## ECE6703J

Computer-Aided Design of Integrated Circuits

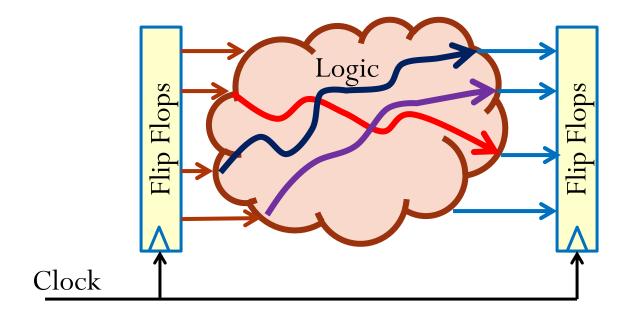
Static Timing Analysis; Interconnect Modeling

#### Outline

- Static Timing Analysis: Algorithm
- Interconnect Timing
  - Electrical Models of Wire Delay
  - Elmore Delay Model

# The Most Typical STA Problem

- Answer this problem: What are all the too-slow paths that violate timing?
- Most useful report:
  - Report paths in order, from slowest to fastest.
  - In other words: **Enumerate** these paths, in delay order.

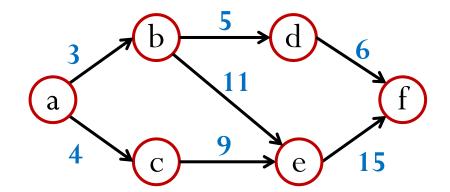


#### What Do We Need?

- Calculate all the **ATs**.
- Calculate all the **RATs**.
- Calculate all the **Slacks**.
- ... do all of these very **efficiently**: Delay graphs are huge!
- Enumerate the violating paths, in worst delay order.

# Computational Strategy

- Topological sorting ("Topsorting") the delay graph.
  - Sort the vertices in the delay graph into one single ordered list.
  - Essential property: if there is an edge from p to s, then p appears before s in sorted order.
- Compute ATs by going **forward** through the sorted list.
- Compute RATs by going backward through the sorted list.



Legal Topsorting Order a, b, c, d, e, f a, b, d, c, e, f

# Compute ATs

```
computeATs() {
  AT(SRC) = 0;
  foreach ( n in topsort order ) {
    AT(n) = -\infty;
    foreach ( node p in pred(n) )
       AT(n) = max(AT(n), AT(p) + \Delta(p,n));
                                                    snk
           src
                                         *
                predecessor
```

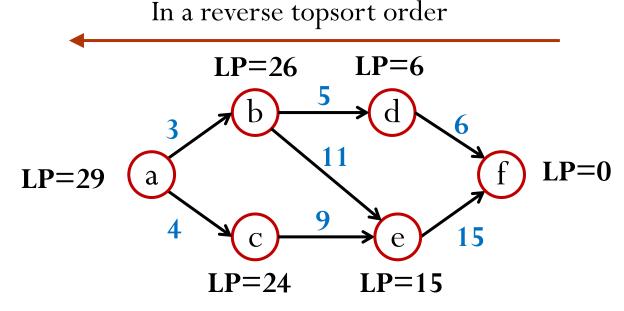
# Compute RATs

• Pretend all edges are **reversed**, they point from SNK to SRC, and walk graph **backwards**.

```
computeRATs() {
  RAT(sink) = CycleTime;
  foreach ( n in reverse topsort order ) {
    RAT(n) = \infty;
    foreach (successor s in succ(n))
      RAT(n) = min( RAT(n), RAT(s) - \Delta(n,s) );
                  Src
                                         (n,s)
                           decessor
```

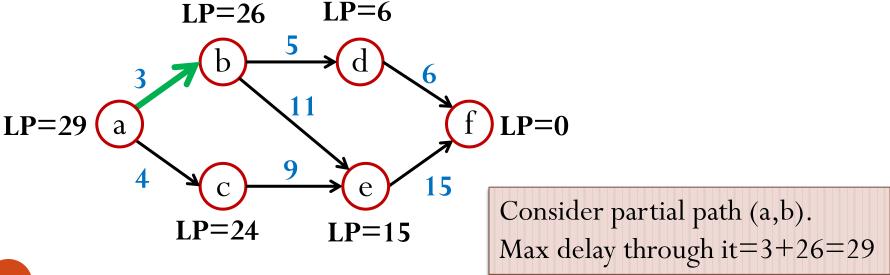
# Path Reporting

- If only the longest path, can use slack property: all nodes on longest path have **same worst slack value**.
- How can we find N worst (longest) paths?
  - We will use **longest path (LP) to sink** for each node (related to RAT)



