

Lab 3: PID control of 4th order plant

The plant's input-output dynamics are represented by:

$$Y(s) = \frac{R_{of1} + R_{of2}}{R_{of1}} \left[\left(\frac{R_{om1}}{R_{om1} + R_{om2}} \right) \frac{\omega_{n1}^2}{s^2 + 2\zeta_1\omega_{n1}s + \omega_{n1}^2} + \left(\frac{R_{om2}}{R_{om1} + R_{om2}} \right) \frac{\omega_{n2}^2}{s^2 + 2\zeta_2\omega_{n2}s + \omega_{n2}^2} \right]$$

The system's two inputs are the control input u and the disturbance input d . The outputs are the output voltage y and its derivative, \dot{y} . Design software by S. B. Fuller 2023.04, based on a Matlab version by M. C. Berg 2017.05.

In [1]:

```
import numpy as np # numerical arrays
import matplotlib.pyplot as plt # plotting
%config InlineBackend.figure_format='retina' # high-res plots
import control.matlab as ctm # matlab layer for control systems library
import control as ct # use regular control library for a few things
ct.set_defaults('statesp', latex_repr_type='separate') # show ABCD matrices
```

state-space plant model

In [2]:

```
# Low-frequency mode
R1 = 160e3 # ohms. "e3" is shorthand for the exponent of ten and means "times 10^3"
R2 = 200e3 # ohms
C1 = 10e-6 # farads
C2 = 0.082e-6 # farads
tau1 = R1 * C1
tau2 = R2 * C2
tau3 = R1 * C2
omegan1 = 1/np.sqrt(tau1 * tau2)
print('Natural frequency of low-frequency vibration mode =', omegan1/(2*np.pi), 'Hz')
zeta1 = (tau2 + tau3) * omegan1/2
print('Damping ratio of low-frequency vibration mode =', zeta1)

# high-frequency mode
R3 = 68e3
R4 = 13e3
C3 = 6.8e-6
C4 = 0.01e-6
tau1 = R3 * C3
tau2 = R4 * C4
tau3 = R3 * C4
omegan2 = 1/np.sqrt(tau1 * tau2);
print('Natural frequency of high-frequency vibration mode =', omegan2/(2*np.pi), 'Hz')
zeta2 = (tau2 + tau3) * omegan2 / 2
print('Damping ratio of high-frequency vibration mode =', zeta2)

# weighting factors
Rof1 = 10e3
Rof2 = 100
Rom1 = 10e3
Rom2 = 1e3
K1 = (Rof1 + Rof2)/Rof1*Rom1/(Rom1+Rom2) # gain of first mode
K2 = (Rof1 + Rof2)/Rof1*Rom2/(Rom1+Rom2) # gain of second mode

AA = np.array(
```

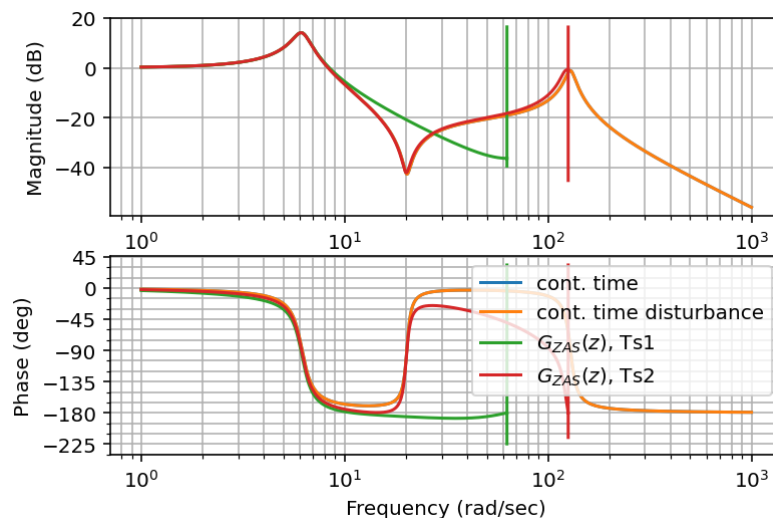
```

[[ 0, 1, 0, 0], # x1
 [-omegan1**2, -2*zeta1*omegan1, 0, 0], # d(x1)/dt
 [ 0, 0, 0, 1], # x2
 [ 0, 0, -omegan2**2, -2*zeta2*omegan2]] # d(x2)/dt
BBu = np.array(
[[ 0,
 [omegan1**2],
 [omegan2**2]])
BBd = np.array(
[[ 0,
 [omegan1**2],
 [omegan2**2]])
BB = np.hstack([BBu, BBd])
CC = np.array(
[[K1, 0, K2, 0], # y
 [0, K1, 0, K2]]) # dy/dt
DD = np.array(
[[0, 0],
 [0, 0]])
plant = ctm.ss(AA, BB, CC, DD)
dcgain = ctm.dcgain(plant[0,0])
print('DC gain of plant u input =', dcgain)
Kff = 1/dcgain

Ts1 = 50/1000 # 50 msec
Ts2 = 25/1000 # 25 msec
plantdisc1 = ctm.c2d(plant[0,0], Ts1, 'zoh')
plantdisc2 = ctm.c2d(plant[0,0], Ts2, 'zoh');
ctm.bode(plant[0,0], label='cont. time')
ctm.bode(plant[0,1], label='cont. time disturbance')
ctm.bode(plantdisc1, label='$G_{ZAS}(z)$, Ts1')
ctm.bode(plantdisc2, label='$G_{ZAS}(z)$, Ts2')
plt.legend();

