

Control Theory homework 3

Kamil Hayrullin
BS18-02

1 Introduction

My Innopolis mail is k.hayrullin@innopolis.ru

My variant is "e"

Link to my GitHub repository is [here](#)

2 Implement given tasks using python

(A) Design PD-controller that tracks time varying reference states i.e. $[x^*(t), x'^*(t)]$ as closely as possible.

Lets consider second order linear ODE

$$x'' + \mu x' + kx = u, \mu = 14, k = 85 \quad (1)$$

To create Proportional-derivative controller we should have P control:

$$u = k_p(x^* - x) \quad (2)$$

And add derivative term to it

$$u = k_p(x^* - x) + k_d(x'^* - x') \quad (3)$$

Where control error equals

$$e = x^* - x \quad (4)$$

Finally, PD controller:

$$u = k_p e + k_d e' \quad (5)$$

Link to python code in Google Colaboratory is [here](#)

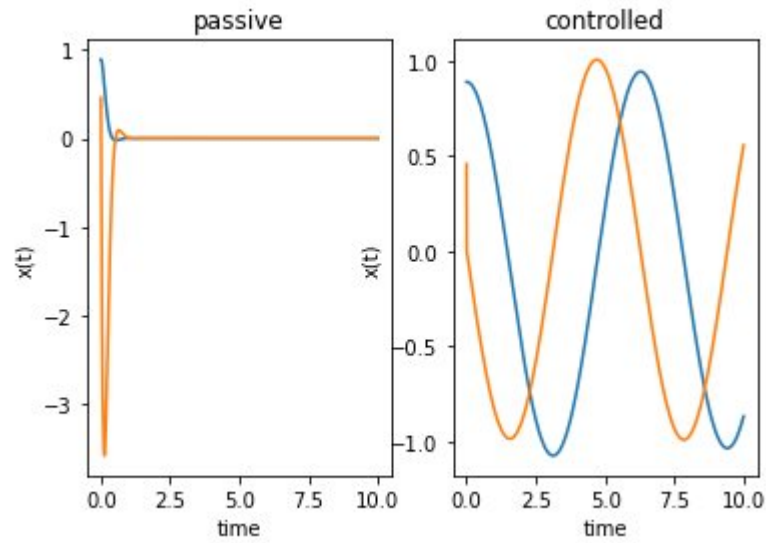


Figure 1: for COS function result is following, $k_p = 1000$, $k_d = 10000$

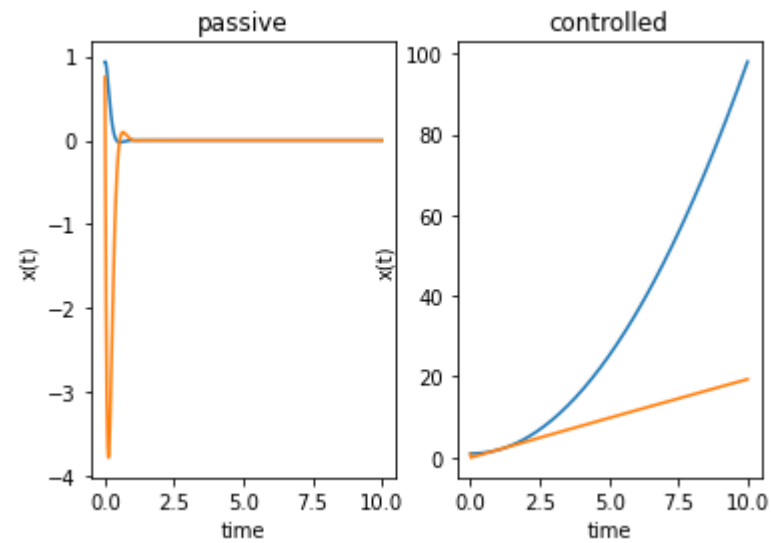


Figure 2: for t^2 function result is following, $k_p = 1000$, $k_d = 10000$

(B) Tune controller gains k_p and k_d . Find gains that provide no oscillations and no overshoot. Prove it with step input.

