Lab 3 – Interrupt I/O

Contents

Overview	2
Question 1	
Question 2	
The 1 Khz to 3 KHz	
Summary	
Exercises	10
Exercise 1	
Exercise 2	13
Traces	17
Exercise 1 Traces	17
Exercise 2 Traces	19
Appendix	22
Exercise 1 Complete Code	22
Exercise 2 Complete Code	

Overview

In this lab session we move our attention to Interrupt service routines with the main focus using a ISR to rectify a input signal using the RT DSP and show this at the output.

Question 1

From Trace 1 (below) we can see that we do indeed get a rectified signal. It is however important to note that due to the audio chip having an inverter at the output, we have inverted output channel (orange) to see this rectified effect clearly. We can also see that the rectified output is centered on 0. This is due to there being a High pass filter at the output of the DSP which removes DC offsets. This can also be explained mathematically, the Fourier series of a full wave rectified sine wave is as shown:

$$a_0 = \frac{2}{\pi} \int_0^{\pi} |\sin x| \, dx = \frac{2}{\pi} \int_0^{\pi} \sin x \, dx = \frac{4}{\pi}$$

$$where \cos n\pi = (-1)^n$$

$$a_n = \frac{2}{\pi} \int_0^{\pi} \sin x \cos nx \, dx$$

$$2 \sin x \cos nx = \sin[(n+1)x] - \sin[(n-1)x]$$

$$a_n = \frac{1}{\pi} \int_0^{\pi} \sin[(n+1)x] \, dx - \frac{1}{\pi} \int_0^{\pi} \sin[(n-1)x] \, dx$$

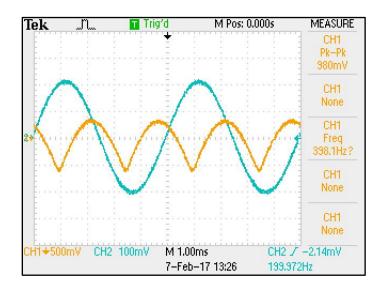
$$a_n = \frac{-1}{\pi} \left[\frac{\cos[(n+1)x]}{n+1} \right] + \frac{1}{\pi} \left[\frac{\cos[(n-1)x]}{n-1} \right] \quad (Both \ limits \ between \ \pi \ \& \ 0)$$

$$a_n = -\frac{1}{\pi} [(-1)^{n+1} - 1] \left[\frac{1}{n+1} - \frac{1}{n-1} \right] = \frac{2[(-1)^{n+1} - 1]}{\pi(n^2 - 1)}$$

$$|\sin x| = \frac{2}{\pi} - \frac{4}{\pi} \sum_{m=1}^{\infty} \frac{\cos 2mx}{(4m^2 - 1)}$$

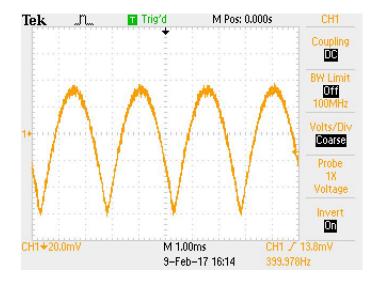
When these harmonics pass through the high pass filter at the output we remove the $\frac{2}{\pi}$ term once again indicating the rectified output must be centered at 0.

We can also see due to processing time that the output is delayed slightly from the input, output frequency has doubled as this wave has been rectified.

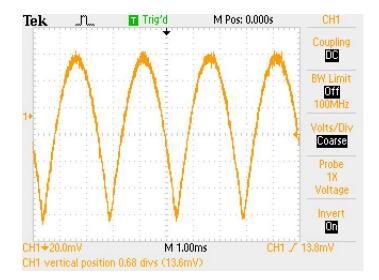


Trace 1: Traces shows a 200 Hz input (blue) alongside the 400 Hz output (orange), At 8 KHz sampling frequency.

Question 2



Trace 2: Shows the output of a 3.8 kHz sine wave, which shows a 400 Hz rectified sine wave. At 8 KHz sampling frequency.



Trace 3: Shows the output of a 200 Hz sine wave, which shows a 400 Hz rectified sine wave. At 8 KHz sampling frequency.

From Trace two and three we can observe that the two outputs have the same frequency. Additionally in order to find the frequency at the output for frequencies above 2 kHz we must use the equation. (The origins of the equation will be explained later on when comparing 200 Hz and 3.8 KHz)

$$2(\frac{F_{Samp}}{2} - Input Frequency).$$

To fully understand why the output waves have a similar frequency and shape it is important to observe these waves in the frequency domain.