```

Natural frequency of low-frequency vibration mode = 0.9825127984059134 Hz
 Damping ratio of low-frequency vibration mode = 0.09111805529092465
 Natural frequency of high-frequency vibration mode = 20.527664581044217 Hz
 Damping ratio of high-frequency vibration mode = 0.05223654379696277
 DC gain of plant u input = 1.01



function to plot lines of constant damping and settle time:

```

In [3]: def plot_constant_lines_z(zeta, omegan, T=1, c=None, plot_omegan=False):
        """plot lines of constant damping ratio (zeta), time constant
        (1/zeta*omegan) and omegan on the z-plane"""

```

```

nyquist_freq = np.pi/T
freqs = np.linspace(-nyquist_freq, nyquist_freq, 201, endpoint=True)
# constant time constant is a vertical line in s-plane at -zeta*omegan
timeconstant_line_s = -zeta*omegan*np.ones_like(freqs) + 1j*freqs
timeconstant_line_z = np.exp(timeconstant_line_s*T)
plt.plot(timeconstant_line_z.real, timeconstant_line_z.imag,
         '--', c=c, label=f'$\zeta\omega_n$={zeta*omegan:.3f}')
# constant damping ratio zeta is a radial line from origin in s-plane
damping_line_splane = -abs(freqs)/np.tan(np.arccos(zeta)) + 1j*freqs
damping_line_zplane = np.exp(damping_line_splane*T)
plt.plot(damping_line_zplane.real, damping_line_zplane.imag, '-', c=c,
         label=f'$\zeta$={zeta:.3f}')
if plot_omegan: # circle centered at origin in s-plane
    angles = np.linspace(np.pi/2, 3./2*np.pi, 31, endpoint=True)
    omegan_line_s = omegan*np.exp(1j*angles)
    omegan_line_z = np.exp(omegan_line_s*T)
    plt.plot(omegan_line_z.real, omegan_line_z.imag,
            c=c, ls='-.', label=f'$\omega_n$={omegan:.3f}')

```

In [4]:

```

# create systems used for interconnections
C_ff = ct.tf2ss(Kff, 1, inputs='r', outputs='uff')
e_summer = ct.summing_junction(['r', '-y'], 'e')
u_summer = ct.summing_junction(['uf', 'uff', '-ub', 'd'], 'u')

```

Controller design

The following cell allows you to iteratively explore how adding or subtracting an amount δK to one of the PID gains K_p , K_i , and K_d affect the step and frequency of the system so that its performance can be tuned. To use, perform the following steps, editing *only* the elements surrounded by `###` :

1. Choose the starting point for your gain values, which gain to vary, direction, and input signal.
2. Run the cell. A figure appears showing a bode plot and step response for $\delta K = 0.1$. The top right figure shows the root locus as δK varies from 0 to ∞ . Click on a branch of the root locus plot to try a different δK ; the associated δK is printed below the cell and the step response and bode plots are updated.
 - remark: you can zoom in using the magnifying glass; you must click the magnifying glass again to be able to choose a gain on the root locus plot again.
3. Once you have found a suitable gain change δK , its value appears as `gain` on the bottom of the printout beneath the cell. *Add or subtract it* to the gain you chose above and return to step 1 for further iteration.
4. Copy your chosen gains including your new gain into the subsequent cell to simulate the full sampled-data system.