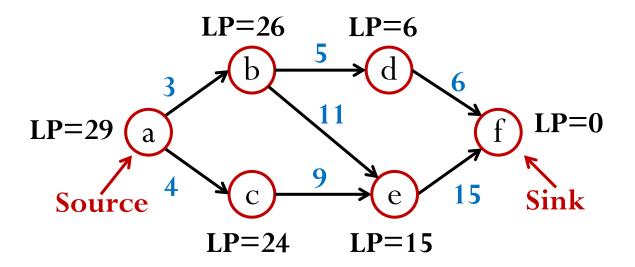
# N-Worst Path Reporting

- We evolve **partial paths**; each partial path stores 3 things: (Path itself, Delay of this path, Max delay **through** this partial path to sink)
  - Max delay through this partial path to sink = Delay of this path
    + longest path (LP) from last node on the path to sink

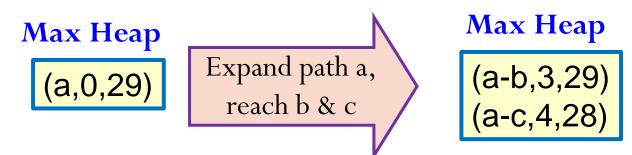


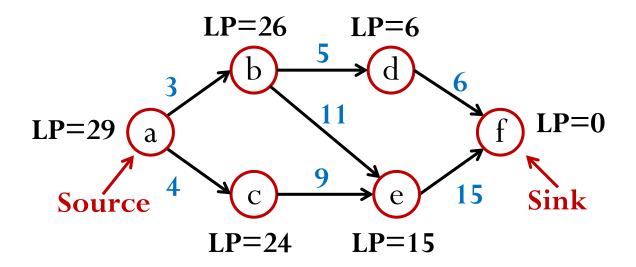
# N-Worst Path Reporting

- We store the partial paths in a **max heap**, which is indexed on the <u>last entry</u>, max delay through the path (i.e., delay+LP).
  - Initially this heap contains only the **source node**.
- Algorithm is quite simple (and just like maze routing!).
  - **Expand**: Pop the partial path that has the largest last entry off the heap
    - Reach target? If its end node is the sink, print out the path
  - Reach: Else add each successor node to make new partial paths, push them back onto the heap, each with (Path, Delay, Delay+LP) labeled.
  - Go pop next partial path until N paths are reported.



- Max heap entry of the form (Path, Delay, Delay+LP)
  - Initially, heap contains only the **source node**.







(a-b,3,29) (a-c,4,28)

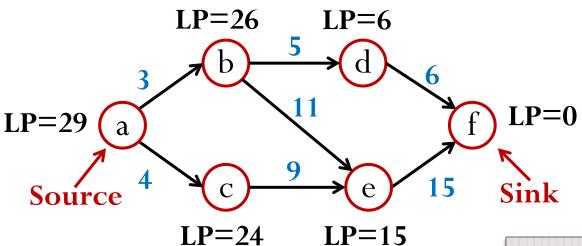
Expand path a-b, reach d & e

#### Max Heap

(a-b-e,14,29)

(a-c,4,28)

(a-b-d,8,14)



#### Max Heap

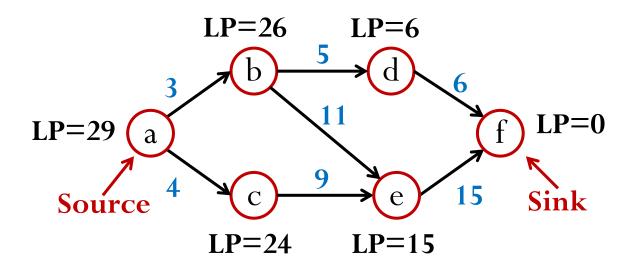
(a-b-e,14,29) (a-c,4,28) (a-b-d,8,14)

Expand path a-b-e, reach f

f is **sink**! Report 1<sup>st</sup> worst path a-b-e-f, with delay=29

#### **Max Heap**

(a-c,4,28) (a-b-d,8,14)

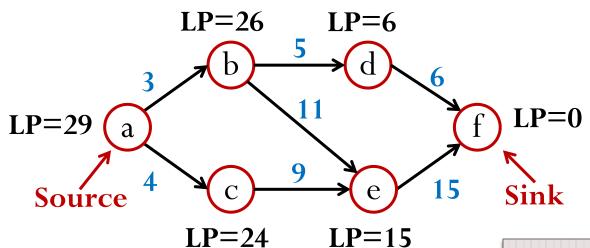




(a-c,4,28) (a-b-d,8,14) Expand path a-c, reach e

#### Max Heap

(a-c-e,13,28) (a-b-d,8,14)



