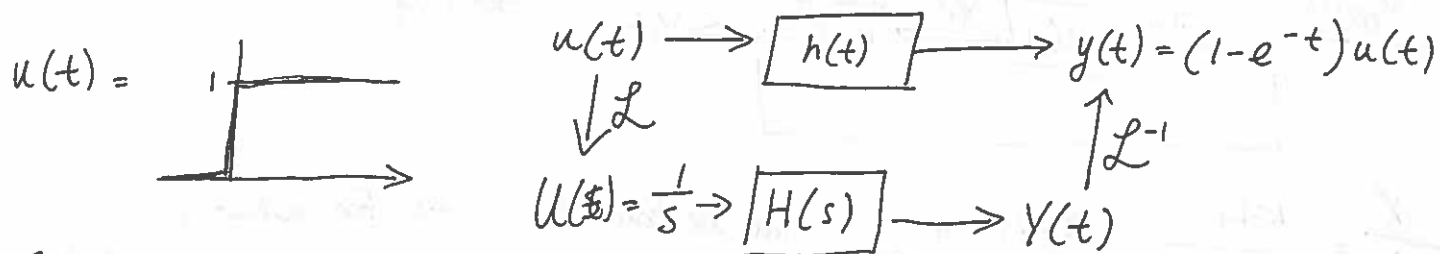


# Problem Set 10 - Jiaying Wei

1. Use the Laplace transform to verify that the step response  $\dot{y} + y = x$  is  $y(t) = (1 - e^{-t})u(t)$ .



$$\mathcal{L}\{\dot{y} + y = x\} = sY(s) + Y(s) = X(s) \quad U(s) = X(s)$$

$$sY(s) + Y(s) = \frac{1}{s}$$

$$Y(s)(s+1) = \frac{1}{s}$$

$$Y(s) = \frac{1}{s(s+1)} \rightarrow \frac{A_1}{s} + \frac{A_2}{s+1}$$

$$\left. \frac{s \cdot 1}{s(s+1)} \right|_{s=0} = \frac{s A_1}{s} + \frac{s A_2}{s+1} \Big|_{s=0}$$

$$1 \neq 0 = A_1 + 0$$

$$\boxed{A_1 = 1}$$

$$\left. \frac{(s+1)}{s(s+1)} \right|_{s=-1} = \frac{(s+1)A_1}{s} + \frac{(s+1)A_2}{(s+1)} \Big|_{s=-1}$$

$$\frac{1}{-1} = \frac{0 \cdot A_1}{-1} + A_2 \quad \boxed{A_2 = -1}$$

$$Y(s) = \frac{1}{s} + \frac{-1}{s+1}$$

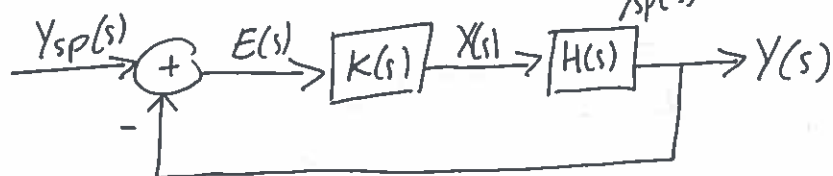
$$y(t) = 1 \cdot e^{-0 \cdot t} u(t) + -1 e^{-1 \cdot t} u(t)$$

$$= u(t) - e^{-t} u(t)$$

$$\boxed{y(t) = (1 - e^{-t})u(t)}$$

## Problem 2

A. Find the DC gain of the system  $\frac{Y(s)}{Y_{sp}(s)}$  if you use an integral controller  $K(s) = \frac{K_I}{s}$  for any  $H(s)$



$$\frac{Y}{Y_{sp}} = \frac{KH}{1+KH} = \frac{K_I/s \cdot H}{1 + K_I/s \cdot H}$$

now we have to solve this for when  $s=0$  because that will be the first coefficient to  $e^0$  thus giving us the DC offset of the system

But  $K_I/0$  is going to ERROR so we need to take the  $\lim_{s \rightarrow 0} \frac{K_I}{s}$

But if we do, we run into  $\infty$  which is also bad so we need to use L'Hospital's Rule

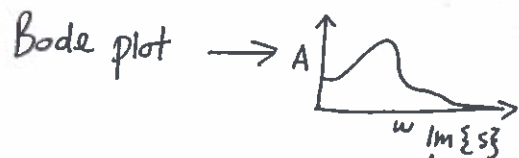
$$\text{L'Hopital said } \lim_{x \rightarrow 0} \frac{f(x)}{g(x)} = \lim_{x \rightarrow 0} \frac{f'(x)}{g'(x)}$$

so:

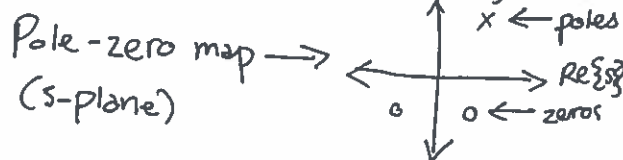
$$\frac{Y}{Y_{sp}} = \lim_{s \rightarrow 0} \frac{d}{ds} \left( \frac{K_I/s \cdot H}{1 + K_I/s \cdot H} \right) = \lim_{s \rightarrow 0} \frac{-HK_I/s^2}{-HK_I/s^2} = 1 \leftarrow \text{DC offset of the system is 1}$$

## Problem 3

Analyze the behavior of the below systems using a Bode plot, a pole-zero map, a step response



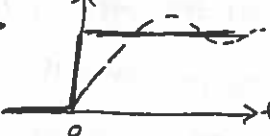
: How does the system react to different  $\omega$



: Shows the poles and zeros of the transfer function

Poles - Values of  $s$  that cause the denominator to go to 0, which if you take the limit of that  $\frac{Y}{Y_{sp}}$  goes to  $\infty$  so it tells you when your system will go off into infinity

Zeros - Values of  $s$  that cause the numerator to go to 0, which means that your system is at a stable point where it is not changing. Not necessarily at the set point though.

Step Response  $\rightarrow$   How the system responds in time to a change in input/setpoint  $u(t)$

$$A. \frac{Y}{Y_{sp}} = \frac{s}{s+1}$$

1. It appears that the step response of this system shows that it will jump almost instantaneously to the setpoint and then ~~exponent~~ exponentially to zero.
2. The Bode plot seems to indicate that it is more sensitive to more rapid change.
3. The poles tell us the system is unstable @  $s = -1 + 0j$  and stable @  $s = 0 + 0j$

$$B. \frac{Y}{Y_{sp}} = \frac{s}{s^2 + 100s + 1}$$

1. This particular system was much slower than the 1st system to getting to the setpoint, but also started decreasing to 0 after reaching the sp though at a much slower rate.
2. The bode plot confirms this, showing greater magnitudes at lower frequencies
3. poles @  $s = -100 + 0j$  ;  $s = -1 + 0j$   
zero @  $s = 0 + 0j$

$$C. \frac{Y}{Y_{sp}} = \frac{s}{s^2 + s + 1}$$

1. The step response shows that even before it ~~star~~ reaches the setpoint the system has an oscillating decay
2. It has a similar Bode plot to B
3. poles @  $s = -0.5 + \approx 0.8j$  ;  $s = -0.5 - \approx 0.8j$   
zero @  $s = 0 + 0j$  appearing to be complex conjugates  
 $\hookrightarrow$  maybe something to do w/ oscillation

$$D. \frac{Y}{Y_{sp}} = \frac{s}{s^2 + 0.15s + 1}$$

1. Has a similar oscillating decay to C, only w/ larger oscillation / slower decay
2. Similar Bode plot to D
3. poles @  $s = -0.1 + 1j$  ;  $s = -0.1 - 1j$   
zero @  $s = 0 + 0j$  does appear that complex conjugates do have something to do w/ oscillation  
★ Maybe a larger  $j$  value means larger oscillations?  
Maybe a smaller  $-s$  value also causes that?

