Linear System Theory

Jun Moon

Lecture 5

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LTI system

$$\dot{x}(t) = ax(t), \ \dot{x}(t) = Ax(t)$$

For $a \in \mathbb{R}$, $x(t) = e^{at}x(0)$. Also,

$$e^{at} = 1 + at + \frac{a^2t^2}{2!} + \frac{a^3t^3}{3!} + \cdots$$

Hence, for $n \times n$ matrix A, $x(t) = e^{At}x(0)$, where

$$e^{At} = I + At + \frac{A^2t^2}{2!} + \frac{A^3t^3}{3!} + \cdots$$

Note that A and e^{At} commute $(Ae^{At} = e^{At}A)$, since

$$\frac{de^{At}}{dt} = A + AAt + A\frac{At}{2!} + \dots = A(I + At + \frac{A^2t^2}{2!} + \dots) = Ae^{At} = e^{At}A$$

Properties of e^{At}

- $ightharpoonup e^{A0} = I$
- $(e^{At})^{-1} = e^{-At}$
- $e^{At}e^{As} = e^{A(t+s)}$ semi-group property

Note that e^{At} is an infinite series!!!!

By using the C-H theorem, there exists coefficients $\{\beta_0(t),...,\beta_{n-1}(t)\}$ such that

$$e^{At} = I + At + \frac{A^2 t^2}{2!} + \frac{A^3 t^3}{3!} + \cdots$$

= $\beta_0(t)I + \beta_1(t)A + \cdots + \beta_{n-1}(t)A^{n-1}$

It is a linear combination of $\{I, A, ..., A^{n-1}\}$

For $a, b \in \mathbb{R}$, $e^{at}e^{at} = e^{(a+b)t}$. How about $e^{At}e^{Bt} \stackrel{?}{=} e^{(A+B)t}$????? $\Rightarrow e^{At}e^{Bt} = e^{(A+B)t}$ if AB = BA (A and B commute) \Rightarrow HW

$$e^{At} = I + At + \frac{A^2t^2}{2!} + \frac{A^3t^3}{3!}$$

$$e^{Bt} = I + Bt + \frac{B^2t^2}{2!} + \frac{B^3t^3}{3!}$$

$$e^{(A+B)t} = I + (A+B)t + \frac{(A+B)^2t^2}{2!} + \frac{(A+B)^3t^3}{3!}$$

LTI system

$$\dot{x}(t) = Ax(t) + Bu(t)$$

Note that $e^{-At}\dot{x}(t) - e^{-At}Ax(t) = e^{-At}Bu(t)$; hence,

$$\begin{split} &\frac{d}{dt}\Big(e^{-At}x(t)\Big)=e^{-At}Bu(t)\\ &\Rightarrow e^{-A\tau}x(t)\Big|_{\tau=0}^t=\int_0^t e^{-At}Bu(\tau)d\tau\\ &\Rightarrow e^{-At}x(t)-x(0)=\int_0^t e^{-A\tau}Bu(\tau)d\tau\\ &\Rightarrow x(t)=e^{At}x(0)+\int_0^t e^{A(t-\tau)}Bu(\tau)d\tau \text{ Solution of the LTI system} \end{split}$$

Then

$$y(t) = Cx(t) = Ce^{At}x(0) + \int_0^t Ce^{A(t-\tau)}Bu(\tau)$$

Note that with
$$X(s) = \int_0^\infty x(t)e^{-st}dt$$
, where $s = \sigma + j\omega$

$$X(s) = (sI - A)^{-1}BU(s), \ Y(s) = C(sI - A)^{-1}BU(s)$$

$$\Leftrightarrow G(s) = \frac{Y(s)}{U(s)} = C(sI - A)^{-1}B$$

We need to compute e^{At} . There are five ways (the four ways are discussed in the textbook, page 106)

- ▶ Use Caley-Hamilton Theorem. That is, first compute eigenvalues of A. Next, find a polynomial $h(\lambda)$ of degree n-1 that equals $e^{\lambda t}$. Then $e^{At} = h(A)$.
- ▶ Use the Jordan form of A and Caley-Hamilton Theorem. Let $A = P\hat{A}P^{-1}$; then $e^{At} = Pe^{\hat{A}t}P^{-1}$, where \hat{A} is the Jordan matrix of A.
- Use the power infinite-series.
- ▶ Use the inverse Laplace transformation. That is $\mathcal{L}(e^{At}) = (sI A)^{-1}$, which implies $e^{At} = \mathcal{L}^{-1}(sI A)^{-1}$.
- Use "expm" in MATLAB

