

Assignment - 3

I affirm that I have neither given nor received help or used any means which would make this assignment unfair.

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OWN

5.31

(a)

$$\cos \omega_0 n \xrightarrow{\text{LTI}} w_0 \cos \omega_0 n ; -\pi \leq w_0 \leq \pi$$

Impulse response $\rightarrow h[n]$

Frequency response $\rightarrow H(e^{jw})$

As seen, on passing through the LTI system, we get a scaled version of the same cos function. By taking DTFT of both sides, we can see that,

$$\pi \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) + \delta(\omega + \omega_0 - 2\pi k)] \xrightarrow{\text{LTI}} w_0 \pi \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) + \delta(\omega + \omega_0 + 2\pi k)]$$

By convolution property, we can say that

$$X(e^{jw}) \xrightarrow{\text{LTI}} H(e^{jw}) X(e^{jw})$$

So, on comparison, we can say that (from $-\pi$ to π), $H(e^{jw})$ is $|w|$, as it scales both the impulses in that range by w_0 , regardless of the sign.

$$\therefore H(e^{jw}) = |w|$$

(b) To obtain impulse response, we can take IDTFT of the frequency response.

$$\begin{aligned}
 h[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \\
 &= \frac{1}{2\pi} \left[\int_{-\pi}^0 -w e^{j\omega n} dw + \int_0^{\pi} w e^{j\omega n} dw \right] \\
 &= \frac{1}{2\pi} \left[\int_0^{\pi} w e^{-j\omega n} dw + \int_0^{\pi} w e^{j\omega n} dw \right] \\
 &= \frac{1}{2\pi} \left[\int_0^{\pi} 2w \cos(\omega n) dw \right] \\
 &= \frac{1}{\pi} \left[w \int \cos(\omega n) dw - \int \int \cos(\omega n) dw \right] \quad (\text{By parts}) \\
 &= \frac{1}{\pi} \left[\frac{w \sin(\omega n)}{n} + \frac{\cos(\omega n)}{n^2} \right]_0^{\pi} \\
 &= \frac{1}{\pi} \left[\frac{\cos(\pi n)}{n^2} - \frac{1}{n^2} \right]
 \end{aligned}$$

5.40

$$y[n] = x[n] * h[n]$$

Using convolution sum,

$$\begin{aligned}
 y[n] &= \sum_{k=-\infty}^{\infty} x[k] h[n-k] \\
 \rightarrow y[0] &= \sum_{k=-\infty}^{\infty} x[k] h[-k] \quad \text{--- (1)}
 \end{aligned}$$

Using convolution property,

$$Y(e^{j\omega}) = X(e^{j\omega}) \cdot H(e^{j\omega})$$

$$\Rightarrow y[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) H(e^{j\omega}) e^{j\omega n} d\omega$$

$$\Rightarrow y[0] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) H(e^{j\omega}) d\omega \quad \text{--- (2)}$$

Now, to ~~prove~~ derive Parseval's relation,

$$\sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

As we can see, we need the modulus of the function squared. This points us towards using the conjugate.

As we need $H(e^{j\omega}) = X^*(e^{j\omega})$, we can take $h[n] = x^*[-n]$, because of the properties of FT.

Using this in ① and ②, and equating, we get.

$$\sum_{k=-\infty}^{\infty} x[k] x^*[k] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) X^*(e^{j\omega}) d\omega \quad (1)$$

$$\Rightarrow \sum_{k=-\infty}^{\infty} |x[k]|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

Now, if $h[n] = z^*[-n]$, we get $H(e^{j\omega}) = Z^*(e^{j\omega})$, which gives us

$$\sum_{k=-\infty}^{\infty} x[k] z^*[k] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Z^*(e^{j\omega}) d\omega$$

∴ Derived.

$$5.43 \quad x[n] \longleftrightarrow X(e^{j\omega})$$

$$g[n] = x[2n] \longleftrightarrow G_r(e^{j\omega})$$

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 (a)

$$\begin{aligned} v[n] &= ((e^{-j\pi n} x[n]) + x[n])/2 \\ &= \frac{x[n] + (-1)^n x[n]}{2} \end{aligned}$$

$$10 \quad \therefore v[n] = \begin{cases} x[n] & ; n \text{ is even} \\ 0 & ; n \text{ is odd} \end{cases} \quad (\text{Interpretation})$$