#### **Max Heap**

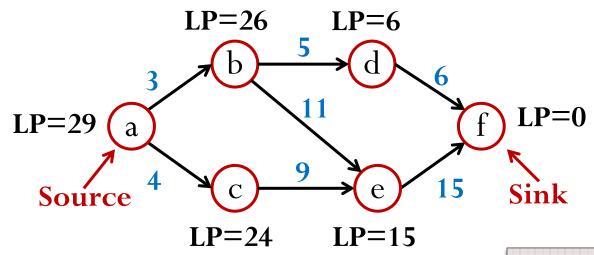
(a-c-e,13,28) (a-b-d,8,14)

Expand path a-c-e, reach f

f is **sink**! Report 2<sup>nd</sup> worst path a-c-e-f, with delay=28

Max Heap

(a-b-d,8,14)



**Max Heap** 

(a-b-d,8,14)

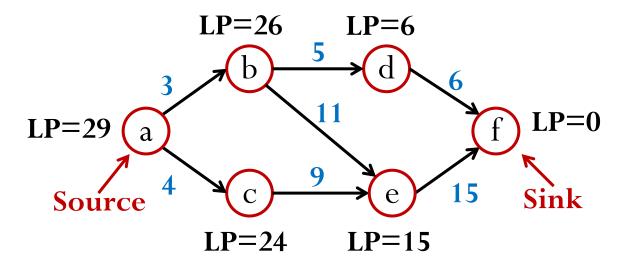
Expand path a-b-d, reach f

f is **sink**! Report 3<sup>rd</sup> worst path a-b-d-f, with delay=14

Max Heap

(EMPTY)

Done!



- We find three paths:
  - a-b-e-f, delay = 29
  - a-c-e-f, delay = 28
  - a-b-d-f, delay = 14.

Note: only 3 possible paths from source to sink in graph, so we found them correctly in delay order!

# Static Timing Analysis: Summary

- STA is a **very important** step in design of complex ASICs.
  - It's a critical "sign off" step, which means: you don't get to fabricate unless you pass.
- Several big ideas
  - Gate level delay models matter, and can be pretty complex in real world.
  - Logical ≠Topological path analysis (i.e., STA).
  - Build delay graph, calculate ATs, RATs, slacks recursively.
  - Concept of **slack** is **important**: lets us locate worst paths, and problem gates on path.
  - A similar idea to maze routing lets us find worst paths in delay order.

# Static Timing Analysis: Aside

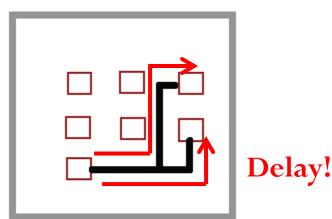
- STA is a **huge** topic several things we did not cover.
- STA for sequential elements
  - How do we model flip flops and latches, so we can verify, e.g., that setup and hold times are met? More tricks with delay graph.
- Early mode versus late mode timing
  - Our development was only so-called late mode timing, where we care about longest path. Early mode focuses on shortest paths, and is critical for more advanced timing.
- Incremental STA
  - In practice, you change 10,000 gates out of 1,000,000 gates, you don't want to **redo** the whole STA analysis. Advanced methods can update **incrementally**.

#### Outline

- Static Timing Analysis: Algorithm
- Interconnect Timing
  - Electrical Models of Wire Delay
  - Elmore Delay Model

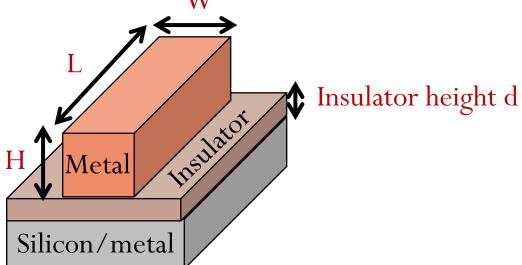
# Interconnect (Wire) Delay Modeling

- You place the logic. Pins are put at a certain distance apart.
- You route the wires. Each wire has an input-to-output delay.
- Questions:
  - Where does the delay come from?
  - How accurately can we predict this delay?
  - How **efficiently** can we **model** this delay for use in layout synthesis or STA?



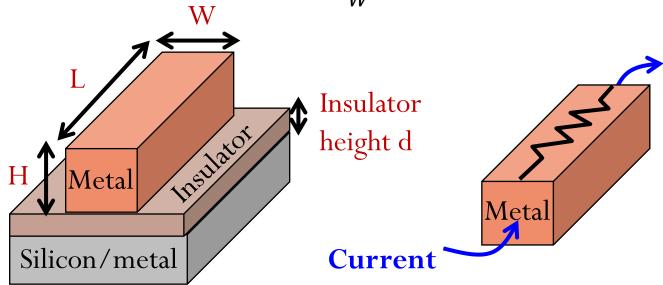
#### Interconnect Model

- We model interconnect as an **electrical circuit** and analyze it as a circuit.
- The model parameters depend critically on **exact geometry** of the wired net.
- Most popular interconnect model used in layout applications:
   RC trees.