Example I: We want to compute $f(J) = e^{Jt}$ via the Caley-Hamilton Theorem

$$J = \begin{pmatrix} \lambda_1 & 1 & 0 & 0 \\ 0 & \lambda_1 & 1 & 0 \\ 0 & 0 & \lambda_1 & 1 \\ 0 & 0 & 0 & \lambda_1 \end{pmatrix}$$
$$h(\lambda) = \beta_0 + \beta_1(\lambda - \lambda_1) + \beta_2(\lambda - \lambda_2)^2 + \beta_3(\lambda - \lambda_3)^3$$
$$\beta_0 = f(\lambda_1), \ \beta_1 = f'(\lambda_1), \ \beta_2 = \frac{f''(\lambda_1)}{2!}, \ \beta_3 = \frac{f'''(\lambda_1)}{3!}.$$

Hence,

$$f(J) = f(\lambda_1)I + \frac{f'(\lambda_1)}{1!}(J - \lambda_1 I) + \frac{f''(\lambda_1)}{2!}(J - \lambda_1 I)^2 + \frac{f'''(\lambda_1)}{3!}(J - \lambda_1 I)^3$$

$$e^{Jt} = \begin{pmatrix} e^{\lambda_1 t} & te^{\lambda_1 t} & t^2 e^{\lambda_1 t} / 2! & t^3 e^{\lambda_1 t} / 3! \\ 0 & e^{\lambda_1 t} & te^{\lambda_1 t} & t^2 e^{\lambda_1 t} / 2! \\ 0 & 0 & e^{\lambda_1 t} & te^{\lambda_1 t} \\ 0 & 0 & 0 & e^{\lambda_1 t} \end{pmatrix}$$

Example II: We want to compute $f(J) = e^{Jt}$ via the Caley-Hamilton Theorem

$$J = egin{pmatrix} \lambda_1 & 1 & 0 & 0 & 0 \ 0 & \lambda_1 & 1 & 0 & 0 \ 0 & 0 & \lambda_1 & 0 & 0 \ 0 & 0 & 0 & \lambda_2 & 1 \ 0 & 0 & 0 & 0 & \lambda_2 \end{pmatrix}$$

$$e^{At} = egin{pmatrix} e^{\lambda_1 t} & te^{\lambda_1 t} & t^2 e^{\lambda_1 t}/2! & 0 & 0 \ 0 & e^{\lambda_1 t} & te^{\lambda_1 t} & 0 & 0 \ 0 & 0 & e^{\lambda_1 t} & 0 & 0 \ 0 & 0 & 0 & e^{\lambda_2 t} & te^{\lambda_2 t} \ 0 & 0 & 0 & 0 & e^{\lambda_2 t} \end{pmatrix}$$

Discrete-Time LTI System

How to discretize? \Rightarrow Sampling t = kT, where k = 0, 1, ... and T is sampling rate

Detailed derivation: Textbook

Discrete-time LTI system

$$x(k+1) = Ax(k) + Bu(k)$$

$$x(1) = Ax(0) + Bu(0)$$

$$x(2) = Ax(1) + Bu(1) = A^{2}x(0) + ABu(0) + Bu(1)$$

Hence,

$$x(k) = A^{k}x(0) + \sum_{m=0}^{n-1} A^{k-1-m}Bu(m)$$

Equivalence System: Similarity Transformation

Let P be an $n \times n$ real nonsingular matrix and let z = Px. Then the state-space equation

$$\dot{z}(t) = \bar{A}z(t) + \bar{B}u(t), \ y(t) = \bar{C}z(t)$$

where

$$\bar{A} = PAP^{-1}, \ \bar{B} = PB, \ \bar{C} = CP^{-1}$$

is said to be equivalent to the system

$$\dot{x}(t) = Ax(t) + Bu(t), \ y(t) = Cz(t)$$

and z = Px is called an equivalence transformation.