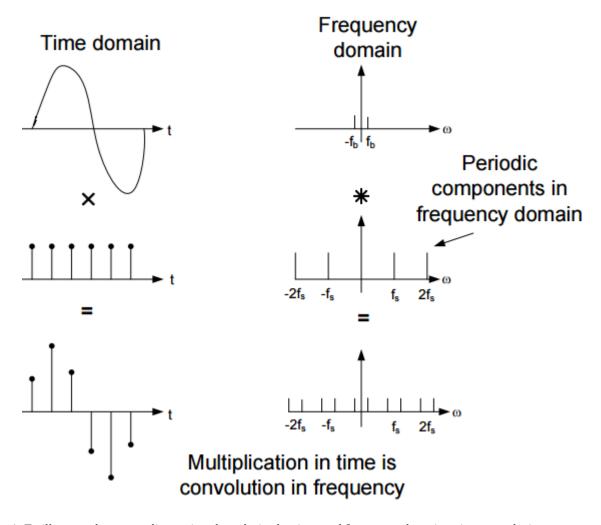


Figure 1: To illustrate how sampling a signal works in the time and frequency domain using convolution.

By using the result from before, the fourier series of a fully rectified sine wave :

$$|\sin t| = \frac{2}{\pi} - \frac{4}{\pi} \sum_{m=1}^{\infty} \frac{\cos 2mt}{(4m^2 - 1)}$$

We then find the Fourier transform of $\cos 2mt$ to plot the frequency spectrum of the full rectified sine wave.

$$\cos 2mt = \pi[\delta(w-2m) + \delta(w+2m)]$$

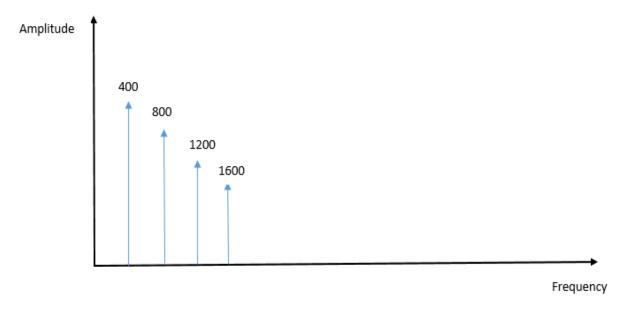


Figure 2: Shows the Fourier transform of the output when the input wave is a 200 Hz sine wave.

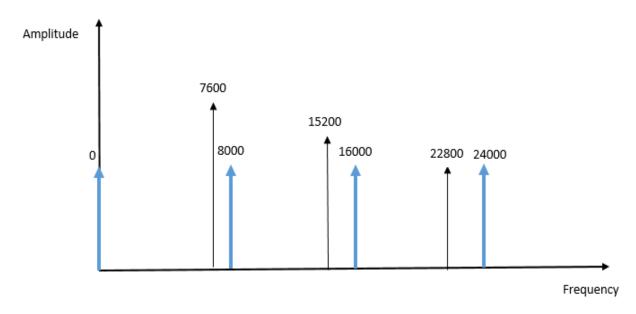


Figure 3: Shows the Fourier transform of the output (Black) when the input wave is a 3.8kHz sine wave & sampling signal (Blue).

Since when we multiple two signals in the time domain this corresponds to convolution in the frequency domain this causes the resultant Fourier transform.

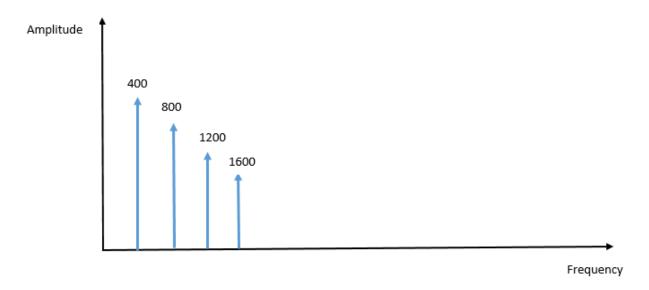


Figure 4: Shows the convolution between the output and sampling signal.

From the observations of figures two and four we can see the Fourier transform of the convoluted signal & the Fourier transform of the output of a 200Hz sine wave are similar thus this explains why the 2 waves have the same frequency and shape.

The 1 KHz to 3 KHz

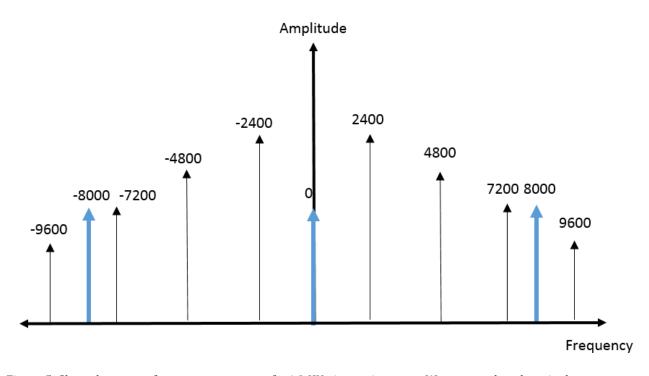


Figure 5: Show the output frequency spectrum of a 1.2 KHz input sine wave. We can see that there is clear symmetry. The blue arrows indicate the sampling frequency and the black indicates the harmonics of the output.