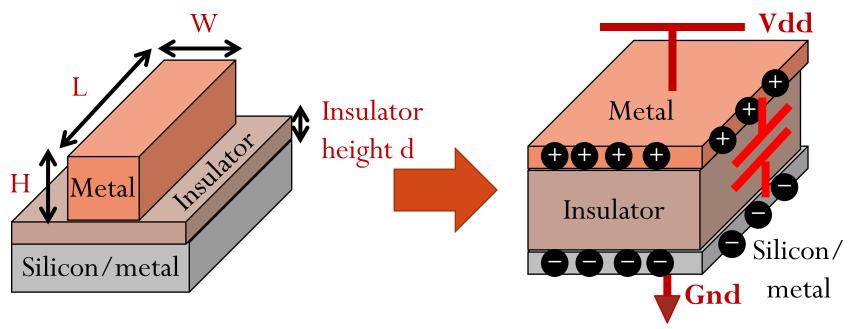
#### Interconnect Model: Resistance

- Metal wire has **resistance R** to current flowing down its length.
  - In physics:  $R = \frac{\rho L}{S} = \frac{\rho L}{WH}$
  - In ASIC design: H is a manufacturing factor. Only W and L can be controlled  $\Rightarrow R = \frac{rL}{W}$



# Interconnect Model: Capacitance

- Metal wire has **capacitance C** to silicon substrate, with insulator between.
  - In physics:  $C = \frac{\varepsilon WL}{d}$
  - In ASIC design: d is a manufacturing factor. Only W and L can be controlled  $\Rightarrow C = cWL$

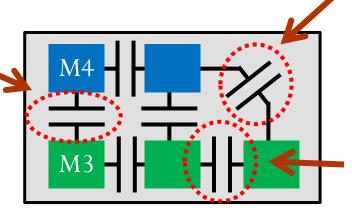


## Aside: About Real Capacitance (Cap)

- Note: this model is very **simplistic**.
- You really get capacitance between any pair of conducting surfaces.
  - So, in a multi-layer metal process you get caps between all the layers.

#### Overlap cap

between 2 adjacent wires on different layers



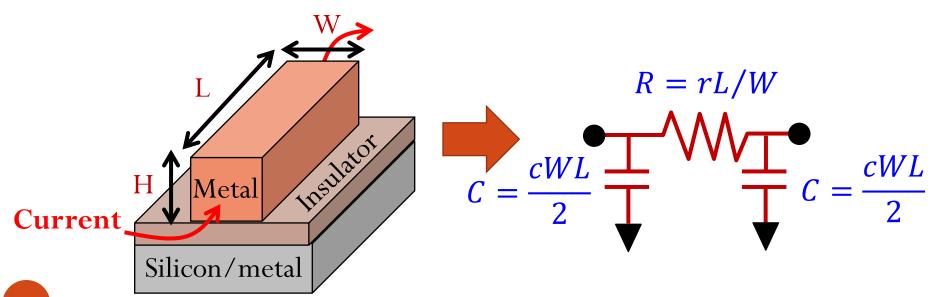
#### Sidewall fringe cap

from side of one layer to the conductors below it

Fringe cap between 2 adjacent wires on the same layers.

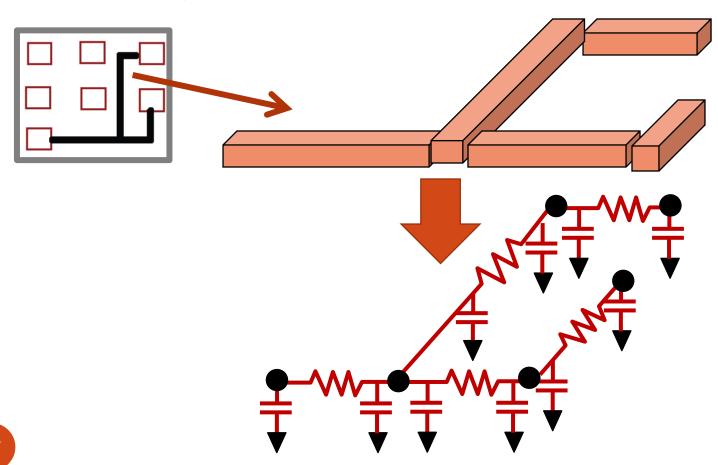
## Typical Interconnect Model: π Model

- Accounts for the resistance R and the capacitance C of wire segment.
- It is a **small** model: only need 2 numbers.
- It is a **symmetric** model: **split** capacitance into two halves.



