

## Preparatory Lab work, Lab 2

$$\frac{\Theta_e}{E_a} = \frac{nK}{s[(R_a + sL_a)(J_{eq}s + B_{eq}) + K_b K]}$$

Where  $J_{eq} = J_m + n^2 J_L$  and  $B_{eq} = B_m + n^2 B_L$

From Appendix A in the lab manual, solving for  $E_a$ ,

$$L_a \frac{dI_a}{dt} + R_a I_a + E_b = E_a,$$

We know  $E_b = K_b \Theta_m$ ,

$$L_a \frac{dI_a}{dt} + R_a I_a + K_b \Theta_m = E_a$$

↳ convert into Laplace domain

$$E_a(s) = I_a(s)[R_a + sL_a] + K_b s \Theta_m(s)$$

From the lab manual, we know:

$$T_m = K_t \Phi I_a$$

but since the transfer function is that of a permanent magnet,

$$T_m = K I_a \rightarrow I_a = \frac{T_m}{K}$$

We can sub this into the previous equation  $\rightarrow$



$$E_a(s) = \frac{[R_a + sL_a] I_m(s)}{K} + K_b s \Theta_m(s)$$

From Appendix A, we also know that:

$$T_m(t) = J_m \frac{d^2 \Theta_m}{dt^2} + B_m \frac{d\Theta_m}{dt} + T,$$

↳ Put in Laplace domain

$$T_m(s) = J_m s^2 \Theta_m + B_m s \Theta_m(s) + T,$$

If we sub this into the previous equation,

$$E_a(s) = \frac{[R_a + sL_a]}{K} [J_m s^2 \Theta_m + B_m s \Theta_m(s)] + K_b \Theta_m(s)$$

$$\frac{E_a(s)}{\Theta_m(s)} = \frac{[R_a + sL_a]}{K} [J_m s^2 + B_m s] + K_b$$

$$\frac{\Theta_m(s)}{E_a(s)} = \frac{[R_a + sL_a] [J_m s^2 + B_m s] + K_b}{K}$$

$$\frac{\Theta_m(s)}{E_a(s)} = \frac{K}{[R_a + sL_a] [J_m s^2 + B_m s] + K_b K}$$

From Appendix A we also know:

$$\Theta_L = n \Theta_m$$

Therefore we can sub in  $\Theta_L \rightarrow$



$$\frac{\Theta_l}{E_a} = \frac{nK}{s[(R_a + sL_a)(J_m s + B_m s) + K_b K]}$$

2.

$$K_a = 0.45 \rightarrow \text{motor \#1}$$

The third-order transfer function model of the motor:

$$G_p(s) = \frac{\Theta(s)}{V_a(s)} : K_a = \frac{0.45 nK}{s[(R_a + sL_a)(J_{eq} s + B_{eq}) + K_b K]}$$

3.

By ignoring Armature Inductance,  $L_a = 0$ :

$$G_p(s) = \frac{\Theta(s)}{V_a(s)} = \frac{0.45 nK}{s[(R_a)(J_{eq} s + B_{eq}) + K_b K]}$$

$$= \frac{0.45 nK}{s[R_a J_{eq} s + R_a B_{eq} + K_b K]}$$

$$= \frac{0.45 nK}{(R_a B_{eq} + K_b K)s \left[ \frac{R_a J_{eq} s}{R_a B_{eq} + K_b K} + 1 \right]}$$

$$= \frac{0.45 nK}{R_a B_{eq} + K_b K} \cdot \frac{1}{s \left[ \frac{R_a J_{eq} s}{R_a B_{eq} + K_b K} + 1 \right]}$$



4.