My choice of gains is:

$$k_p = 1000, k_d = 10000 \quad (6)$$

Link to python code in Google Colaboratory is [here](#)

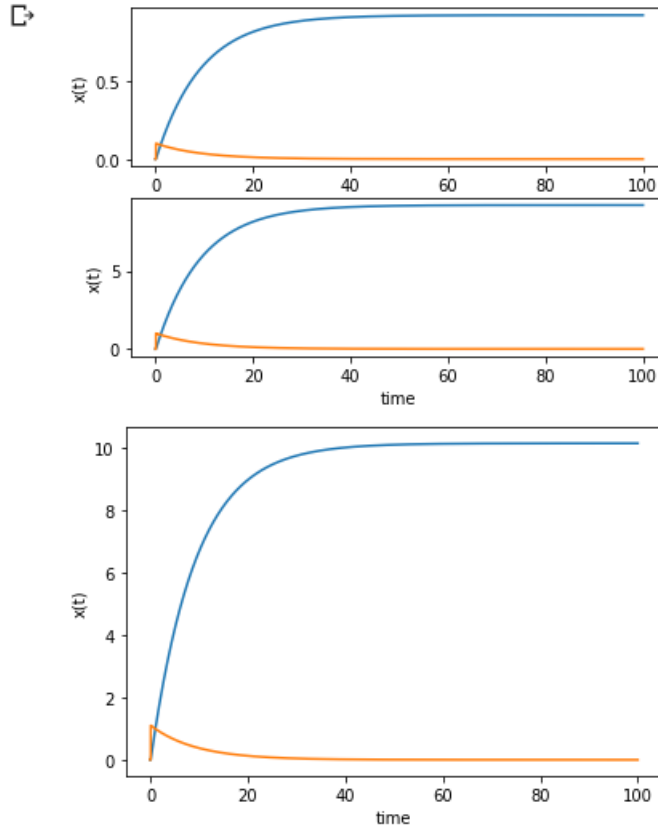


Figure 3: As we can see, no oscillations and no overshoot

(C) Prove that controlled oscillator dynamics is stable for your choice of k_p , k_d

Lets consider second order linear ODE

$$x'' + 14x' + 85x = 1000 * (x^* - x) + 10000 * (x'^* - x') \quad (7)$$

$$x'' = -14x' - 85x + 1000 * x^* - 1000 * x + 10000 * x'^* - 10000 * x' \quad (8)$$

$$\begin{pmatrix} x'' \\ x' \end{pmatrix} = \begin{pmatrix} -14 - 10000 & -85 - 1000 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} x' \\ x \end{pmatrix} + \begin{pmatrix} 10000 & 1000 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} x'^* \\ x^* \end{pmatrix} \quad (9)$$

Now let's find eigenvalues of matrix A

$$\det(A - \lambda I) = (-10014 - \lambda)(-\lambda) + 1085 = \lambda^2 + 10014\lambda + 1085 = 0 \quad (10)$$

$$\lambda_1, \lambda_2 = -0.10834, -10013.89 \quad (11)$$

Both eigenvalues less than zero, system is **stable**

(D) Think of how you would implement PD control for a linear system:

$$\begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} = \begin{pmatrix} 10 & 3 \\ 5 & -5 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} \quad (12)$$

,

To implement PD controller I would add control law

$$u = k_p e + k_d e' \quad (13)$$

$$\begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} = \begin{pmatrix} 10 & 3 \\ 5 & -5 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} 1 \\ 0 \end{pmatrix} (k_p e + k_d e') \quad (14)$$

I would translate this state space form into Linear Time Varying system and apply 1st task's code to this LTV.

(E) Implement a PI/PID controller for the system: $\ddot{x} + 14\dot{x} + 85x + 9.8 = u$. Test your controller on different trajectories, at least two.

Link to python code in Google Colaboratory is [here](#)

