Homework 3 in EL2450 Hybrid and Embedded Control Systems

First name1 Last name1 person number email

First name3 Last name3
person number
email

First name2 Last name2 person number email

First name4 Last name4 person number email

Task 1

Since

$$u_{\omega} = \frac{u_r + u_l}{2}$$
$$u_{\Psi} = u_r - u_l$$

 \Leftrightarrow

$$u_l = u_\omega - \frac{u_\Psi}{2}$$
$$u_r = u_\omega + \frac{u_\Psi}{2}$$

- Task 2
- Task 3
- Task 4
- Task 5
- Task 6
- Task 7

$$\begin{split} \theta[k+1] &= \theta[k] + \frac{T_s R}{L} u_{\Psi}^R[k] \\ &= \theta[k] + \frac{T_s R}{L} K_{\Psi}^R(\theta^R - \theta[k]) \\ &= \theta[k] (1 - \frac{T_s R}{L} K_{\Psi}^R) + \frac{T_s R}{L} K_{\Psi}^R \theta^R \end{split}$$

By subtracting θ^R from both sides one gets

$$\theta[k+1] - \theta^{R} = \theta[k] (1 - \frac{T_{s}R}{L} K_{\Psi}^{R}) + (\frac{T_{s}R}{L} K_{\Psi}^{R} - 1) \theta^{R}$$

$$= (\theta[k] - \theta^{R}) (1 - \frac{T_{s}R}{L} K_{\Psi}^{R})$$
(1)

We now define state θ' as

$$\theta'[k] = \theta[k] - \theta^R$$

Then, equation 1 becomes

$$\theta'[k+1] = \theta'[k](1 - \frac{T_s R}{L} K_{\Psi}^R)$$

In order for this system to be stable, that is, $|\theta - \theta^R| \to 0$ as $k \to \infty$, the solution of the equation inside the parentheses should lie inside the unit circle:

$$\left| 1 - \frac{T_s R}{L} K_{\Psi}^R \right| < 1$$

$$-1 < 1 - \frac{T_s R}{L} K_{\Psi}^R < 1$$

$$-2 < -\frac{T_s R}{L} K_{\Psi}^R < 0$$

$$0 < \frac{T_s R}{L} K_{\Psi}^R < 2$$

$$0 < K_{\Psi}^R < \frac{2L}{T_s R}$$
(2)

Hence, the maximum value K_{Ψ}^R can take for the system to be marginally stable is $K_{\Psi,max}^R=\frac{2L}{T_sR}$

Theoretically, the value of K_{Ψ}^{R} can be chosen to be any value inside the interval defined in inequality 2. However, first, it would be wise to choose a value that is far enough from the maximum value so as to avoid overshoot, but close enough to it, so that convergence happens in reasonable time. Hence, in practice, it is reasonable that one would need to experiment with different values and choose one that results in balancing a small angular error, a minimal overshoot, if any, and a quick enough settling time.

Eventually, the choice made for K_{Ψ}^{R} was $K_{\Psi}^{R} = \frac{1}{2}K_{\Psi,max}^{R}$.

Task 8

For the purpose of simulating this part of the controller, the initial point of the robot was taken to be $I(x_0, y_0) \equiv (0, 0)$. The goal was set to $G(x_g, y_g) \equiv (-0.37, 1.68)$, which is node 1 in the simulation environment. The angle between the line connecting I and G and the x-axis is hence $\theta_R = tan^{-1}(1.68/-0.37) = 102.42$ degrees.

Figures 2, 4, 6, 8, 10 show the angular response of the robot for different values of K_{Ψ}^{R} inside the interval set by inequality 2. Figure 12 verifies that the upper limit for K_{Ψ}^{R} is indeed $K_{\Psi,max}^{R} = \frac{2L}{T_{s}R}$ by showing that the angular response of the robot cannot converge for $K_{\Psi}^{R} > K_{\Psi,max}^{R}$.

Here, one can see that the smaller the value of K_{Ψ}^{R} is, the larger the settling time, the lower the rise time and the smoother the response is. However, as the value of K_{Ψ}^{R} increases, the steady-state response begins to oscillate, with the amplitude of this oscillation proportional to the value of K_{Ψ}^{R} .

