

Noise in Digital Signals

ENCE361 Embedded Systems 1

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Where we're going today

- **Sources of noise**
- Signal-to-noise ratio
- Noise margin and hardware interfacing
- Homework

Sources of Noise: Overview

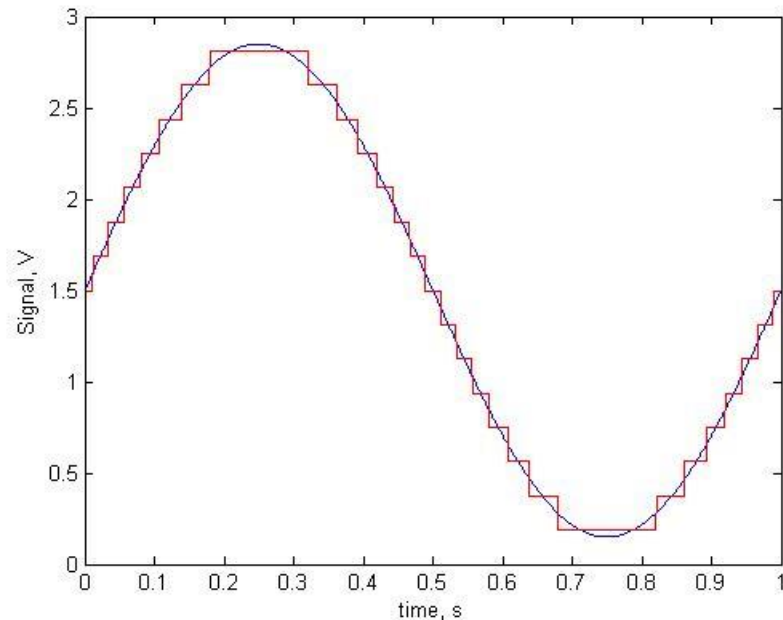
- Definition of noise
 - Disturbance that are **unwanted and unrelated** to information content in a digital signal
- **Random noise** vs. systematic error
 - **Random noise** is **stochastic**
 - Quantization error
 - Sampling jitter
 - Thermal effect in electronics
 - Interference from other signals
 - System noise is **deterministic**
 - Measurement offset
 - Non-linear response
 - Generally can be calibrated



Quantization Error

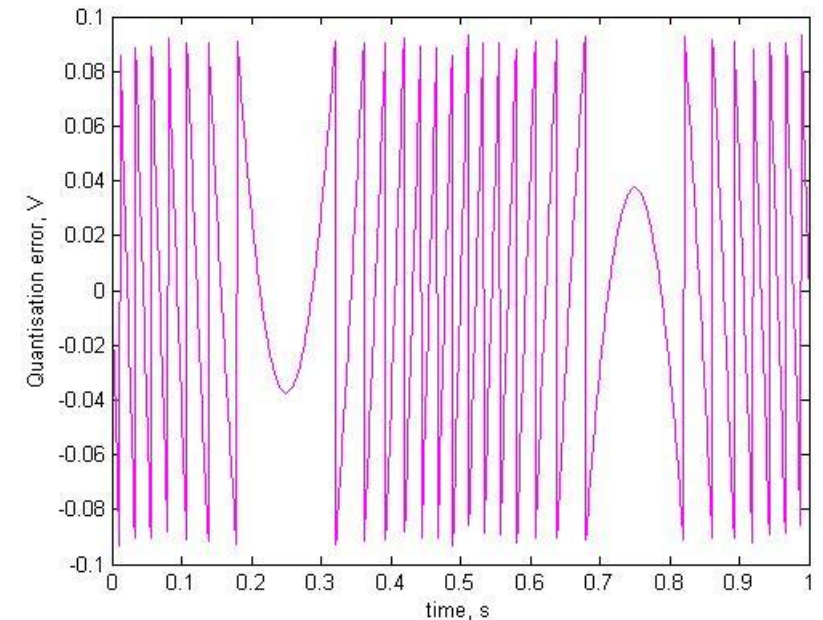
- **Quantization** maps a sample value to one of the possible discrete values

Original signal vs. its **ZOH-sampled & quantised version**



- 4-bit uniform quantization with 16 discrete values
- **Quantization step** $\Delta = 3/16 = 0.188$ Volt