$$\equiv \frac{Y}{Y_s} = \frac{s^2 - 0.01s + 1}{s^2 + 0.01s + 1}$$

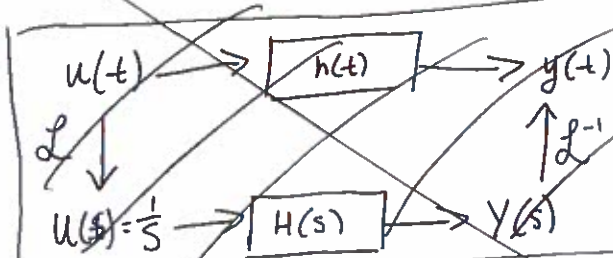
- A. For the first time it seems that the resultant  $y(t)$  is following the set point, albeit some oscillations
- B. The magnitude Bode plot seems to indicate that it treats fast and slow inputs equally
- C. I can't really tell by eyeballing it where the poles / zeros are since they're overlapping kinda. But both appear to be in complex conjugate pairs.  
Not sure what having complex conjugate zeros do

$$F. \frac{Y}{Y_s} = \frac{s^2 + 0.1s + 1}{s^2 + 0.11s + 1}$$

- A. The system follows the setpoint pretty spot on. There appears to be just a little oscillation but my eyes can't really tell.
- B. Bodeplot shows that it reacts greater to higher frequency changes
- C. Similar looking poles and zeros to E

#### Problem 4 - Stabilize

Plant!  $\rightarrow H(s) = \frac{1}{s^2 - 0.01s + 1}$



So I wasn't supposed to do part A by hand...



$$\frac{Y}{Y_{sp}} = \frac{KH}{1+KH}$$

$$Y(s) = U(s)H(s) = \frac{1}{s^2 - 0.01s^2 + s} = \frac{1}{s(s^2 - 0.01s + 1)}$$

quadratic formula this

$$s = \frac{0.01 \pm \sqrt{(-0.01)^2 - 4(1)(1)}}{2(1)}$$

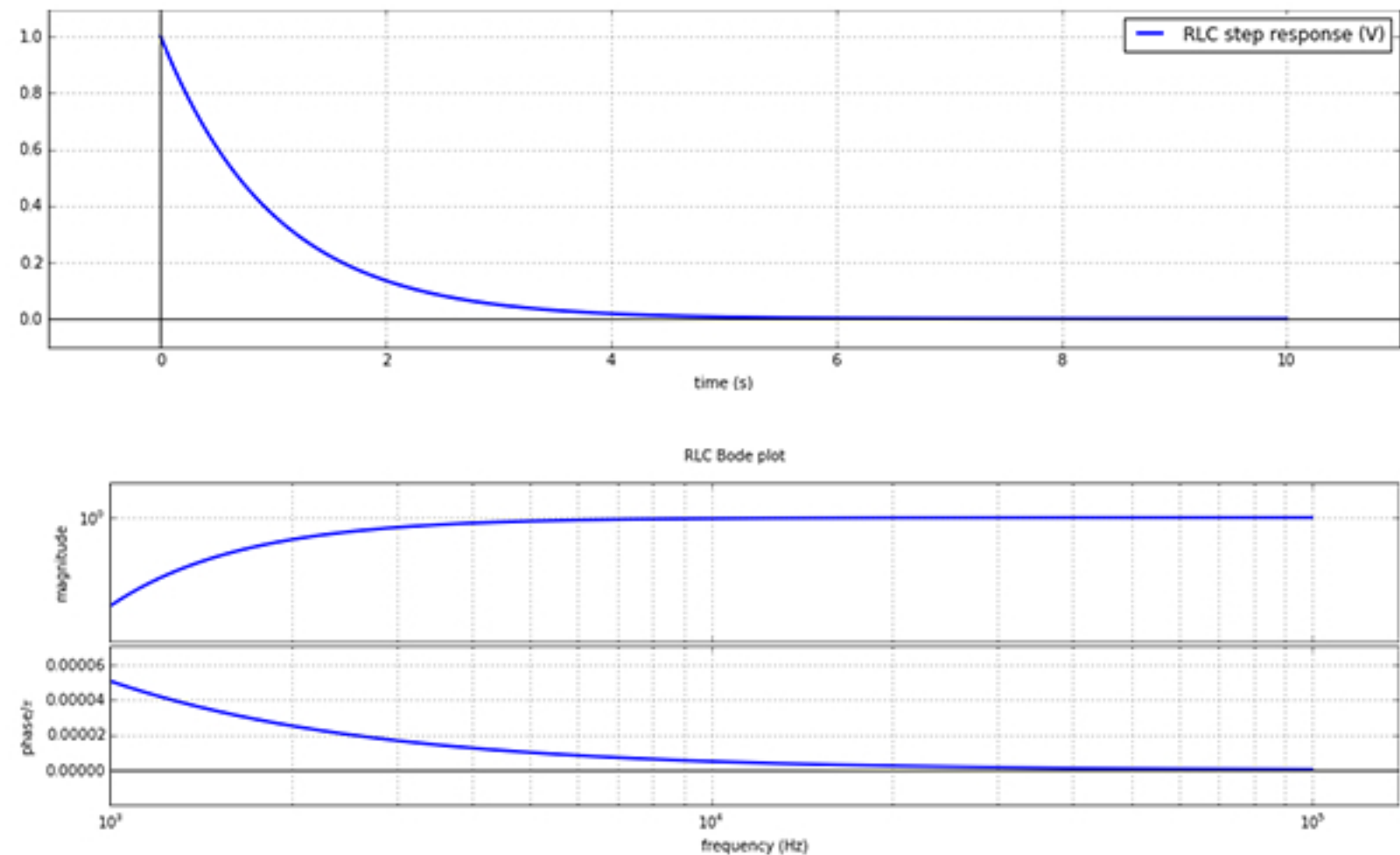
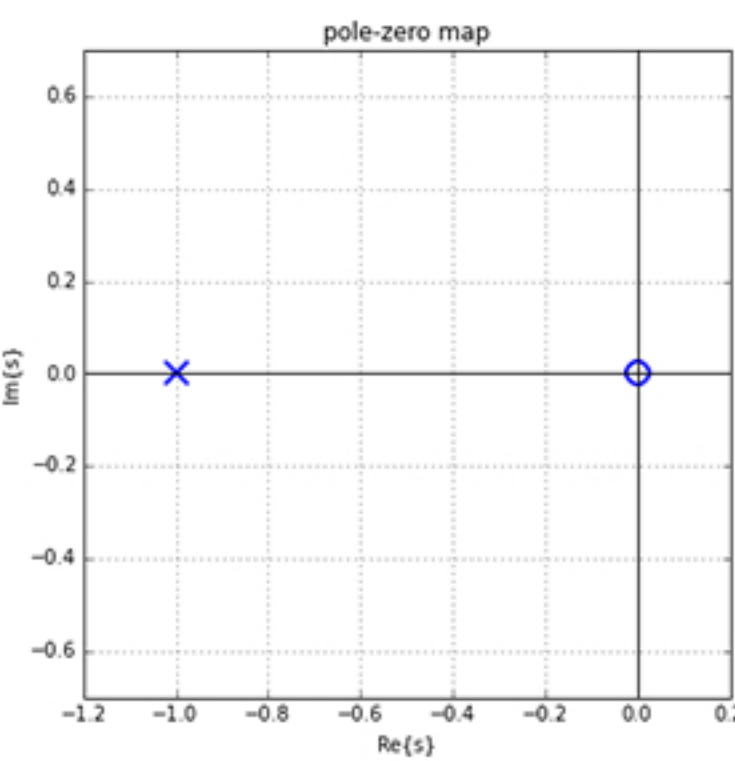
$$s = \frac{0.01 \pm \sqrt{0.0001 - 4}}{2} = \frac{0.01 \pm \sqrt{-3.9999}}{2}$$