Figures 3, 5, 7, 9 and 3 focus on the steady-state value of the aforementioned responses. As it is evident, none of the responses converge to the value $\theta_R = 102.42$. This is reasonable since with only a purely proportional control signal, as the angular error, i.e. $e(\theta) = \theta^R - \theta$, tends to zero, the product of K_{Ψ}^R and $e(\theta)$ isn't large enough to force the robot to rotate exactly θ^R degrees.

Another way to look at this is by looking at the steady-state response of the system, which is linear, for a step input of magnitude θ^R . Figure 1 shows the structure of the system. The z-transform of the input is then $R(z) = \frac{\theta^R}{1-z^{-1}}$ and the equation of the closed-loop system is

$$Y(z) = \frac{K_{\Psi}^R G(z)}{1 + K_{\Psi}^R G(z)} R(z)$$

The steady-state response is

$$\lim_{t \to \infty} y(t) = \lim_{z \to 1} (1 - z^{-1}) \frac{\theta^R}{1 - z^{-1}} \frac{K_{\Psi}^R G(z)}{1 + K_{\Psi}^R G(z)} = \theta^R \cdot \lim_{z \to 1} \frac{K_{\Psi}^R G(z)}{1 + K_{\Psi}^R G(z)}$$

The steady-state response cannot reach exactly θ^R as the above limit cannot converge to 1 under our limitations for K_{Ψ}^R and the dynamics of G(z).

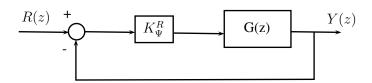


Figure 1: The structure of the system under rotational control. G(z) is the discretized transfer function of the linear system whose state-space equation is $\dot{\theta} = \frac{R}{L} u_{\Psi}$

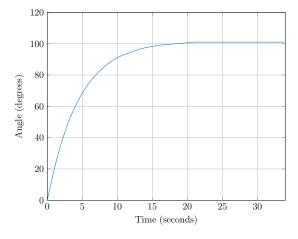


Figure 2: The orientation of the robot over time for $K_{\Psi}^R=0.1K_{\Psi,max}^R$

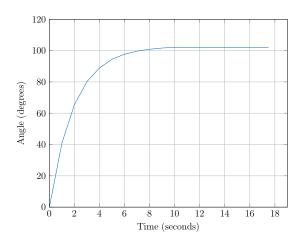


Figure 4: The orientation of the robot over time for $K_{\Psi}^R=0.2K_{\Psi,max}^R$

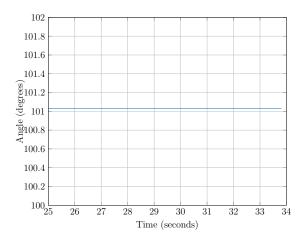


Figure 3: The steady state orientation of the robot for $K_{\Psi}^{R}=0.1K_{\Psi,max}^{R}$

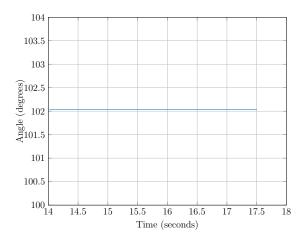


Figure 5: The steady state orientation of the robot for $K_{\Psi}^R=0.2K_{\Psi,max}^R$

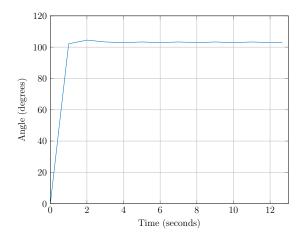


Figure 6: The orientation of the robot over time for $K_{\Psi}^R = 0.5 K_{\Psi,max}^R$

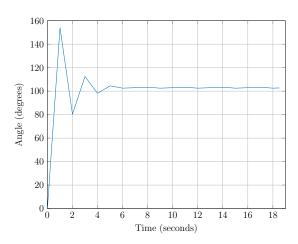


Figure 8: The orientation of the robot over time for $K_{\Psi}^R=0.75K_{\Psi,max}^R$

