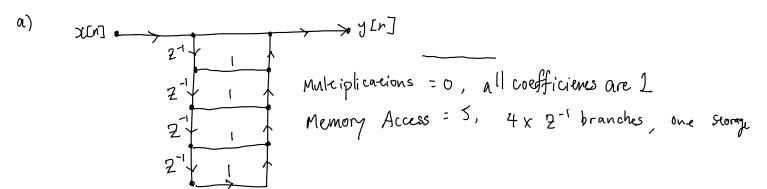
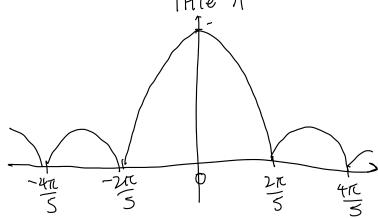
| Y[n] = x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4]



b)
$$\overline{Y(2)} = \overline{X(2)} + 2^{-1} \overline{X(2)} + 2^{-2} \overline{X(2)} + 2^{-3} \overline{X(2)} + 2^{-4} \overline{X(2)}$$

 $\overline{Y(2)} = \frac{Y(2)}{X(2)} = [1 2^{-1} 1 2^{-2} + 2^{-3} + 2^{-4}]$

c)
$$H(e^{i\omega}) = \sum_{n=0}^{4} e^{-j\omega n}$$
, $|H(e^{i\omega})| = \left|\frac{\sin(5\omega/2)}{\sin(\omega/2)}\right|$, $\frac{5\omega}{2} = k\pi$
 $|H(e^{i\omega})|$



d)
$$H(Z) = \frac{2^4 + Z^3 + 2^2 + 2 + 1}{2^4}$$
, pole is at $Z = 0$, within the unit circle (|Z|<1), filter is stable

- · 2 multiplications (-0.125, 0.75)
- . 2 memory re-trievals, 2 branches

b)
$$H(2) = \frac{\overline{Y}(2)}{\overline{X}(2)}$$

 $Y(2) = \overline{X}(2) + 0.75 \overline{Y}(2) 2^{-1} - 0.125 \overline{Y}(2) 2^{-2}$
 $Y(2) (1 - 0.75 x^{-1} + 0.125 x^{-2}) = \overline{X}(2)$
 $H(2) = 1 - 0.75 x^{-1} + 0.125 x^{-2}$

c) set denominator of
$$H(z)=0$$

$$1-0.95z^{-1} + 0.125z^{-2} = 0$$
 $(z-0.5)(x-0.25)$, both values are $|z| < 0.95z + 0.125 = 0$ $|z| < 1$, therefore stable

The same can be said for $H(e^{j\alpha})$ therefore there are a that would produce a O value.

3. a)
$$y[n]: x[n] - \frac{1}{4}x[n-2] + \frac{1}{2}y[n-1]$$

$$x[n] \qquad y[n]$$

$$y[n] + \frac{1}{2}y[n-1]$$

$$y[n] + \frac{1}{2}y[n-1]$$

$$\frac{1}{1+\frac{1}{2}y[n-1]} + \frac{1}{2}y[n-1] = \frac{1}{2}y[n] + \frac{$$

$$\neg H(Z) = H \pm \chi^{-1}$$

 $\overline{Y(\chi)} = \overline{X(\chi)} + \frac{1}{2} \overline{X(\chi)} \chi^{-1}$
 $Y[n] = \chi[n] + \frac{1}{2} \chi[n-1]$

Memory access and multiplications are both halved

4. a)
$$H(2) = \left[(1-0.9e^{\frac{i\pi}{4}} / 2^{-1})(1-0.9e^{\frac{i\pi}{4}} / 2^{-1}) \right]^{-1} + (2) = \frac{\overline{\Upsilon(2)}}{\overline{\chi(2)}}, \quad \overline{\Upsilon(2)} = \overline{H(2)} \overline{\chi(2)}$$

$$1 - (0.9e^{\frac{i\pi}{4}} / 0.9e^{-\frac{i\pi}{4}} / 2^{-1} + 0.81e^{-2})$$

$$2 \cdot Re \left\{ 0.9e^{\frac{i\pi}{4}} \right\} = 2 \cdot 0.9 \cos^{\frac{\pi}{4}} = 1.8 \cdot \frac{\sqrt{2}}{2}$$

$$H(2) : (1-0.9\sqrt{2} 2^{-1} + 0.81 2^{-2})^{-1}$$

$$\Rightarrow \quad \overline{\Upsilon(2)} \left[1-0.9\sqrt{2} 2^{-1} + 0.81 2^{-2} \right] = \overline{\chi(2)}$$

$$\overline{\Upsilon(2)} = 0.9\sqrt{2} \overline{\Upsilon(2)} = 0.81 \overline{\Upsilon(2)} = 2^{-2} + \overline{\chi(2)}$$

$$\overline{\Upsilon(2)} = 0.9\sqrt{2} \overline{\Upsilon(2)} = 0.81 \overline{\Upsilon(2)} = 2^{-2} + \overline{\chi(2)}$$

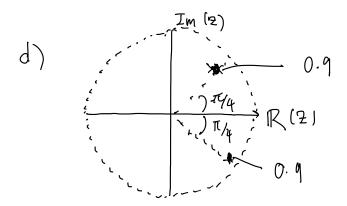
b) No, the coefficients of the filter will determine whether the resulting output would be real-valued or complex.

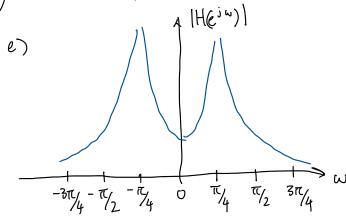
$$x[n] = \frac{0.9\sqrt{2}}{2^{-1}} \sqrt{2^{-1}}$$

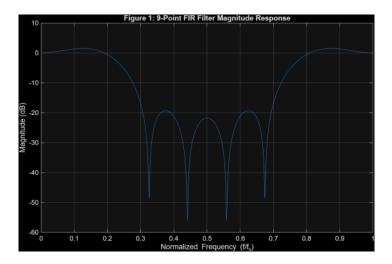
Multiplications: 2 complex multiplications = 4

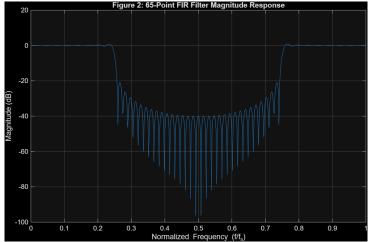
Memory: 2 × each access of y[n-1] and y[n-2] = 4

intermediare signals are not guranteed real









Part 3 - Normalization:

The normalization step as described in the exercise ensures that the filter has a gain of 0 dB at ω = 0; ensuring that a constant signal as an input through the filter would result in a constant with unchanged amplitude. The new coefficients after normalization is 1, a linear gain of 1 equates to a logarithmic gain of 0 dB.

Comparing both figures, figure 2 shows faster gain transitions whereas figure 1 with the 9 point FIR shows a slow and wider transition between the gain drop from pass to stop band. The sides of the 65-point FIR also shows lower levels of gain compared to the 9-point FIR.

The multiplication or rather cost required for both filters alludes to their names, 9 points for 9 multiplications and 65 points for 65 multiplications.