Quantization error



- Random error with maximum magnitude = $\Delta/2 = 0.094$ Volt

Sampling Jitter

- **Sampling jitter** makes sampling interval no long uniform
- **Effect** of sampling jitter
 - Cause sampling amplitude **uncertainty**
 - Example:

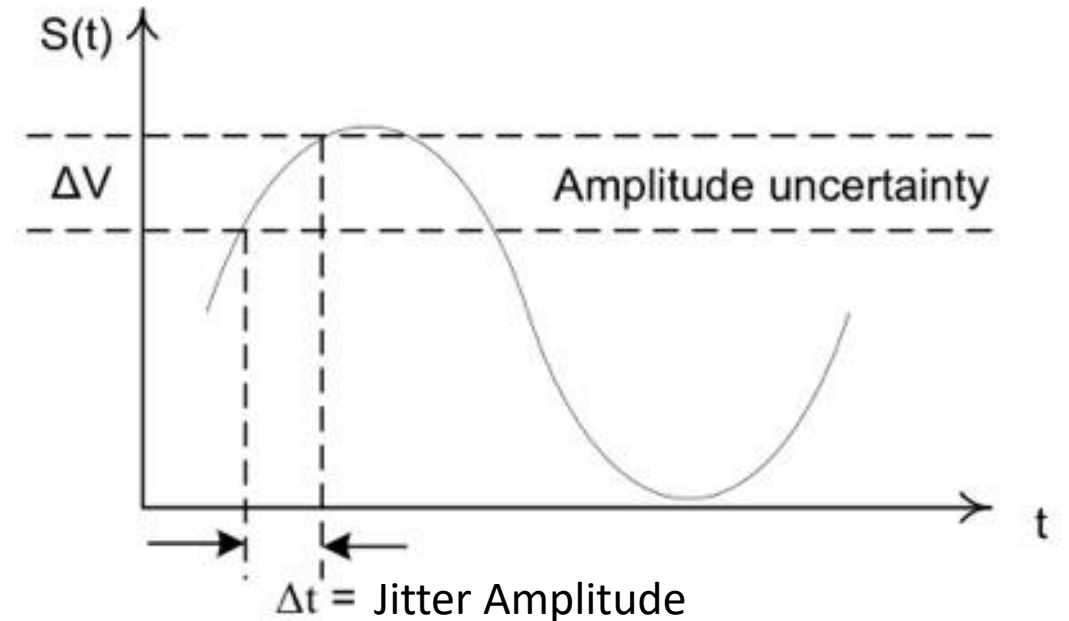
for a sinewave with amplitude A and frequency f Hz, maximum error magnitude is

$$\Delta V = 2\pi f A \Delta t$$

Derivation:

$A \sin(2\pi f t)$ has an instantaneous changing rate $2\pi f A \cos(2\pi f t)$.

Maximum amplitude change after sampling jitter is $2\pi f A \cos(2\pi f t) \Delta t \leq 2\pi f A \Delta t$



Johnson Noise

- Thermal noise

- Due to random thermal motion of charge carriers (usually electrons)
- Occurs regardless of any applied voltage

