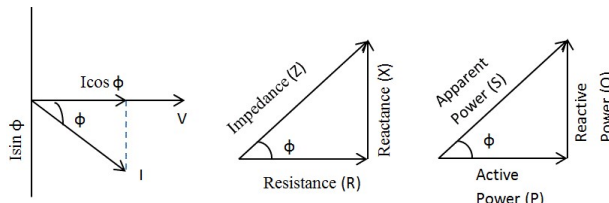# General Equations

- KCL: $\sum I_{in} = \sum I_{Out} \rightarrow$ Node's Input Current = Node's Output Current
- KVL: $\sum V = 0 \rightarrow$ Voltage across a loop totals to 0.
- Ohm's Law: $V = IR$

Phasors

RMS/Complex Power/Max Power Transfer

- $X_{rms} = \sqrt{\frac{1}{T}\int_0^T x(t)^2 dt}$
- $P_{avg} = \frac{1}{2}\operatorname{Re}\{\mathbf{VI}^*\} = \frac{1}{2}V_m I_m \cos(\theta_v - \theta_i)$
- $\mathbf{S} = I_{rms}^2 \mathbf{Z} = \frac{V_{rms}^2}{\mathbf{Z}^*} = \mathbf{V}_{rms}\mathbf{I}_{rms}^*$
- $\sum_{k=1}^n S_k$
- $C = \frac{Q_C}{\omega V_{rms}^2} = \frac{P(\tan\theta_1 - \tan\theta_2)}{\omega V_{rms}^2}$
- $L = \frac{V_{rms}^2}{\omega(Q_1 - Q_2)}$

Name	Symbol	Equation(s)	Units
Complex Power	$\mathbf{S}$	$\frac{P}{Pf}\angle \arccos(Pf) = P + jQ = \mathbf{V}_{rms}\mathbf{I}_{rms}^* = |\mathbf{V}_{rms}||\mathbf{I}_{rms}|\angle(\theta_v - \theta_i)$	VA
Apparent Power	S	$\|\mathbf{S}\| = |\mathbf{V}_{rms}||\mathbf{I}_{rms}| = \sqrt{P^2 + Q^2}$	VA
Real Power	P	$\operatorname{Re}\{\mathbf{S}\} = S * Pf \cos[\arccos(Pf)] = S\cos(\theta_v - \theta_i)$	W
Reactive (Imaginary) Power	Q	$\operatorname{Im}\{\mathbf{S}\} = S * Pf \sin[\arccos(Pf)] = S\sin(\theta_v - \theta_i)$	VAR
Power Factor	Pf	$\frac{P}{S} = \cos(\theta_v - \theta_i)$	Lead/Lag

Elements

Methods to Solve Equations

Nodal Analysis

1. # of Nodes? $\rightarrow n$
2. Make one node the reference node. Assign $n - 1$ nodal voltages
3. For a **voltage** source, write a CONSTRAINT EQUATION (Con. Eq.). If there is a voltage source between 2 non-reference nodes, make that a **SUPERNODE**.
4. Write KCL at each node. $(n - 1)$ equations.
5. Solve Equations.

Mesh/Loop Analysis

1. # of Nodes? $\rightarrow n$ # of Branches? $\rightarrow b$ # of meshes/loops? $\rightarrow b - n + 1 = l$
2. Assign l loop currents.
3. For **current** sources, write a CONSTRAINT EQUATION (Con. Eq.). If there is a current source between 2 meshes, that's a **SUPERMESH**.
4. Write KVL for each mesh.
5. Solve Equations.

Superposition

- # of sources, n, determines the number of equations you will have.
- Shut off each source, one at a time, solving for the term that you want.
  - Voltage Source = S.C.
  - Current Source = O.C.
- Sum each of the individual terms together. $\sum_{i=1}^n x_i$

- **THIS IS THE ONLY WAY TO SOLVE FOR A CIRCUIT WITH MULTIPLE SOURCES!!**
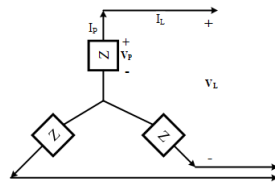
(a) 3 Phase Y
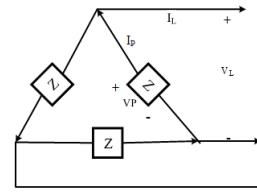
(b) 3 Phase Delta

$$I_L = I_P \qquad V_L = \sqrt{3}V_P\angle 30° \qquad Z_Y = \frac{Z_\Delta}{3}$$

$$\mathbf{S} = \sqrt{3}\mathbf{V}_L\mathbf{I}_L \qquad \mathbf{S}_P = 3\mathbf{V}_P\mathbf{I}_P^* \qquad \phi = \theta_{V_P} - \theta_{I_P}$$

$$\mathbf{S} = \|\mathbf{S}\|\angle\phi \qquad P = \|\mathbf{S}\|\cos(\phi) \qquad Q = \|\mathbf{S}\|\sin(\phi)$$

$$I_L = \sqrt{3}I_P\angle -30° \qquad V_L = V_P \qquad Z_\Delta = 3Z_Y$$

$$\|\mathbf{S}\| = \sqrt{3}V_L I_L \qquad \|\mathbf{S}\| = 3V_P I_P \qquad \phi = \theta_{V_P} - \theta_{I_P}$$

$$\mathbf{S} = \|\mathbf{S}\|\angle\phi \qquad P = \mathbf{S}\cos(\phi) \qquad Q = \mathbf{S}\sin(\phi)$$

Source Transformations

ALL source transformations obey Ohm's Law. $V = IR$. This will **ONLY** work on impedances in series with **VOLTAGE** sources, or impedances in parallel with **CURRENT** sources.