Assuming a ZOH of the function  $G_p(s)$ :

$$G(z) = (1 - z^{-1}) Z \left( \frac{G(s)}{s} \right)$$

$$= (1 - z^{-1}) Z \left( \frac{0.45 nK / R_a B_{eq} + K_b K}{s^2 \left[ \frac{R_a J_{eq} s}{R_a B_{eq} + K_b K} + 1 \right]} \right)$$

↳ Sub in variable values

$$= (1 - z^{-1}) Z \left[ \frac{0.45 (1/17.2) (6.09 \times 10^{-3})}{3 (1.23 \times 10^{-7}) + (6.09 \times 10^{-3}) (6.09 \times 10^{-3})} \right]$$

$$= (1 - z^{-1}) Z \left[ \frac{4.25}{s^2 (0.025s + 1)} \right]$$

$$\frac{4.25}{s^2 (0.025s + 1)} = \frac{A}{s^2} + \frac{B}{s + 40} + \frac{C}{s}$$

↳ Using partial fraction expansion,

$$A = 4.25 \quad \rightarrow \quad G(z) = (1 - z^{-1}) Z \left[ \frac{4.25}{s^2} + \frac{0.10625}{s + 40} - \frac{0.10625}{s} \right]$$

$$B = 0.10625$$

$$C = -0.10625$$

↳

$$G(z) = \frac{z-1}{z} \left[ \frac{4.25z}{(z-1)^2} + \frac{0.10625z}{z - e^{-40t}} - \frac{0.10625z}{z-1} \right]$$

$$= \frac{4.225z - 3.226}{z^2 - 1.765z + 0.765}$$

$$1/150 = 0.0067$$



## Part B

1.

The relationship between the PID parameters is that they are the same parameters rewritten.

Equation 2.1

$$G_c(s) = K_p + \frac{K_i}{s} + K_d s$$

Equation 2.4

$$G_c(s) = K_p \left( 1 + \frac{1}{T_i s} + T_d s \right)$$

In equation 2.1, the gain is already multiplied by the values of  $K_p$  and it is filtering it out in equation 2.4. For the integrator value, in equation 2.1 and equation 2.4, the denominator both contain the parameter  $s$ , the value of  $K_i = K_p / T_i$ .  $T_i$  acts as the integral action time or reset rate. The final value  $K_d = K_p \cdot T_d$ , they both contain  $s$  in the numerator and act as the derivative action time or derivative rate.

2.

From the frequency response,

$$K_u = 10^{GM/20} \rightarrow K_u = 10^{37.4/20} \rightarrow K_u = 74$$

The frequency of oscillation  $P_u$ ,

$$\begin{aligned} P_u &= \frac{2\pi}{W_{pc}} \rightarrow \text{from the frequency response } W_{pc} = 105 \\ &= \frac{2\pi}{105} \rightarrow P_u \approx 0.06 \end{aligned}$$

$\rightarrow$  From the ultimate sensitivity method,

$$K_p = 0.6 K_u \rightarrow 0.6(74) \rightarrow K_p = 44.4$$

$$T_i = P_u / 2 \rightarrow (0.06) / 2 \rightarrow T_i = 0.03$$

$$T_d = P_u / 8 \rightarrow (0.06) / 8 \rightarrow T_d = 0.0075$$



Part C

$$\Phi_{CL}(s) = \frac{G_p(s) G_c}{1 + G_p(s) G_c}$$

$$G_p(s) \cdot G_c(s) = \frac{K_m}{s(T_m s + 1)} \cdot \left( K_p + \frac{K_i}{s} + \frac{K_d s}{1 + \beta s} \right)$$

$$= \frac{K_m K_p}{s(T_m s + 1)} + \frac{K_m K_i}{s^2(T_m s + 1)} + \frac{K_m K_d s}{s(T_m s + 1)(1 + \beta s)}$$

$$= \frac{K_m K_p}{T_m s^2 + s} + \frac{K_m K_i}{T_m s^3 + s^2} + \frac{K_m K_d s}{T_m s^2 + s + T_m \beta s + \beta s^2}$$