# From Wire Segments to RC Tree

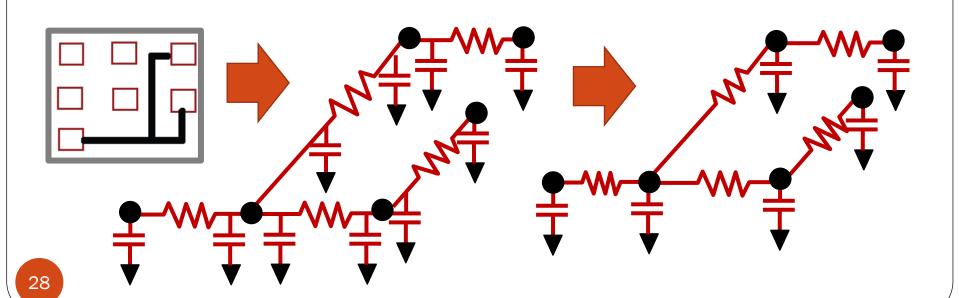
• Big idea: Replace every straight wire segment with  $\pi$  model.



#### From Wire Segments to Final RC Tree

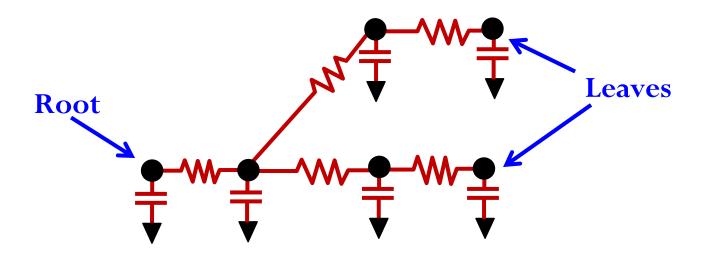
- Simplification: Recall a rule from basic circuits:
  - ullet Parallel capacitors can be replaced by 1 capacitor with  $\Sigma \mathcal{C}_i$

$$C1 \stackrel{\longleftarrow}{\downarrow} C2 \stackrel{\longleftarrow}{\downarrow} C3 \stackrel{\longleftarrow}{\downarrow} C1 + C2 + C3$$



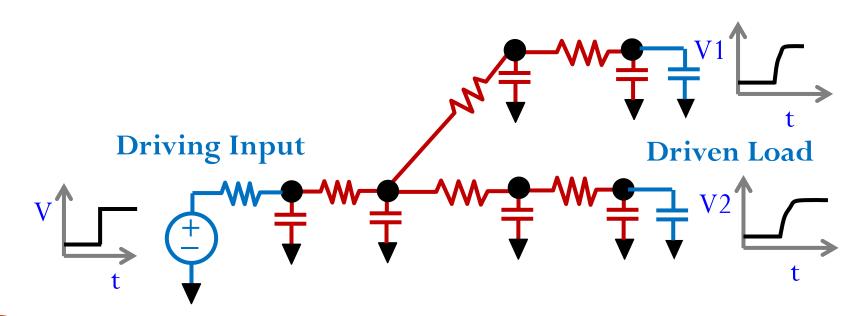
#### General Form of RC Tree

- Capacitors "hanging off" all tree nodes.
- Tree edges are **resistors**.
- Root of tree is where signal is input.
- Leaves of tree are the driven outputs.

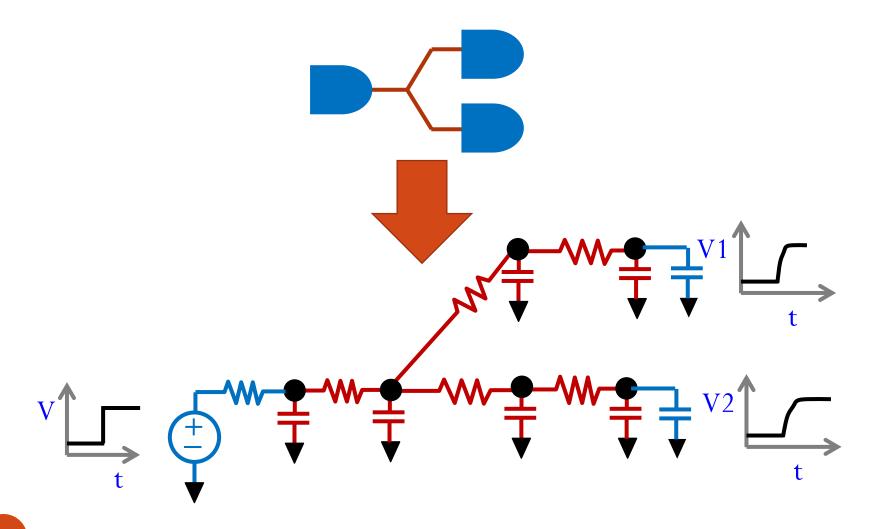


# Model the Driving and Driven Gates

- We also need to model driving gate and driven gates.
  - Voltage source + resistor as input at root. This models driving gate.
  - Capacitor as load at each leaf. Each models a driven gate.



