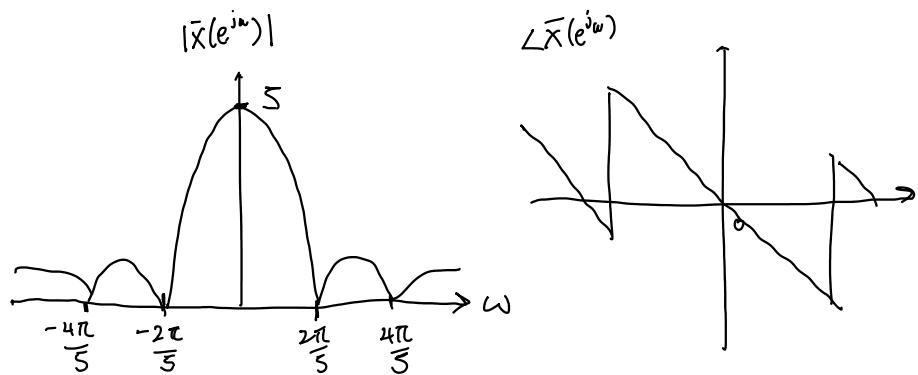


Problem 1.

A) $x[n] = u[n] - u[n-5]$

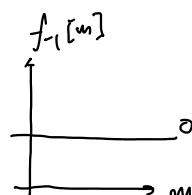
a) $\bar{X}(e^{j\omega}) = e^{-j\omega \cdot 4/2} \cdot \frac{\sin(5\omega/2)}{\sin(\omega/2)}$



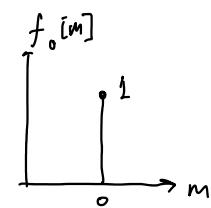
b) $f_n[m] = w[m] \chi[n+m]$

$$= \bar{j}[m] \chi[n+m]$$

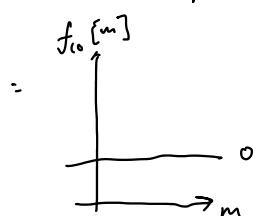
$$f_{-1}[m] = w[m] \chi[-1+m] \\ = \bar{j}[m] \chi[-1+m]$$