$$\Phi_{CL}(s) = \frac{\frac{K_m K_p}{T_m s^2 + s} + \frac{K_m K_i}{T_m s^3 + s^2} + \frac{K_m K_d s}{T_m s^2 + s + T_m \beta s + \beta s^2}}{1 + \frac{K_m K_p}{T_m s^2 + s} + \frac{K_m K_i}{T_m s^3 + s^2} + \frac{K_m K_d s}{T_m s^2 + s + T_m \beta s + \beta s^2}}$$

$$\Phi_{CL} = \frac{K_m K_p + K_m K_i + K_m K_d s}{(T_m s^2 + s + T_m s^3 + s^2 + T_m s^2 + s + T_m \beta s^2 + \beta s^2) + (K_m K_p + K_m K_i + K_m K_d s)}$$

↳ By expanding and simplifying the terms →

$$\Phi_{CL} = s^4 + \left( \frac{1}{T_m} + \frac{1}{\beta} \right) s^3 + \left( \frac{K_m K_p}{T_m} + \frac{K_m K_d}{T_m \beta} + \frac{1}{T_m \beta} \right) s^2 + \left( \frac{K_m K_p}{T_m \beta} + \frac{K_m K_i}{T_m} \right) s + \left( \frac{K_m K_i}{T_m \beta} \right)$$



2.

$$\begin{aligned}
 a) \zeta &= \frac{-\ln(PO/100)}{\sqrt{\pi^2 + (\ln(PO/100))^2}} \\
 &= \frac{-\ln(15/100)}{\sqrt{\pi^2 + (\ln(15/100))^2}} \\
 &= 0.5169 \\
 &\approx 0.517
 \end{aligned}$$

$$b) T_s(1\%) = \frac{4.6}{\zeta \omega_n}$$

$$0.25(0.517) \omega_n = 4.6$$

$$\omega_n = 35.59$$

Therefore, the pair of dominant closed loop poles which satisfy the design requirement are:

$$\begin{aligned}
 s_d &= -\zeta \omega_n \pm j \omega_n \sqrt{1 - \zeta^2} \\
 &= -18.396 \pm j 30.465
 \end{aligned}$$

3.

The 2 additional closed loops are:

$$s = -a = -4\zeta \omega_n = -4(0.517)(35.59) = -73.6$$

$$s = -b = -6\zeta \omega_n = -6(0.517)(35.59) = -110.4$$

By equating equation 2.7 and 2.8,

$$\begin{aligned}
 s^4 + \left(\frac{1}{T_m} + \frac{1}{\beta}\right)s^3 + \left(\frac{K_m K_p}{T_m} + \frac{K_m K_d}{T_m \beta} + \frac{1}{T_m \beta}\right)s^2 + \\
 \left(\frac{K_m K_p}{T_m \beta} + \frac{K_m K_i}{T_m}\right)s + \left(\frac{K_m K_i}{T_m \beta}\right) = (s+a)(s+b)(s^2 + 2\zeta \omega_n s + \omega_n^2)
 \end{aligned}$$

→



We know the values of  $K_m, T_m, a, b, \zeta, \omega_n$ :

$$K_m = \frac{0.45 nK}{R_a B_{eq} + K_{bk}} = \frac{0.45 (1/172) (6.09 \times 10^{-3})}{3(1.23 \times 10^{-7}) + (6.09 \times 10^{-3})(6.09 \times 10^{-3})} = 4.25$$

$$T_m = \frac{R_a J_{eq}}{R_a B_{eq} + K_{bk}} = \frac{3(0.25 \times 10^{-6} + (1/172)^2 (2 \times 10^{-5}))}{3(1.23 \times 10^{-7}) + (6.09 \times 10^{-3})(6.09 \times 10^{-3})} = 0.025$$

$$a = -73.6$$

$$b = 110.4$$

$$\zeta = 0.517$$

$$\omega_n = 35.59$$

$$s^4 + \left( \frac{1}{0.025} + \frac{1}{\beta} \right) s^3 + \left( \frac{4.25 K_p}{0.025} + \frac{4.25 K_d}{0.025 \beta} + \frac{1}{0.025 \beta} \right) s^2 + \left( \frac{4.25 K_p}{0.025 \beta} + \frac{4.25 K_i}{0.025} \right) s + \left( \frac{4.25 K_i}{0.025 \beta} \right) = (s + 73.6)(s + 110.4)(s^2 + 2(0.517)(35.59)s + (35.59)^2)$$