$$Y(s) = \frac{1}{s(0.005)}$$

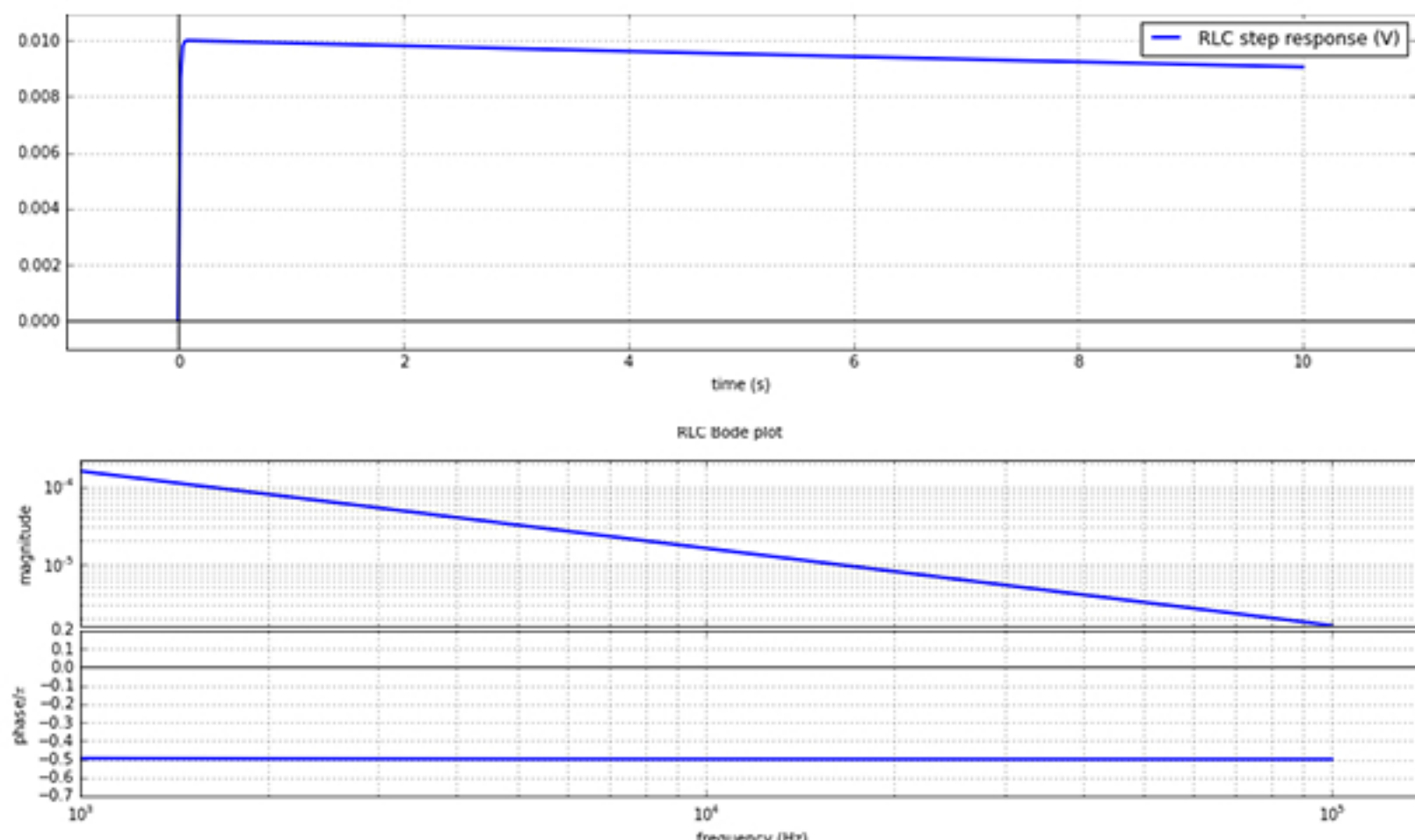
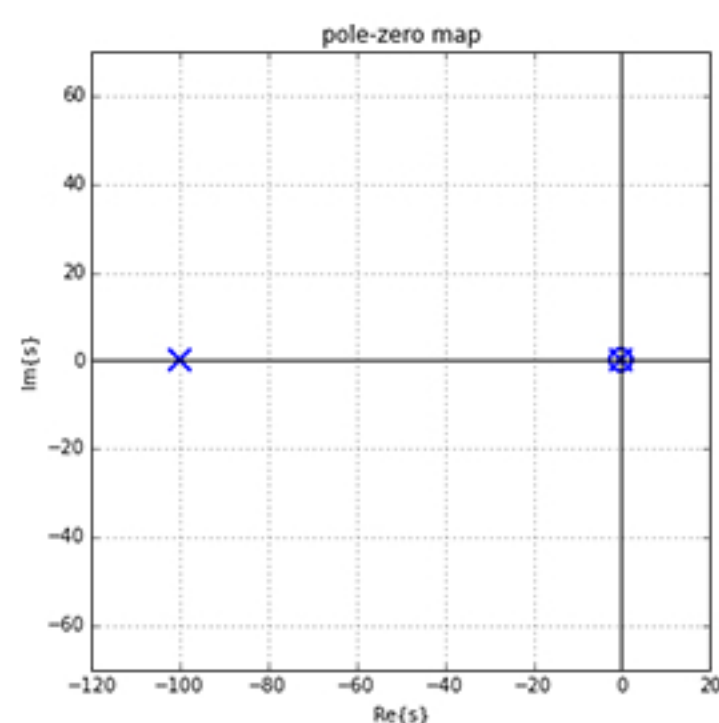