# Summary: RC Tree Is Built from Gates + Wires

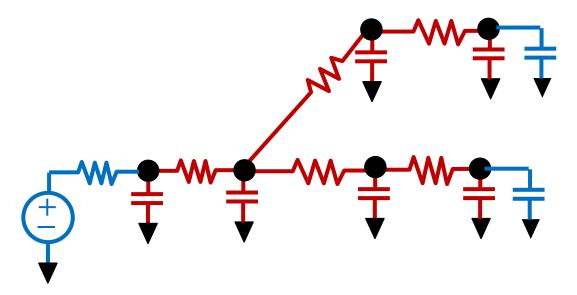


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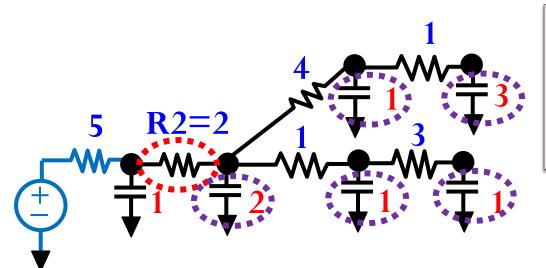
# Calculating Delay

- How to calculate a **delay** number for **each output** of tree?
- Famous formula: "Elmore" delay
  - Derived in 40s for circuits applications.
  - Resurrected in 80s by Penfield et al. for RC trees.
  - A very **simple**, but very **useful** computational recipe.



# Elmore Delay $\tau$ : Tree Walk Computing Recipe

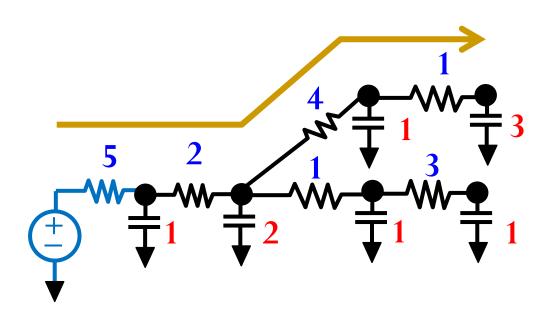
- Set  $\tau = 0$ . Walk down path of resistors from **Root** to **Leaf** where you want to calculate delay.
- At each resistor, let  $\tau = \tau + R \cdot \Sigma$  (all capacitors **downstream**)
  - **Downstream capacitor** = any C that is reachable in tree below this resistor.



Example: at R2 resistor in RC tree, the downstream capacitors are 2, 1, 1, 3, 1

# Calculating Elmore Delay: Example

- Set  $\tau = 0$ . Walk down path of resistors from **Root** to **Leaf** where you want to calculate delay.
- At each resistor, let  $\tau = \tau + R \cdot \Sigma$  (all capacitors downstream)