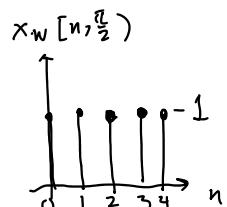
$f_0[m] : \bar{j}[0] \chi[m]$



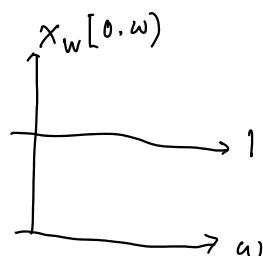
$f_{10}[m] = \bar{j}[m] \chi[m+10]$



c) $\bar{X}_w[n, \frac{\pi}{2}] = x[n] =$



d) $\bar{X}_w[0, \omega] = x[0] = 1$

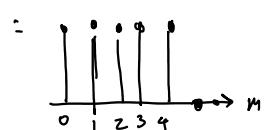


II.B. $f_n[m] = x[n+m]$

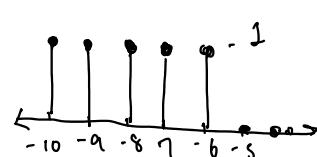
a) $f_{-1}[m] = x[m-1] :$



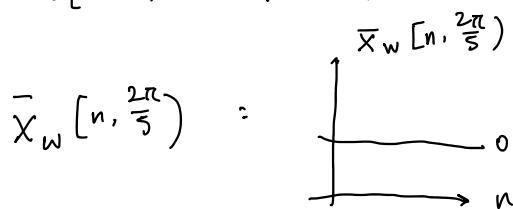
$f_0[m] = x[m]$



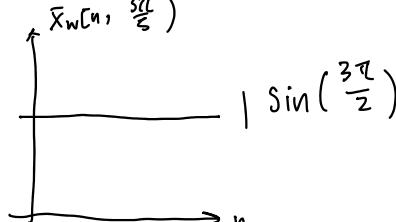
$f_{10}[m] = x[10+m]$



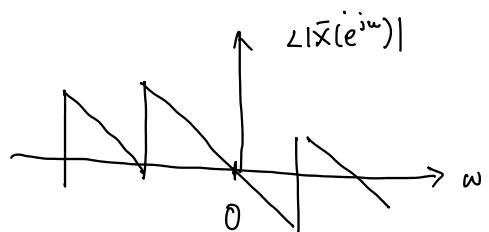
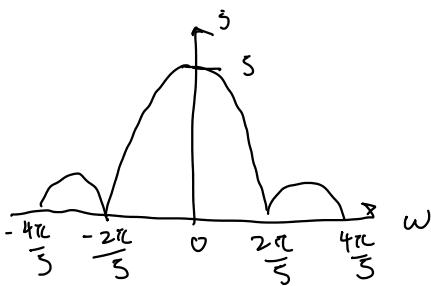
b) $X_w[n, \omega] = e^{j\omega n} \bar{X}(e^{j\omega})$



c) $\bar{X}_w[n, \frac{3\pi}{5}]$



$$d) X_w[0, \omega] = \bar{X}(e^{j\omega}) \quad ; \quad |\bar{X}(e^{j\omega})|$$



11.2 a) TDFT is an output of a filter system. For a fixed ω ,
The TDFT $\bar{X}_w[n, \omega]$ is obtained by : $x[n] * h_w[n]$
 $\xrightarrow{\text{Convolution}} w[n] e^{j\omega n}$

Because the operation is convolution with a fixed filter $h_w[n]$
 $x[n]$ can be mapped as the input to the output $\bar{X}_w[n, \omega]$
which is an LTI system.

For any time-invariant system, shift in the input would cause a shift
to the output

b) TDFT is the DFT of $f_n[m] = x[m] w[m]$

$$\bar{X}_w[n, \omega] = \sum_{m=-\infty}^{\infty} f_n[m] e^{-j\omega m}$$

The signal $x[n]$ is real, the window $w[n]$ is also real, therefore their
product $f_n[m]$ is also a real-valued sequence.

Real-valued DTFT exhibits conjugate symmetry: $\vec{H}(\omega) = H^*(-\omega)$

$$\text{which would apply to } \bar{X}_w[n, \omega] = \bar{X}^*_{-n}[-\omega]$$

$$c) \text{ Let } w[n] = \delta[n], \quad \bar{X}_w[n, \omega] = X[n], \quad Y[n, \omega] = y[n] \\ = \delta[n] \quad \quad \quad = \delta[n-1]$$

$$\text{TDFT } \{x * y\} = x[n] \cdot y[n]$$

$$= \delta[n] \cdot \delta[n-1] = 0 \text{ which is not true}$$

$$11.3) \quad H(z) = G \left/ \left| + \sum_{k=1}^{10} a_k z^{-k} \right. \right. \quad H\left(e^{\frac{j2\pi k}{256}}\right) = G \left/ \left| + \sum_{l=1}^{10} a_l e^{(-j\frac{2\pi k}{256})l} \right. \right.$$