# Problem 3

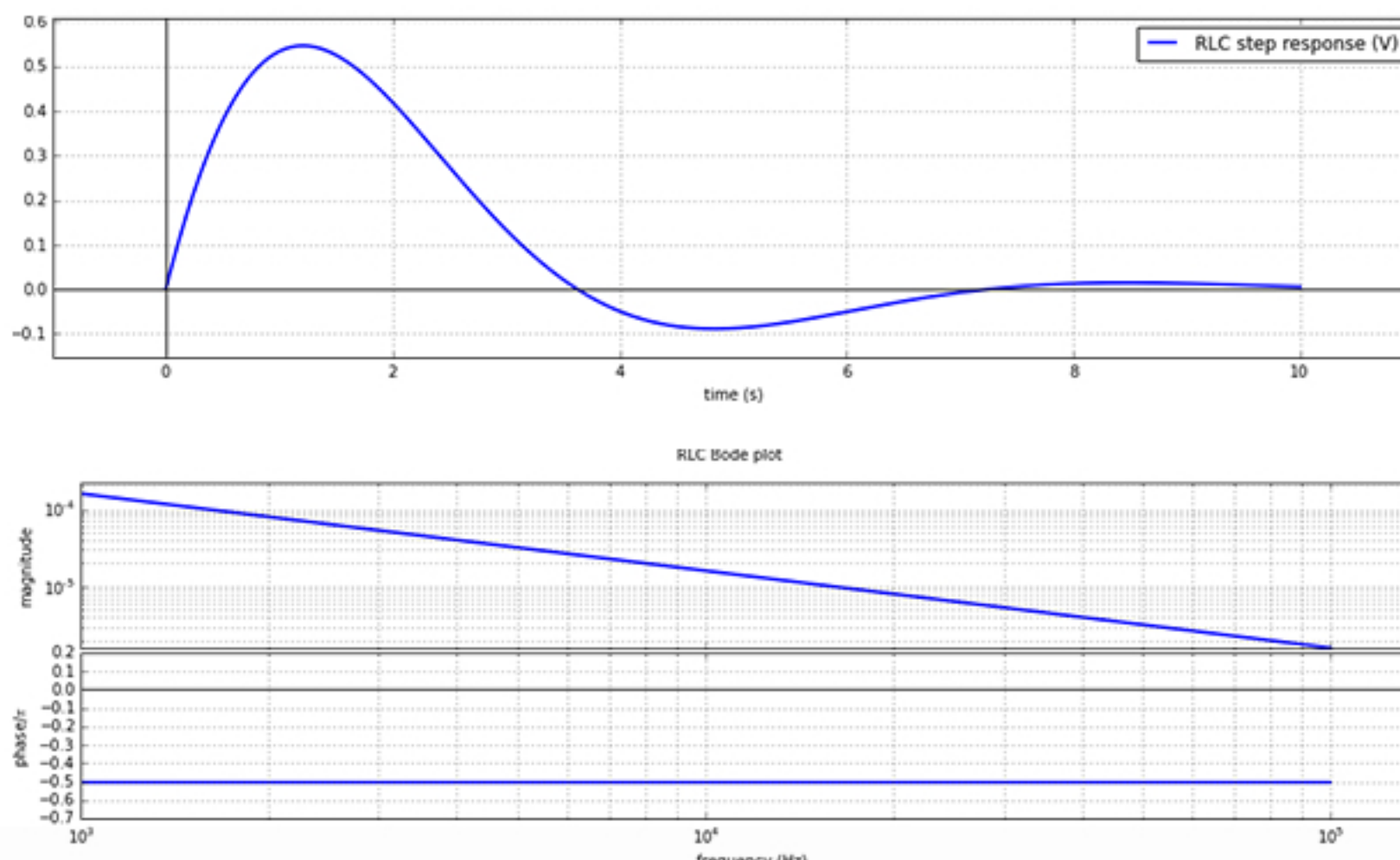
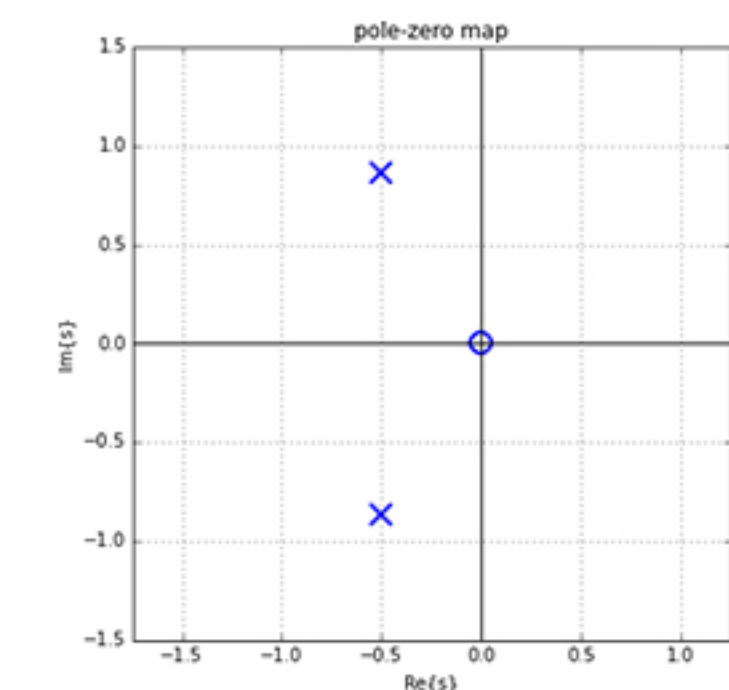
## System A