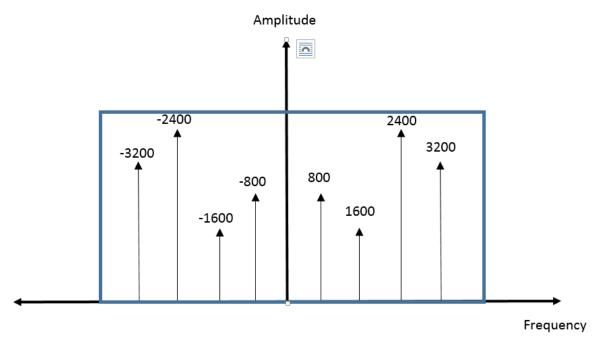
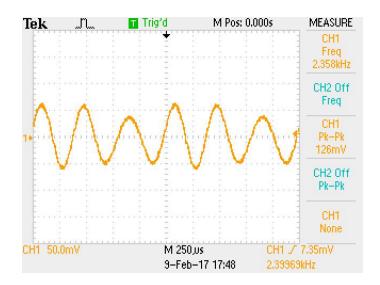


Figure 6: Shows the Anti – aliasing filter at half the sampling frequency, which is around the 4 KHz mark.



Trace 4: Shows a 2.4 KHz output wave from a 1.2 KHz at 8 Hz sampling frequency. From here we can clearly see that it is not a fully rectified waves.

From the frequency spectrum it shows that the output does not look like staircase formation similar to figure 4 thus the output from 1000-3000Hz does not look like a fully rectified sine wave instead it looks like a distorted waveform.

Summary

To conclude from our results the system only generates a full-wave rectified version of the system when the input frequency is between 0-1000Hz and 3000-4000Hz. Also, to note when the frequency is above 4000Hz, the output will simply be a shifted version of the system between 0-4000Hz however due to the properties of the anti-aliasing filter beyond 4000Hz, there will no output.

Exercises

In order to proceed with the exercises we first must understand and be able to explain all core initialisation steps.

```
// initialize board and the audio port
init_hardware();

/* initialize hardware interrupts */
init_HWI();

/* loop indefinitely, waiting for interrupts */
while(1)
{ };
```

In our main function, we begin by initialising the hardware and hardware interrupts. It also consists on an infinite loop in order for the program to run indefinitely so we can make best use of our future interrupt functions.

```
void init hardware()
   // Initialize the board support library, must be called first
   DSK6713 init();
   // Start the AIC23 codec using the settings defined above in config
   H_Codec = DSK6713_AIC23_openCodec(0, &Config);
   /* Function below sets the number of bits in word used by MSBSP (serial port) for
   receives from AIC23 (audio port). We are using a 32 bit packet containing two
   16 bit numbers hence 32BIT is set for receive */
   MCBSP_FSETS(RCR1, RWDLEN1, 32BIT);
   /* Configures interrupt to activate on each consecutive available 32 bits
   from Audio port hence an interrupt is generated for each L & R sample pair */
   MCBSP_FSETS(SPCR1, RINTM, FRM);
   /* These commands do the same thing as above but applied to data transfers to
   the audio port */
   MCBSP_FSETS(XCR1, XWDLEN1, 32BIT);
   MCBSP FSETS (SPCR1, XINTM, FRM);
```

Within the initialising hardware function we must first call **DSK6713_init** (); which ensure all hardware is running at default settings.

We then call **H_Codec = DSK6713_AIC23_openCodec(0, &Config)** which essentially configures all hardware, more importantly it sets the audio ports bit resolution and sampling frequency which will later on be crucial for both exercises as they use the audio chip and the pre-defined function of **mono_read 16Bit()** and **mono_write_16Bit()**.

The four **MCBSP_FSETS** functions are used to set up the MCBSP ports allowing communication between the AIC23 audio port and the DSP. These functions are pre-defined.

We conceptually think of these predefined functions as black box's, we know their operation but don't know the inner workings.

We then begin to configure the interrupt settings through **init_HWI**.

The pre-defined function <code>IRQ_globalDisable()</code> is critical in avoiding interrupt corruption due to interrupting critical sections of code, which are essentially sections of code that must have complete and undisturbed access to a block of data. Technically we only need to disable certain interrupts that could interrupt a critical section, but <code>IRQ_globalDisable()</code> takes a heavy handed solution and disables all interrupts which is a lot simpler to implement. This does ultimately make the entire system less responsive than it should be, however in this lab its effects will be negligible.

Exercise 1

The task required for exercise was to read a sample from the codec, these would be samples of a digital signal generator on a Windows computer then fully rectify the wave and then send out the samples to the left and right channels all within an interrupt function.

```
void Audio_ISR (void) {
    /* Takes a sample from the digital signal generator into a 16 bit short variable,
        to ensure we get a recified signal we take the absolute vale of each sample.
        It is then pass on to Mono_write where we output the rectified signal */
        short read_sample = abs( mono_read_16Bit());
        mono_write_16Bit(read_sample);
}
```

Figure 7: Shows the interrupt service routine that reads a sample, rectifies and sends a copy to the output.