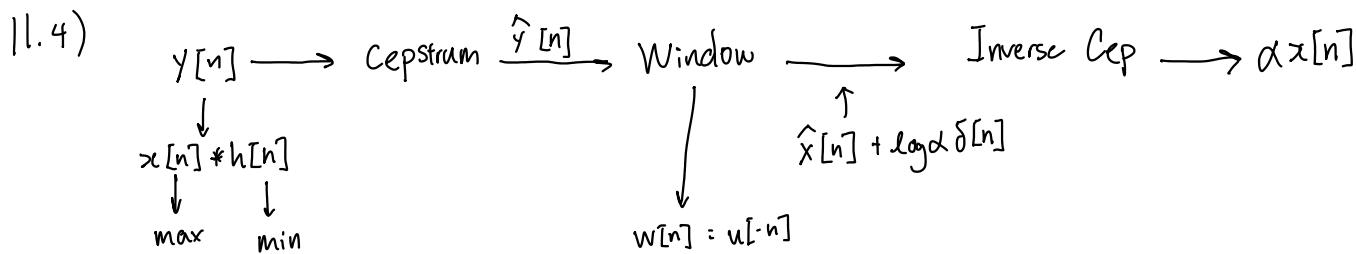
$\underbrace{\hspace{10em}}$

$D[k] = \sum_{n=0}^{255} d[n] e^{-j\frac{2\pi k}{256} n}$, $d[n]$ is 0 for $n > 10$

$\Rightarrow 1 + \sum_{n=1}^{10} a_n e^{-j\frac{2\pi k}{256} n} \rightarrow H\left(e^{\frac{j2\pi k}{256}}\right) = \frac{G}{D[k]}$

let this be $d[n]$

$\begin{cases} 1, l=0 \\ a_l, 1 \leq l \leq 10 \\ 0, \text{ otherwise} \end{cases}$



Convolution in time = Addition in cepstral $\rightarrow \hat{y}[n] = \hat{x}[n] + \hat{h}[n]$

$\downarrow 0 \text{ for } n < 0 \quad \uparrow 0 \text{ for } n > 0$

Let $\hat{k}[n] = \hat{y}[n] \cdot w[n] = (\hat{x}[n] + \hat{h}[n]) \cdot u[-n]$

$\rightarrow \hat{k}[n] : \begin{cases} \text{for } n < 0 : u[-n] = 1, \hat{h}[n] = 0, \hat{k}[n] = \hat{x}[n] \\ \text{for } n > 0 : u[-n] = 0, \hat{k}[n] = 0 \\ n=0 : u[-n] = 1 \quad \hat{k}[n] = \hat{x}[0] + \hat{h}[0] \end{cases} \quad \left. \begin{array}{l} \hat{k}[n] = \hat{x}[n] + \hat{h}[0] \delta[n] \\ = \hat{x}[n] + \log \alpha \delta[n] \end{array} \right\}$

$\log \alpha = \hat{h}[0]$ Inverse Cep of $\hat{k}[n]$

$= \log(\bar{K}(z)) = \log(\bar{X}(z)) + C \rightarrow C = \hat{h}[0] = \log \alpha \delta[n]$

$\rightarrow \bar{K}(z) = \bar{X}(z) \cdot e^C$

$\rightarrow \hat{k}[n] = e^C \cdot x[n]$

Output $= \underline{x[n] \cdot e^{\hat{h}[0]}}$,

$\rightarrow \alpha = e^{\hat{h}[0]}$

11.5)

$$a) \quad x_c[n] = -2\delta[n+2] + 4\delta[n+1]$$

$$x_c[-2] = -2$$

$$x_c[-1] = 4$$

$$x_c[n] = 0 \text{ for } n \neq -2, -1$$

$$\bar{x}_c(z) = -2z^2 + 4z$$

checking zeros: $\bar{x}_c(z) = -2z^2 + 4z = -2z(z-2)$ $z = 0, 2$

\downarrow outside unit circle
within unit circle
requires shifting

$$\text{let } y[n] = x_c[n-2], \quad y[0] = -2, \quad y[1] = 4$$

$$\bar{Y}(z) = z^{-2} \bar{x}_c(z) = z^{-2}(-2z^2 + 4z) = -2 + 4z^{-1}$$

Scaling: should scale by -0.5

$$b) \quad x_b[n] = (0.5)^n u[n]$$

$$\begin{cases} \hat{g}_b[n] + \hat{x}_b[n] = -\frac{(0.5)^n}{n} u[n-1] \\ \bar{x}_b(z) = \sum_{n=0}^{\infty} (0.5 z^{-1})^n = \frac{1}{1-0.5 z^{-1}} \end{cases}$$