Example: to examine the effect of varying K_p starting from an initial value of 10, use `gain = 'P'` and `Kp0 = 10` . If a δK value of 5 is found to give satisfactory performance, then use `Kp0 = 15` on the next iteration.

for online notebook users

If you are using an online notebook like Google colab, you cannot use live, clickable plots. Instead, comment out the first line on the next cell to use inline plots, and you can repeatedly

run this cell with different values δK by setting the value of the `initial_gain` argument in `sisotool` at the bottom of this cell to your desired δK .

```
In [5]: # comment out the following line if you are using an online notebook like colab (see
%matplotliblib

#### choose between plantdisc1 or plantdisc2, which have different sampling times
plantdisc = ctm.ss(plantdisc1, inputs='u', outputs='y')
# starting point gains
Kp0 = 0.7003
Ki0 = 0.14519
Kd0 = 0.3528
# choose which gain to vary, choose one of: 'P', 'I', or 'D'
gain = 'd'
# pick direction to vary the gain
sign = -1
# pick input, one of 'r' for reference or 'd' for disturbance
input_signal = 'r'
###

# see lecture slides for explanation for the following construction
Ts = plantdisc.dt
prop = ctm.tf2ss(1, 1, inputs='e', outputs='prop_e')
integ = ctm.tf([Ts, 0], [1, -1], Ts, inputs='e', outputs='int_e')
deriv = ctm.tf([1, -1], [Ts, 0], Ts, inputs='y', outputs='deriv_y')
Kpgain = ctm.tf2ss(Kp0, 1, inputs='prop_e', outputs='uf')
Kigain = ctm.tf2ss(Ki0, 1, inputs='int_e', outputs='uf')
Kdgain = ctm.tf2ss(Kd0, 1, inputs='deriv_y', outputs='ub')
if gain == 'P' or gain == 'p':
    Kpgain = ctm.ss([], [], [], [[0, 1], [-sign, Kp0]],
                    inputs=['input', 'prop_e'], outputs=['output', 'uf'])
elif gain == 'I' or gain == 'i':
    Kigain = ctm.ss([], [], [], [[0, 1], [-sign, Ki0]],
                    inputs=['input', 'int_e'], outputs=['output', 'uf'])
elif gain == 'D' or gain == 'd':
    Kdgain = ctm.ss([], [], [], [[0, 1], [-sign, Kd0+1e-5]],
                    inputs=['input', 'deriv_y'], outputs=['output', 'ub'])
else:
    raise ValueError(gain_selection + ' gain not recognized.')

loop = ct.interconnect((plantdisc, Kpgain, Kigain, Kdgain,
                        prop, integ, deriv, C_ff, e_summer, u_summer),
                       inputs=['input', input_signal], outputs=['output', 'y'])

ctm.sisotool(loop, initial_gain=0.1)
plt.sca(plt.gcf().get_axes()[1]) # switch to root locus plot
plot_constant_lines_z(.7797, 5.13, Ts)
plt.axis((-1,1,-1,1));
```

Using matplotlib backend: Qt5Agg

C:\Users\ROG\anaconda3\lib\site-packages\control\iosys.py:1503: UserWarning: Unused input(s) in InterconnectedSystem: (9, 3)=sys[17].d
warn(msg)

simulate sampled-data system (digital+analog):

```
In [6]: def sampled_data_system(sysd, simulation_dt):
        """Create a (discrete-time, non-linear) sampled-data system.

        The returned system models the behavior of a sampled-data system
        `sysd`, such as a controller consisting of a sampler and a
```

digital-to-analog converter. The returned system is discrete-time, with a timebase `simulation_dt` smaller than or equal to the sampling interval `sysd.dt`. Using a small `simulation_dt` will insure that continuous-time dynamics of the interconnected systems are accurately simulated. The system returned by this function must be interconnected with other systems that also have a timebase equal to `simulation_dt`. `sysd.dt` must be greater than or equal to `simulation_dt`, and an integral multiple of it. Connect to a continuous-time system such as a plant `P` by first converting `P` to a discrete-time system using e.g. `P_sim = P.sample(simulation_dt)`. To implement a sampled static gain (e.g. a proportional control law), insure that `sysd` has at least one state, even if it has no effect on the output. For example, for a gain `Kp`, use `sysd = ct.ss(0,0,0,Kp,Ts)`.

```

"""
assert ct.isdtime(sysd, True), "sysd must be discrete-time"
sysd = ct.ss(sysd) # convert to state-space if not already
nsteps = int(round(sysd.dt / simulation_dt))
assert np.isclose(nsteps, sysd.dt/simulation_dt), \
    "simulation_dt must be an integral multiple of sysd.dt"
st = 0
y = np.zeros((sysd.noutputs, 1))
def updatefunction(t, x, u, params):
    nonlocal st
    if st == 0: # is it time to sample?
        x = sysd._rhs(t, x, u)
    st += 1
    if st == nsteps:
        st = 0
    return x
def outputfunction(t, x, u, params):
    nonlocal y
    if st == 0: # is it time to sample?
        y = sysd._out(t, x, u)
    return y
return ct.ss(updatefunction, outputfunction, dt=simulation_dt,
             name=sysd.name, inputs=sysd.input_labels,
             outputs=sysd.output_labels, states=sysd.state_labels)