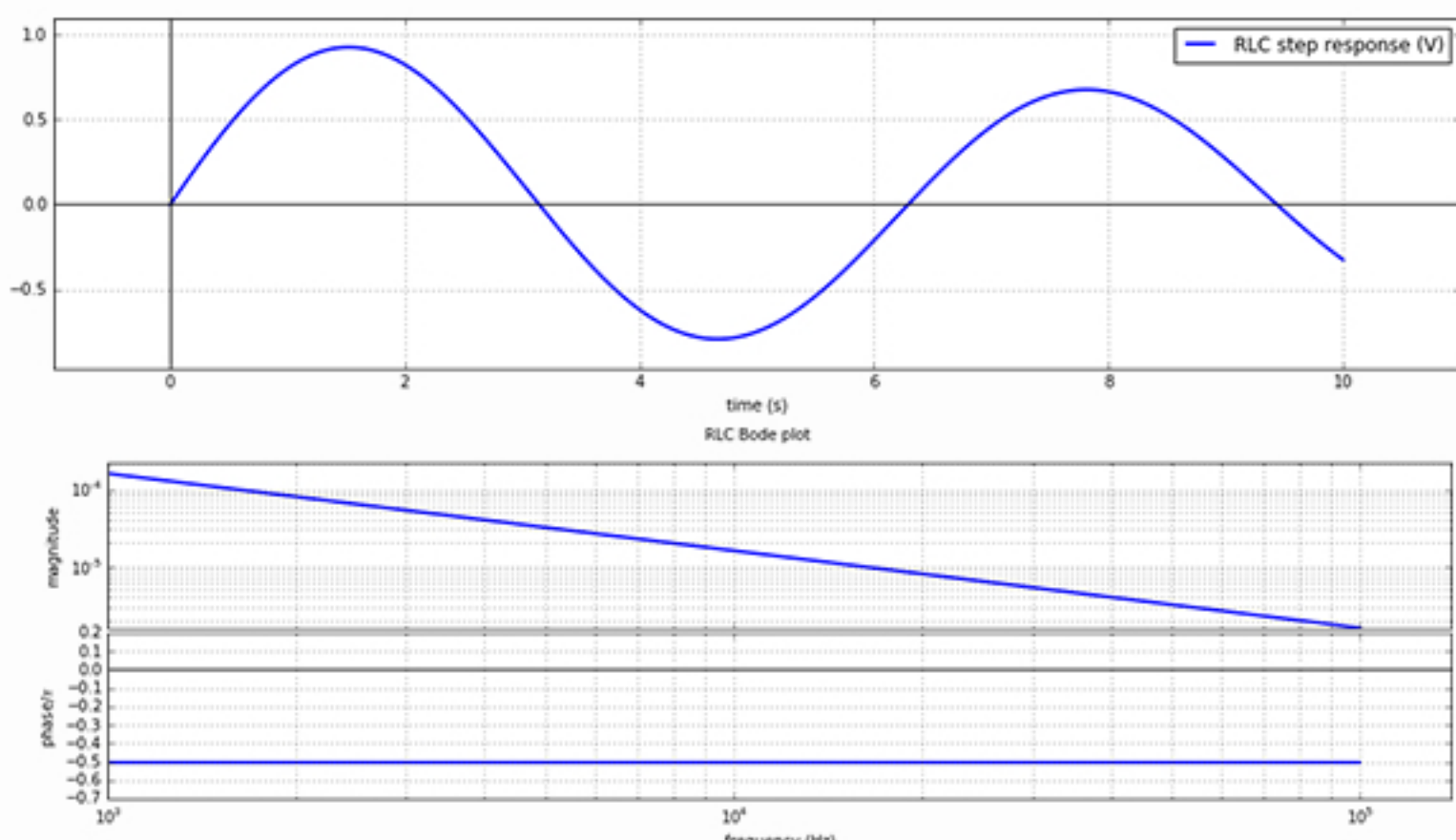
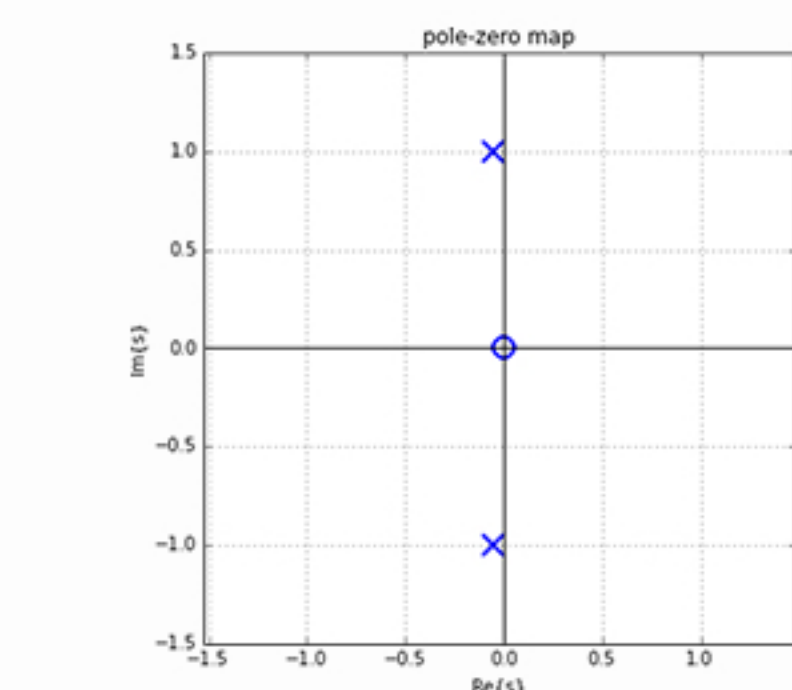
## System B



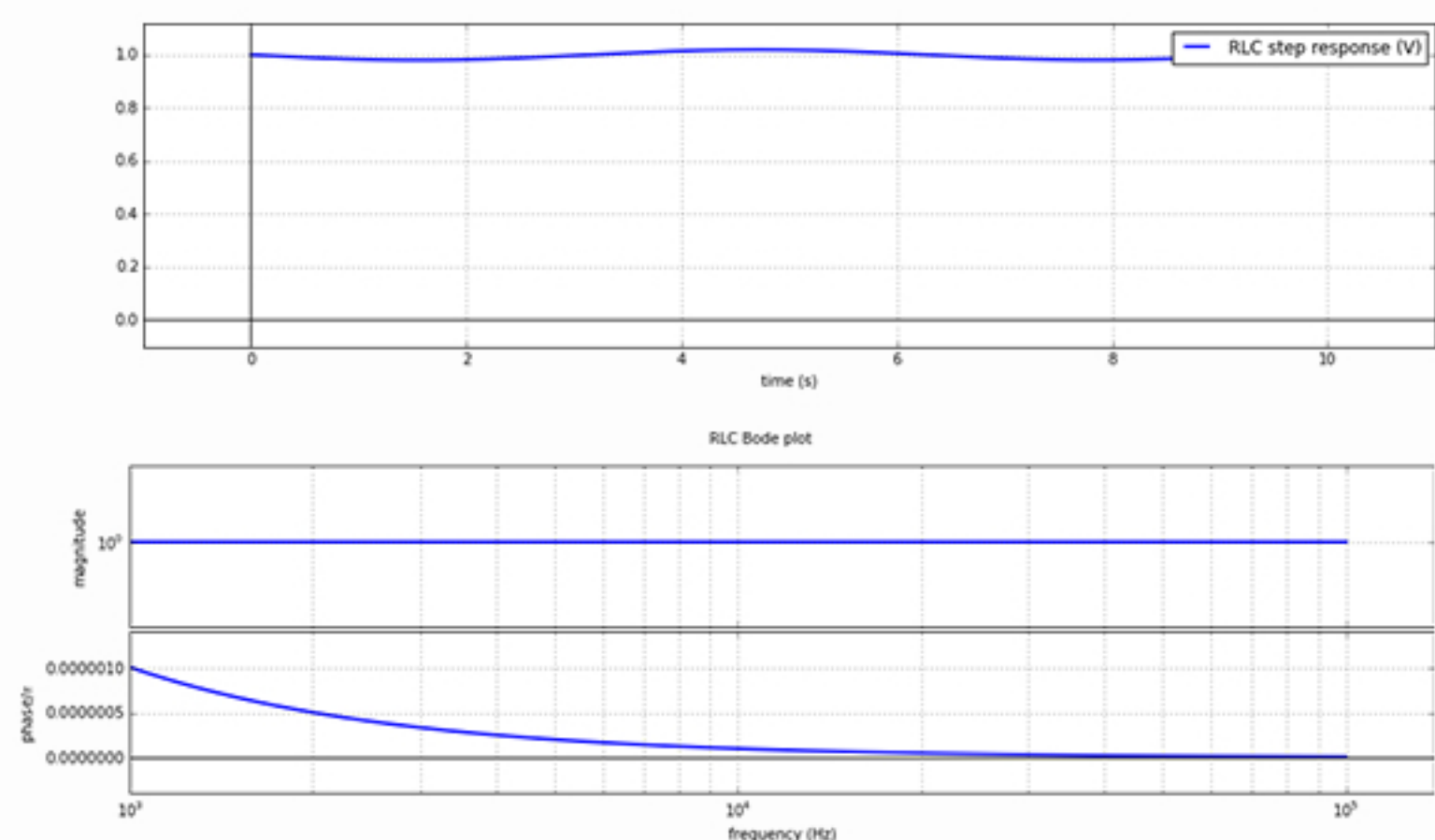
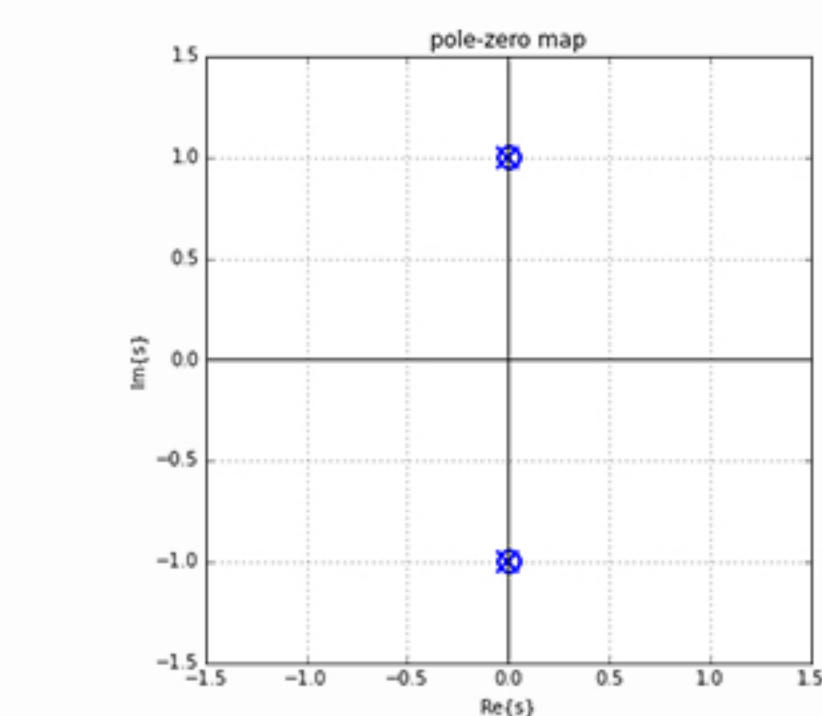
## System C



