 $s^1$ 
 $4 - \frac{2K}{5}$ 
 0

  $s^0$ 
 $2K$ 
 0

To make the system stable:

$$4 - \frac{2K}{5} > 0, \qquad 2K > 0$$

Thus,

To minimize steady-state error:

$$K = 10$$

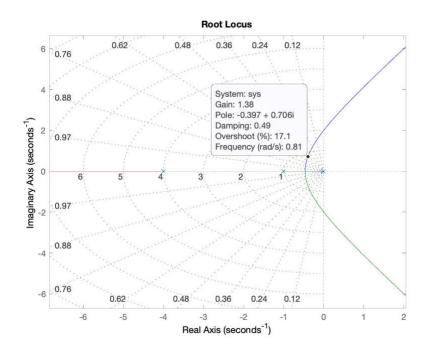
1.c)

$$G(s) = \frac{2}{s(s+1)(s+4)}$$

$$n = 3$$
,  $m = 0$ ,  $n - m = 3$ 

$$\phi_{Asym} = \pm 60^{\circ}$$
,  $180^{\circ}$ 

**Root Locus:** 



When 
$$\zeta = 0.5$$
,

$$\beta = \cos^{-1} 0.5 = 60^{\circ}, \qquad s = -0.4 + j0.7$$

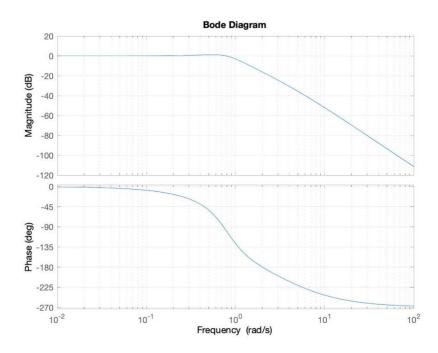
Since we have  $\left|\frac{2K}{s(s+1)(s+4)}\right| = 1$ ,

$$K = 1.34$$

# 1.d)

$$T_R(s) = \frac{2*1.34}{s(s+1)(s+4) + 2*1.34} = \frac{2.68}{s^3 + 5s^2 + 4s + 2.68}$$

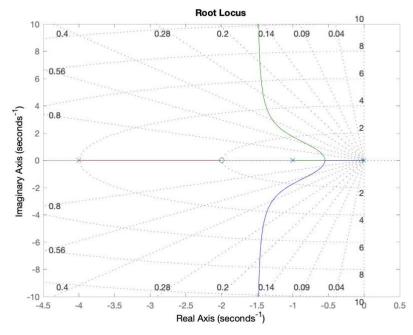
Bode plots:



2.a)

$$L_s = \frac{2K(s+2)}{s(s+1)(s+4)}$$

**Root Locus:** 



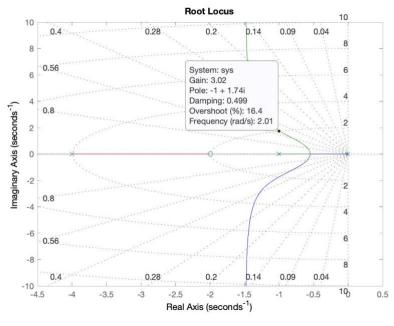
It is noted that we now have  $\phi_{Asym}=\pm 90^o$  since n-m=2. Meanwhile, the crossing point of asymptotes is shifted to the left due to the new zero.

PD controller makes the system faster and more stable.

2.b)

From rlocus in 2.a):

$$T_s = 4 s$$
,  $\zeta \omega_n = 1$ 

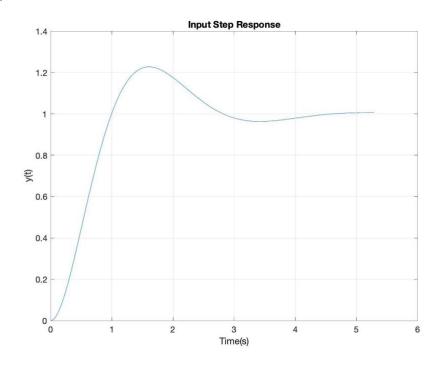


$$K = 3.02$$

## 2.c)

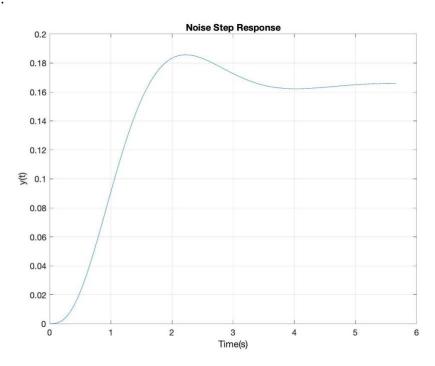
$$T_R(s) = \frac{2K(s+2)}{s^3 + 5s^2 + 4s + 2K(s+2)} = \frac{6.04s + 12.08}{s^3 + 5s^2 + 10.04s + 12.08}$$