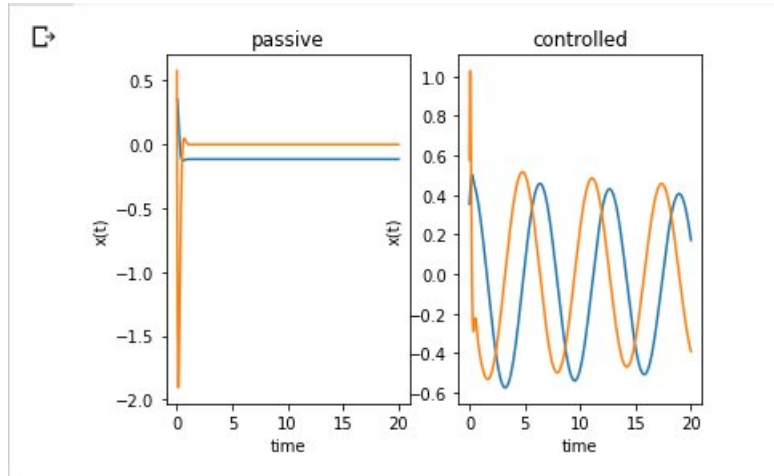


Figure 4: for COS function result is following, $k_p = 2$, $k_d = 100$

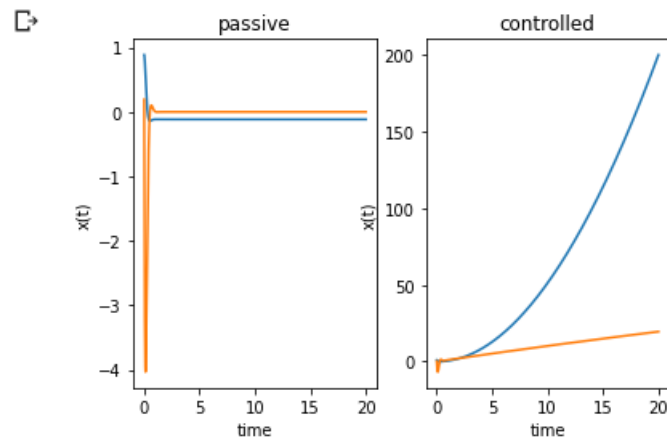


Figure 5: for t^2 function result is following, $k_p = 2$, $k_d = 100$

- 3 **Design a PID controller. Use step input function and try to improve rise time, overshoot and steady-state error, comparing with no controller system**

$$W(s) = \frac{s + 2}{2s^2 + 7} \quad (15)$$

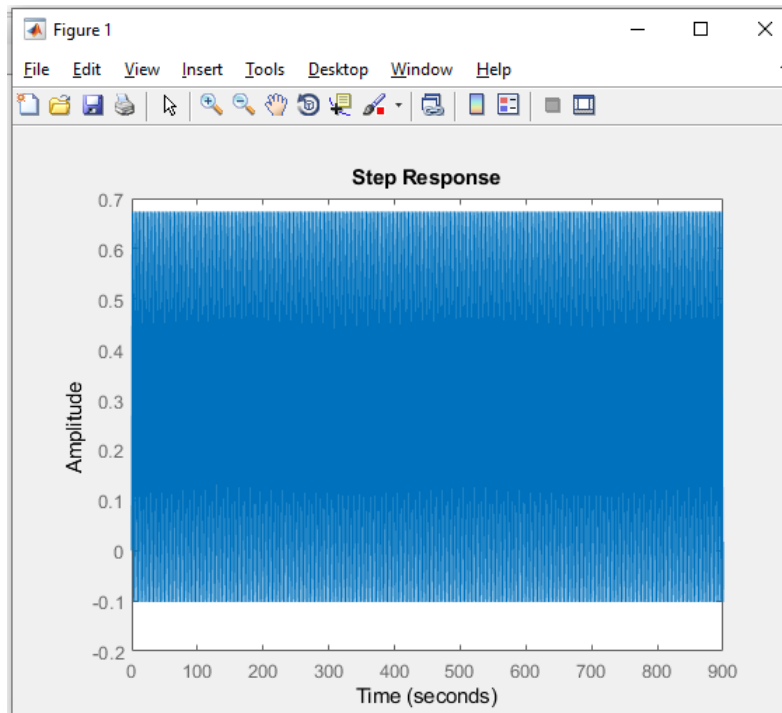
Let's write such code in matlab

```

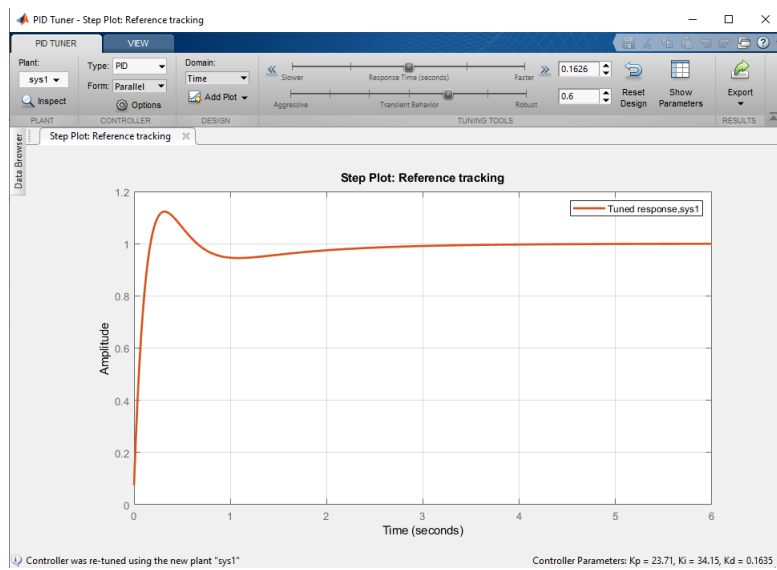
numerator = [1, 2];
denominator = [2, 0, 7];
sys1 = tf(numerator, denominator)
step(sys1)