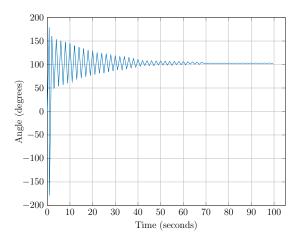


Figure 10: The orientation of the robot over time for $K_{\Psi}^{R} = K_{\Psi,max}^{R}$. This is the upper limit value of K_{Ψ}^{R} before the system becomes unstable

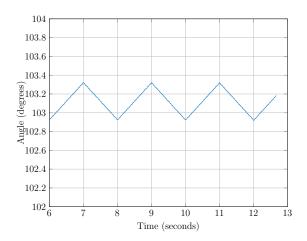


Figure 7: The steady state orientation of the robot for $K_{\Psi}^R=0.5K_{\Psi,max}^R$

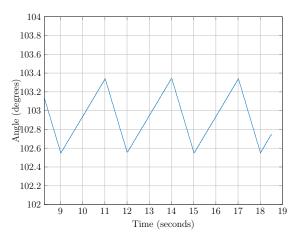


Figure 9: The steady state orientation of the robot for $K_{\Psi}^R=0.75K_{\Psi,max}^R$

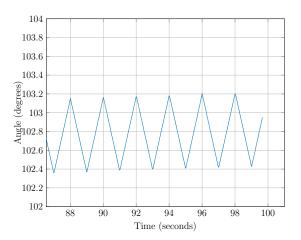


Figure 11: The steady state orientation of the robot for $K_{\Psi}^R = K_{\Psi,max}^R$

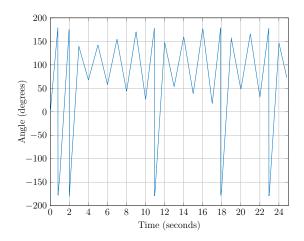


Figure 12: The orientation of the robot over time for $K_{\Psi}^{R} = K_{\Psi,max}^{R} + 1$. The system is indeed unstable

Task 9

$$\begin{split} d_0[k] &= \cos(\theta[k])(x_0 - x[k]) + \sin(\theta[k])(y_0 - y[k]) \\ &= \cos(\theta[k])(x_0 - x[k-1] - T_sRu_\omega^R[k-1]\cos(\theta[k-1])) \\ &+ \sin(\theta[k])(y_0 - y[k-1] - T_sRu_\omega^R[k-1]\sin(\theta[k-1])) \\ &= \cos(\theta^R)(x_0 - x[k-1] - T_sRu_\omega^R[k-1]\cos(\theta^R)) \\ &+ \sin(\theta^R)(y_0 - y[k-1] - T_sRu_\omega^R[k-1]\sin(\theta^R)) \\ &= \cos(\theta^R)(x_0 - x[k-1]) + \sin(\theta^R)(y_0 - y[k-1]) - T_sRK_\omega^T d_0[k-1] \\ &= d_0[k-1] - T_sRK_\omega^T d_0[k-1] \\ &= (1 - T_sRK_\omega^R)d_0[k-1] \end{split}$$

Hence

$$d_0[k+1] = (1 - T_s R K_\omega^R) d_0[k]$$

In order for this system to be stable, that is, $d_0 \to 0$ as $k \to \infty$, the solution of the equation inside the parentheses should lie inside the unit circle:

$$\left| 1 - T_s R K_{\omega}^R \right| < 1$$

$$-1 < 1 - T_s R K_{\omega}^R < 1$$

$$-2 < -T_s R K_{\omega}^R < 0$$

$$0 < T_s R K_{\omega}^R < 2$$

$$0 < K_{\omega}^R < \frac{2}{T_s R}$$
(3)

Hence, the maximum value K_ω^R can take for the system to be marginally stable is $K_{\omega,max}^R=\frac{2}{T_sR}.$

Theoretically, the value of K_{ω}^{R} can be chosen to be any value inside the interval defined in inequality 3. However, first, it would be wise to choose a value that is far enough from the maximum value so as to avoid overshoot, but close enough to it, so that convergence happens in reasonable time. Hence, in practice, it is reasonable that one would need to experiment with different values and choose one that results in balancing a small angular error, a minimal overshoot, if any, and a quick enough settling time.