```

In [7]:

```

# back to in-line plots:
%matplotlib inline

### choose between plantdisc1 or plantdisc2 (different sampling times)
plantdisc = ctm.ss(plantdisc1, inputs='u', outputs='y')
# final gains
Kp = 0.6194
Ki = 0.1
Kd = 0.35796
###

# first simulate discrete-time model from GZAS model of plant, for stem plot
Ts = plantdisc.dt
# the following two have zero A and B matrices to insure that it has states so
# that sampling works correctly, see sampled_data_system function above
C_ff = ctm.ss(0, 0, 0, Kff, Ts, inputs='r', outputs='uff')
prop = ctm.ss(0, 0, 0, 1, Ts, inputs='e', outputs='prop_e')
integ = ctm.tf([Ts, 0], [1, -1], Ts, inputs='e', outputs='int_e')
deriv = ctm.tf([1, -1], [Ts, 0], Ts, inputs='y', outputs='deriv_y')
Kpgain = ctm.tf2ss(Kp, 1, inputs='prop_e', outputs='uf')
Kigain = ctm.tf2ss(Ki, 1, inputs='int_e', outputs='uf')
Kdgain = ctm.tf2ss(Kd, 1, inputs='deriv_y', outputs='ub')
# r to y
Gyr_discrete = ct.interconnect((plantdisc, Kpgain, Kigain, Kdgain,

```

```

                                prop, integ, deriv, C_ff, e_summer, u_summer),
                                inputs='r', outputs='y')
# d to y
Gyd_discrete = ct.interconnect((plantdisc, Kpgain, Kigain, Kdgain,
                                prop, integ, deriv, C_ff, e_summer, u_summer),
                                inputs='d', outputs='y')

plt.figure(1)
plt.subplot(2,1,1)
y, t = ctm.step(Gyr_discrete, T=2)
plt.stem(t, y, 'C1.', markerfmt='C1.', label='discrete-time model')

plt.figure(2)
plt.subplot(2,1,1)
y, t = ctm.step(Gyd_discrete, T=30)
plt.plot(t, y, 'C1.', markersize=5, label='discrete-time model')

# simulate full sampled-data system by simulating with a much shorter dt.
simulation_dt = 0.005

# constituent parts
# note: pure gains Kpgain, Kigain, and Kdgain don't need to be converted
prop_sim = sampled_data_system(prop, simulation_dt)
integ_sim = sampled_data_system(integ, simulation_dt)
deriv_sim = sampled_data_system(deriv, simulation_dt)
C_ff_sim = sampled_data_system(C_ff, simulation_dt)
plant_sim = ctm.ss(ctm.c2d(plant[0,0], simulation_dt, 'zoh'), inputs='u', outputs='y')

# closed-loop r to y and u
Gr_sim = ct.interconnect((plant_sim, Kpgain, Kigain, Kdgain,
                            prop_sim, integ_sim, deriv_sim,
                            C_ff_sim, e_summer, u_summer),
                            inputs='r', outputs=['y', 'u'])

# closed-loop d to y and u
Gd_sim = ct.interconnect((plant_sim, Kpgain, Kigain, Kdgain,
                            prop_sim, integ_sim, deriv_sim,
                            C_ff_sim, e_summer, u_summer),
                            inputs='d', outputs=['y', 'u'])

# simulate
plt.figure(1)
plt.subplot(2,1,1)
time_sim = np.arange(0, 2, simulation_dt)
input_sim = np.ones_like(time_sim)
t, y = ct.input_output_response(Gr_sim, time_sim, input_sim)
plt.plot(t, y[0], label='full simulation')
plt.xlabel('time (s)')
plt.ylabel('ref. step resp. $y$ (V)')
plt.legend()
plt.axhline(1.02, c='k', linestyle="--", linewidth=0.75)
plt.axhline(.98, c='k', linestyle="--", linewidth=0.75)
plt.axvline(1, c='k', linestyle="--", linewidth=0.75)

plt.subplot(2,1,2)
plt.plot(t, y[1], label='full simulation')
plt.xlabel('time (s)')
plt.ylabel('$u$ (V)')
plt.axhline(0, c='k', linestyle="--", linewidth=0.75)
plt.legend()

plt.figure(2)
time_sim = np.arange(0, 30, simulation_dt)
input_sim = np.ones_like(time_sim)
t2, y2 = ct.input_output_response(Gd_sim, time_sim, input_sim)