```

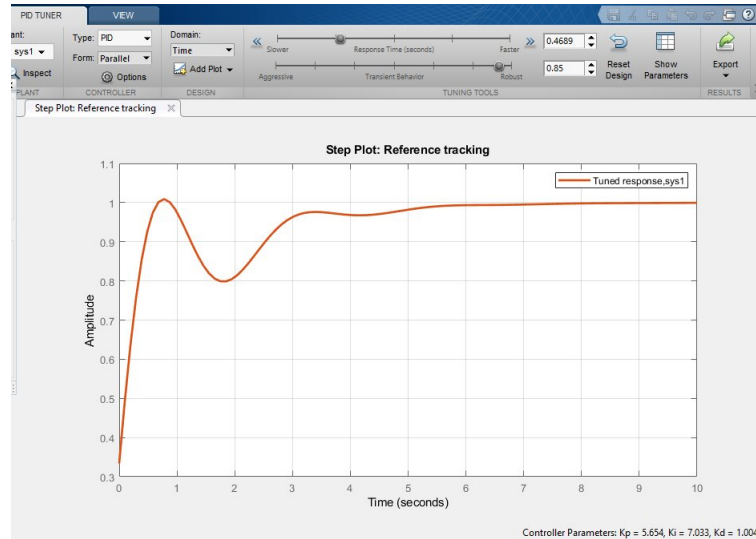
non controlled step function



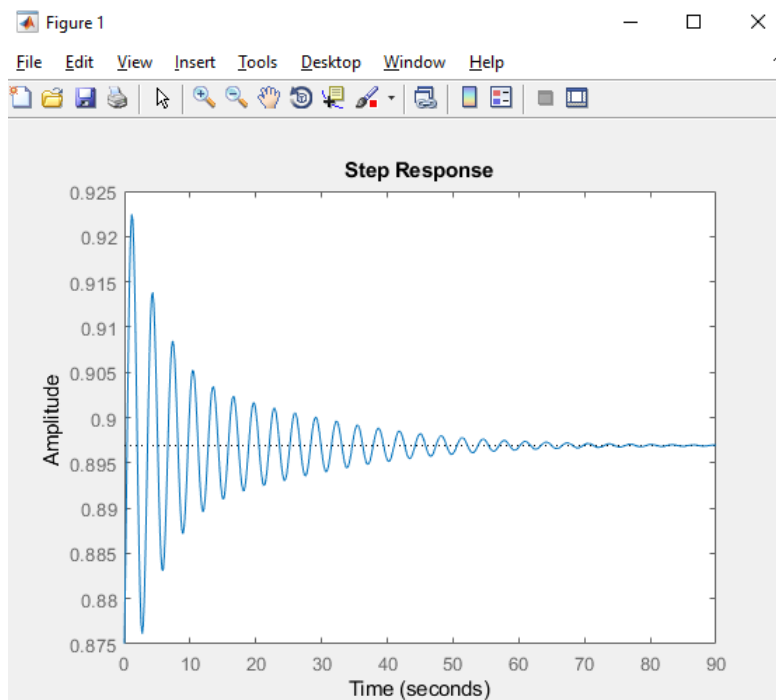
In PID Tuner it looks like that



Now we tune it a bit



and get this step function



Was:

Controller Parameters	
	Tuned
Kp	4.2362
Ki	5.9182
Kd	0.59223
Tf	n/a
Performance and Robustness	
	Tuned
Rise time	0.486 seconds
Settling time	5.15 seconds
Overshoot	2.71 %
Peak	1.03
Gain margin	Inf dB @ NaN rad/s
Phase margin	64.6 deg @ 3.46 rad/s
Closed-loop stability	Stable

Now:

Controller Parameters	
	Tuned
Kp	5.6536
Ki	7.0333
Kd	1.004
Tf	n/a
Performance and Robustness	
	Tuned
Rise time	0.458 seconds
Settling time	5.23 seconds
Overshoot	0.894 %
Peak	1.01
Gain margin	Inf dB @ NaN rad/s
Phase margin	89.9 deg @ 4.27 rad/s
Closed-loop stability	Stable

Results became better.

- 4 Design a lag or lead compensator (if applicable), play with zero and pole to find optimal values (of overshoot, peak time, transient process time, stationary error, etc.) for transient process.

$$W(s) = \frac{s + 4}{s^2 + 3s + 15} \quad (16)$$

Lag or lead compensator looks like that:

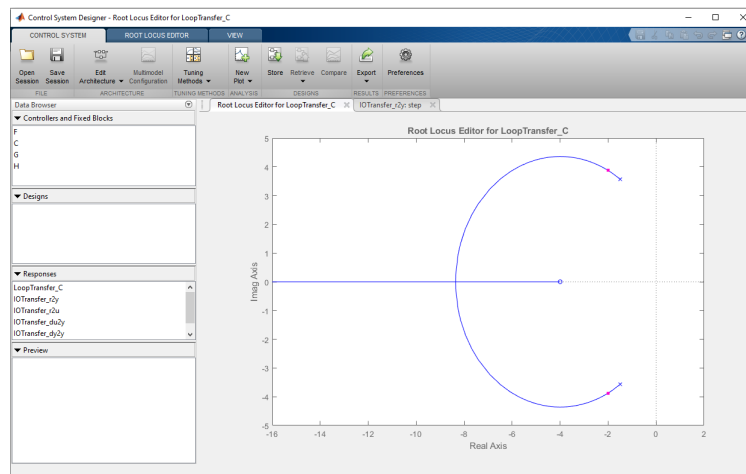
$$C(s) = K_c \frac{(s - z_0)}{(s - p_0)}$$

Figure 6: first-order lead compensator $C(s)$

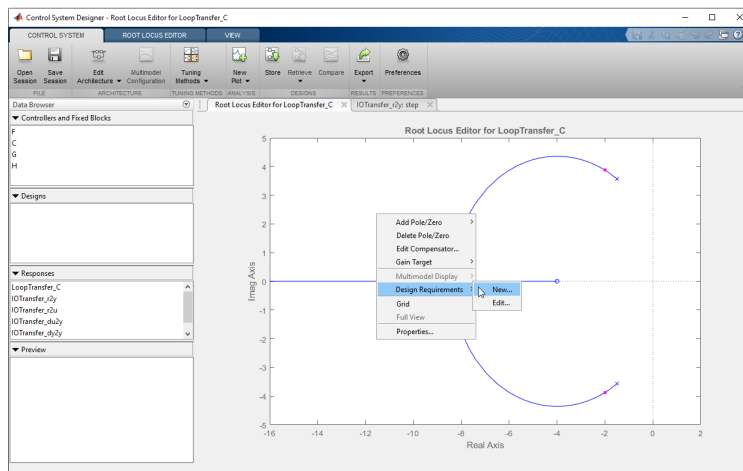
Let's run such program in matlab:

```
numerator = [1, 4];
denominator = [1, 3, 15];
sys = tf(numerator, denominator)
rltool(sys)
```

We can see, that Root Locus of our transfer function in Control System Designer.



Let's add some design requirements



New Design Re...