- A resistor of resistance R generates thermal noise with a RMS voltage

$$v_n = \sqrt{4kTBR}$$

$$v(t) = v(t+T) \quad T = \text{period}$$

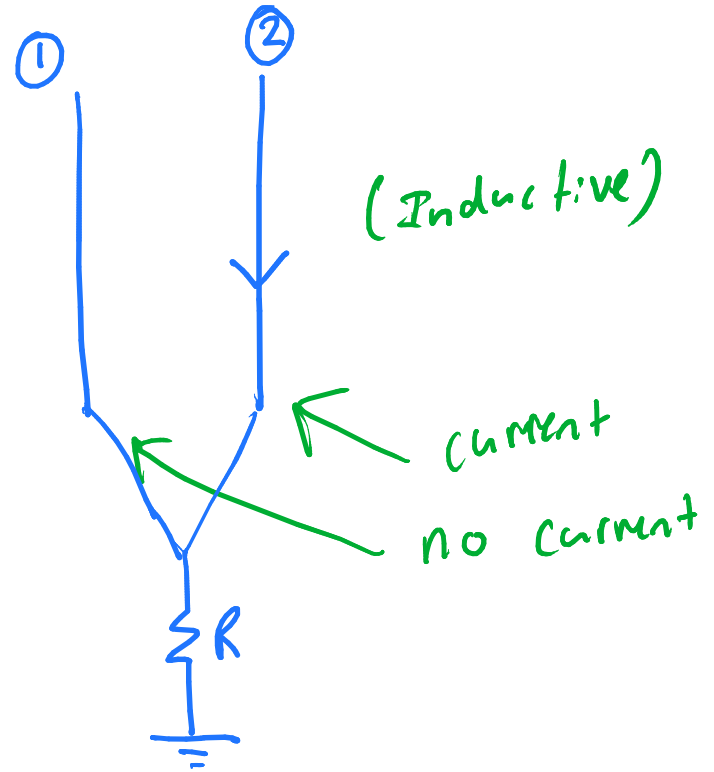
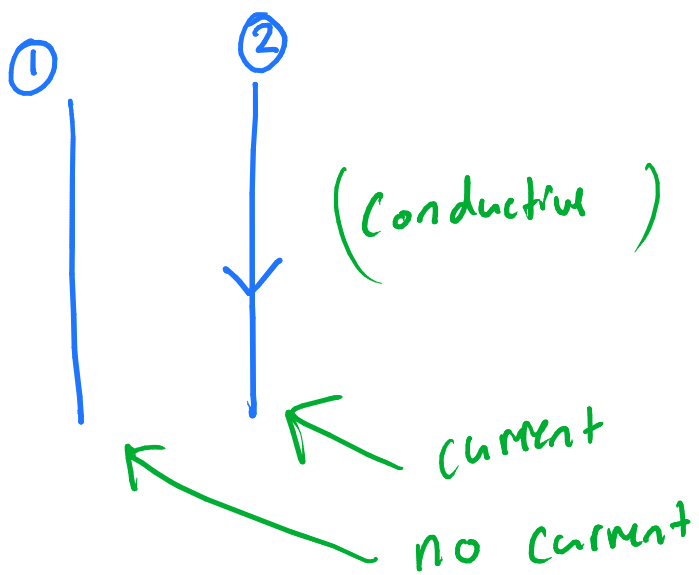
- $k = 1.38 \times 10^{-23} \text{ J/}^\circ\text{K}$ Boltzmann's constant
- T : absolute temperature in $^\circ\text{K}$ (Kelvin)
- B : noise bandwidth in Hz
- R : resistance in Ohm

$$\left[\frac{1}{T} \int_0^T v^2(t) dt \right]^{1/2}$$

RMS voltage

Interference

- 3 sources of interference that could be present simultaneously
 - Conductive coupling: noise current caused by a changing voltage in a nearby circuit
 - Cross-talk between closely spaced circuits
 - Inductive coupling: noise voltage caused by a changing current in a nearby circuit
 - Resistive coupling: occurs when high-level signals share a wire with low-level signals
 - Direct coupling



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Signal-to-Noise Ratio (1)

- Signal-to-noise ratio (SNR)
 - Expressed in **decibel (dB)**

0 dB = noise power & signal power
is the same

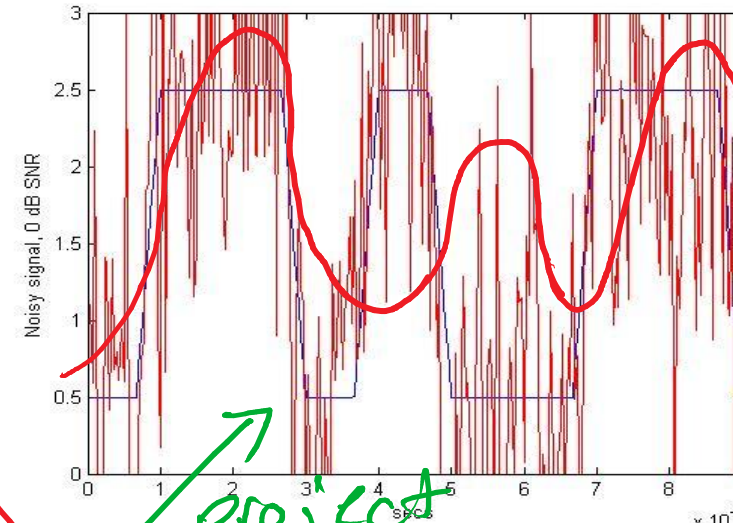
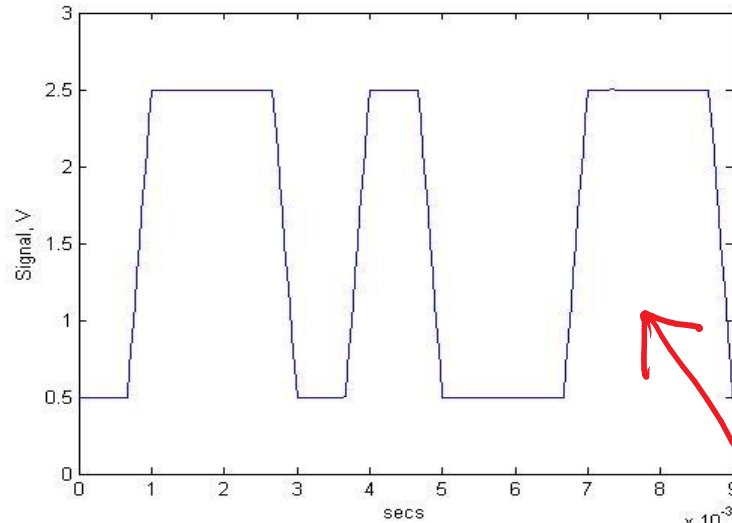
- Definition 1: $SNR = 10 \log \left(\frac{\text{Average Signal Power}}{\text{Average Noise Power}} \right)$