$$\log(\bar{x}_b(z)) = \log \frac{1}{1-0.5 z^{-1}}$$

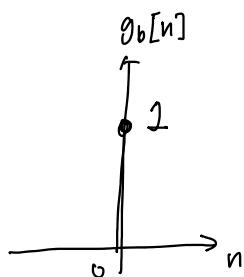
$$= -\log(1-0.5 z^{-1})$$

$$= \sum_{n=1}^{\infty} \frac{(0.5)^n}{n} z^{-n}$$

$$\hat{x}_b[n] = \frac{(0.5)^n}{n} u[n-1], \quad \hat{g}_b[n] + \hat{x}_b[n] = \hat{x}_b[n]; \quad \hat{g}_b[n] = 0 \rightarrow \log(\bar{G}_b(z)) = 0$$

$$\Rightarrow \bar{G}_b(z) = e^0 = 1$$

$$\therefore g_b[n] = \delta[n]$$



```
% Part A
[y, Fs] = audioread('sad.wav');
N = length(y);
t = (0:N-1) / Fs;

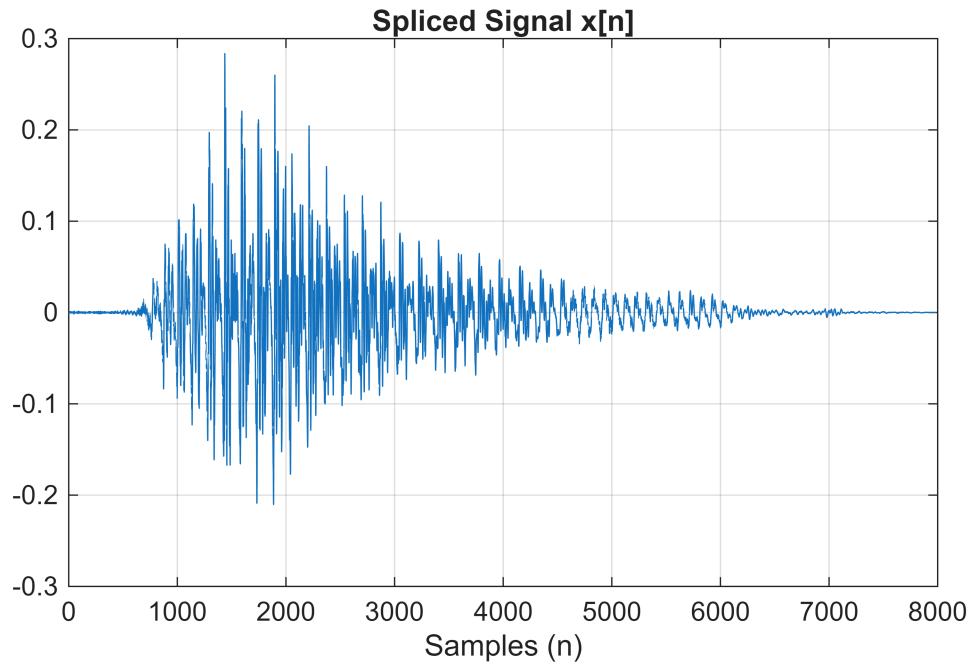
t_start = 0.6011787;
t_end = 1.1004;

idx_start = round(t_start * Fs);
idx_end = round(t_end * Fs);

if idx_start < 1
    idx_start = 1;
end

spliced = y(idx_start : idx_end);

% Plot the spliced signal
figure;
plot(spliced);
title('Spliced Signal x[n]');
xlabel('Samples (n)');
grid on;
```



```
% Part B
% a)
```