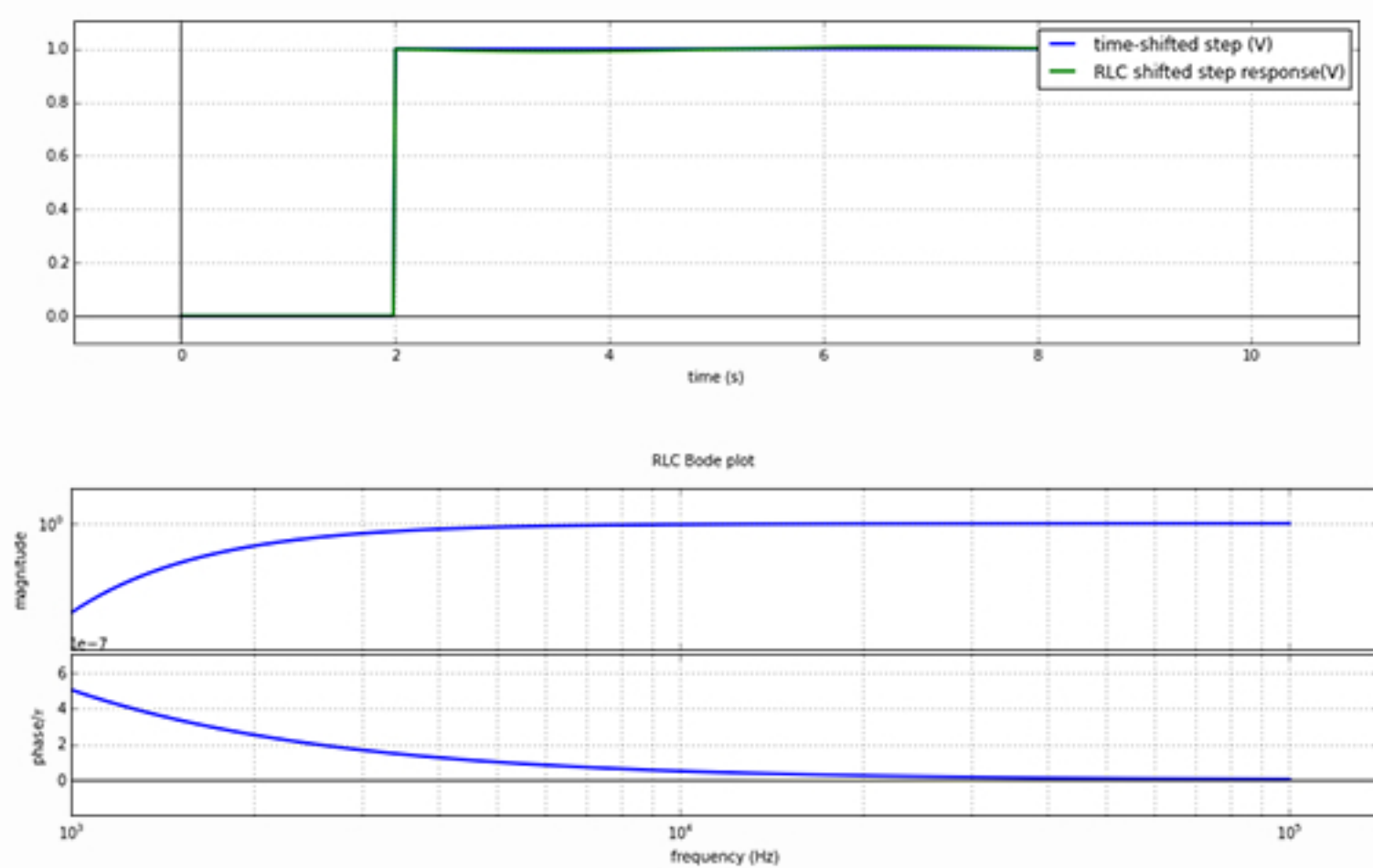
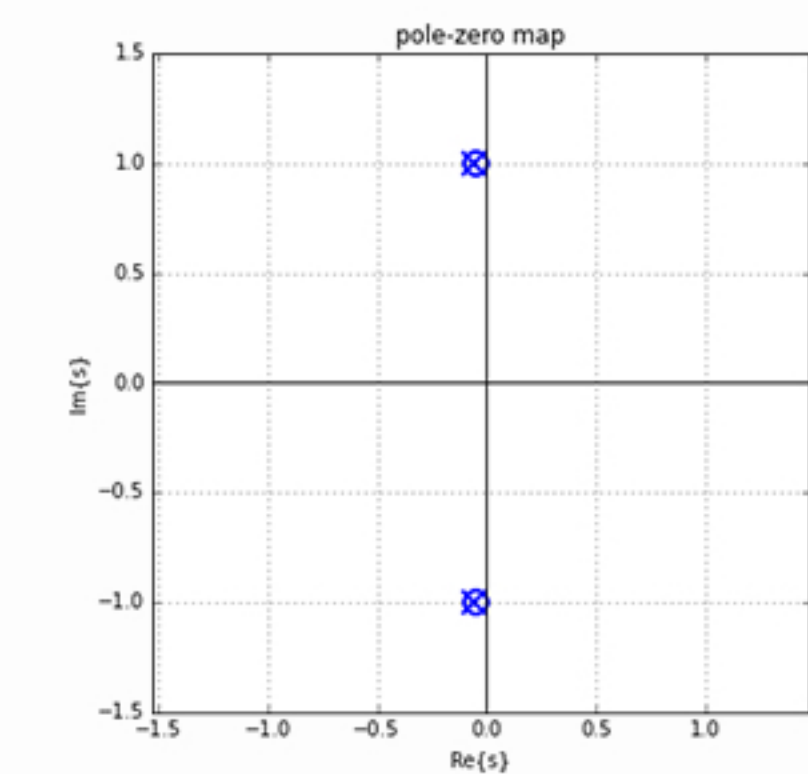
## System D