- Definition 2: $SNR = 10 \log \left(\frac{(\text{Signal RMS Voltage})^2}{(\text{Noise RMS Voltage})^2} \right) = 20 \log \left(\frac{\text{Signal RMS Voltage}}{\text{Noise RMS Voltage}} \right)$

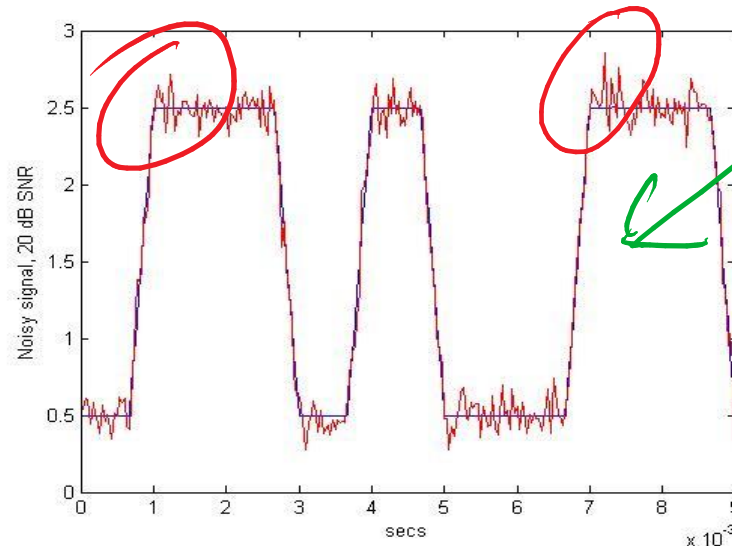
- Use of decibel (dB)
 - Logarithm used to better quantify very large or small values such as SNR

Signal-to-Noise Ratio (2)