$$\tau = 0$$

$$+ 5 \times (1 + 2 + 1 + 3 + 1 + 1)$$

$$+ 2 \times (2 + 1 + 3 + 1 + 1)$$

$$+ 4 \times (1 + 3)$$

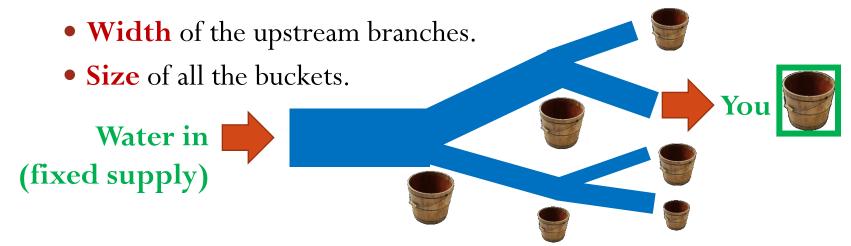
$$+ 1 \times 3$$

$$= 45 + 16 + 16 + 3$$

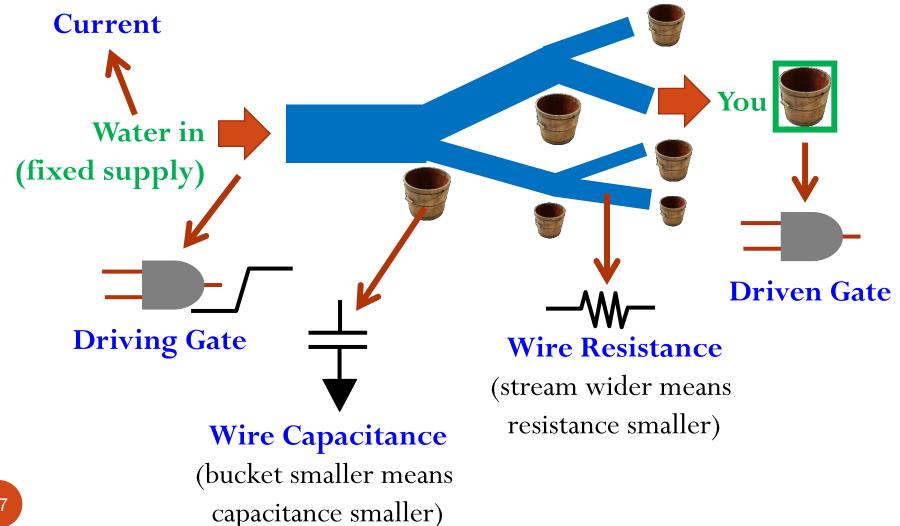
$$= 80$$

# Insight: Stream Analogy

- Think of RC tree like branching stream, current like water.
  - Goal: You are downstream, trying to fill your bucket. How **fast** can you fill it?
  - However, at **every** branch point, somebody else has a **bucket**.
  - The **farther** you are downstream, the **less** water you get from upstream.
- What matters here?



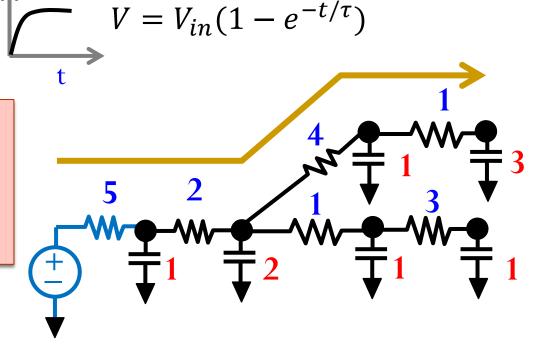
#### From Stream to Circuit



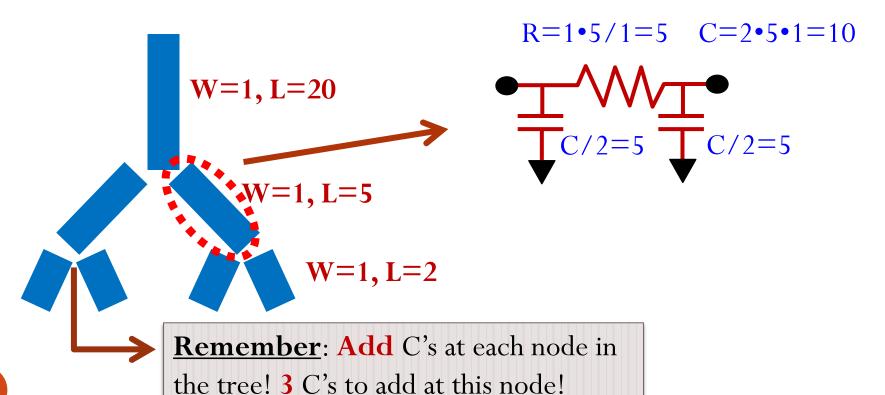
## Circuit Aside: What Is Elmore Delay?

- If you want to model the path from input to an output as a simplified circuit with **exactly one** R and **one** C, the **best approximation** you have will have its R•C = Elmore delay.
  - Note: for a circuit with one R and one C, R•C is the **time** constant  $\tau$ .

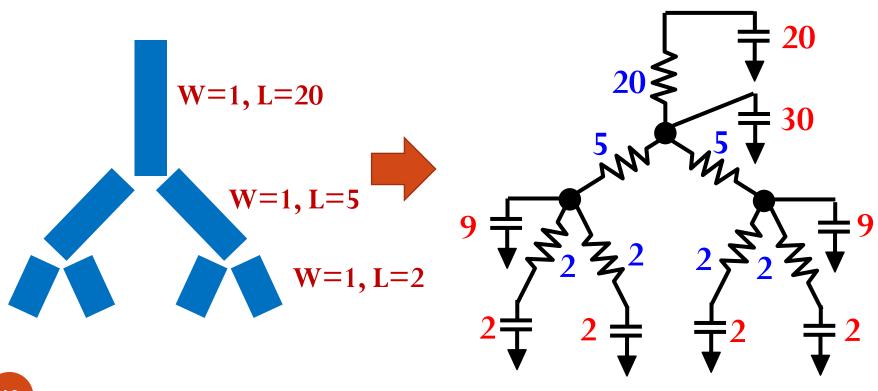
Elmore delay is the time constant of the best approximating circuit with one R and one C