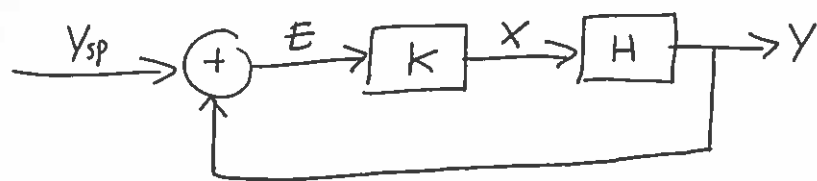
## System E



## System F



# Problem 4



$$H(s) = \frac{1}{s^2 - 0.01s + 1}$$

A. Plot the poles i ~~system~~ step response (see graphs)

System is naturally wanting to oscillate -1 to 1

B. Show the effects of using proportional control on the system

$$\frac{Y}{Y_{sp}} = \frac{K_p H}{1 + K_p H} = \frac{K_p \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)}{1 + K_p \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)} = \frac{K_p}{K_p + s^2 - 0.01s + 1}$$

So it doesn't look like you can stabilize this system using proportional control. You just get various frequencies of oscillations.

C. Show the effects of using integral control  $K(s) = K_I / s$

$$\frac{Y}{Y_{sp}} = \frac{K_I / s \cdot H}{1 + K_I / s \cdot H} = \frac{K_I / s \cdot \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)}{1 + K_I / s \cdot \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)} = \frac{K_I / s}{K_I / s + s^2 - 0.01s + 1} \quad (s)$$

$$\frac{Y}{Y_{sp}} = \frac{K_I}{K_I + s^3 - 0.01s^2 + s}$$

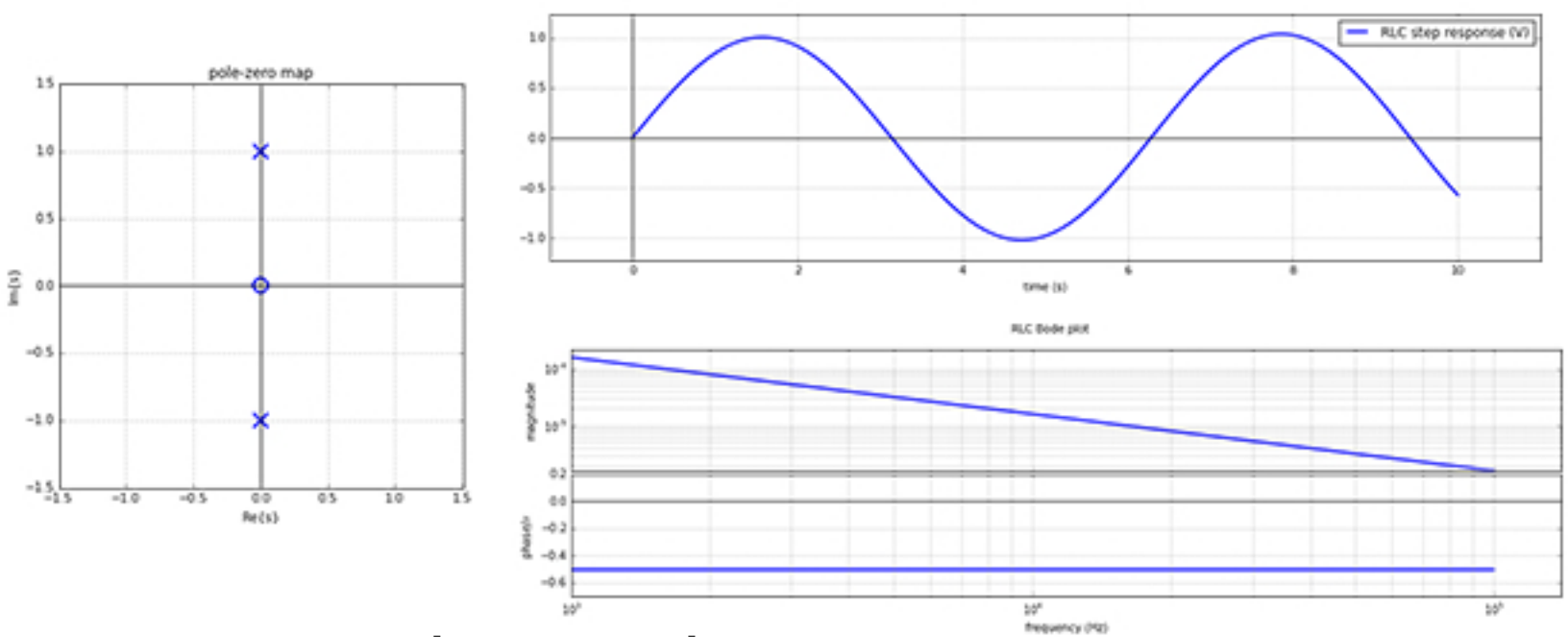
It looks like integral control doesn't help much either

D. Show the effects of using derivative control  $K(s) = K_D s$

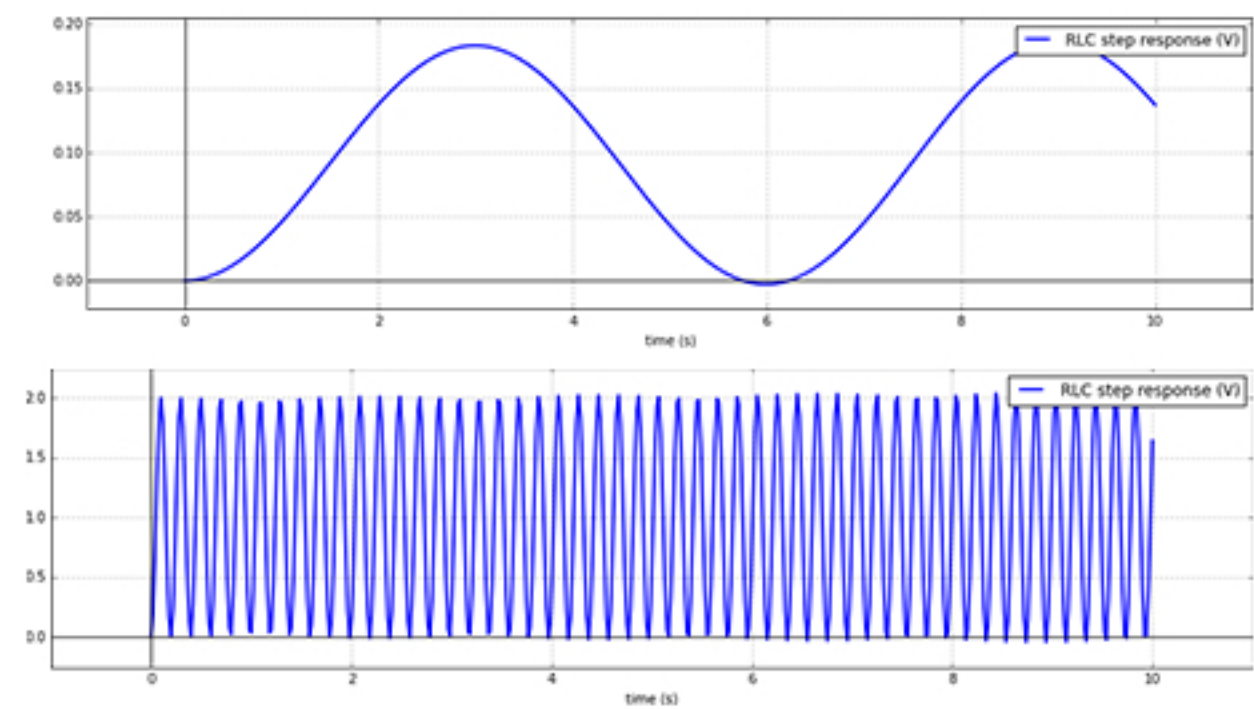
$$\frac{Y}{Y_{sp}} = \frac{K_D s H}{1 + K_D s H} = \frac{K_D s \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)}{1 + K_D s \left( \frac{1}{s^2 - 0.01s + 1} \right) \left( \frac{s^2 - 0.01s + 1}{s^2 - 0.01s + 1} \right)} = \frac{K_D s}{K_D s + s^2 - 0.01s + 1}$$

It looks like you need to use derivative control for this system. For high values of  $K_D$  it basically daves/follows the step.