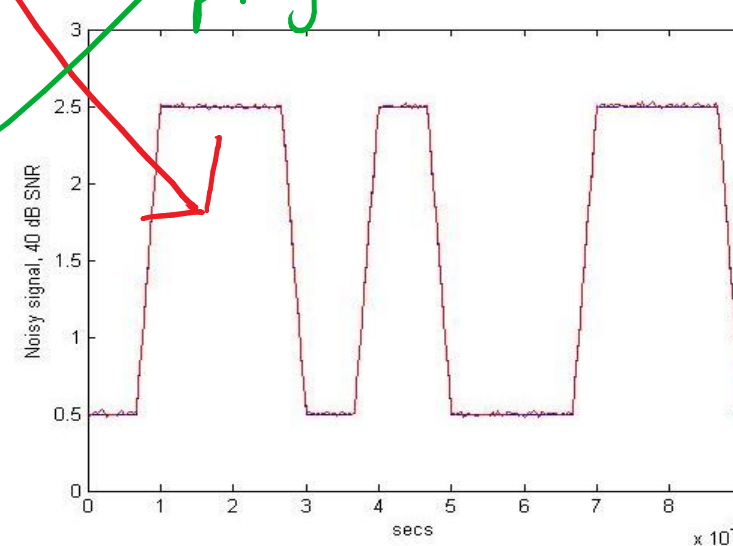
“Clean” signal



SNR = 0 dB
Signal & noise have
SAME power



SNR = 20 dB
Signal power is 100
times higher than
noise power



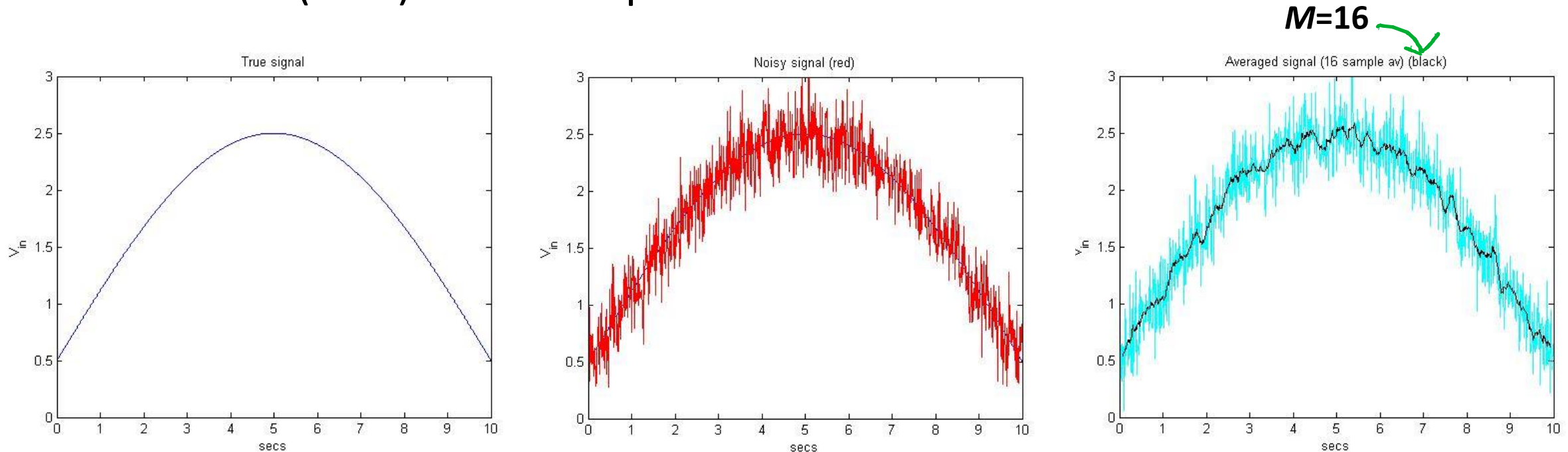
SNR = 40 dB
Signal power is
100,00 times higher
than noise power

Signal Averaging Revisited (1)

- Digital signal conditioning via **signal averaging (slide 15 in previous lecture)**

