

## T4 - Audio Amplifier

Integrated Master in Physics Engineering

João Lehodey (96538), Jorge Silva (96545), Pedro Monteiro (93156)

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### 1 Introduction

In this laboratory assignment we seek to build an audio amplifier circuit , with the goal of maximizing the merit figure,  $M$ , given by:

$$M = \frac{voltageGain * bandwidth}{cost * lowerCutoffFreq}$$

being that the voltage gain is given by the quotient between the output voltage and the input voltage; the bandwidth is the magnitude of the interval of frequencies for which the circuit functions; and the lower cut off frequency is given by the lowest frequency for which the circuit functions. The cost equals the sum of the cost of each individual component.

The circuit consists only of resistors, voltage sources (AC as well as DC), capacitors and bipolar junction transistors (more specifically one PNP transistor and one NPN transistor)

The circuit used was the following:

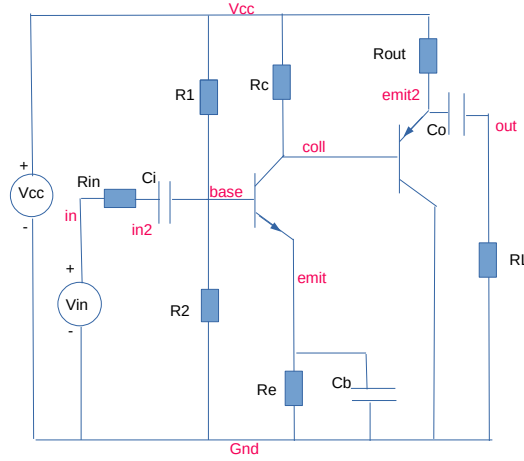


Figure 1: Circuito utilizado

In Section 2, a theoretical analysis is presented. In Section 3, the circuit is analysed by simulation, and the results are compared to the theoretical results obtained in Section 2. The conclusions of this study are outlined in Section 4.

## 2 Theoretical Analysis

### 2.1 First part

The theoretical analysis is split into three parts. In the first part we compute the operating point using the Ebers-Moll model for the transistor. Taking into account the fact that capacitors act as an open circuit on the operating point, and using Thevenin's theorem to simplify the circuit, we get the following equations:

$$R_{th} = \frac{1}{\frac{1}{R_1} + \frac{1}{R_2}}$$

$$V_{th} = \frac{R_2 V_{CC}}{R_1 + R_2}$$

$$I_{B1} = \frac{V_{th} - V_{BEON}}{R_{th} + (1 + \beta_{FN})R_e}$$

$$I_{C1} = \beta_{FN} I_{B1}$$

$$I_{E1} = (1 + \beta_{FN}) I_{B1}$$

$$V_{E1} = R_e * I_{E1}$$

$$V_{O1} = V_{CC} - R_1 I_{C1}$$

$$V_{CE} = V_{O1} - V_{E1}$$

being that  $I_{B1}$  is the current going **into** the base of the first transistor;  $I_{E1}$  is the current going **out** of the emitter of the first transistor;  $V_{O1}$  is the voltage at the 'coll' node;  $V_{CE}$  is the potential difference between the collector and emitter of the first transistor.

Using the following values:

RE1	100.000000
RC1	1000.000000
RB1	80000.000000
RB2	20000.000000
VCC	12.000000
RS	100.000000

Table 1: Values used as parameters for the circuit studied.

We get:

IB1	0.000050
IC1	0.008943
IE1	0.008993
VE1	0.899293
VO1	3.057109
VCE	2.157816
gm1	0.357716
rpi1	499.558824
ro1	7793.900392
ZI1	484.433630
ZO1	886.284816
AV1	262.790895
VI2	3.057109
IE2	0.082429
IC2	0.082068
VO2	3.757109
gm2	3.282714
go2	0.002206
gpi2	0.014442
ge2	0.010000
AV2	0.991948
ZI2	8598.855359
ZO2	0.302173
a	0.828894
b	0.906561
c	0.963603

Table 2: Operating point theoretically

## 2.2 Second part

In this part we compute the gain, input and output impedances separately for the 2 stages. The model used for the transistors here is the same.

$$g_{m1} = I_{C1}/V_T$$

$$r_{pi1} = \beta F N / g_{m1}$$

$$r_{o1} = V_{AFN} / I_{C1}$$

Using this, we can calculate the voltage gain

$$A_{V1} = \frac{g_{m1} Z_{O1} Z_{I1}}{Z_{I1} + R_S}$$

$$R_{E1} = 100$$

For calculating the entrance impedance of transistor 1, we've come to the following equations

$$Z_{I1} = r_{pi} || R_2 || R_1$$

$$Z_{I1} = \frac{R_{B1}R_{B2}r_{pi1}}{R_{B1}R_{B2} + R_{B1}r_{pi1} + R_{B2}r_{pi1}}$$

For calculating the exit impedance

$$Z_{O1} = r_{o1} || R_{C1}$$

Repeating the same calculations, but for transistor 2, we get that

$$g_{m2} = I_{C2}/V_T$$

$$g_{o2} = I_{C2}/V_{AFP}$$

$$g_{pi2} = g_{m2}/\beta_{FP}$$

$$g_{e2} = 1/R_{E2}$$

For calculating the entrance impedance of transistor 2, we've come to the following equation

$$A_{V2} = \frac{g_{m2}}{g_{m2} + g_{pi2} + g_{o2} + g_{e2}}$$

for calculating the exit impedance of transistor 2, we get

$$Z_{O2} = 1/(g_{m2} + g_{pi2} + g_{o2} + g_{e2})$$

The following ratio should be equal to one. if that's the case, then we've confirmed there's a small loss of signal between the transistors

$$b = Z_{I2}/(Z_{I2} + Z_{O1})$$

Similarly, the following ratio should be one if there's a small loss of signal between Vin and the first transistor

$$a = Z_{I1}/(Z_{I1} + R_S)$$

Finally, to ensure that there's no signal loss between the speaker and the second transistor, the following ratio should be approximately one

$$c = R_L/(R_L + Z_{O2})$$

### 3 Simulation Analysis

Name	Value
base	1.365942e+00
coll	4.821057e+00
emit	6.570592e-01
emit2	5.649825e+00
in	0.000000e+00
in2	0.000000e+00
out	0.000000e+00
vcc	1.200000e+01

Table 3: Operating point.

The impedances measured are the following:

Name	Value
Error(parse.c—checkvalid): gain: no such vector.	
zi	-5.63461e-01
z0	-4.79476e+01

Table 4: Impedances

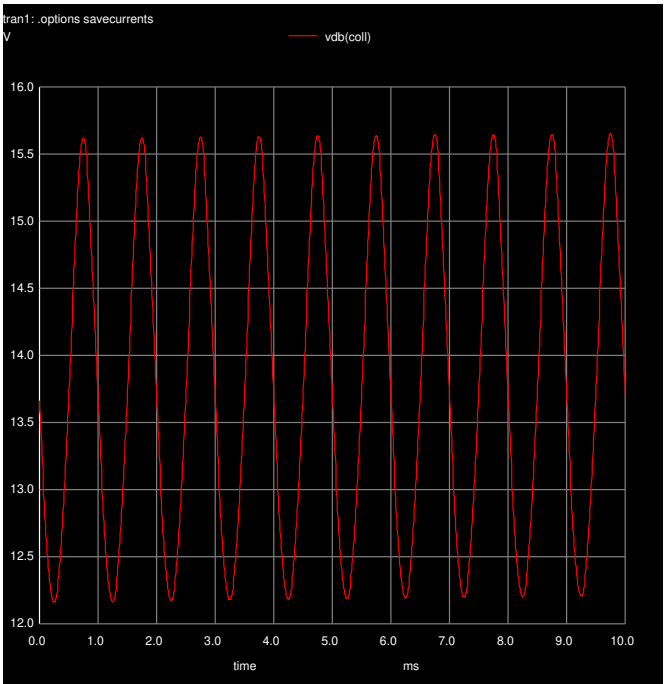


Figure 2: ———

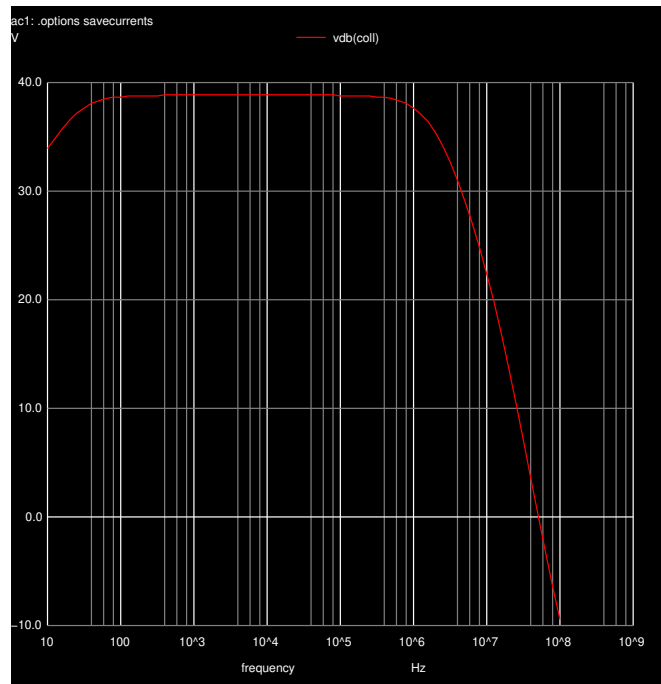


Figure 3: —

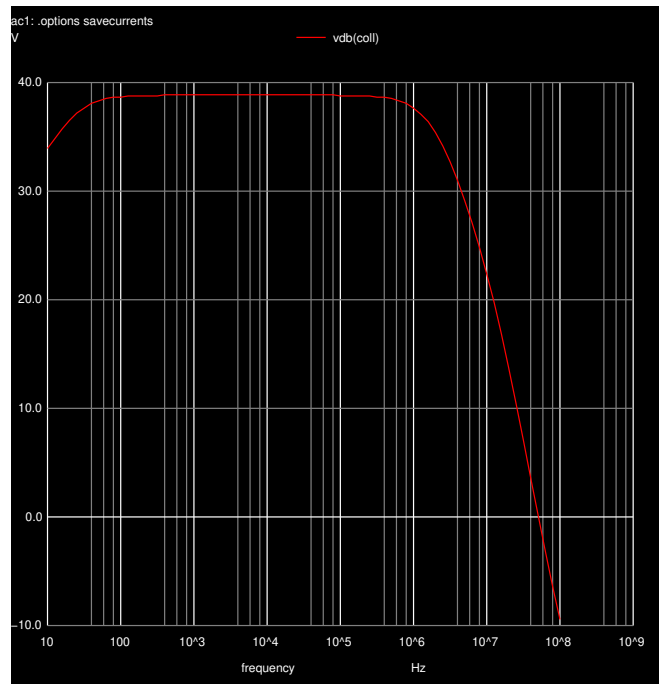


Figure 4: —————

## 4 Conclusion

As we can see in the operating point analysis there are some differences between the theoretical values and simulation values. This is due to the approximations used in the theoretical analysis, ngspice uses a much more sophisticated model.