Eventually, the choice made for K_{ω}^{R} was $K_{\omega}^{R} = \frac{1}{2}K_{\omega,max}^{R}$.

Task 10

This part of the controller is responsible for compensating translational errors during rotation. Since the rotational speed u_{Ψ} is zero, then it is expected that the robot will not move away from its origin. Figure 13 plots the robot's distance from its origin over time for $K_{\omega}^{R} = 0.5 K_{\omega,max}^{R}$.

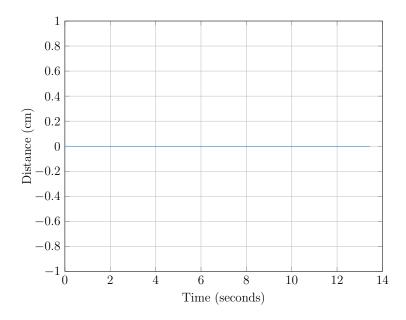


Figure 13: The distance of the robot from its origin position over time for $K_{\omega}^{R}=0.5K_{\omega,max}^{R}$

Task 11

Figures 14, 16, 18, 20, 22 show the bearing error of the robot for different values of K_{Ψ}^{R} inside the interval set by inequality 2. Figures 15, 17, 19, 21 and 15 focus on the steady-state bearing error of the aforementioned figures.

The evolution of the bearing and distance error is the same when both of the rotational controllers are enabled compared to when only one of them is enabled. This happens because the behaviour of each controller does not affect the behaviour of the other, since this is an ideal system. In reality, we expect that the distance error will be non-zero.

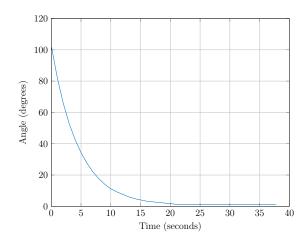


Figure 14: The error in orientation of the robot over time for $K_{\Psi}^{R}=0.1K_{\Psi,max}^{R}$

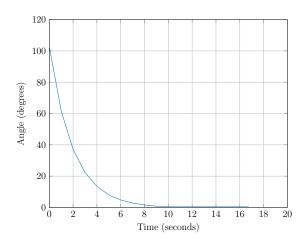


Figure 16: The error in orientation of the robot over time for $K_{\Psi}^{R}=0.2K_{\Psi,max}^{R}$

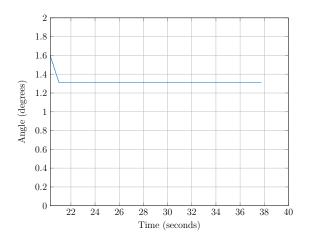


Figure 15: The steady state error in orientation of the robot for $K_{\Psi}^R=0.1K_{\Psi,max}^R$

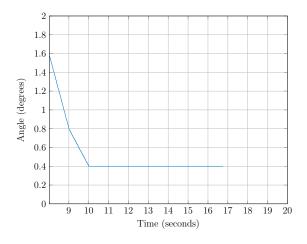


Figure 17: The steady state error in orientation of the robot for $K_{\Psi}^R=0.2K_{\Psi,max}^R$

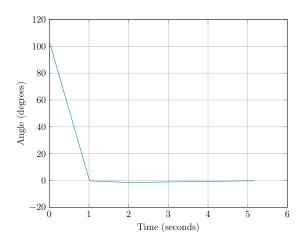


Figure 18: The error in orientation of the robot over time for $K_{\Psi}^{R} = 0.5 K_{\Psi,max}^{R}$

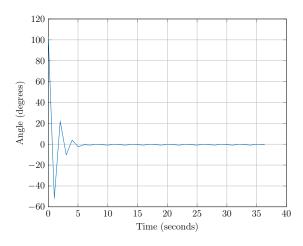


Figure 20: The error in orientation of the robot over time for $K_{\Psi}^R=0.75K_{\Psi,max}^R$

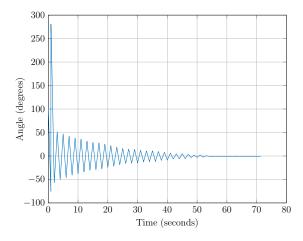


Figure 22: The error in orientation of the robot over time for $K_{\Psi}^{R} = K_{\Psi,max}^{R}$.