$$z(nT_s) = \frac{1}{M} \sum_{m=0}^{M-1} y((n-m)T_s)$$

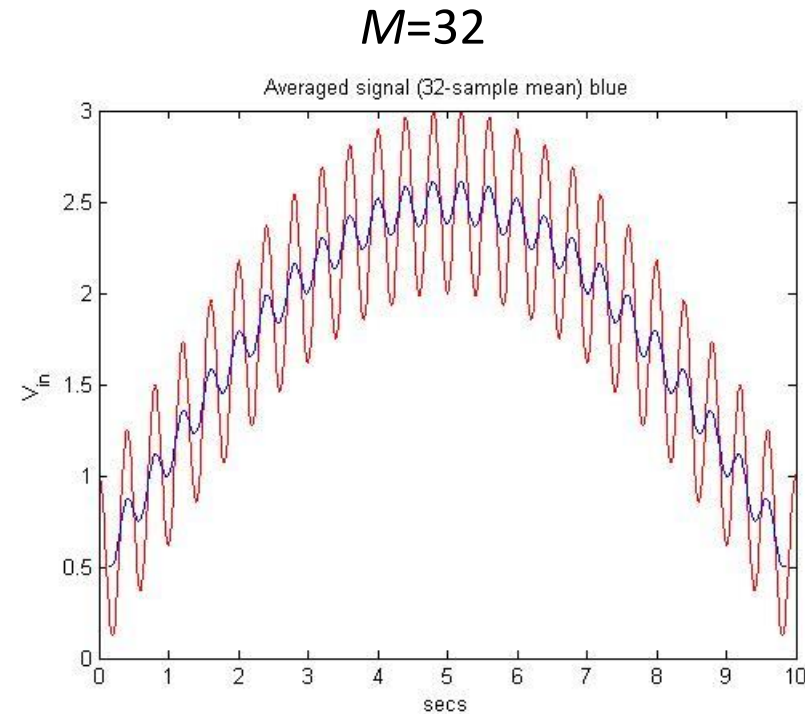
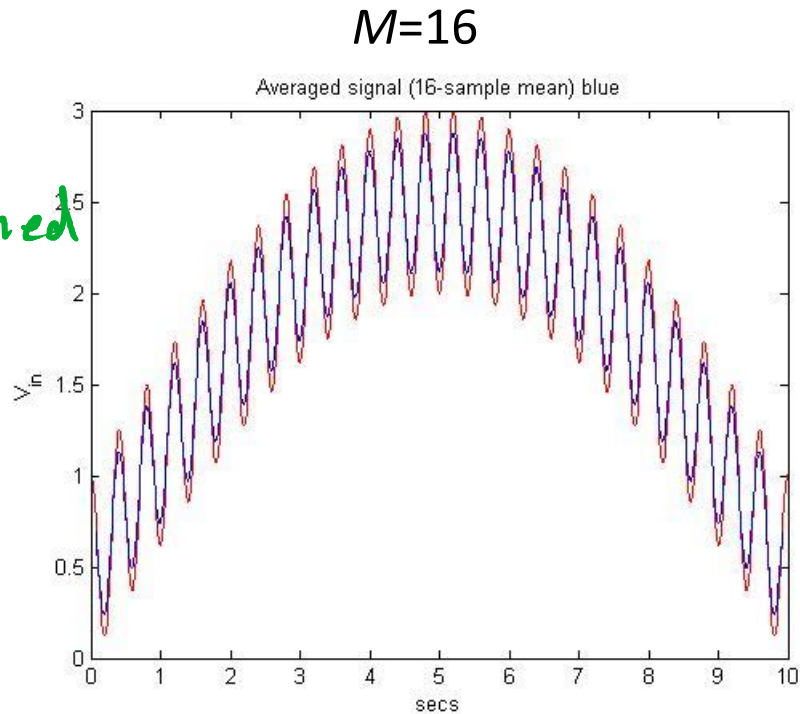
- Reduce noise power by a factor of $1/M$ for independent & identically distributed (i.i.d.) noise samples



$\sin(2\pi ft)$ $f = 2.5 \text{ Hz}$ $t = 0.01 \Rightarrow \frac{1}{2.5} \times 100 = 40$
Signal Averaging Revisited (2) $M = 16$

- If the noise is **NOT** random, the performance of signal averaging may depend on the noise period (in samples) and its relation to M

Red = corrupt
Blue = obtained



- Sample the signal at a sampling rate $f_s = 100 \text{ Hz}$ ($T_s = 0.01\text{s}$)
- The noise is a sinewave at 2.5Hz

Where we're going today

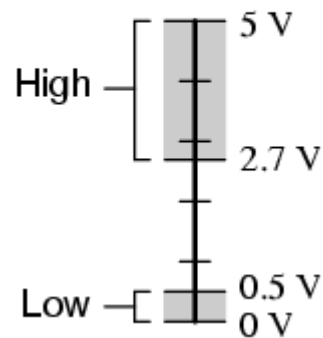
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Noise Margin (1)

tTL = transistor to transistor logic

- Logic gate circuits output and input **logic-1 ('High')** and **logic-0 ('Low')**
 - Ideally, logic-1 is represented by full power supply voltage (say, 5 V)
 - Ideally, logic-0 is represented by zero voltage
- Ideal voltages are **rarely** attained in practice

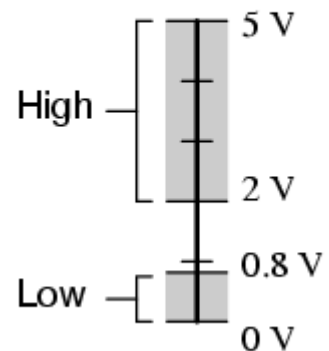
Acceptable TTL gate
output signal levels