Design requirement type: Damping ratio

Design requirement parameters

Damping ratio > 0.6000

Ok Close Help

New Design Requirement

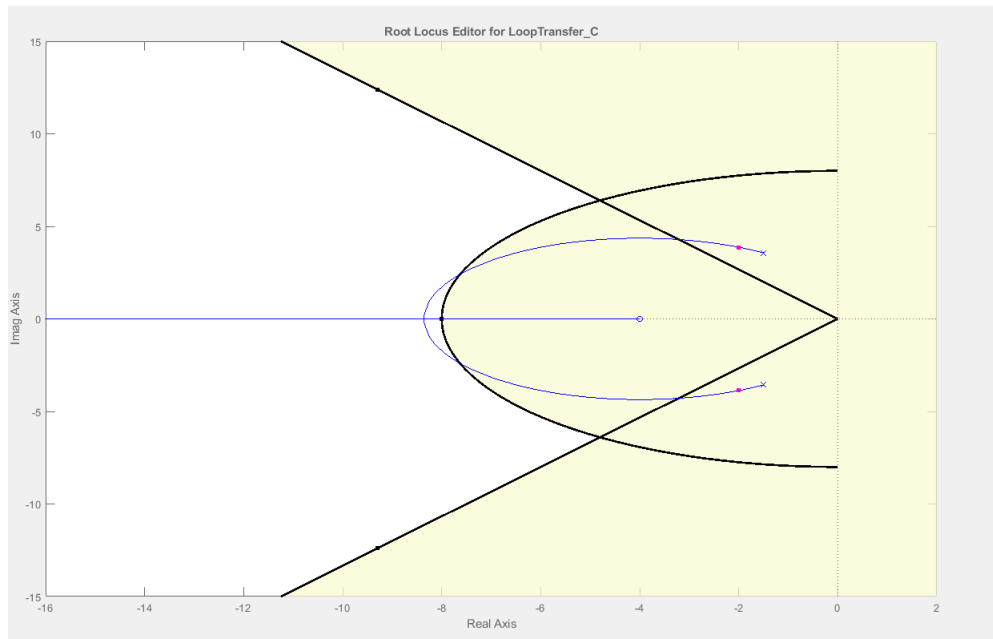
Design requirement type: Natural frequency

Design requirement parameters

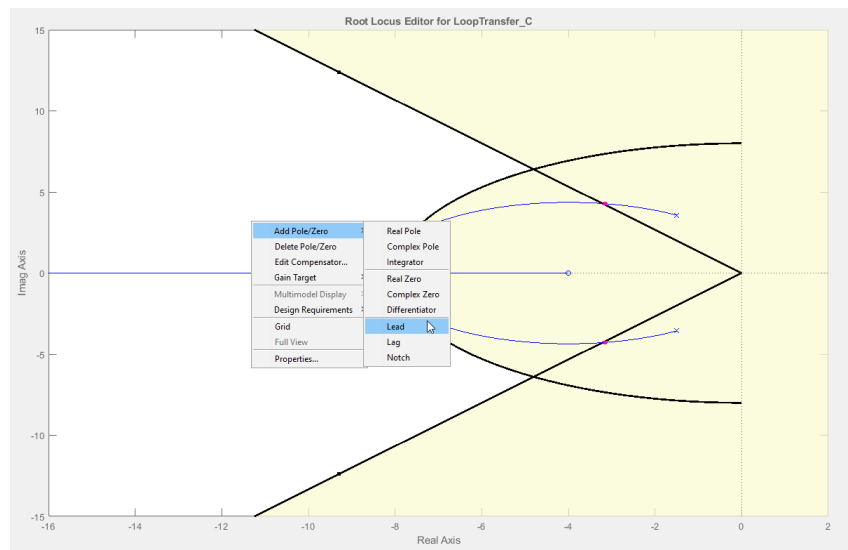
Natural frequency at least 8 rad/s

Ok Close Help

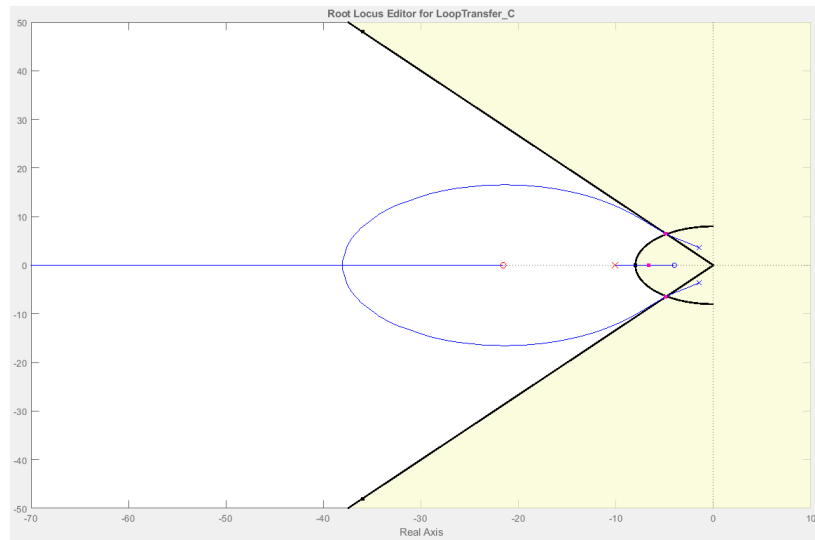
We have:



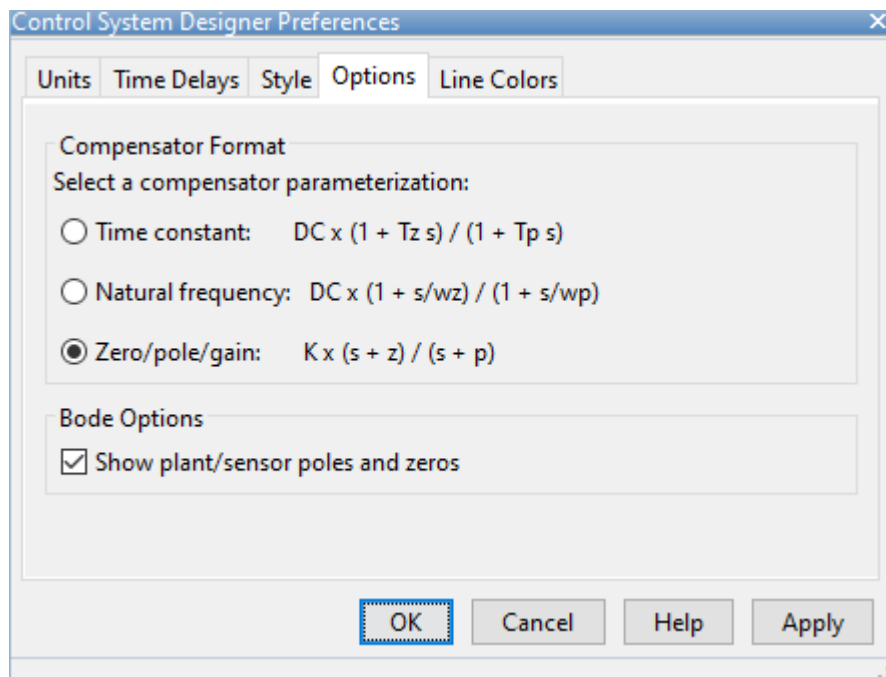
Our goal is to place red squares in intersection of damping ratio and natural frequency graph.
To do this let's add Lead



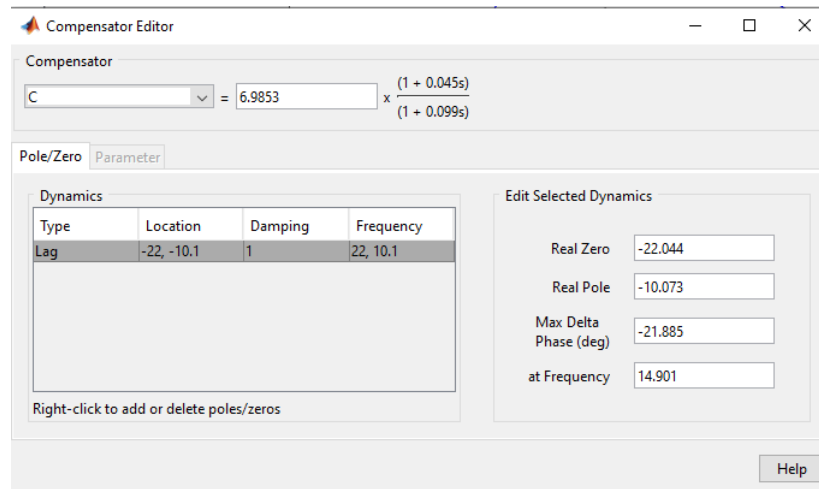
Now we move new zero point



In preferences let's change compensator format to Zero/Pole/Gain



Now in C controller we have our solution.



Here overshoot is 29.7
Peak amplitude is 0.844 at time 0.339
transient process time 0.906
After some manipulations with poles and zeros and setting Real zero = -3.04,
Real Pole = -1.07 overshoot is 7.36
Peak amplitude is 0.69 at time 0.52
transient process time 1.61