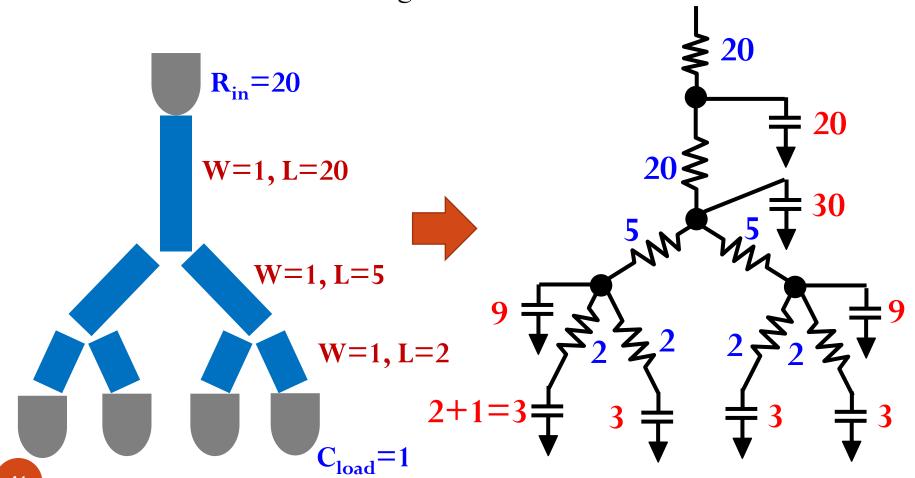
- Simple tree with 4 leaf nodes
  - For each segment, total R=rL/W, C=cWL.
  - Electrical parameters: r=1, c=2.



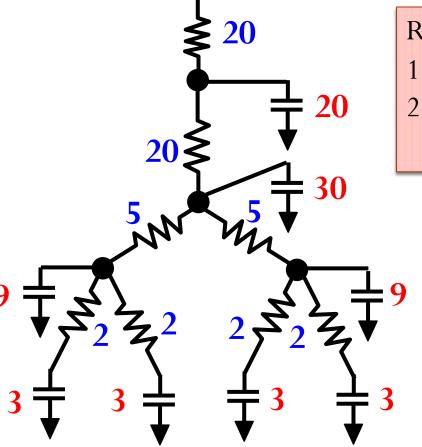
- RCTree for the interconnect alone
  - Remember to add up C's hanging off each internal node of tree



• Add driver and driven gates



- Now we can compute delay to each leaf.
  - Since symmetric, only need to compute 1 path.



#### Remember the recipe:

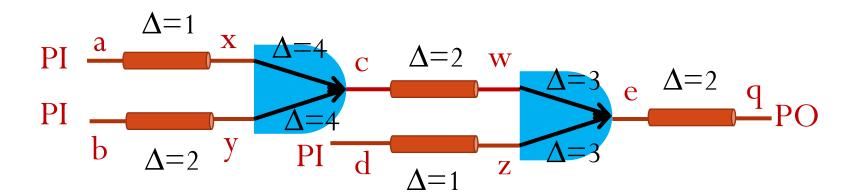
- 1. Set  $\tau$ = 0, walk from root to leaf
- 2. At each R,

$$\tau += R \cdot \Sigma$$
 (all C's downstream)

$$\tau = 0$$
+20×(20+30+2×9+4×3)
+20×(30+2×9+4×3)
+5×(9+2×3)
+2×3
= 2881

# Elmore Delay Applications

- Timing verification
  - Can use this to give realistic post-layout **wire delays** for final STA.



# Elmore Delay Applications

- During placement
  - Estimate wire shape (e.g., a simple Steiner tree), then you can get very quick delay estimate.
  - Analytical placers use to adjust **weights** on wires, force critical wires to be **short**.

# Summary

- Interconnect has a **huge** impact on chip speed.
  - Cannot ignore delays caused by the electrical properties of real wires
- Individual wires are today modeled as complex circuits.
  - RC tree is the most useful model; Elmore delay is easiest to compute
  - Can use for both verification, and for layout optimizations
  - Accurate enough that they beat simple length-based schemes.
  - Unfortunately, not so accurate that you can avoid later verification with "higher order" models that incorporate more than one time constant.
  - There are sophisticated estimators beyond Elmore, but they takes more time.