2.7-5 V: High

0-0.5 V: Low

Acceptable TTL gate
input signal levels

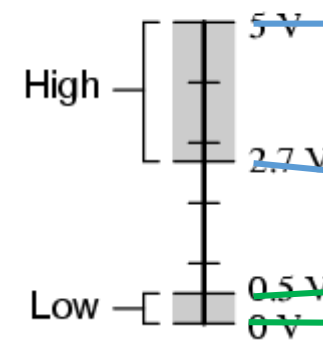


2-5 V: High

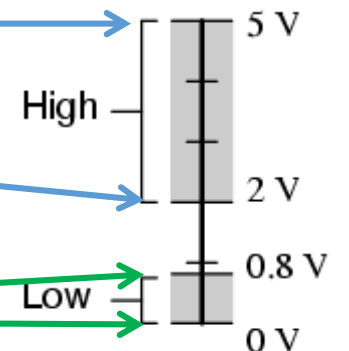
0-0.7 V: Low

0.7-2 V: No certain response!

Acceptable TTL gate
output signal levels

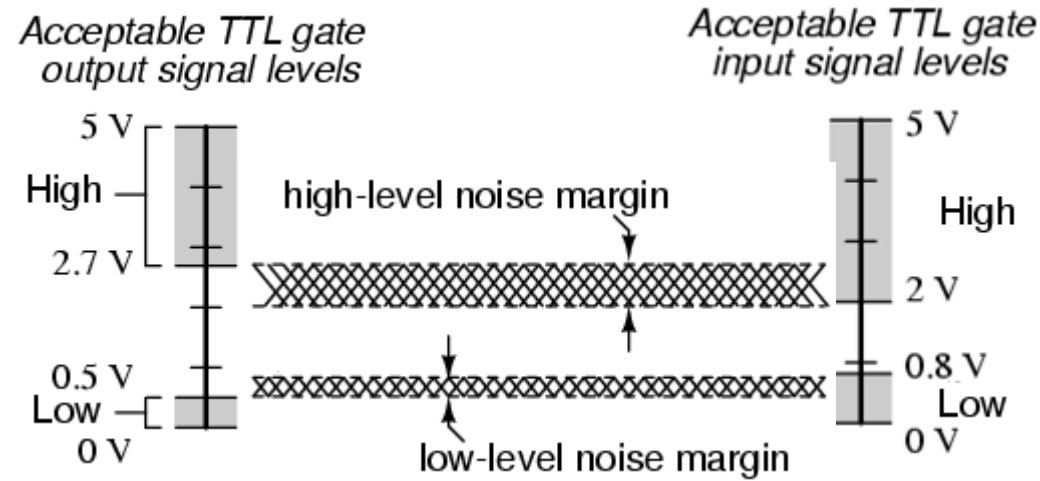


Acceptable TTL gate
input signal levels

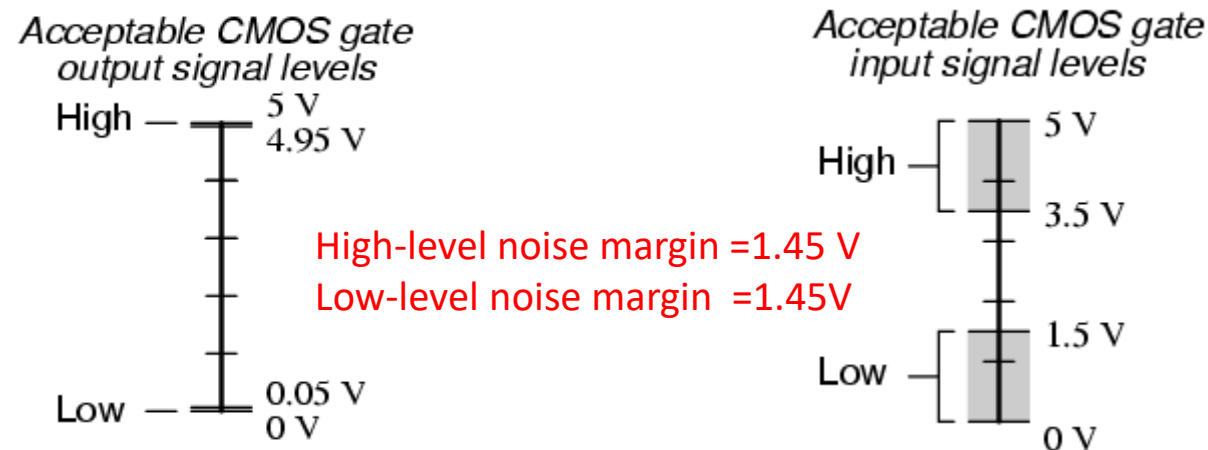


Connect a TTL output to a TTL input

Noise Margin (2)