We have

$$\det(\lambda I - A) = \det(\lambda I - P\bar{A}P^{-1})$$
$$= \det(P)\det(\lambda I - \bar{A})\det(P)^{-1} = \det(\lambda I - \bar{A})$$



Definition

A rational (transfer) function $\hat{g}(s)$ is said to be proper if $\hat{g}(\infty) < \infty$. If $\hat{g}(\infty) = 0$, then $\hat{g}(s)$ is strictly proper. A rational matrix $\hat{G}(s)$ is said to be proper if $\hat{G}(\infty) < \infty$. If $\hat{G}(\infty) = 0$, then $\hat{G}(s)$ is strictly proper.

- We consider the LTI system
- G(t) is an impulse response of $\hat{G}(s)$

Definition:

A matrix transfer function $\hat{G}(s)$ is realizable if there exists a finite-dimensional state-space equation, i.e., $\{A, B, C, D\}$ such that

$$\hat{G}(s) = \frac{Y(s)}{U(s)} = C(sI - A)^{-1}B + D$$

- $\{A, B, C, D\}$ is a realization of G(s)
- ▶ Realization is not unique, since for the SISO case, $\hat{g}(s) = \hat{g}^T(s)$. For the general matrix case, the realization is not also unique. See page 130 and Problem 4.13 of the textbook.

Theorem (Generalized version of Theorem 4.2 of the textbook)

- A matrix transfer function $\hat{G}(s)$ is realizable by a finite-dimensional LTI system if and only if $\hat{G}(s)$ is a proper rational matrix function
- (same as Theorem 4.2 of the textbook) For a SISO system, $\hat{g}(s)$ is realizable by a finite-dimensional SISO LTI system if and only if $\hat{g}(s)$ is a proper rational transfer function

Note that

- ▶ If $\hat{G}(s)$ is strictly proper, then D = 0
- ▶ If $\hat{g}(s)$ is strictly proper, then D = 0

Example (Example 4.5.2): SISO LTI system

$$\hat{g}(s) = \frac{4s+3}{40s^3 + 30s^2 + 9s + 3}$$

Example (Example 4.5.1): SISO LTI system

$$\hat{g}(s) = \frac{3s^4 + 5s^3 + 24s^2 + 23s - 5}{2s^4 + 6s^3 + 15s^2 + 12s + 5}$$

Example (Example 4.5.3): MIMO LTI system

$$\begin{split} \hat{G}(s) &= \begin{pmatrix} \frac{4s-10}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{s+1}{(s+2)^2} \end{pmatrix} = \begin{pmatrix} 2 & 0 \\ 0 & 0 \end{pmatrix} + \begin{pmatrix} \frac{-12}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{s+1}{(s+2)^2} \end{pmatrix} \\ &= \frac{1}{s^3 + 4.5s^2 + 6s + 2} \begin{pmatrix} -6(s+2)^2 & 3(s+2)(s+0.5) \\ 0.5(s+2) & (s+1)(s+0.5) \end{pmatrix} + \begin{pmatrix} 2 & 0 \\ 0 & 0 \end{pmatrix} \end{split}$$

Example (Example 4.5.3): Two-input Two-Output LTI system

$$\begin{split} \hat{G}(s) &= \begin{pmatrix} \frac{4s-10}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{3}{s+2} \end{pmatrix} = \begin{pmatrix} 2 & 0 \\ 0 & 0 \end{pmatrix} + \begin{pmatrix} \frac{-12}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{s+1}{(s+2)^2} \end{pmatrix} \\ &= \frac{1}{s^3 + 4.5s^2 + 6s + 2} \begin{pmatrix} -6(s+2)^2 & 3(s+2)(s+0.5) \\ 0.5(s+2) & (s+1)(s+0.5) \end{pmatrix} + \begin{pmatrix} 2 & 0 \\ 0 & 0 \end{pmatrix} \\ &= \frac{1}{s^3 + 4.5s^2 + 6s + 2} \begin{pmatrix} \begin{pmatrix} -6 & 3 \\ 0 & 1 \end{pmatrix} s^2 + \begin{pmatrix} -24 & 7.5 \\ 0.5 & 1.5 \end{pmatrix} s + \begin{pmatrix} -24 & 3 \\ 1 & 0.5 \end{pmatrix} \end{pmatrix} \\ &+ \begin{pmatrix} 2 & 0 \\ 0 & 0 \end{pmatrix} \end{split}$$

Strictly proper transfer function

$$\hat{g}(s) = \frac{b_1 s^3 + b_2 s^2 + b_3 s + b_4}{s^4 + a_2 s^3 + a_3 s^2 + a_4 s + a_5}$$