Thevenin and Norton Equivalencies

- ONLY independent sources - Zero all sources, find $\mathbf{Z}_{eq}$.
    - 0-ing Current Sources = O.C., 0-ing Voltage Sources = S.C.
    - Look at circuit from load's perspective for $\mathbf{Z}_{eq}$
    - $\mathbf{V}_{Th} = \mathbf{V}_{OC}$, $\mathbf{I}_N = \mathbf{I}_{SC}$
- BOTH dependent and independent sources
    - Find $\mathbf{V}_{Th} = \mathbf{V}_{OC}$, $\mathbf{I}_N = \mathbf{I}_{SC}$
    - Solve $\mathbf{Z}_{Th} = \frac{\mathbf{V}_{OC}}{\mathbf{I}_{SC}}$
- ONLY dependent sources
    - $\mathbf{V}_{Th} = 0$, $\mathbf{I}_N = 0$
    - $\mathbf{Z}_{Th} = \mathbf{Z}_N \rightarrow$ Attach test source @ load.
        * If voltage test source, find current. If current test source, find voltage
    - $\mathbf{Z}_{Th} = \frac{\mathbf{V}_{Test}}{\mathbf{I}_{Test}}$

Maximum Power Transfer - AC

- $\mathbf{Z}_{Load} = \mathbf{Z}_{Th}^*$, $R_{Th} = \text{Re}\{\mathbf{Z_{Th}}\}$, $R_L = |\mathbf{Z}_{Th}| = \sqrt{R_{Th}^2 + (X_{Th} + X_L)^2}$
- $P_{max} = \frac{|\mathbf{V}_{Th}|^2}{8R_{Th}}$

3-Phase Circuits

You want to get everything into Y formation, because the common neutral allows you to do single-phase analysis.

Mutual Inductance

Equivalent Mutual Inductance

|		
Series-**Aiding** Connection	$L = L_1 + L_2 + 2M$	
Series-**Opposing** Connection	$L = L_1 + L_2 - 2M$	

Dot Convention

There are 2 cases:
1. Current enters through dotted side on 1 inductor $\longrightarrow$ **POSITIVE VOLTAGE ON DOTTED SIDE OF OTHER INDUCTOR**

- Current flows into the dotted side of one inductor
- Current flows out of the un-dotted side of second inductor, just like the first

2. Current enters through **NON**-dotted side of 1 inductor $\longrightarrow$ **POSITIVE VOLTAGE ON UN-DOTTED SIDE OF OTHER INDUCTOR**
   - Current flows into the un-dotted side of one inductor
   - Current flows out of the dotted side of the second inductor, just like the first

Solving Disjoint Coupled Circuits

1. Apply KVL
2. Don't forget about the Mutual Inductance Voltage Difference because of the first current
3. There is a second way to thing about these, shown in Figures 2a and 2b, below.
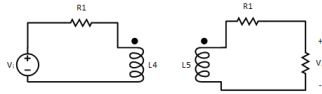
(a) Original Circuit
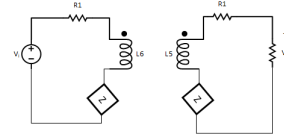
(b) Circuit "Simplified" by Adding Dependent Sources

The sign on the dependent sources depends on which side of the inductor the current is going into. Use the Dot Convention to determine which direction the source's voltage should go.

Transformers

These elements consume no power, and convert voltages and currents.
- $\frac{v_1}{v_2} = \frac{N_1}{N_2} \longleftarrow$ Voltage Change
- $\frac{i_1}{i_2} = -\frac{N_2}{N_1} \longleftarrow$ Current Change

Representations for Turns

There are 3 common ways to represent the number of turns in a transformer:

1. $N_1 : N_2$
   - Both N_1 and N_2 are integers
2. $1 : n$
   - The first term might not be 1, if there isn't perfect division, i.e. $2 : 5$ will not be reduced to $1 : \frac{5}{2}$
   - $n = \frac{N_2}{N_1}$
   - This is the form generally used by our textbook

Reflecting Elements

There are only 3 equations:

1. $\frac{v_1}{v_2} = \frac{N_1}{N_2} \longleftarrow$ Voltage Change
2. $\frac{i_1}{i_2} = -\frac{N_2}{N_1} \longleftarrow$ Current Change
3. $Z_1 = \frac{Z_2}{n^2}$, as seen in Figure 3
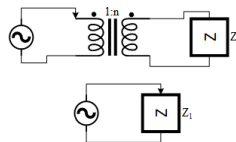4. A negative can be in any one of these, depending on the dot orientation

Figure 3: Transformer Reflecting Elements

Transfer Functions/Bode Plots

- Basic form of a Transfer function is $H(\omega) = \frac{X_{Out}}{X_{In}}$
  - $H(\omega) = \frac{V_{Out}}{V_{In}}$
  - $H(\omega) = \frac{I_{Out}}{I_{In}}$
  - $H(\omega) = \frac{V_{Out}}{I_{In}}$
  - $H(\omega) = \frac{I_{Out}}{V_{In}}$

We replace $s = \omega j \to \omega = \frac{s}{j}$.

When we have the transfer function, and plug in the equivalency $s = \omega j$, we end up with something like:

$$H(s) = \frac{k(s + z_1)(s + z_2)(s + z_3)\cdots}{(s + p_1)(s + p_2)(s + p_3)\cdots}$$

Now to make the bode plot:

$$\|H(s)\| = \frac{k\|s + z_1\|\|s + z_2\|\|s + z_3\|\cdots}{\|s + p_1\|\|s + p_2\|\|s + p_3\|\cdots}$$

$$\|H(\omega)\|(dB) = 20\log(k) + 20\log(j\omega + z_1) + 20\log(j\omega + z_2) + 20\log(j\omega + z_3)$$
$$- 20\log(j\omega + p_1) - 20\log(j\omega + p_2) - 20\log(j\omega + p_3)$$

$$\angle\varphi = \arctan(k) + \arctan(j\omega + z_1) + \arctan(j\omega + z_2) + \arctan(j\omega + z_3)$$
$$- \arctan(j\omega + p_1) - \arctan(j\omega + p_2) - \arctan(j\omega + p_3)$$

Factor	Magnitude $\|H(\omega)\|(dB)$	Phase $\angle\varphi$
K	$20\log_{10} K$	$0°$
$(j\omega)^N$	$20N$dB/decade (Passes through 1 and continues)	$90N°$
$\frac{1}{(j\omega)^N}$	$-20N$dB/decade (Passes through 1 and continues)	$-90N°$
$\left(1 + \frac{j\omega}{z}\right)^N$	$\begin{cases} 0 & x \le z \\ 20N\text{dB/decade} \end{cases}$	$\begin{cases} 0 & < \frac{z}{10} \\ \frac{1}{2}(90N)° & = z \\ 90N° & \ge 10z \end{cases}$
$\frac{1}{(1 + j\omega/p)^N}$	$\begin{cases} 0 & x \le z \\ -20N\text{dB/decade} \end{cases}$	$\begin{cases} 0 & < \frac{z}{10} \\ \frac{1}{2}(-90N)° & = z \\ -90N° & \ge 10z \end{cases}$
$\left[1 + \frac{2j\omega\zeta}{\omega_n} + \left(\frac{j\omega}{\omega_n}\right)^2\right]^N$	$\begin{cases} 0 & \le \omega_n \\ 40N\text{dB/decade} & > \omega_n \end{cases}$	$\begin{cases} 0 & \le \frac{\omega_n}{10} \\ \frac{1}{2}(180N)° & = \omega_n \\ 180N° & \ge 10\omega_n \end{cases}$
$\frac{1}{\left[1 + \frac{2j\omega\zeta}{\omega_k} + (j\omega/\omega_k)^2\right]^N}$	$\begin{cases} 0 & \le \omega_k \\ -40N\text{dB/decade} & > \omega_k \end{cases}$	$\begin{cases} 0 & \le \frac{\omega_n}{10} \\ \frac{1}{2}(-180N)° & = \omega_k \\ -180N° & \ge 10\omega_k \end{cases}$

Resonant Frequencies

Remember, $\omega = 2\pi f$. Also, $\text{Im}\{Z_{eq}\} = 0$ and $\text{Im}\{Y_{eq}\} = 0$.

- $\omega_0 = \frac{1}{\sqrt{LC}}$ - Imaginary portion of Transfer function vanishes
- Half-Power Frequencies - Frequencies where power dissipated is $1/2$ of that dissipated at resonant frequency

    - $\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$
    - $\omega_2 = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$
    - $\omega_0 = \sqrt{\omega_1\omega_2}$

- $B = \omega_2 - \omega_1 = \frac{R}{L}$ - Bandwidth is the frequency band between half-power frequencies

    - $B = \frac{R}{L}$ - Series Impedance Circuit
    - $B = \frac{1}{RC}$ - Parallel Impedance Circuit

- $Q = \frac{\omega_0}{B}$ - Quality Factor: Sharpness of resonance peak

    - $Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 RC}$ - Series Impedance Circuit
    - $Q = \omega_0 RC = \frac{R}{\omega_0 L}$ - Parallel Impedance Circuit