```

```

plt.subplot(2,1,1)
plt.plot(t2, y2[0], label='full simulation')
plt.xlabel('time (s)')
plt.ylabel('disturb. step resp. $y$ (V)')
plt.axhline(.1, c='k', linestyle="--", linewidth=0.75)
plt.axvline(30, c='k', linestyle="--", linewidth=0.75)
plt.legend()

plt.subplot(2,1,2)
plt.plot(t2, y2[1], label='full simulation')
plt.xlabel('time (s)')
plt.ylabel('$u$ (V)')
plt.axhline(0, c='k', linestyle="--", linewidth=0.75)
plt.legend();

```

C:\Users\ROG\anaconda3\lib\site-packages\control\iosys.py:1503: UserWarning: Unused input(s) in InterconnectedSystem: (9, 3)=sys[17].d

warn(msg)

C:\Users\ROG\anaconda3\lib\site-packages\control\iosys.py:1503: UserWarning: Unused input(s) in InterconnectedSystem: (7, 0)=sys[55].r; (8, 0)=sys[16].r

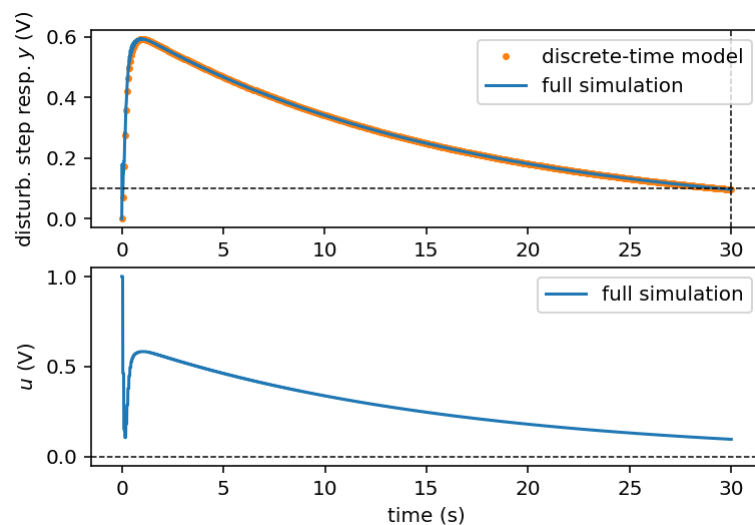
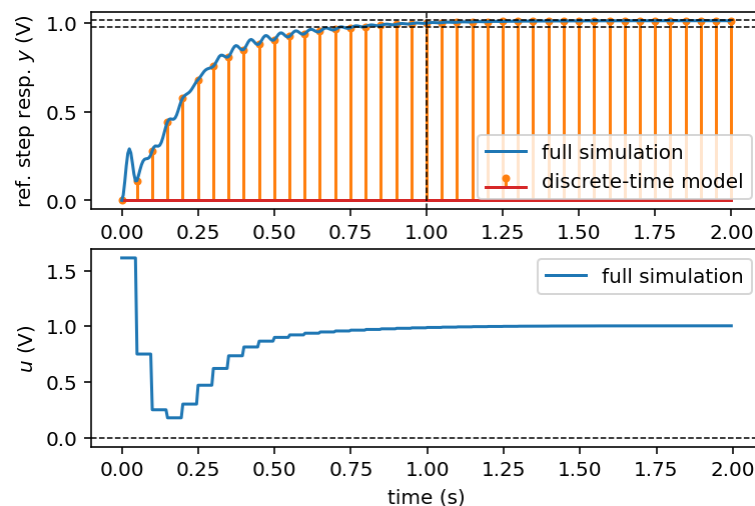
warn(msg)

C:\Users\ROG\anaconda3\lib\site-packages\control\iosys.py:1503: UserWarning: Unused input(s) in InterconnectedSystem: (9, 3)=sys[17].d

warn(msg)