Controllable canonical form

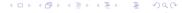
$$\dot{x}(t) = \begin{pmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -a_5 & -a_4 & -a_3 & -a_2 \end{pmatrix} x(t) + \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} u(t)$$

$$y(t) = \begin{pmatrix} b_4 & b_3 & b_2 & b_1 \end{pmatrix} x(t)$$

Observable canonical form

$$\dot{x}(t) = A^{T}x(t) + C^{T}u(t), \ y(t) = B^{T}x(t)$$

Modal form: Similarity transformation



One-dimensional LTV system

$$\begin{aligned} &x(t)=a(t)x(t),\ a(t):\ \text{continuous}\\ &x(t)=e^{\int_0^t a(\tau)d\tau}x(0)\\ &\text{check:}\ \frac{d}{dt}e^{\int_0^t a(\tau)d\tau}=a(t)e^{\int_0^t a(\tau)d\tau}=e^{\int_0^t a(\tau)d\tau}a(t) \end{aligned}$$

LTI approach holds!!!

Let's see general LTV systems

$$\dot{x}(t) = A(t)x(t)$$

A(t) needs to be a continuous function (why????)

Can we do?

$$x(t) = e^{\int_0^t A(\tau)d\tau}x(0)$$

No!! Consider

$$e^{\int_0^t A(\tau)d\tau} = I + \int_0^t A(\tau)d\tau + \frac{1}{2} \int_0^t A(\tau)d\tau \int_0^t A(\tau)d\tau + \cdots$$

$$\frac{d}{dt} e^{\int_0^t A(\tau)d\tau} = A(t) + \frac{1}{2} A(t) \int_0^t A(s)ds + \frac{1}{2} \int_0^t A(s)ds A(t) + \cdots$$

$$\neq A(t) e^{\int_0^t A(s)ds}$$

The LTI approach does not hold for LTV systems. Namely, we cannot extend the solution of the scalar time-varying equations to the matrix case, and must use a different approach.

Note that, since A(t) is continuous, we must have a unique solution of x(t) with a given initial condition (why??? think about the Lipschitz condition discussed in class)

We can have n linearly independent initial conditions, $x_i(0)$, i=1,2,...,n, and for each $x_i(0)$, there exists a unique solution, denoted by $x_i(t)$, i=1,2,...,n. Then due to the linearity, $x_i(t)$, i=1,2,...,n form an n-dimensional vector space over \mathbb{R} . That is, $x_i(t)$, i=1,2,...,n, are linearly independent. Let

$$X(t) = (x_1(t), x_2(t), \dots, x_n(t)), \ X(0) = (x_1(0), \dots, x_n(0))$$

Then we can show that

$$\dot{X}(t) = A(t)X(t)$$

X(t) is called a fundamental matrix. Note that X(t) is not unique, since we can choose n linearly independent initial conditions arbitrarily. We also note that X(t) is always invertible for all t, since $x_i(t)$, i=1,2,...,n are linearly independent.

Definition (Definition 4.2 of the textbook):

Let X(t) be a fundamental matrix of $\dot{x} = A(t)x$. Then

$$\Phi(t,0) := X(t)X^{-1}(0)$$

is called the state transition matrix of $\dot{x} = A(t)x$. The state transition matrix is also the unique solution of

$$\frac{d}{dt}\Phi(t,0)=A(t)\Phi(t,0),$$

where $\Phi(0,0) = I$.

Properties of the state transition matrix

- $\blacktriangleright \Phi(t,t) = I$
- $\Phi^{-1}(t,0) = (X(t)X^{-1}(0))^{-1} = X(0)X^{-1}(t) = \Phi(0,t)$
- $\Phi(t,0) = \Phi(t,s)\Phi(s,0)$ (semi-group property)

The properties of the state transition matrix can be extended to when the initial time is t_0

- $ightharpoonup \Phi(t,t) = I$
- $\Phi(t, t_0) = \Phi(t, s)\Phi(s, t_0)$ (semi-group property)