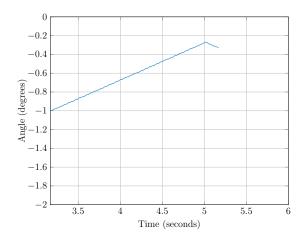


Figure 19: The steady state error in orientation of the robot for $K_{\Psi}^{R}=0.5K_{\Psi,max}^{R}$

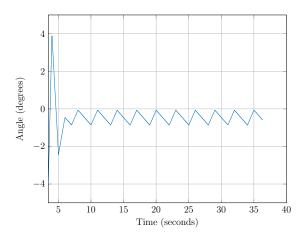


Figure 21: The steady state error in orientation of the robot for $K_{\Psi}^{R}=0.75K_{\Psi,max}^{R}$

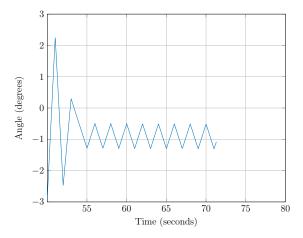


Figure 23: The steady state error in orientation of the robot for $K_{\Psi}^R = K_{\Psi,max}^R$

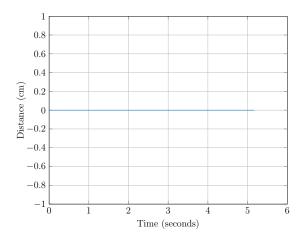


Figure 24: The distance of the robot from its origin position over time for $K_{\omega}^{R}=0.5K_{\omega,max}^{R}$

Task 12

$$\begin{split} d_g[k] &= \cos(\theta_g)(x_g - x[k]) + \sin(\theta_g)(y_g - y[k]) \\ &= \cos(\theta_g)(x_g - x[k-1] - T_s R u_\omega^T[k-1] \cos(\theta_g)) + \\ &\sin(\theta_g)(y_g - y[k-1] - T_s R u_\omega^T[k-1] \sin(\theta_g)) \\ &= \cos(\theta_g)(x_g - x[k-1]) + \sin(\theta_g)(y_g - y[k-1]) - T_s R K_\omega^T d_g[k-1] \\ &= d_g[k-1] - T_s R K_\omega^T d_g[k-1] \\ &= (1 - T_s R K_\omega^T) d_g[k-1] \end{split}$$

Hence

$$d_g[k+1] = (1 - T_s R K_{\omega}^T) d_g[k]$$

In order for this system to be stable, that is, $d_g \to 0$ as $k \to \infty$, the solution of the equation inside the parentheses should lie inside the unit circle:

$$\left| 1 - T_s R K_{\omega}^T \right| < 1$$

$$-1 < 1 - T_s R K_{\omega}^T < 1$$

$$-2 < -T_s R K_{\omega}^T < 0$$

$$0 < T_s R K_{\omega}^T < 2$$

$$0 < K_{\omega}^T < \frac{2}{T_s R}$$
(4)

Hence, the maximum value K_ω^T can take for the system to be marginally stable is $K_{\omega,max}^T=\frac{2}{T_sR}$

Theoretically, the value of K_{ω}^{T} can be chosen to be any value inside the interval defined in inequality 4. However, first, it would be wise to choose a value that is far enough from the maximum value so as to avoid overshoot, but close enough to it, so that convergence

happens in reasonable time. Hence, in practice, it is reasonable that one would need to experiment with different values and choose one that results in balancing a small angular error, a minimal overshoot, if any, and a quick enough settling time.

Eventually, the choice made for K_{ω}^{T} was $K_{\omega}^{T} = \frac{1}{2}K_{\omega,max}^{T}$.