Audio_ISR is the interrupt function we used to implement this function. Very few lines of code is required to execute the required task. The essential part of the ISR is the two special functions **mono_write_16Bit** and **mono_read_16Bit**. **Mono_read_16Bit** provides a mono input by reading left and right samples from an audio port, divides by two and sums them up. The result is then returned as a 16 bit integer.

We store this mono result in a short variable named read_sample. Short is used as its 16 bit signed allowing us to store the returned mono sample. To ensure we have fully rectified the input signal we have to make all negative value samples from the input turn into positives at the same magnitude. This can be easily done by simply just finding the absolute value using the **abs** () function.

Read_sample is then passed through **mono_write_16Bit** where in which it sends the rectified samples to left and right output channels to achieve a mono output.

Exercise 2

In this exercise we are given the task of fusing sections of lab 2 into lab 3 where in which we are required to use a interurupt service routine to generate a sine-wave using a look up table. Here is how we impletement this.

Firstly we have to fill a look up table with one cycle of a sin wave. This is done within the main which calls a function **sine_init()**.

```
void main() {

    // initialize board and the audio port
    init_hardware();

    sine_init(); //fill the look up table with sine wave samples

    /* initialize hardware interrupts */
    init_HWI();

    /* loop indefinitely, waiting for interrupts */
    while(1)
    { };
}
```

Figure 8 : Shows the main function, hardware interrupts are initalised along side filling the look up table with a one cycle sinwave.

Within this function we are filling a look-up table with samples of a one cycle sin wave. The size of the look up table is of size 256 hence a For loop is required to cycle through the entire table with each array block being filled by a sample from equation: $\frac{\sin 2i\pi}{256}$. Here is a example of how it visually it does this.

I	62	63	64	65	66
	0.9987954	0.9996988	1.0	0.9996988	0. 9987954
$\frac{\sin 2i\pi}{256}$					
	*	+	*	*	*
Table[i]					

These values again can be seen from the 'watch' window on the code composer studio.

Local (1) 🥸 Watch (1) 🛭			□ × ¾ ⋄ ጭ		· 🗆
Name	Value	Address	Туре	Format	•
(x)= [61]	0.9972904	0x000099BC	float	Natural	
(x)= [62]	0.9987954	0x000099C0	float	Natural	
(x)= [63]	0.9996988	0x000099C4	float	Natural	
(×)= [64]	1.0	0x000099C8	float	Natural	
(x)= [65]	0.9996988	0x000099CC	float	Natural	
(x)= [66]	0.9987954	0x000099D0	float	Natural	
(x)= [67]	0.9972904	0x000099D4	float	Natural	
(×)= [68]	0.9951847	0x000099D8	float	Natural	
(x)= [69]	0.9924796	0x000099DC	float	Natural	
(x)= [70]	0.9891765	0x000099E0	float	Natural	
(×)= [71]	0.9852777	0x000099E4	float	Natural	Ŧ

Figure 9 : Shows a snippet of ten values [index from 61 – 71] within the lookup table.

Figure 10 : Shows the function that implements and fills a look up table of size 256 with samples of one cycle of a sine wave

Further to this, to actually generate a sine wave from the look up table we have to edit the interrupt service routine to give a rectified output signal.

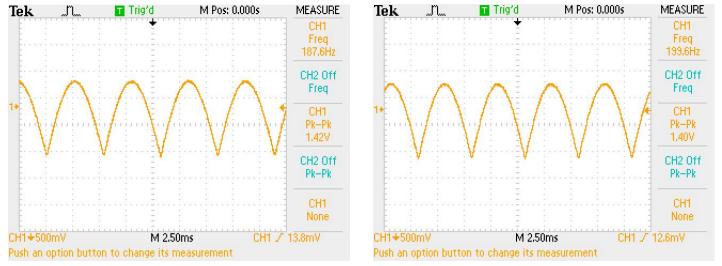
Figure 11: Shows the ISR that will read from the look up table, rectifies and does a mono write.

In order to get our desired frequency we have to determine how many sample we have to skip. To do this we use equation:

$$\frac{Sampling\ freq\ *\ skipped\ samples}{Sine_table_size} = Sine\ frequency$$

This has been implemented with line 'realstepsize += ((sine_freq*SINE_TABLE_SIZE)/(sampling_freq))' which equates the global variable 'realstepsize' of type float to a certain sample of the one cycle sine wave. Furthermore due to the '+=' we increment 'realstepsize' by the amount we need to skip by determined by the equation above.

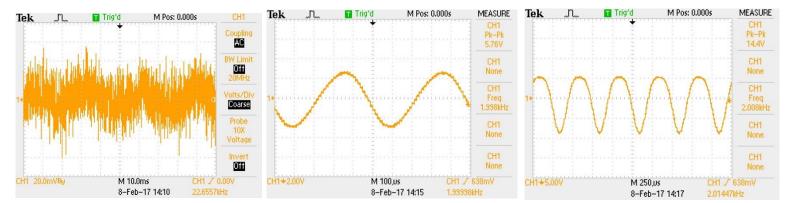
The reason we use type float for 'realstepsize' is because for certain frequencies we need to skip in decimal numbers. An example would be at a 100 Hz where, using the above equation we get $\frac{100*256}{8000} = 3.2$ skipping steps. If integer was used the numbers after the decimal would be ignored, i.e in this case 0.2 would be ignored and would just skip 3. This becomes a problematic issue if say the number we need to skip is 3.9, in integer form we again would just skip 3, resulting in us not always getting our desired frequency. To solve this we used 'indexcounter = (realstepsize < 0 ? (realstepsize - 0.5) : (realstepsize) + 0.5);' which does this rounding function.