# Original System - Step response & Poles



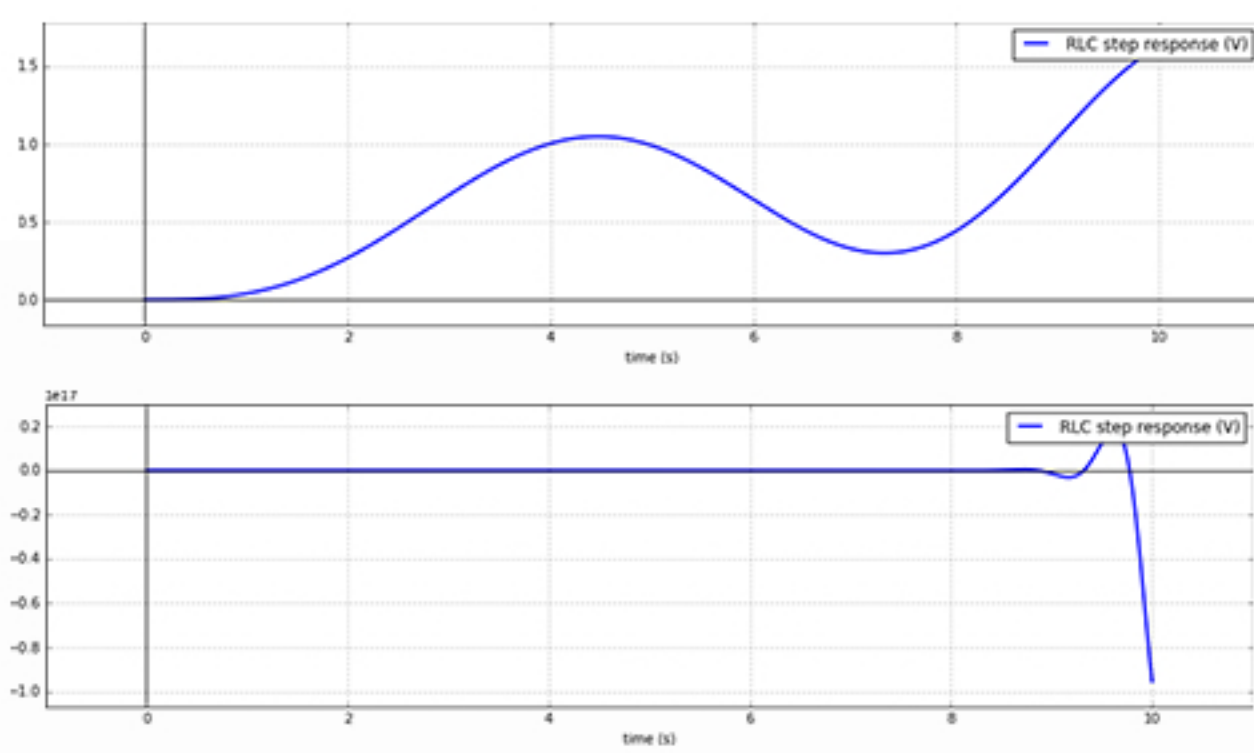
## Proportional Control



$K_p = 1$

$K_p = 1000$

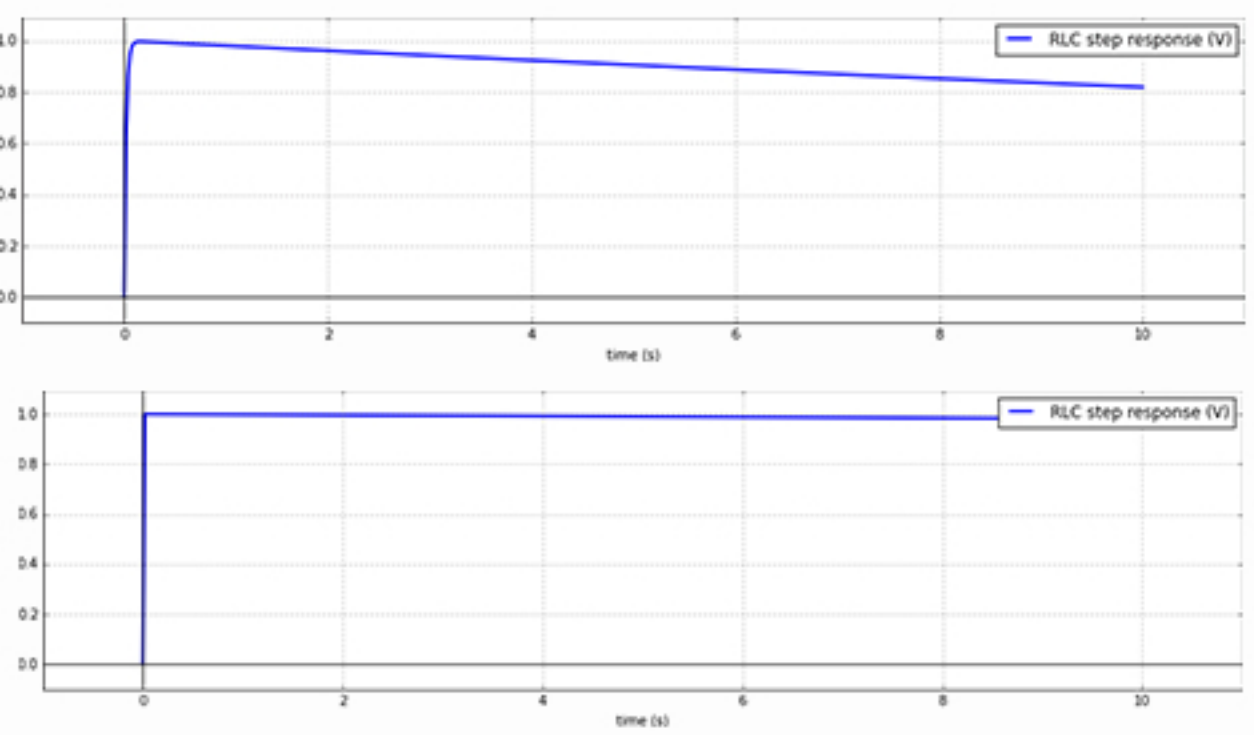
## Integral Control



$K_I = 25$

$K_I = 500$

## Derivative Control



$K_D = 50$

$K_D = 500$