Input Step Response:



$$T_D(s) = \frac{2}{s^3 + 5s^2 + 4s + 2K(s+2)} = \frac{2}{s^3 + 5s^2 + 10.04s + 12.08}$$

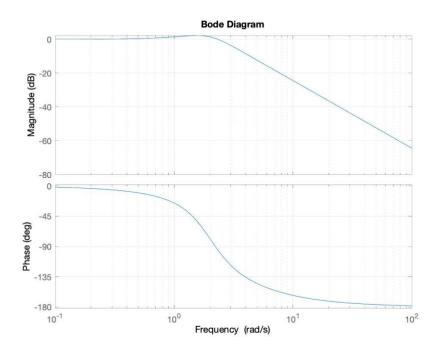
Nosie Step Response:



2.d)

$$T_R(s) = \frac{6.04s + 12.08}{s^3 + 5s^2 + 10.04s + 12.08}$$

Bode Plot:



This is the end of the first section. Thanks for reading it.

### **Code**

# Problem 1. CP7.10

>> sys=tf([0 1 0 0],[1 2 5 1])

sys =

s^2 -----

 $s^3 + 2 s^2 + 5 s + 1$ 

Continuous-time transfer function.

>> rlocus(sys)

#### Problem 2. DP7.12

>> sys=tf([0 0 0 1 1.5],[1 17 84 148 80])

sys =

s^4 + 17 s^3 + 84 s^2 + 148 s + 80

Continuous-time transfer function.

>> rlocus(sys)

>> K=100

K =

100

>> vpasolve(1+K\*(s+1.5)/((s+1)\*(s+2)\*(s+4)\*(s+10))==0)

ans =

-11.376224942507200415181627931047

-1.4460198425824747094898010247309

- 2.0888776074551624376642855221109 3.1013131610271484903489641911771i
- 2.0888776074551624376642855221109 + 3.1013131610271484903489641911771i

>> K=300

K =

300

```
>> vpasolve(1+K*(s+1.5)/((s+1)*(s+2)*(s+4)*(s+10))==0)
ans =
                   -12.939320814675075614406929484264
                   -1.4821505578182696046435102197956
- 1.2892643137533273904747801479703 + 5.0964300428123383917409735291857i
- 1.2892643137533273904747801479703 - 5.0964300428123383917409735291857i
>> K=600
K =
 600
>> vpasolve(1+K*(s+1.5)/((s+1)*(s+2)*(s+4)*(s+10))==0)
ans =
                    -14.444522307111860227899924413701
                    -1.4911078580399334553177469171597
- 0.53218491742410315839116433456944 - 6.7243612669765139076940964681778i
- 0.53218491742410315839116433456944 + 6.7243612669765139076940964681778i
>> sys=tf([100 150], [1 17 84 248 230])
sys =
       100 s + 150
 _____
 s^4 + 17 s^3 + 84 s^2 + 248 s + 230
Continuous-time transfer function.
>> [y,t]=step(sys)
>> plot(t,y)
>> grid
>> title('Step Response K=100')
>> xlabel('Time(s)')
>> ylabel('y(t)')
>> sys2=tf([300 450], [1 17 84 448 530])
```

### Problem 3. DP8.1

>> sys=tf([1 2],[1 12 0 0]) sys = s + 2s^3 + 12 s^2 Continuous-time transfer function. >> bode(sys) >> grid >> sys1=tf([1 2],[1 12 1 2]) sys1 = s + 2 $s^3 + 12 s^2 + s + 2$ Continuous-time transfer function. >> bode(sys1) >> sys=tf([50 100],[1 12 0 0]) sys = 50 s + 100 ----s^3 + 12 s^2 Continuous-time transfer function. >> bode(sys) >> sys1=tf([50 100],[1 12 50 100]) sys1 = 50 s + 100 s^3 + 12 s^2 + 50 s + 100 Continuous-time transfer function.

>> bode(sys1)

## Problem 4. DP8.3

Continuous-time transfer function.

>> bode(sys)
>> grid

### Problem 5. CP8.6

sys =

Continuous-time transfer function.

Continuous-time transfer function.