Traces 4,5: Show how the rounding feature results in more accurate outputs at 100 Hz input at 8 KHz sampling frequency. The left trace does not use the rounding function whereas the right trace does indeed use the rounding function.

To ensure that the index counter always remains between 0 – 255 and not extend the reach of the table size we run a condition. The If loop checks if the **'indexcounter'** and **'realstepsize'** is larger than the table size of 256, if it is the we subtract the size value of the look up table in order for the **'indexcounter'** and **'realstepsize'** to keep incrementing in our desired skip size.

When we go ahead to write the sample, we are doing several things all at once. Firstly we have to take note that the look up table is of type float which is 32 Bit signed but the pre-defined function **mono_write_16Bit** can only take a copy of a 16 bit value. Hence there is a compatibility issue. To solve this issue we use function **(short) floor(...)** which reduces a value from the table to a 16 bit short solving the compatibility issue. However as the values from the look up table are too small to be seen on an oscilloscope we have to have some gain. In this case we multiply by 32767 which is the max number of a 16 bit signed value.

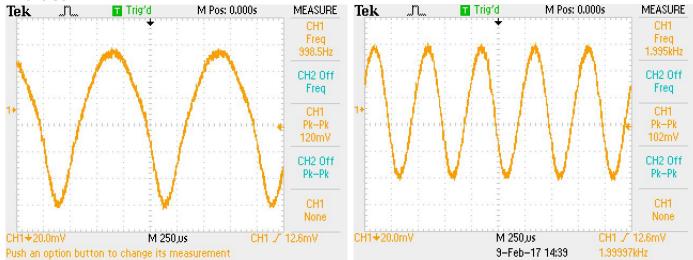


Figures 12: Show how multiplying by a gain makes the signal clear to read of an oscilloscope. Taken at 1000 Hz

Furthermore to get a rectified output signal we take the absolute value of read samples just in the same fashion we did in exercise 1. We could have done this rectification at other stages also in the code. The Function **mono_write_16Bit** then sends copies to the left and right sections channels.

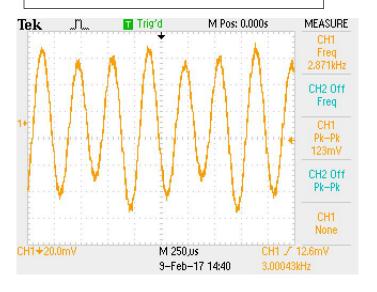
Traces



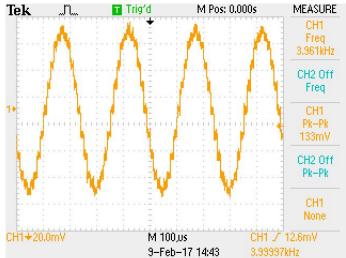


Shows a 1000 Hz output from a 500 Hz input generated by a digital signal generator

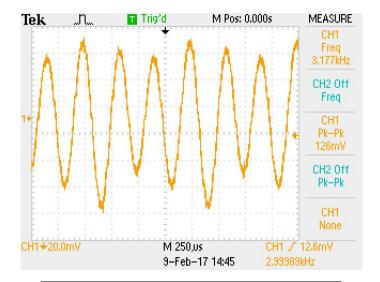


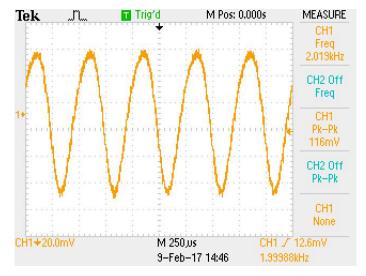


Shows a 3000 Hz output from a 1500 Hz input generated by a digital signal generator



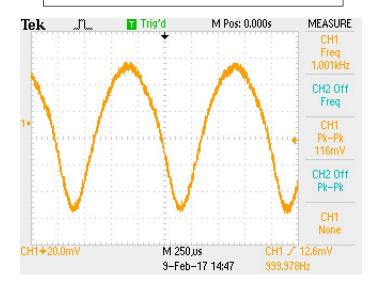
Shows a 4000 Hz output from a 2000 Hz input generated by a digital signal generator

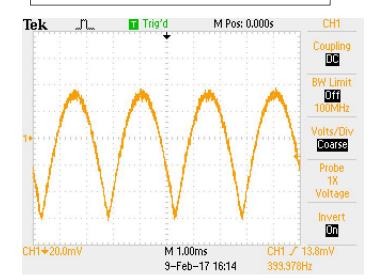




Shows a 3000 Hz output from a 2500 Hz input generated by a digital signal generator