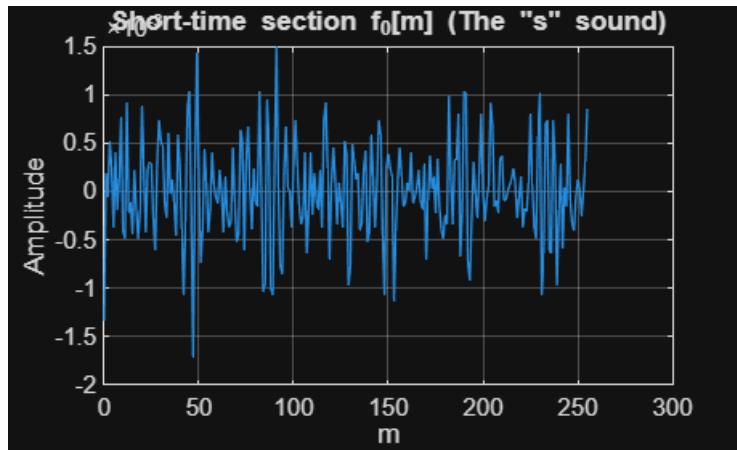
```
M = 256;
m = 0:M-1;
```

```

f0 = spliced(1:M);

% Plot
figure;
plot(m, f0);
title('Short-time section f_0[m] (The "s" sound)');
xlabel('m'); ylabel('Amplitude');
grid on;

```

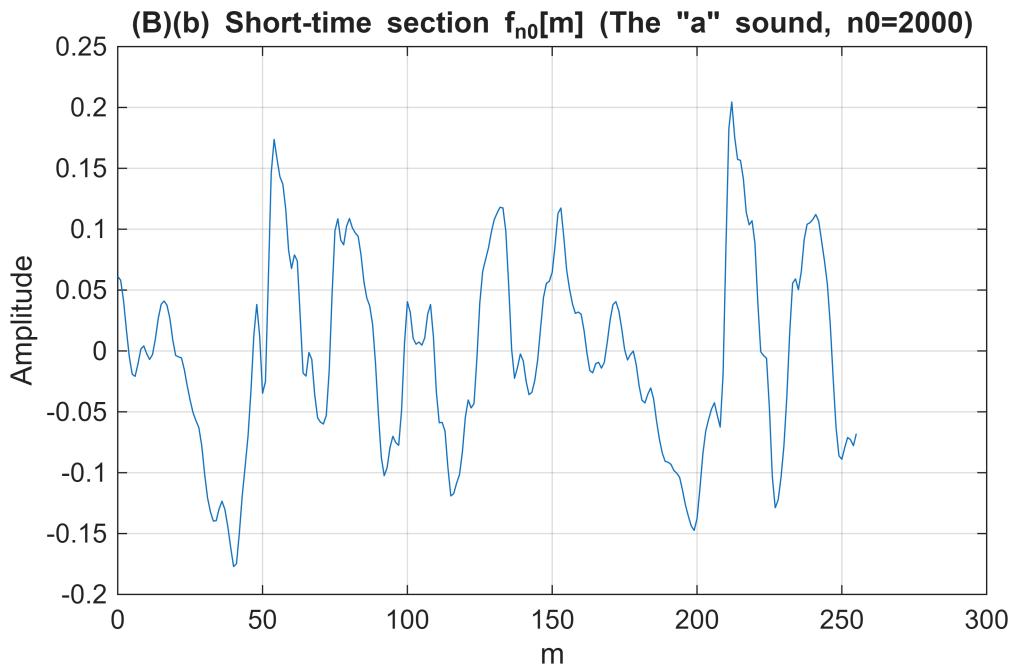


```

% Part B
% b)
n0 = 2000;
f_n0 = spliced(n0+1 : n0+M);

figure;
plot(m, f_n0);
title([(B)(b) Short-time section f_{n0}[m] (The "a" sound, n0=', num2str(n0),
')]);
xlabel('m'); ylabel('Amplitude');
grid on;

```



```
% Part B
% c)

N_fft = 512;
X_k = fft(f_n0, N_fft);

k = 0:N_fft-1;
freq_axis = k * (Fs / N_fft);

half_N = N_fft/2;

figure;
plot(freq_axis(1:half_N), abs(X_k(1:half_N)));
title('(B)(c) Magnitude of 512-point DFT of  $f_{n0}[m]$ ');
xlabel('Frequency (Hz)'); ylabel('Magnitude');
grid on;
```

(B)(c) Magnitude of 512-point DFT of $f_{n0}[m]$