Task 13

For the purpose of simulating this part of the controller, the initial point of the robot was taken to be $I(x_0, y_0) \equiv (0, 0)$. The goal was set to $G(x_q, y_q) \equiv (1.0, 0.0)$.

Figures 25, 27, 29, 31 and 33 show the displacemental response of the robot for various values of K_{ω}^{T} inside the interval set by inequality 4. Figure 35 verifies that the upper limit for K_{ω}^{T} is indeed $\frac{2}{T_{sR}}$ by showing that the displacemental response of the robot cannot converge for $K_{\omega}^{T} > K_{\omega,max}^{T}$.

Here, one can see that the smaller the value of K_{ω}^{T} is, the larger the settling time, the lower the rise time and the smoother the response is. However, as the value of K_{ω}^{T} increases, the steady-state response begins to oscillate, with the amplitude of this oscillation proportional to the value of K_{ω}^{T} .

Figures 26, 28, 30, 32 and 34 focus on the steady-state value of the aforementioned responses. In contrast to the proportional rotational controller, the robot *can* arrive to its reference signal. The equations that govern the robot's translational movement are non-linear, as opposed to the one that governs its rotational movement, which in turn means that the translational system's behaviour is not bounded within the laws that govern control with proportional controllers on linear systems.

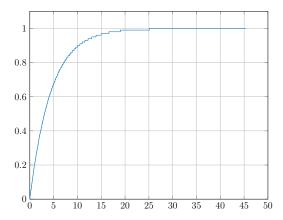


Figure 25: The orientation of the robot over time for $K_{\omega}^{T} = 0.1 K_{\omega,max}^{T}$

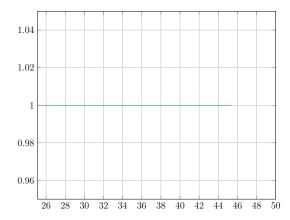


Figure 26: The steady state orientation of the robot for $K_{\omega}^T=0.1K_{\omega,max}^T$

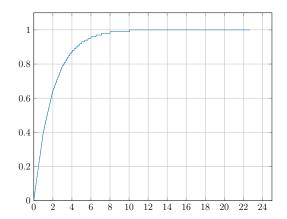


Figure 27: The orientation of the robot over time for $K_{\omega}^T=0.2K_{\omega,max}^T$

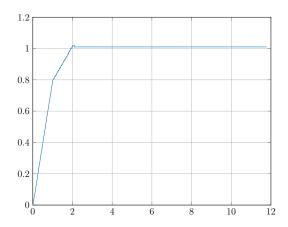


Figure 29: The orientation of the robot over time for $K_{\omega}^T=0.5K_{\omega,max}^T$

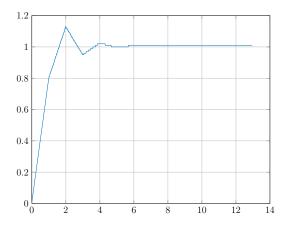


Figure 31: The orientation of the robot over time for $K_{\omega}^T=0.75K_{\omega,max}^T$

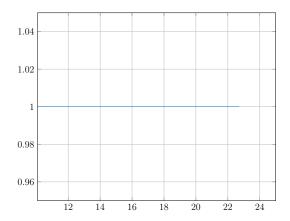


Figure 28: The steady state orientation of the robot for $K_{\omega}^T=0.2K_{\omega,max}^T$

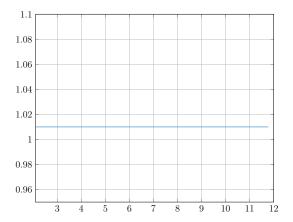


Figure 30: The steady state orientation of the robot for $K_{\omega}^T=0.5K_{\omega,max}^T$

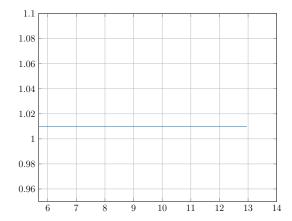


Figure 32: The steady state orientation of the robot for $K_{\omega}^T=0.75K_{\omega,max}^T$

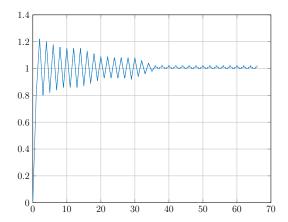


Figure 33: The orientation of the robot over time for $K_{\omega}^T = K_{\omega,max}^T$.