Shows a 2000 Hz output from a 3000 Hz input generated by a digital signal generator

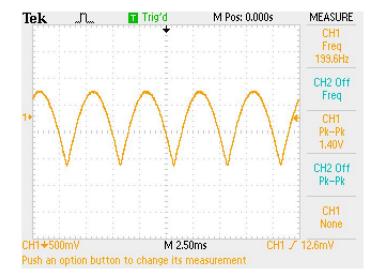




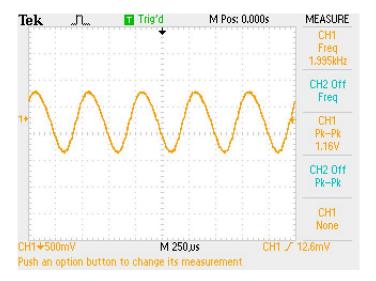
Shows a 1000 Hz output from a 3500 Hz input generated by a digital signal generator

Shows a 400 Hz output from a 3800 Hz input generated by a digital signal generator

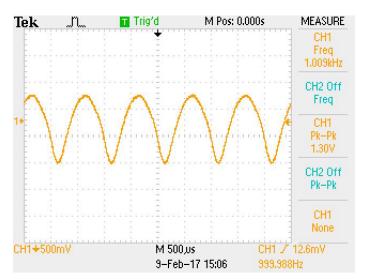
Exercise 2



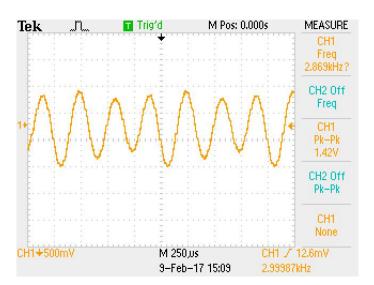
Shows a 200 Hz output from a 100 Hz input generated by a digital signal generator



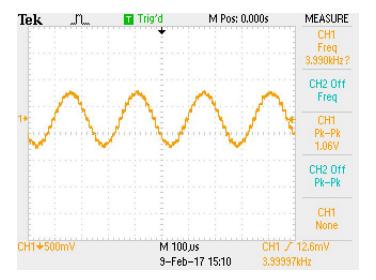
Shows a 2000 Hz output from a 1000 Hz input generated by a digital signal generator

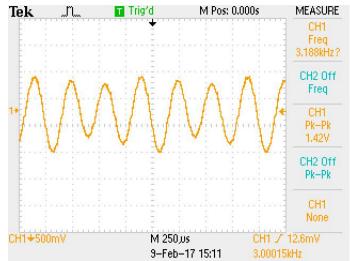


Shows a $1000~\mathrm{Hz}$ output from a $500~\mathrm{Hz}$ input generated by a digital signal generator



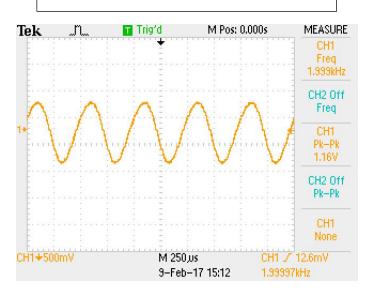
Shows a 3000 Hz output from a 1500 Hz input generated by a digital signal generator

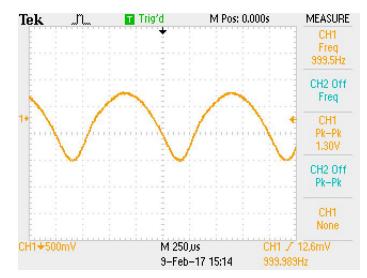




Shows a $4000~\mathrm{Hz}$ output from a $2000~\mathrm{Hz}$ input generated by a digital signal generator

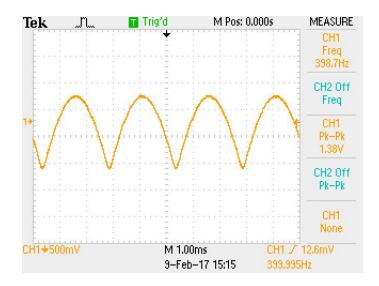
Shows a 3000 Hz output from a 2500 Hz input generated by a digital signal generator





Shows a 2000 Hz output from a 3000 Hz input generated by a digital signal generator

Shows a 1000 Hz output from a 3500 Hz input generated by a digital signal generator



Shows a 400 Hz output from a 3800 Hz input generated by a digital signal generator

