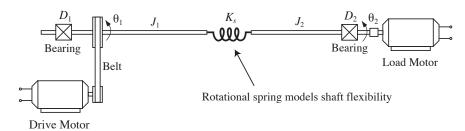
## AA/EE/ME 581 Digital Control System Design

Homework #6
Reading: Chapter 6

1:



The apparatus for a DC motors experiment is represented schematically above. The voltage applied to the Drive Motor's amplifier,  $e_1$ , is the control input. The voltage applied to the Load Motor's amplifier,  $e_2$ , is the disturbance input. A controller is to be designed to make the angle,  $\theta_2$ , of the load motor end of the flexible shaft follow a reference angle,  $\theta_{2ref}$ , in spite of the disturbance  $e_2$ . Only  $\theta_{2ref}$  and  $\theta_2$  are measured. The dynamics of the system are represented by:

$$\begin{split} L_1 \frac{di_1}{dt} &= - \left( R_{A1} + R_1 \right) i_1 - n K_1 \frac{d\theta_1}{dt} + K_{A1} e_1 \\ L_2 \frac{di_2}{dt} &= - \left( R_{A2} + R_2 \right) i_2 - K_2 \frac{d\theta_2}{dt} + K_{A2} e_2 \\ \left( J_1 + n^2 J_{m1} \right) \frac{d^2 \theta_1}{dt^2} &= n K_1 i_1 - \left( D_1 + n^2 D_{m1} \right) \frac{d\theta_1}{dt} - K_s \left( \theta_1 - \theta_2 \right) \\ \left( J_2 + J_{m2} \right) \frac{d^2 \theta_2}{dt^2} &= K_2 i_2 - \left( D_2 + D_{m2} \right) \frac{d\theta_2}{dt} + K_s \left( \theta_1 - \theta_2 \right) \end{split}$$

where

 $L_1$ ,  $L_2$  are the drive and load motor armature inductances (H)

 $i_1$ ,  $i_2$  are the drive and load motor currents (A)

 $R_{A1}$ ,  $R_{A2}$  are the drive and load motor amplifier resistances ( $\Omega$ )

 $R_1$ ,  $R_2$  are the drive and load motor armsture resistances ( $\Omega$ )

 $K_1$ ,  $K_2$  are the drive and load motor back EMF constants (V/rad/sec)

 $K_{A1}$ ,  $K_{A2}$  are the drive and load motor amplifier gains (dimensionless)

 $J_{m1}$ ,  $J_{m2}$  are the drive and load motor rotational mass moment of inertias (kg m<sup>2</sup>)

n is the gear ratio of the pulley and belt assembly (dimensionless)

 $D_{m1}$ ,  $D_{m2}$  are the drive and load motor viscous damping constants (N m/rad/sec)

 $D_1$ ,  $D_2$  are the drive and load motor bearing viscous damping constants (N m/rad/sec)

 $J_1$ ,  $J_2$  are the drive and load motor shaft section rotational mass moments of inertia (kg m<sup>2</sup>)

or, in state equation form,

$$\frac{dx}{dt} = Ax(t) + Bu(t)$$

with

$$x = \begin{bmatrix} i_1 & i_2 & \theta_1 & \frac{d\theta_1}{dt} & \theta_2 & \frac{d\theta_2}{dt} \end{bmatrix}^T$$

$$u = \begin{bmatrix} e_1 & e_2 \end{bmatrix}^T$$

$$A = \begin{bmatrix} -\frac{R_{A1} + R_1}{L_1} & 0 & 0 & -\frac{nK_1}{L_1} & 0 & 0 \\ 0 & -\frac{R_{A2} + R_2}{L_2} & 0 & 0 & 0 & -\frac{K_2}{L_2} \\ 0 & 0 & 0 & 1 & 0 & 0 \\ \frac{nK_1}{J_1 + n^2 J_{m1}} & 0 & -\frac{K_s}{J_1 + n^2 J_{m1}} & -\frac{D_1 + n^2 D_{m1}}{J_1 + n^2 J_{m1}} & \frac{K_s}{J_1 + n^2 J_{m1}} & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & \frac{K_2}{J_2 + J_{m2}} & \frac{K_s}{J_2 + J_{m2}} & 0 & -\frac{K_s}{J_2 + J_{m2}} & -\frac{D_2 + D_{m2}}{J_2 + J_{m2}} \end{bmatrix}$$

$$B = \begin{bmatrix} \frac{K_{A1}}{L_1} & 0 \\ 0 & \frac{K_{A2}}{L_2} \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix}$$

Matlab users: the script dcmtrs.m gives these matrices for a particular set of parameters: [A,B] = dcmtrs;

Python users: use dcmtrs.ipynb

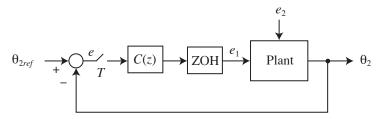


Figure 1. Closed-loop control system

1. In Figure 1, the Plant block represents the dynamics of the DC motors apparatus defined by the dcmtrs file.

The controller transfer function is

$$C(z) = \frac{55}{100} \frac{z - 0.46415}{z - 0.2} \frac{z - 0.991}{z - 1}$$

and the sampling period is T = 0.02 sec.

- (a) Determine the Figure 1 system's gain and phase margins.
- (b) Determine the smallest positive gain for the sensor (larger than the nominal gain for the sensor, which is one) that generates the measurement of  $\theta_2$  for the Figure 1 system's feedback path such that the system will be unstable.
- (c) Use Matlab and Simulink or Python-Control to generate a plot that has superimposed on it the first 1 sec of the  $\theta_1(t)$  and  $\theta_2(t)$  responses of the Figure 1 system to a unit step  $\theta_{2ref}$  input when the sensor that generates the measurement of  $\theta_2$  for the feedback path has the destabilizing gain you determined for part (b).
- (d) Suppose that the sensor that generates the measurement of  $\theta_2$  for the Figure 1 system's feedback path generates its measurement with a time delay. Determine the smallest such time delay (larger than the nominal time delay for the sensor, which is zero) such that Figure 1 system will be unstable. 1
- (e) Use Matlab and Simulink or Python-Control to generate a plot that has superimposed on it the first 1 sec of the  $\theta_1(t)$  and  $\theta_2(t)$  responses of the Figure 1 system to a unit step  $\theta_{2ref}$  input when the sensor that generates the measurement of  $\theta_2$  for the system's feedback path generates its measurement with the time delay you determined for part (d).

<sup>&</sup>lt;sup>1</sup> Such a time delay could also model, equally well, the computation time for a digital computer to update the control signal following each sampling instant.