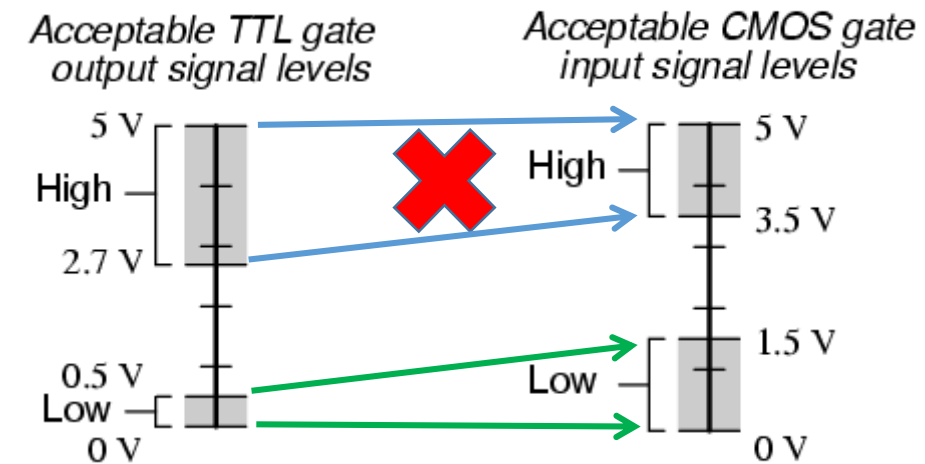
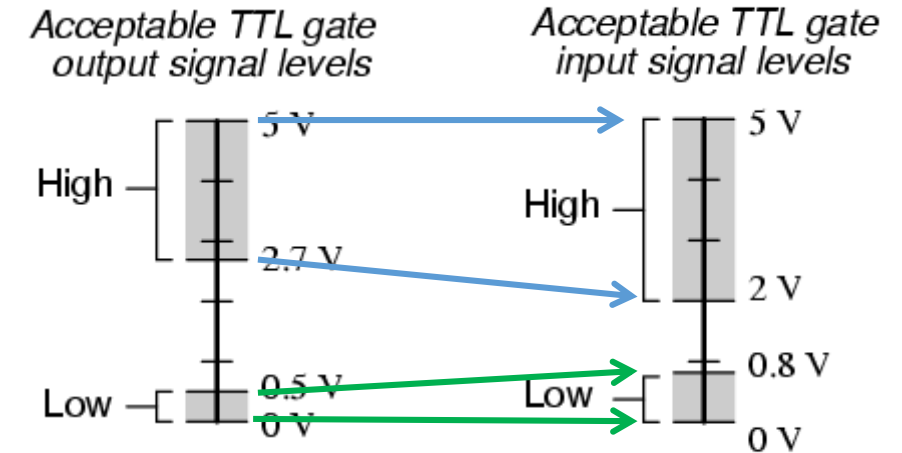


- High-level noise margin in TTL (transistor-transistor logic) = $2.7 - 2 = 0.7 \text{ V}$
- Low-level noise margin in TTL (transistor-transistor logic) = $0.8 - 0.5 = 0.3 \text{ V}$



Hardware Interfacing

- **Compatible** hardware interfacing
 - High-level output range \subset High-level input range
 - Low-level output range \subset Low-level input range
- **Incompatible** hardware interfacing
 - Cause uncertain response at CMOS gate when TTL gate outputs logic-1 even without noise



Homework

1. A primary source of noise in resistors is due to thermal (Johnson) noise
 - What is the noise RMS voltage of a $1\text{ M}\Omega$ resistor at 300° K over a 100 kHz bandwidth?
 - What is the noise RMS voltage of a $1\text{ M}\Omega$ resistor in parallel with a $1\text{ k}\Omega$ resistor at 300° K over a 100 kHz bandwidth?
2. In the example with $\text{SNR} = 20\text{ dB}$ shown on slide 12, the average power of the signal is 0.9 W (ignoring the DC offset). What is the average power of the noise?
3. In slide 13, explain why the noise amplitude on the right (obtained via signal averaging with $M=32$) is significantly less than that on the left (obtained via signal averaging with $M=16$).