Appendix: Exercise 1 complete code:

```
DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING
 3
                             IMPERIAL COLLEGE LONDON
 4
 5
                     EE 3.19: Real Time Digital Signal Processing
 6
                         Dr Paul Mitcheson and Daniel Harvey
 8
                              LAB 3: Interrupt I/O
9
10
                          ****** I N T I O. C *******
11
12
      Demonstrates inputing and outputing data from the DSK's audio port using interrupts.
13
14
      15
                Updated for use on 6713 DSK by Danny Harvey: May-Aug 2006
16
                Updated for CCS V4 Sept 10
    17
   □/*
18
     * You should modify the code so that interrupts are used to service the
19
20
21
    22
23
24
   #include <stdlib.h>
25
   // Included so program can make use of DSP/BIOS configuration tool.
26
   #include "dsp bios cfg.h"
27
28 -/* The file dsk6713.h must be included in every program that uses the BSL. This
29
      example also includes dsk6713 aic23.h because it uses the
30
      AIC23 codec module (audio interface). */
   #include "dsk6713.h"
31
    #include "dsk6713 aic23.h"
32
33
34
    // math library (trig functions)
35
    #include <math.h>
36
37
    // Some functions to help with writing/reading the audio ports when using interrupts.
    #include <helper_functions_ISR.h>
38
39
    40
41
42
   -/* Audio port configuration settings: these values set registers in the AIC23 audio
43
       interface to configure it. See TI doc SLWS106D 3-3 to 3-10 for more info. */
44
   DSK6713 AIC23 Config Config = { \
              45
              /* REGISTER
46
                                 FUNCTION
                                                    SETTINGS
47
              0x0017, /* 0 LEFTINVOL Left line input channel volume 0dB
48
                                                                   */\
                                                                  */\
49
        0x0017, /* 1 RIGHTINVOL Right line input channel volume 0dB
        0x01f9, /* 2 LEFTHPVOL Left channel headphone volume 0dB
                                                                   */\
        0x01f9, /* 3 RIGHTHPVOL Right channel headphone volume 0dB
        0x0011, /* 4 ANAPATH Analog audio path control
                                                  DAC on, Mic boost 20dB*/\
                          Digital audio path control
53
        0x0000, /* 5 DIGPATH
                                                  All Filters off
54
        0x0000, /* 6 DPOWERDOWN Power down control
                                                  All Hardware on
                                                                   */\
        0x0043, /* 7 DIGIF Digital audio interface format 16 bit
55
                                                                   */\
        0x008d, /* 8 SAMPLERATE Sample rate control
56
                                                  8 KHZ
                                                                   */\
       0x0001 /* 9 DIGACT Digital interface activation On
```

```
59
    13;
60
61
62
     // Codec handle:- a variable used to identify audio interface
63
    DSK6713_AIC23_CodecHandle H_Codec;
64
     65
66
    void init_hardware(void);
67
    void init_HWI(void);
68
     void Audio_ISR (void); // ISR function
69
70
     71
   □void main(){
72
73
74
        // initialize board and the audio port
75
     init_hardware();
76
77
      /* initialize hardware interrupts */
78
      init_HWI();
79
80
      /* loop indefinitely, waiting for interrupts */
81
82
      while(1)
83
      { };
84
85
86
     87
88
     void init_hardware()
89
90
        // Initialize the board support library, must be called first
91
        DSK6713_init();
92
93
        // Start the AIC23 codec using the settings defined above in config
94
       H Codec = DSK6713 AIC23 openCodec(0, &Config);
95
       /* Function below sets the number of bits in word used by MSBSP (serial port) for
96
   receives from AIC23 (audio port). We are using a 32 bit packet containing two
97
98
        16 bit numbers hence 32BIT is set for receive */
99
        MCBSP FSETS (RCR1, RWDLEN1, 32BIT);
100
101
   /* Configures interrupt to activate on each consecutive available 32 bits
102
        from Audio port hence an interrupt is generated for each L & R sample pair */
103
        MCBSP_FSETS(SPCR1, RINTM, FRM);
104
105
   /st These commands do the same thing as above but applied to data transfers to
106
        the audio port */
107
        MCBSP_FSETS(XCR1, XWDLEN1, 32BIT);
108
        MCBSP_FSETS(SPCR1, XINTM, FRM);
109
110
111
112
```

```
DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING
 3
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 4
 5
                      EE 3.19: Real Time Digital Signal Processing
 6
                         Dr Paul Mitcheson and Daniel Harvey
 7
8
                               LAB 3: Interrupt I/O
9
                           ****** I N T I O. C *******
10
11
12
       Demonstrates inputing and outputing data from the DSK's audio port using interrupts.
13
14
      15
                 Updated for use on 6713 DSK by Danny Harvey: May-Aug 2006
16
                 Updated for CCS V4 Sept 10
17
18
   □ /*
19
     * You should modify the code so that interrupts are used to service the
     * audio port.
20
21
     22
23
24
25
     // Included so program can make use of DSP/BIOS configuration tool.
26
    #include "dsp bios cfg.h"
27
28
   -/* The file dsk6713.h must be included in every program that uses the BSL. This
29
       example also includes dsk6713_aic23.h because it uses the
30
       AIC23 codec module (audio interface). */
31
   #include "dsk6713.h"
32
    #include "dsk6713 aic23.h"
33
34
    // math library (trig functions)
35
    #include <math.h>
36
37
    // Some functions to help with writing/reading the audio ports when using interrupts.
38
    #include <helper_functions_ISR.h>
39
40
     // PI defined here for use in your code
41
    #define PI 3.141592653589793
42
    #define SINE_TABLE_SIZE 256
43
44
     45
46
   /* Audio port configuration settings: these values set registers in the AIC23 audio
47
       interface to configure it. See TI doc SLWS106D 3-3 to 3-10 for more info. */
48
   DSK6713 AIC23 Config Config = { \
               /-----/
49
               /* REGISTER
50
                                   FUNCTION
                                                       SETTINGS
               51
52
        0x0017,
              /* 0 LEFTINVOL Left line input channel volume 0dB
        0x0017, /* 1 RIGHTINVOL Right line input channel volume 0dB
53
                                                                     */\
        0x01f9, /* 2 LEFTHPVOL Left channel headphone volume 0dB
                                                                     */\
54
        0x01f9, /* 3 RIGHTHPVOL Right channel headphone volume 0dB
55
        0x0011, /* 4 ANAPATH Analog audio path control DAC on, Mic boost 20dB*/\
56
        0x0000, /* 5 DIGPATH Digital audio path control
57
                                                   All Filters off
```

```
0x0000, /* 6 DPOWERDOWN Power down control
         0x0043, /* 7 DIGIF Digital audio interface format 16 bit
         0x008d, /* 8 SAMPLERATE Sample rate control
               /* 9 DIGACT Digital interface activation On
               63
64
65
66
     // Codec handle:- a variable used to identify audio interface
     DSK6713_AIC23_CodecHandle H_Codec;
68
69
    -/* Sampling frequency in HZ. Must only be set to 8000, 16000, 24000
    _32000, 44100 (CD standard), 48000 or 96000 */
70
71
     int sampling_freq = 8000;
72
73
     // Holds the value of the current sample
74
     float sample;
75
76
   -/* Use this variable in your code to set the frequency of your sine wave
       be carefull that you do not set it above the current nyquist frequency! */
78
    float sine_freq = 1000.0;
79
    float table [SINE_TABLE_SIZE];
     int indexcounter = 0;
80
81
     float realstepsize=0;
82
      83
84
     void init hardware(void);
85
     void init HWI (void);
86
     void Audio ISR (void);
87
     void sine_init(void);
88
     89
90
   -void main() {
91
92
93
        // initialize board and the audio port
94
      init_hardware();
95
        sine init(); //fill the look up table with sine wave samples
97
98
       /* initialize hardware interrupts */
99
       init HWI();
100
101
       /* loop indefinitely, waiting for interrupts */
102
103
      while(1)
104
      { };
105
106
107
     108
109
     void init hardware()
110
    □ {
111
         // Initialize the board support library, must be called first
112
        DSK6713_init();
113
      // Start the AIC23 codec using the settings defined above in config
```

```
/* Configures interrupt to activate on each consecutive available 32 bits
                      from Audio port hence an interrupt is generated for each L & R sample pair */
                    MCBSP FSETS(SPCR1. RINTM. FRM);
                    /\!\!\!\!\!\!^* These commands do the same thing as above but applied to data transfers to
                     the audio port */
                    MCBSP_FSETS(XCR1, XWDLEN1, 32BIT);
                    MCBSP_FSETS(SPCR1, XINTM, FRM);
         void init HWI (void)
                   /****** WRITE YOUR INTERRUPT SERVICE ROUTINE HERE************/
□void Audio_ISR (void) {
                     realstepsize += ((sine_freq*SINE_TABLE_SIZE)/(sampling_freq)); // determines the number of samples skipped in
                    indexcounter = (realstepsize < 0 ? (realstepsize - 0.5 ) : (realstepsize) + 0.5); /*Since the index of the table only takes integer values
                                                                                                                                                                                                                                                      we have to do rounding on the index for certain frequencyies in order to get accurate output */
                    if (indexcounter >= SINE_TABLE_SIZE) /* if condition ensures the index doesn't exceed the
                                                                                                                             size of the look up table for the next sample reading.*/
                    realstepsize = realstepsize - SINE_TABLE_SIZE; //^^
                    indexcounter = indexcounter- SINE TABLE SIZE;
                    mono_write_16Bit(abs((short)floor(table [indexcounter] * 32767))); // Reduces float values in table to shorts and multiplies by 32767 which is the heighest
                                                                                                                                                                                                                               // in a short in order to display on ossciliscope. We also take the absolute value
                                                                                                                                                                                                                              // in orderto rectify the signal
            void sine init(void) // initialises all values on the look up table. i.e generates a whole cycle of a sine wave.
  ₽ {
                     int i; // counter variable for the For loop below.
                       \textbf{for (i = 0 ; i < SINE\_TABLE\_SIZE ; i++)} \ / \text{`Size of For loop is limited to the size of the look up table, to a size of the look up table, to the size of table. The look up table, to the look up table, the look up table,
                                                                                                                                                      ensure there are no redudant values. */
                                 table[i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills the entire look up table with 256 sample of table [i] = sin((2*i*PI)/SINE_TABLE_SIZE); // Fills table [i] = sin((2*i*
```

(All snippets is in high resolution, please zoom in if its appears too small)