↳ expanding the right side

$$\begin{aligned} &= (s + 73.6)(s + 110.4)(s^2 + 36.8s + 1266.65) \\ &= (s^2 + 184s + 8125.44)(s^2 + 36.8s + 1266.65) \\ &= s^4 + 220.8s^3 + 16163.3s^2 + 532080s + 1.02921 \times 10^7 \end{aligned}$$

↳ Since the  $s^3$  has 1 missing variable, solve for  $\beta$

$$\left( \frac{1}{0.025} + \frac{1}{\beta} \right) s^3 = 220.8 s^3 \quad \rightarrow \quad \beta = 0.00553097 \approx 0.0055$$

$$40 + \frac{1}{\beta} = 220.8 \quad \rightarrow$$



→ Since  $s^0$  has 1 missing variable solve for  $K_i$

$$\left( \frac{4.25 K_i}{0.025 \times \beta} \right) = 1.02921 \times 10^7 \quad \left. \vphantom{\frac{4.25 K_i}{0.025 \times \beta}} \right\} K_i = 332.98$$
$$\left( \frac{4.25 (K_i)}{0.025 (0.0055)} \right) = 1.02921 \times 10^7$$

→ Since  $s^1$  has 1 missing variable solve for  $K_p$

$$\left( \frac{4.25 K_p}{0.025 \beta} + \frac{4.25 K_i}{0.025} \right) s = 532.0808$$
$$\left( \frac{4.25 (K_p)}{(0.025)(0.0055)} + \frac{4.25 (332.98)}{0.025} \right) = 532.080$$
$$K_p = 15.383$$

→ Since  $s^2$  has 1 missing variable solve for  $K_d$

$$\left( \frac{4.25 K_p}{0.025} + \frac{4.25 K_d}{0.025 \beta} + \frac{1}{0.025 \beta} \right) s^2 = 16.163.3 s^2$$
$$\left( \frac{4.25 (15.383)}{0.025} + \frac{4.25 (K_d)}{0.025 (0.0055)} + \frac{1}{0.025 (0.0055)} \right) = 16.163$$
$$K_d = 0.20302$$

Part D

→ By using the forward and backward rectangle rule;

$$G_c(s) = K_p \left( 1 + \frac{1}{T_i s} + T_d s \right)$$

$$U(s) = K_p \left( 1 + \frac{1}{T_i s} + T_d s \right) E(s)$$

apply rectangle rule =  $K_p s E(s) + \frac{K_p}{T_i} E(s) + K_p T_d s^2 E(s)$

$$\hookrightarrow \frac{U(t) - U(t-1)}{T} = \left( K \frac{e(t) - e(t-1)}{T} \right) + \frac{K_p}{T_i} + \boxed{T_d \left( \frac{\dot{e}(t) - \dot{e}(t-1)}{T} \right)}$$

→ for the derivative term, b/c of the  $s^2$ , apply rectangle rule again

$$T_d \left( \frac{\frac{e(t) - e(t-1)}{T} - \frac{e(t-1) - e(t-2)}{T}}{T} \right)$$

→ Simplify the equation

$$\frac{U(t) - U(t-1)}{T} = \left[ K \left( \frac{e(t) - e(t-1)}{T} \right) \right] - \frac{K_p}{T_i} + T_d \left[ K \left[ \frac{e(t) - e(t-1)}{T^2} - \frac{e(t-1) - e(t-2)}{T^2} \right] \right]$$