C:\Users\ROG\anaconda3\lib\site-packages\control\iosys.py:1503: UserWarning: Unused input(s) in InterconnectedSystem: (7, 0)=sys[55].r; (8, 0)=sys[16].r

warn(msg)



In [8]:

```

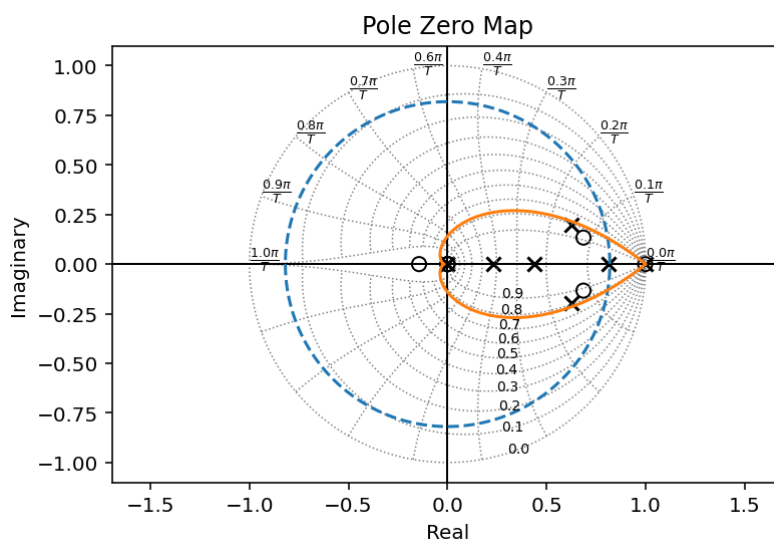
ctm.pzmap(Gyr_discrete, grid=True)
plot_constant_lines_z(.7797, 5.13, Ts)

```

```
ctm.damp(Gyr_discrete);
```

```
C:\Users\ROG\anaconda3\lib\site-packages\control\lti.py:122: RuntimeWarning: divide
by zero encountered in log
    splane_poles = np.log(poles.astype(complex))/self.dt
C:\Users\ROG\anaconda3\lib\site-packages\control\lti.py:122: RuntimeWarning: invalid
value encountered in true_divide
    splane_poles = np.log(poles.astype(complex))/self.dt
C:\Users\ROG\anaconda3\lib\site-packages\control\lti.py:126: RuntimeWarning: invalid
value encountered in true_divide
    Z = -real(splane_poles)/wn
```

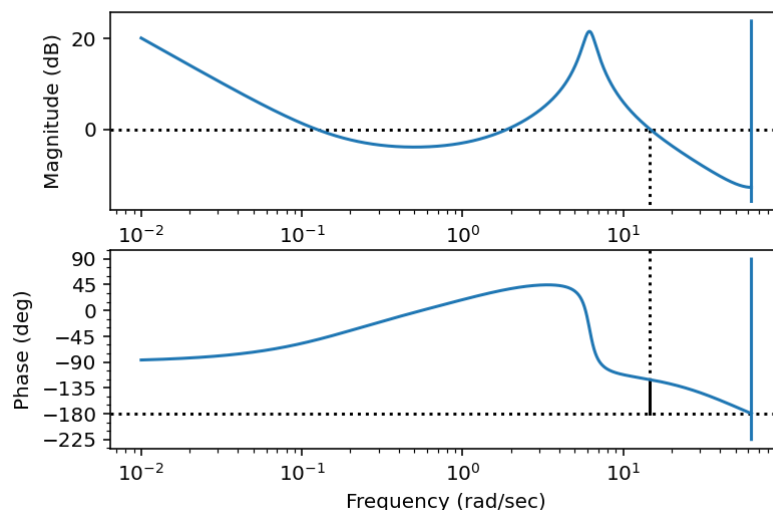
Eigenvalue	Damping	Frequency
0.2282	1	-0.2282
0.4359	1	-0.4359
0.6267 +0.1977j	0.8086	10.39
0.6267 -0.1977j	0.8086	10.39
0.9969	1	-0.9969
0.8115	1	-0.8115
0	1	-0
0	1	-0



In [9]:

```
loopdisc = (Kp + Ki * integ + Kd * deriv) * plantdisc
ctm.bode(loopdisc, margins=True);
```

Gm = inf dB (at nan rad/s), Pm = 59.38 deg (at 14.68 rad/s)



In []: