

# **SYDE252 - lecture notes**

09/01/18

Presented by: John Zelek  
Systems Design Engineering  
note: some material (figures) borrowed from various sources



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**FACULTY OF ENGINEERING**

# 4. Laplace

09/11/18

Presented by: John Zelek  
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# inspiration

- “You can certainly use the Laplace transform on this problem, but that would be akin to slicing your bread with a chainsaw: it’s overkill, it’s difficult to set up, and if you let your 5-year-old try it someone is going to end up calling 911. Well, maybe not that last one.” — Electrical engineering professor



# Laplace

The Laplace transform is a tool to map signal and system behaviours from the time domain into the frequency domain. We break  $x(t)$  into exponential components of the form  $e^{st}$ , where  $s$  is the complex frequency  $s = \alpha + j\omega$ . The Laplace transform applies to general signals and not just sinusoids (e.g., the Fourier only applies to sinusoids and is a subset of the Laplace).

The Laplace transform is:

$$X(\alpha, \omega) = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

and the inverse Laplace transform is

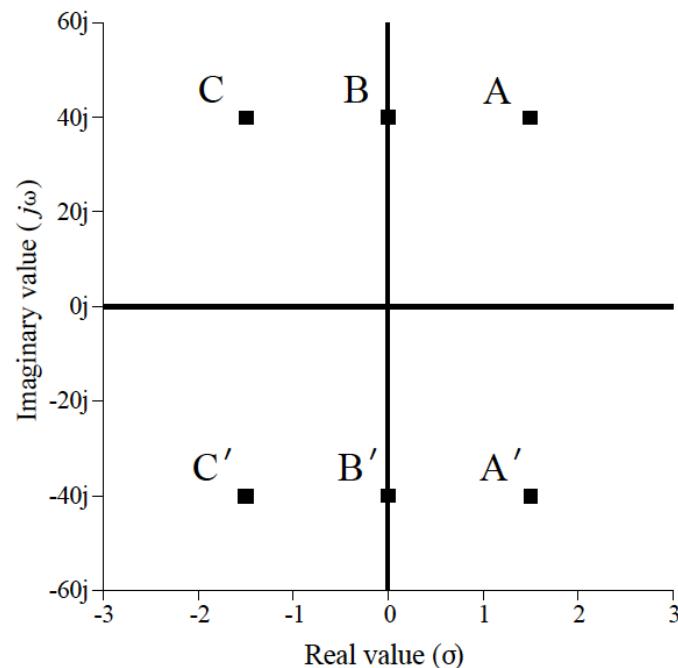
$$x(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} X(s)e^{st} ds$$

where  $c$  is a constant chosen to ensure convergence of the first integral. The path of integration along  $c + i\omega$  with  $\omega$  varying from  $-\infty$  to  $\infty$  must lie in the ROC (Region of Convergence).

Let us assume that all signals we will be dealing with are **causal**, meaning they are on-sided (or unilateral) Laplace transform.



## s-Domain



## Associated Waveforms

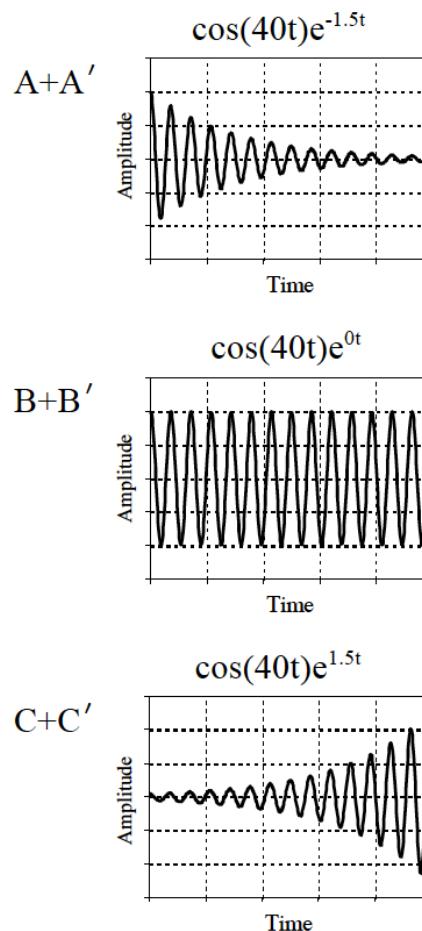


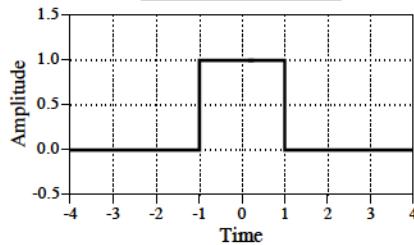
FIGURE 32-2

Waveforms associated with the s-domain. Each location in the s-domain is identified by two parameters:  $\sigma$  and  $\omega$ . These parameters also define two waveforms associated with each location. If we only consider pairs of points (such as: A&A', B&B', and C&C'), the two waveforms associated with each location are sine and cosine waves of frequency  $\omega$ , with an exponentially changing amplitude controlled by  $\sigma$ .

[dspguide.com](http://dspguide.com)



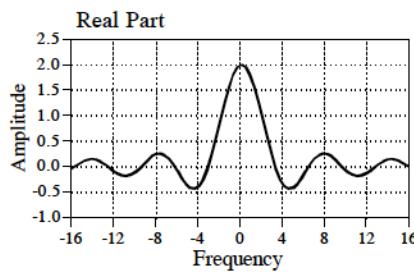
## Time Domain



*Fourier  
Transform*

*Laplace  
Transform*

## Frequency Domain



## s-Domain

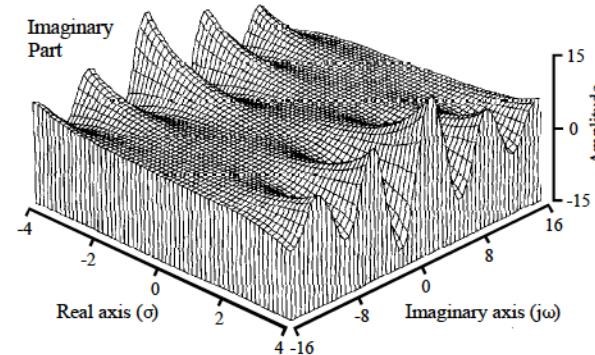
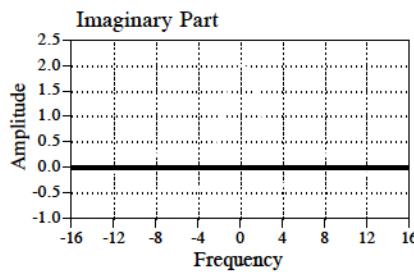
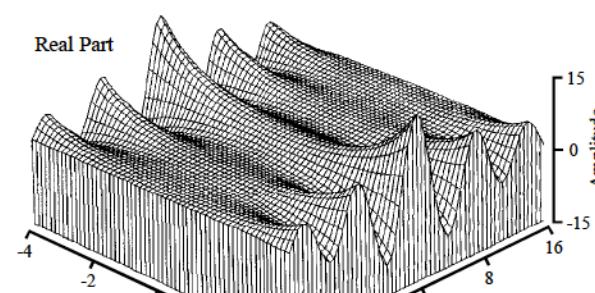


FIGURE 32-3

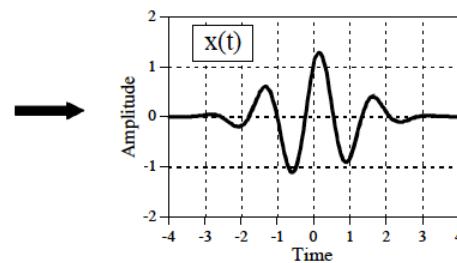
Time, frequency and s-domains. A time domain signal (the rectangular pulse) is transformed into the frequency domain using the Fourier transform, and into the s-domain using the Laplace transform.

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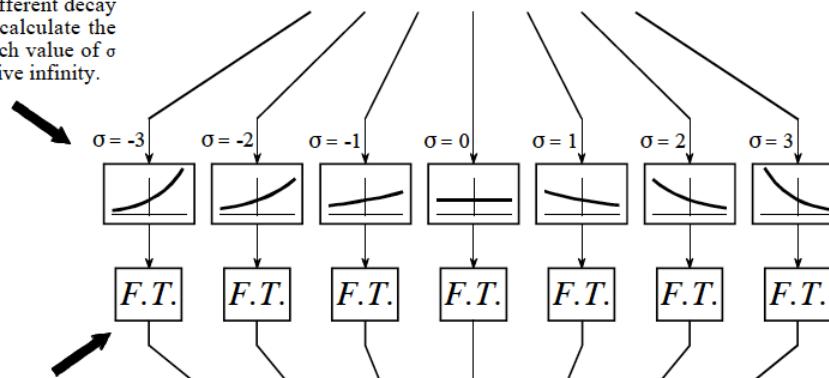


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STEP 1  
Start with the time domain signal called  $x(t)$



STEP 2  
Multiply the time domain signal by an infinite number of exponential curves, each with a different decay constant,  $\sigma$ . That is, calculate the signal:  $x(t) e^{-\sigma t}$  for each value of  $\sigma$  from negative to positive infinity.

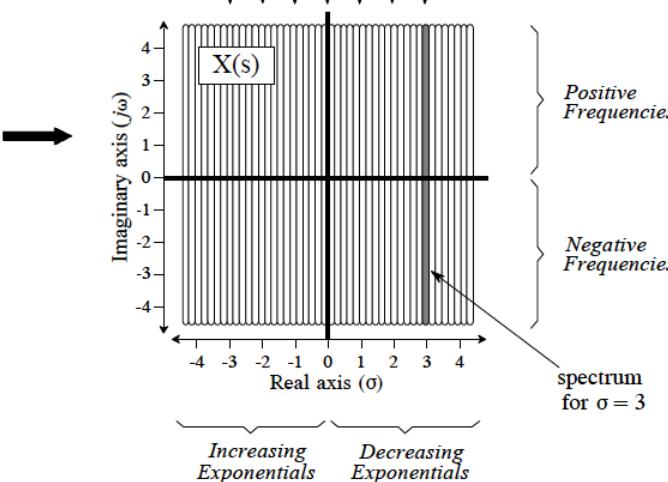


STEP 3  
Take the complex Fourier Transform of each exponentially weighted time domain signal. That is, calculate:

$$\int_{-\infty}^{\infty} [x(t) e^{-\sigma t}] e^{-j\omega t} dt$$

for each value of  $\sigma$  from negative to positive infinity.

STEP 4  
Arrange each spectrum along a vertical line in the s-plane. The positive frequencies are in the upper half of the s-plane while the negative frequencies are in the lower half.



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# Laplace transform

The Laplace transform is:

$$X(\alpha, \omega) = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

and the inverse Laplace transform is

$$x(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} X(s)e^{st} ds$$

where  $c$  is a constant chosen to ensure convergence of the first integral. The path of integration along  $c + j\omega$  with  $\omega$  varying from  $-\infty$  to  $\infty$  must lie in the ROC (Region of Convergence).

$$s = \alpha + j\omega$$

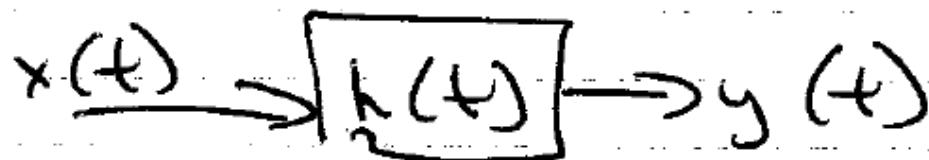
Let us assume that all signals we will be dealing with are **causal**, meaning they are on-sided (or unilateral) Laplace transform.



# Laplace transform

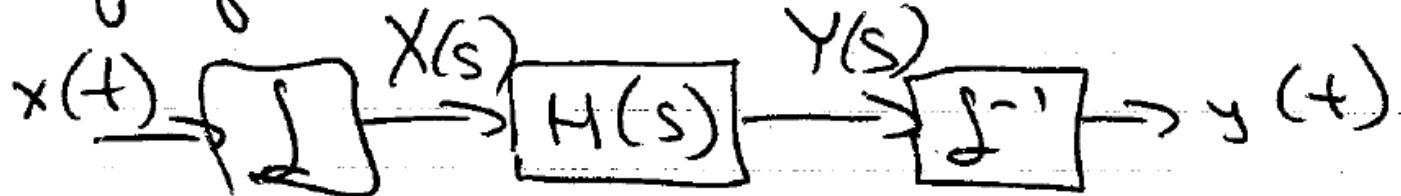
time

$$y(t) = x(t) * h(t)$$



freq  $\omega$

$$Y(s) = X(s)H(s)$$



# Laplace transform examples

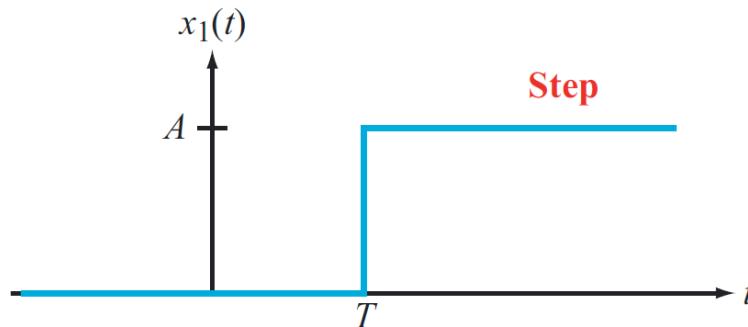
## Example 3-1: Laplace Transforms of Singularity Functions

Determine the Laplace transforms of the signal waveforms displayed in Fig. 3-1.

### Solution:

(a) The step function in Fig. 3-1(a) is given by

$$x_1(t) = A u(t - T).$$



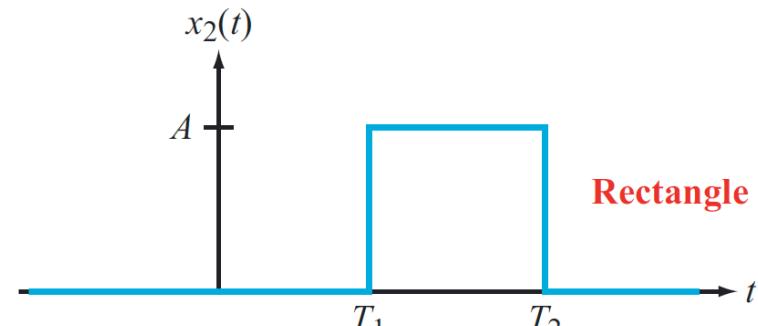
$$\begin{aligned} X_1(s) &= \int_{0^-}^{\infty} x_1(t) e^{-st} dt \\ &= \int_{0^-}^{\infty} A u(t - T) e^{-st} dt \\ &= A \int_T^{\infty} e^{-st} dt \\ &= -\frac{A}{s} e^{-st} \Big|_T^{\infty} = \frac{A}{s} e^{-sT}. \end{aligned}$$

For the special case where  $A = 1$  and  $T = 0$  (i.e., the step occurs at  $t = 0$ ), the transform pair becomes

$$u(t) \leftrightarrow \frac{1}{s}. \quad (3.6)$$



# Laplace transform examples



(b) The rectangle function in Fig. 3-1(b) can be constructed as the sum of two step functions:

$$x_2(t) = A[u(t - T_1) - u(t - T_2)],$$

and its Laplace transform is

$$\begin{aligned} \mathbf{X}_2(\mathbf{s}) &= \int_{0^-}^{\infty} A[u(t - T_1) - u(t - T_2)]e^{-st} dt \\ &= A \int_{0^-}^{\infty} u(t - T_1) e^{-st} dt - A \int_{0^-}^{\infty} u(t - T_2) e^{-st} dt \\ &= \frac{A}{s} [e^{-sT_1} - e^{-sT_2}]. \end{aligned}$$



# Laplace transform examples

$$(D^2 + D)y(t) = D \times (t)$$

$$(s^2 + s)Y(s) = sX(s)$$

(c) The impulse function in Fig. 3-1(c) is given by

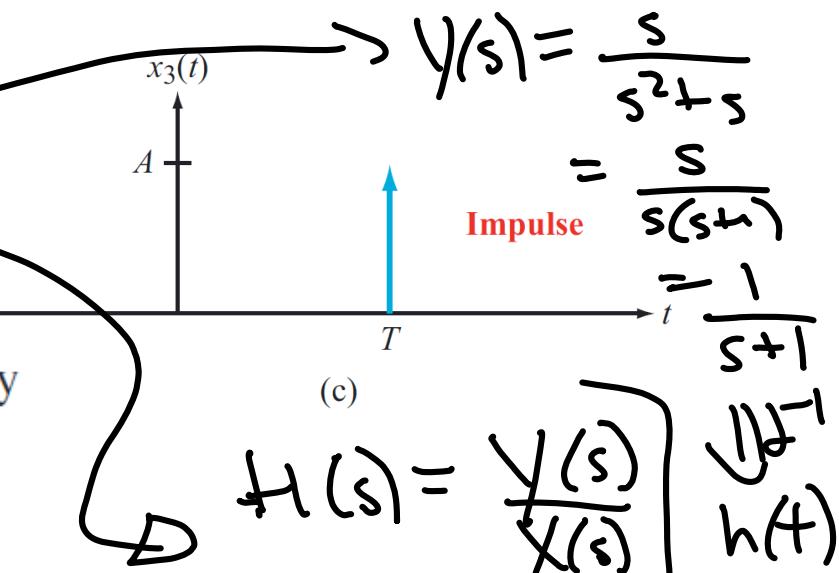
$$x_3(t) = A \delta(t - T),$$

and the corresponding Laplace transform is

$$X_3(s) = \int_{0^-}^{\infty} A \delta(t - T) e^{-st} dt = Ae^{-sT},$$

defined by Eq. (1.29). For the special case where  $A = 1$  and  $T = 0$ , the Laplace transform pair simplifies to

$\delta(t) \leftrightarrow 1.$



$$H(s) = \frac{Y(s)}{1}$$

$$h(t) = \mathcal{J}^{-1} H(s)$$



# Laplace transform examples

## Example 3-2: Laplace Transform Pairs

Obtain the Laplace transforms of (a)  $x_1(t) = e^{-at} u(t)$  and (b)  $x_2(t) = [\cos(\omega_0 t)] u(t)$ .

**Solution:**

(a) Application of Eq. (3.1) gives

$$\begin{aligned} X_1(s) &= \int_{0^-}^{\infty} e^{-at} u(t) e^{-st} dt \\ &= \frac{e^{-(s+a)t}}{-(s+a)} \Big|_0^{\infty} = \frac{1}{s+a}. \end{aligned}$$

Hence,

$$e^{-at} u(t) \leftrightarrow \frac{1}{s+a}. \quad (3.8)$$

(b) We start by expressing  $\cos(\omega_0 t)$  in the form

$$\cos(\omega_0 t) = \frac{1}{2}[e^{j\omega_0 t} + e^{-j\omega_0 t}].$$

Next, we take advantage of Eq. (3.8):

$$\begin{aligned} X_2(s) &= \mathcal{L}[\cos(\omega_0 t) u(t)] \\ &= \frac{1}{2} \mathcal{L}[e^{j\omega_0 t} u(t)] + \frac{1}{2} \mathcal{L}[e^{-j\omega_0 t} u(t)] \\ &= \frac{1}{2} \frac{1}{s - j\omega_0} + \frac{1}{2} \frac{1}{s + j\omega_0} \\ &= \frac{s}{s^2 + \omega_0^2}. \end{aligned}$$

Hence,

$$[\cos(\omega_0 t)] u(t) \leftrightarrow \frac{s}{s^2 + \omega_0^2}.$$



# Laplace transform examples (1)

$$x(t) = e^{-at}u(t) \quad , \text{a is real}$$
$$\mathcal{L}\{x(t)\} = X(s) = \int_{-\infty}^{\infty} e^{-at}u(t)e^{-st}dt = \int_{0^+}^{\infty} e^{-(s+a)t}dt$$
$$= -\frac{1}{s+a} \Big|_{0^+}^{\infty} = \frac{1}{s+a}, \text{Re}(s) > a$$

because  $\lim_{t \rightarrow \infty} e^{-(s+a)t} = 0$  only if  $\text{Re}(s+a) > 0$   
 $\text{or } \text{Re}(s) > -a$



# Laplace transform examples (2)

$$x(t) = -e^{-at}u(-t) \quad , \text{ a is real}$$

$$X(s) = \frac{1}{s+a} \quad , \text{ Re}(s) < -a$$



# Laplace transform examples (3+)

$$x(t) = \delta(t)$$

$$\mathcal{L}[\delta(t)] = 1 \quad \delta(t) \leftrightarrow 1$$

$$x(t) = u(t)$$

$$u(t) \leftrightarrow \frac{1}{s} \quad \text{Re } s > 0$$

$$x(t) = e^{at}u(t)$$

$$e^{at}u(t) \leftrightarrow \frac{1}{s-a}$$

$$x(t) = \cos \omega_0 t u(t)$$

$$\cos \omega_0 t u(t) \leftrightarrow \frac{s}{s^2 + \omega_0^2}$$



# Laplace transform pairs

Laplace transform pairs

$$x(t) \leftrightarrow X(s)$$

$$\delta(t) \Leftrightarrow 1$$

$$u(t) \Leftrightarrow \frac{1}{s}$$

$$tu(t) \Leftrightarrow \frac{1}{s^2}$$

$$t^n u(t) \Leftrightarrow \frac{n!}{s^{n+1}}$$

$$e^{\lambda t} u(t) \Leftrightarrow \frac{1}{s-\lambda}$$

$$t^n e^{\lambda t} u(t) \Leftrightarrow \frac{n!}{(s-\lambda)^{n+1}}$$

**Inverse Laplace** is best done by using the tables available.

You will need to use partial fraction expansion before you can use the tables. See ~~Appendix B for a review of partial fraction expansion.~~ If the order of the numerator is equal to order of the denominator then you will need to add coefficient of the highest power in the numerator for partial fraction expansion.



**Table 3-2:** Examples of Laplace transform pairs. Note that  $x(t) = 0$  for  $t < 0^-$  and  $T \geq 0$ .

Laplace Transform Pairs			
	$x(t)$	$X(s) = \mathcal{L}[x(t)]$	
1	$\delta(t)$	$\leftrightarrow$	1
1a	$\delta(t - T)$	$\leftrightarrow$	$e^{-Ts}$
2	$u(t)$	$\leftrightarrow$	$\frac{1}{s}$
2a	$u(t - T)$	$\leftrightarrow$	$\frac{e^{-Ts}}{s}$
3	$e^{-at} u(t)$	$\leftrightarrow$	$\frac{1}{s + a}$
3a	$e^{-a(t-T)} u(t - T)$	$\leftrightarrow$	$\frac{e^{-Ts}}{s + a}$
4	$t u(t)$	$\leftrightarrow$	$\frac{1}{s^2}$
4a	$(t - T) u(t - T)$	$\leftrightarrow$	$\frac{e^{-Ts}}{s^2}$
5	$t^2 u(t)$	$\leftrightarrow$	$\frac{2}{s^3}$
6	$te^{-at} u(t)$	$\leftrightarrow$	$\frac{1}{(s + a)^2}$
7	$t^2 e^{-at} u(t)$	$\leftrightarrow$	$\frac{2}{(s + a)^3}$
8	$t^{n-1} e^{-at} u(t)$	$\leftrightarrow$	$\frac{(n-1)!}{(s + a)^n}$
9	$\sin(\omega_0 t) u(t)$	$\leftrightarrow$	$\frac{\omega_0}{s^2 + \omega_0^2}$
10	$\sin(\omega_0 t + \theta) u(t)$	$\leftrightarrow$	$\frac{s \sin \theta + \omega_0 \cos \theta}{s^2 + \omega_0^2}$
11	$\cos(\omega_0 t) u(t)$	$\leftrightarrow$	$\frac{s}{s^2 + \omega_0^2}$
12	$\cos(\omega_0 t + \theta) u(t)$	$\leftrightarrow$	$\frac{s \cos \theta - \omega_0 \sin \theta}{s^2 + \omega_0^2}$
13	$e^{-at} \sin(\omega_0 t) u(t)$	$\leftrightarrow$	$\frac{\omega_0}{(s + a)^2 + \omega_0^2}$
14	$e^{-at} \cos(\omega_0 t) u(t)$	$\leftrightarrow$	$\frac{s + a}{(s + a)^2 + \omega_0^2}$
15	$2e^{-at} \cos(bt - \theta) u(t)$	$\leftrightarrow$	$\frac{e^{j\theta}}{s + a + jb} + \frac{e^{-j\theta}}{s + a - jb}$
15a	$e^{-at} \cos(bt - \theta) u(t)$	$\leftrightarrow$	$\frac{(s + a) \cos \theta + b \sin \theta}{(s + a)^2 + b^2}$
16	$\frac{2t^{n-1}}{(n-1)!} e^{-at} \cos(bt - \theta) u(t)$	$\leftrightarrow$	$\frac{e^{j\theta}}{(s + a + jb)^n} + \frac{e^{-j\theta}}{(s + a - jb)^n}$



# inverse Laplace (1,2)

$$X(s) = \frac{7}{s^2}$$

$$X(s) = \frac{2s - 1}{s^2 + 4}$$

$$\mathcal{L}^{-1}\left\{\frac{7}{s^2}\right\} = 7 \mathcal{L}^{-1}\left\{\frac{1}{s^2}\right\}$$

(entry 4)  $= \frac{7}{2!} t^2 u(t) = \frac{7}{2} t^2 u(t)$

$$\mathcal{L}^{-1}\left\{\frac{2s - 1}{s^2 + 4}\right\} = 2 \mathcal{L}^{-1}\left\{\frac{s}{s^2 + 4}\right\} - \mathcal{L}^{-1}\left\{\frac{1}{s^2 + 4}\right\}$$

Table entry 8a, 8b

$$= (2\cos 2t - \frac{1}{2}\sin 2t) u(t)$$



(6)

## inverse Laplace (3)

$$X(s) = \frac{7s - 6}{s^2 - s - 6}$$

$$\mathcal{L}^{-1} \left\{ \frac{7s - 6}{s^2 - s - 6} \right\}$$

$$X(s) = \frac{7s - 6}{(s+2)(s-3)} = \frac{k_1}{s+2} + \frac{k_2}{s-3}$$

use partial fraction expansion

$$k_1 = \frac{7s - 6}{(s+2)(s-3)} \Big|_{s=-2} = 4 \quad \therefore X(s)$$

$$= \frac{4}{s+2} - \frac{3}{s-3}$$

$$k_2 = \frac{7s - 6}{(s+2)(s-3)} \Big|_{s=3} = -3$$

$$\frac{4}{s+2} + \frac{3}{s-3} = 1 \quad \checkmark$$

check  $s=0$  both sides

$$\frac{7(0) - 6}{(0^2 - 0 - 6)} = 1 \quad \checkmark$$



# inverse Laplace (3) (ctn)

✓

$$X(s) = \frac{7s - 6}{s^2 - s - 6}$$

check

$$x(t) = 4e^{-2t}u(t) - 3e^{3t}u(t)$$



## inverse Laplace (4)

$$X(s) = \frac{2s^2 + 5}{(s+1)(s+2)}$$

$$\mathcal{L}^{-1} \left\{ \frac{2s^2 + 5}{(s+1)(s+2)} \right\}$$

$$\therefore X(s) = 2 + \frac{k_1}{s+1} + \frac{k_2}{s+2}$$

$$k_1 = \frac{2s^2 + 5}{(s+1)(s+2)} \Big|_{s=-1} = 7 \quad k_2 = \frac{2s^2 + 5}{(s+1)(s+2)} \Big|_{s=-2} = -13$$

$$\therefore X(s) = 2 + \frac{7}{s+1} - \frac{13}{s+2}$$

$N \geq D$

$$\boxed{\begin{aligned} \text{LHS} &\rightarrow \frac{0+s}{1(s)} = \frac{5}{2} = 2.5 \checkmark \\ \text{RHS} &\rightarrow 2 + 7 - 6.5 = 2.5 \checkmark \end{aligned}}$$

$$\text{use } \mathcal{S}(t) \leftrightarrow 1 \quad e^{xt} u(t) \leftrightarrow \frac{1}{s-x}$$

$$\therefore x(t) = 2 \mathcal{S}(t) + (7e^{-t} - 13e^{-2t}) u(t)$$



# Laplace properties

## *Time Shifting*

if  $x(t) \Leftrightarrow X(s)$   
then  $x(t - t_0) \Leftrightarrow X(s)e^{-st_0}$

## *Frequency Shifting*

if  $x(t) \Leftrightarrow X(s)$   
then  $x(t)e^{s_0 t} \Leftrightarrow X(s - s_0)$

## *Time Differentiation*

if  $x(t) \Leftrightarrow X(s)$   
then  $\frac{dx}{dt} \Leftrightarrow sX(s) - x(0^-)$   
and  $\frac{d^2x}{dt^2} \Leftrightarrow s^2X(s) - sx(0^-) - \dot{x}(0^-)$   
and  $\frac{d^n x}{dt^n} \Leftrightarrow s^n X(s) - \sum_{k=1}^n s^{n-k} x^{k-1}(0^-)$

## *Frequency Differentiation*

if  $x(t) \Leftrightarrow X(s)$   
then  $tx(t) \Leftrightarrow -\frac{d}{ds}X(s)$



# Laplace properties (2)

## *Time Integration*

if  $x(t) \Leftrightarrow X(s)$   
then  $\int_{0^-}^t x(\tau)d\tau \Leftrightarrow \frac{X(s)}{s}$

## *Scaling*

if  $x(t) \Leftrightarrow X(s)$   
then  $x(at) \Leftrightarrow \frac{1}{a}X(\frac{s}{a})$  for all  $a$

## *Time Convolution*

if  $x_1(t) \Leftrightarrow X_1(s)$  and  $x_2(t) \Leftrightarrow X_2(s)$   
then  $x_1(t) * x_2(t) \Leftrightarrow X_1(s)X_s(s)$

## *Frequency Convolution*

if  $x_1(t) \Leftrightarrow X_1(s)$  and  $x_2(t) \Leftrightarrow X_2(s)$   
then  $x_1(t)x_2(t) \Leftrightarrow \frac{1}{2\pi j}X_1(s) * X_s(s)$



# Laplace properties (3)

*Initial Value Theorem*

if  $x(t) \Leftrightarrow X(s)$

then  $\lim_{t \rightarrow 0} x(t) = x(0^+) = \lim_{s \rightarrow \infty} sX(s)$

*Final Value Theorem*

if  $x(t) \Leftrightarrow X(s)$

then  $\lim_{t \rightarrow \infty} x(t) = x(0^+) = \lim_{s \rightarrow 0} sX(s)$



# Laplace properties - examples (1) - time shift

$$x(t) = (t - 1)[u(t - 1) - u(t - 2)] + [u(t - 2) - u(t - 4)]$$



# Laplace properties - examples (2) - frequency shift

$$x(t) = e^{-at} \cos(bt)u(t)$$



# Laplace properties - examples (3) - time differentiation

$$x(t) = t[u(t) - u(t-2)] + (3-t)[u(t-2) - u(t-3)]$$



# Laplace properties - examples (3) - time differentiation

$$x(t) = t[u(t) - u(t-2)] + (3-t)[u(t-2) - u(t-3)]$$



# Laplace properties - examples (4) - time convolution

$$c(t) = e^{at}u(t) * e^{bt}u(t)$$



# Laplace properties - examples (5) - initial, final value

$$Y(s) = \frac{5(3s + 1)}{s(s + 3)(s + 2)}$$



# Laplace - solving differential equations - ZI +ZS

$$\frac{d^2y}{dt^2} + 9y(y) = e^{4t}u(t)$$

I.C.  $y(0) = 1$

$$\dot{y}(0) = 0$$



# Laplace - solving differential equations - ZI +ZS [2]

$$\frac{d^2y}{dt^2} + 9y(y) = e^{4t}u(t)$$

I.C.  $y(0) = 1$

$$\dot{y}(0) = 0$$



# Laplace - solving differential equations - ZI +ZS [3]

$$\frac{d^2y}{dt^2} + 9y(y) = e^{4t}u(t)$$

I.C.  $y(0) = 1$

$$\dot{y}(0) = 0$$



# Laplace - solving differential equations - ZI +ZS - example 2

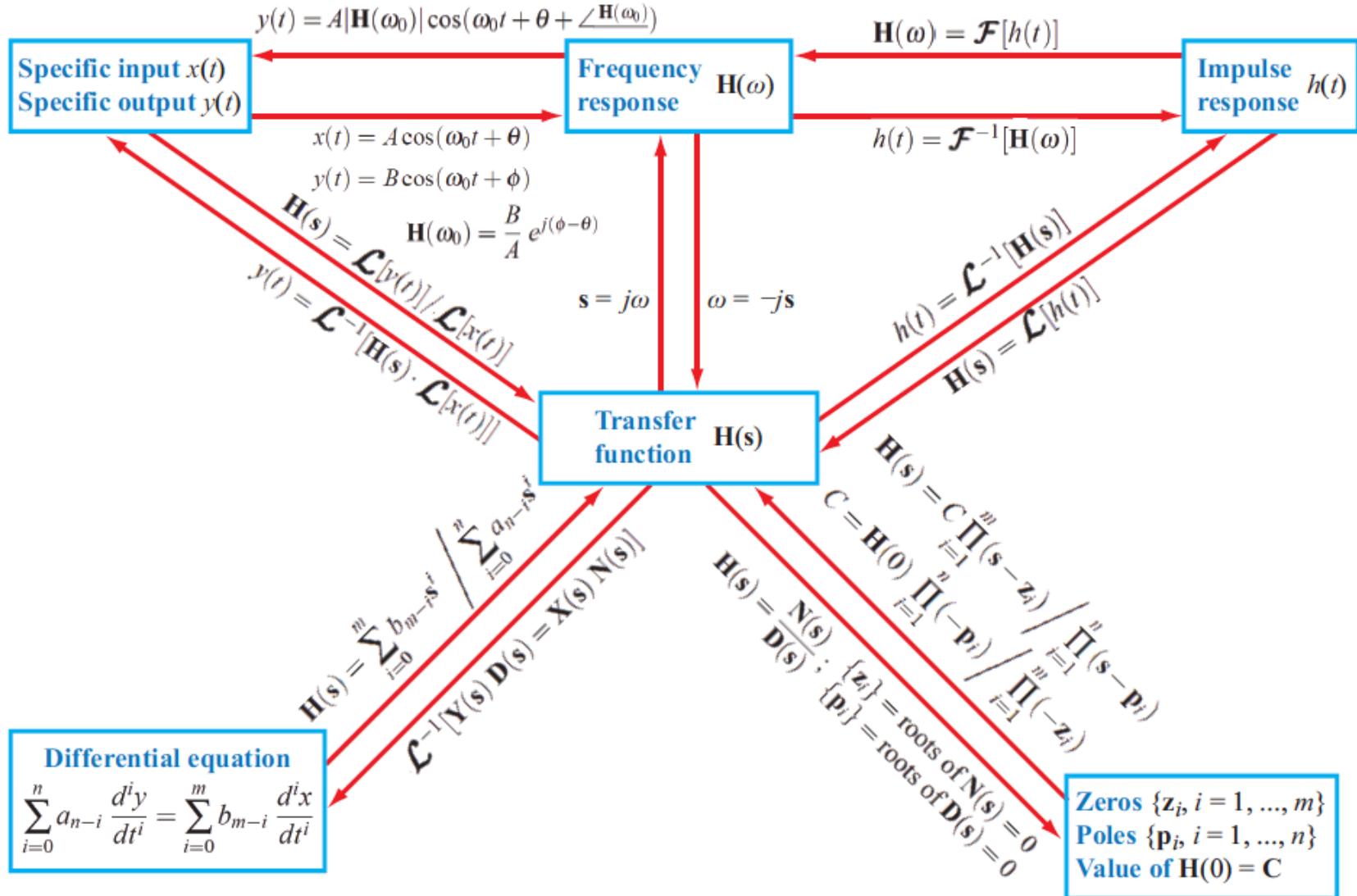
$$(D^2 + 5D + 6)y(t) = (D + 1)x(t)$$

I.C.     $y(0) = 2$      $\dot{y}(0) = 1$

$$x(t) = e^{-4t}u(t)$$



# System Description relationships





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