For LTI system, the state transition matrix $\Phi(t,t_0)$ can be viewed as $\Phi(t,t_0)=e^{A(t-t_0)}=e^{At}e^{-At_0}=X(t)X^{-1}(t_0)$

$$\frac{d}{dt}e^{A(t-t_0)} = Ae^{A(t-t_0)}, \ e^{A(t_0-t_0)} = I$$

- $\Phi(t,t) = e^{A(t-t)} = I$
- $\qquad \qquad \Phi^{-1}(t,t_0) = (e^{A(t-t_0)})^{-1} = e^{-A(t-t_0)} = e^{A(t_0-t)} = \Phi(t_0,t)$
- $\Phi(t,t_0) = \Phi(t,s)\Phi(s,t_0) = e^{A(t-t_0)} = e^{A(t-s)}e^{A(s-t_0)}$

Example (Examples 4.6.1 and 4.6.2 of the textbook)

$$\dot{x} = \begin{pmatrix} 0 & 0 \\ t & 0 \end{pmatrix} x(t), \ \dot{x}_1(t) = 0, \ \dot{x}_2(t) = tx_1(t)
x_1(t) = x_1(0), \ \dot{x}_2(t) = tx_1(0),
x_2(t) = \int_0^t \tau x_1(0) d\tau + x_2(0) = 0.5t^2 x_1(0) + x_2(0)$$

Hence, we have

$$\begin{aligned} x(0) &= \begin{pmatrix} 1\\0 \end{pmatrix} \Rightarrow x(t) = \begin{pmatrix} 1\\0.5t^2 \end{pmatrix} \\ x(0) &= \begin{pmatrix} 1\\2 \end{pmatrix} \Rightarrow x(t) = \begin{pmatrix} 1\\0.5t^2 + 2 \end{pmatrix} \\ X(t) &= \begin{pmatrix} 1\\0.5t^2 & 0.5t^2 + 2 \end{pmatrix}, \ X(t) \text{ is invertible for all } t!!!! \end{aligned}$$

Example (Examples 4.6.1 and 4.6.2 of the textbook)

$$X(t) = \begin{pmatrix} 1 & 1 \\ 0.5t^2 & 0.5t^2 + 2 \end{pmatrix}, \ X(t) \text{ is invertible!!!!}$$
 $X(0) = \begin{pmatrix} 1 & 1 \\ 0 & 2 \end{pmatrix}, \ X^{-1}(0) = \begin{pmatrix} 1 & -0.5 \\ 0 & 0.5 \end{pmatrix}$
 $\Phi(t,0) = X(t)X^{-1}(0)$

The solution to the LTI system

$$\dot{x}(t) = A(t)x(t), \ x(t) = \Phi(t,0)x(0)$$

This can be checked easily via the definition of the state transition matrix Also,

$$\begin{aligned} \dot{x}(t) &= A(t)x(t) + B(t)u(t) \\ x(t) &= \Phi(t,0)x_0 + \int_0^t \Phi(t,\tau)B(\tau)u(\tau)d\tau \\ &= \Phi(t,0)\Big(x_0 + \int_0^t \Phi(0,\tau)B(\tau)u(\tau)d\tau\Big) \end{aligned}$$

Note that $\Phi(t,0)$ is the state transition matrix of $\dot{x}(t) = A(t)x(t)$. We now check the above solution

Note that

$$x(0) = \Phi(0,0)x_{0} + \int_{0}^{0} \Phi(0,\tau)B(\tau)u(\tau)d\tau = x_{0}$$

$$\frac{d}{dt}x(t) = \frac{d}{dt} \Big(\Phi(t,0)x_{0} + \int_{0}^{t} \Phi(t,\tau)B(\tau)u(\tau)d\tau \Big)$$

$$= A(t)\Phi(t,0)x_{0} + \int_{0}^{t} A(t)\Phi(t,\tau)B(\tau)u(\tau)d\tau + B(t)u(t)$$

$$= A(t) \Big(\Phi(t,0)x_{0} + \int_{0}^{t} \Phi(t,\tau)B(\tau)u(\tau)d\tau \Big) + B(t)u(t)$$

$$= A(t)x(t) + B(t)u(t)$$

Hence, we have the solution.

The LTV system for the discrete-time case is fairly easy, since we can just enumerate the state transition matrix $\Phi(k,0) = A(k)A(k-1)\cdots A(0)$. The detailed discussion is provided in the textbook (page 140).

Conclusions

- Solutions of LTI, LTV systems
- State transition matrix and matrix exponential
- ► Caley-Hamilton theorem
- ▶ Next class: Lyapunov stability of nonlinear systems