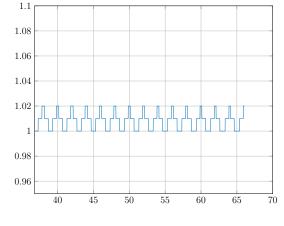


Figure 34: The steady state orientation of the robot for $K_{\omega}^{T}=K_{\omega,max}^{T}$

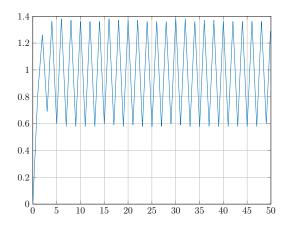


Figure 35: The orientation of the robot over time for $K_{\omega}^{T} = K_{\omega,max}^{T} + 1$. The system is marginally stable

Task 14

$$d_p[k+1] = p(\theta[k+1] - \theta_g) \tag{5}$$

But

$$\theta[k+1] = \theta[k] + T_s \frac{R}{L} u_{\Psi}^T[k]$$

Hence equation 5 becomes

$$\begin{split} d_{p}[k+1] &= p(\theta[k] + T_{s} \frac{R}{L} u_{\Psi}^{T}[k] - \theta_{g}) \\ &= p(\theta[k] + T_{s} \frac{R}{L} K_{\Psi}^{T} d_{p}[k] - \theta_{g}) \\ &= pT_{s} \frac{R}{L} K_{\Psi}^{T} d_{p}[k] + p(\theta[k] - \theta_{g}) \\ &= pT_{s} \frac{R}{L} K_{\Psi}^{T} d_{p}[k] + d_{p}[k] \\ &= d_{p}[k] (1 + pT_{s} \frac{R}{L} K_{\Psi}^{T} d_{p}[k]) \end{split}$$

In order for this system to be stable, that is, $d_p \to 0$ as $k \to \infty$, the solution of the equation inside the parentheses should lie inside the unit circle:

$$\left| 1 + pT_s R K_{\omega}^T \right| < 1$$

$$-1 < 1 + pT_s R K_{\omega}^T < 1$$

$$-2 < pT_s R K_{\omega}^T < 0$$

$$-\frac{2}{pT_s R} < K_{\omega}^T < 0$$
(6)

Hence, the maximum value K_{ω}^{T} can take for the system to be marginally stable is $K_{\omega,max}^{T}=0$.

Theoretically, the value of K_{ω}^{T} can be chosen to be any value inside the interval defined in inequality 3. However, first, it would be wise to choose a value that is far enough from the maximum value so as to avoid overshoot, but close enough to it, so that convergence happens in reasonable time. Hence, in practice, it is reasonable that one would need to experiment with different values and choose one that results in balancing a small angular error, a minimal overshoot, if any, and a quick enough settling time.

Eventually, the choice made for K_{ω}^{T} was $K_{\omega}^{T} = \frac{1}{2}K_{\omega,min}^{T} = -\frac{1}{pT_{s}R}$.

- Task 15
- Task 16
- Task 17
- Task 18
- Task 19
- Task 20
- Task 21
- Task 22

References

- [1] Hassan K Khalil. *Nonlinear systems*. Prentice Hall, Upper Saddle river, 3. edition, 2002. ISBN 0-13-067389-7.
- [2] Tobias Oetiker, Hubert Partl, Irene Hyna, and Elisabeth Schlegl. The Not So Short Introduction to $\not\!\! ET_E\!\! X \not\!\! 2_{\mathcal E}$. Oetiker, OETIKER+PARTNER AG, Aarweg 15, 4600 Olten, Switzerland, 2008. http://www.ctan.org/info/lshort/.
- [3] Shankar Sastry. Nonlinear systems: analysis, stability, and control, volume 10. Springer, New York, N.Y., 1999. ISBN 0-387-98513-1.