>> bode(sys)

### Problem 6. DP10.1

Continuous-time transfer function.

```
>> [y,t]=step(comp)
>> plot(t,y)
>> grid
>> s = tf('s');
>> [y,t]=step(comp/s)
>> plot(t,y)
>> grid
```

#### Problem 7. CP10.2

```
>> sys=tf([0 0 24.2],[1 8 24.2])

sys =

24.2

------

s^2 + 8 s + 24.2
```

Continuous-time transfer function.

```
>> margin(sys*5),grid;
>> margin(sys*6),grid;
>> margin(sys*7),grid;
```

### Problem 8. CP10.4

```
>> sys=tf([0 0 -10 -10.1 -0.1],[1 2.02 2.0501 0.0602 0.0202])
sys =
       -10 s^2 - 10.1 s - 0.1
 s^4 + 2.02 s^3 + 2.05 s^2 + 0.0602 s + 0.0202
Continuous-time transfer function.
>> margin(sys)
>> clear
>> sys=tf([0 0 0 100 101 1],[1 12.02 22.2501 20.5612 0.6222 0.202])
sys =
           100 s^2 + 101 s + 1
 s^5 + 12.02 s^4 + 22.25 s^3 + 20.56 s^2 + 0.6222 s + 0.202
Continuous-time transfer function.
>> margin(sys)
>> grid
>> sys1=(sys*tf([4 8],[1 22]))
sys1 =
          400 s^3 + 1204 s^2 + 812 s + 8
 s^6 + 34.02 s^5 + 286.7 s^4 + 510.1 s^3 + 453 s^2 + 13.89 s
                                + 4.444
Continuous-time transfer function.
>> margin(sys1);grid
b)
>> closed_sys=feedback(sys1,1)
closed_sys =
```

```
400 s^3 + 1204 s^2 + 812 s + 8
```

\_\_\_\_\_

```
s^6 + 34.02 s^5 + 286.7 s^4 + 910.1 s^3 + 1657 s^2 + 825.9 s + 12.44
```

Continuous-time transfer function.

```
>> t=[0:0.02:5]
>> step(n*closed_sys,t)
```

### Problem 9. CP11.3

```
a=[0 1; -1 -2];
b=[1; 1];
c=[1 -1];
d=[0];
p=[-1; -2];
K=acker(a, b, p);
```

#### **Problem 10. DP11.7**

>> step(sys1)

>> grid

### **Problem 11**

```
>> sys=tf([0 1 1 0],[1 1 0 1])

sys =

s^2 + s

-----

s^3 + s^2 + 1
```

Continuous-time transfer function.

```
>> rlocus(sys)
>> grid
alpha=[1 1.1 1.2];
for k=1:3
sys=tf([1 alpha(k)],[1 1+alpha(k) alpha(k) 1]);
pzmap(sys);
hold on
end
legend({'\alpha = 1.0','\alpha = 1.1','\alpha = 1.2'},'Location','northwest');
grid;
alpha=[1 1.1 1.2];
for k=1:3
sys=tf([1 alpha(k)],[1 1+alpha(k) alpha(k) 1]);
step(sys);
hold on
end
legend('\alpha = 1.0','\alpha = 1.1','\alpha = 1.2');
grid;
```

### **Problem 12**

```
% gains
k=[3 6 9];
% plant
g=tf([0 \ 0 \ 0 \ 2],[1 \ 5 \ 4 \ 0]);
% input STEP RESPONSE
step(feedback(g*k(1),1));
legend('Step-Res for k=3 with D(s)=0');
figure;
grid;
step(feedback(g*k(2),1));
legend('Step Res for k=6 with D(s)=0');
figure;
grid;
step(feedback(q*k(3),1));
legend('Step_Res for k=9 with D(s)=0')
figure;
```

```
grid;
% noise STEP RESPONSE
step(feedback(g,k(1)));
legend('Step Res for k=3 with R(s)=0')
figure;
grid;
step(feedback(g,k(2)));
legend('Step Res for k=6 with R(s)=0')
figure;
grid;
step(feedback(g,k(3)));
legend('Step Res for k=9 with R(s)=0')
figure;
grid;
>> sys=tf([0 0 0 2],[1 5 4 0])
sys =
     2
 s^3 + 5 s^2 + 4 s
Continuous-time transfer function.
>> rlocus(sys)
>> grid
>> sys=tf([0 0 0 2.68],[1 5 4 2.68])
sys =
      2.68
 s^3 + 5 s^2 + 4 s + 2.68
Continuous-time transfer function.
>> bode(sys)
>> grid
>> sys=tf([0 0 2 4],[1 5 4 0])
sys =
    2s + 4
 s^3 + 5 s^2 + 4 s
```

Continuous-time transfer function.

```
>> rlocus(sys)
>> grid
>> sys=tf([0 0 6.04 12.08],[1 5 10.04 12.08])
sys =
     6.04 s + 12.08
 -----
 s^3 + 5 s^2 + 10.04 s + 12.08
Continuous-time transfer function.
>> [y,t]=step(sys)
>> plot(t,y)
>> grid
>> title('Input Step Response')
>> xlabel('Time(s)')
>> ylabel('y(t)')
>> sys1=tf([0 0 0 2],[1 5 10.04 12.08])
sys1 =
         2
 s^3 + 5 s^2 + 10.04 s + 12.08
Continuous-time transfer function.
>> [y,t]=step(sys1)
>> plot(t,y)
>> grid
>> title('Noise Step Response')
>> xlabel('Time(s)')
>> ylabel('y(t)')
```

>> bode(sys)
>> grid