As DTFT is linear, we have the given below relation
 Because of frequency shift property.

$$V(e^{j\omega}) = \text{Im } X(e^{j\omega}) + X(e^{j(\omega+\pi)})$$

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 (b)

$$v[2n] = y[n]$$

$$\therefore Y(e^{j\omega}) = \sum_{k=-\infty}^{\infty} v[2k] e^{-j\omega k}$$

(odd samples are zero anyway).

Since this gives us even samples of v_n on changing $2k$ to p , we get,

$$\begin{aligned} Y(e^{j\omega}) &= \sum_{p=-\infty}^{\infty} v[p] e^{-j\omega p/2} \\ &\underline{=} V(e^{j\omega/2}) \end{aligned}$$

∴ Proved.

- (c) $x[2n] \rightarrow$ only even samples of $x[n]$
 $v[n] \rightarrow$ odd samples = 0
 even samples = $x[n]$.

$\therefore v[2n]$ will also be only even samples of $x[n]$, as $2n$ is always even.

As seen in (b), $v[2n] \longleftrightarrow V(e^{j\omega_2})$
 $= \underline{x(e^{j\omega_2})} + \underline{x(e^{j(\omega_2 + \pi)})}$

$\therefore A, v[2n] = x[2n]$,
 $g[n] \geq v[2n]$ (From (a))

$$\Rightarrow g[n] \longleftrightarrow G(e^{j\omega}) = \underline{V(e^{j\omega_2})} \\ = \underline{\underline{x(e^{j\omega_2})}} + \underline{\underline{x(e^{j(\omega_2 + \pi)})}}$$

PM

2.25. $y(n) = a^n u(n)$ ($|a| < 1$)

$$(a) y(n) = \sum_{k=-\infty}^n c_k y(n-k) \\ = \sum_{k=-\infty}^n c_k a^{n-k} \quad (\because u(n) = 0 \quad \forall n < 0).$$

Now, as we need to find c_k , we can do it by subtracting all the other terms, that is,

$$y(k) = \sum_{j=-\infty}^k c_j a^{k-j} \quad \text{--- (1)}$$

Now, if we take $y(k-1)$,

$$x(k-1) = \sum_{j=-\infty}^{k-1} c_j a^{k-j-1}$$

Now, we need to scale $x(k-1)$ by a , as all c_j for $j < k$ have a^{k-j-1} , which will not make them zero on subtraction.

$$a x(k-1) = \sum_{j=-\infty}^{k-1} c_j a^{k-j}. \quad \text{--- (2)}$$

Now, if we do $\textcircled{1} - \textcircled{2}$, we get,

$$\begin{aligned} x(k) - a x(k-1) &= \sum_{j=-\infty}^k c_j a^{k-j} - \sum_{j=-\infty}^{k-1} c_j a^{k-j} \\ &= \boxed{c_k} a^0 \\ &= \boxed{c_k} \end{aligned}$$

\therefore As c_k exists, we can represent $x(n)$ as that summation, and c_k will be defined as,

$$c_k = \underline{x(k) - a x(k-1)}$$

$$(b) y(n) = T[x(n)] ; g(n) = T[x(n)]$$

$$= T \left[\sum_{k=-\infty}^n c_k r(n-k) \right]$$

Using linearity,

$$y(n) = \sum_{k=-\infty}^n c_k T[r(n-k)]$$

Using time invariance, we get,

$$y(n) = \sum_{k=-\infty}^{\infty} c_k g(n-k)$$

(C) $\delta(n)$, $h(n) = T[\delta(n)]$

As $x(n) = \sum_{k=-\infty}^n c_k r(n-k)$, we can construct a delta

function using $x(n)$.

As $\delta(n)$ in discrete time is defined as 1 at 0 and zero everywhere else, we can obtain this by making ~~coeff~~ finding c_k .

To find coefficients we use

$$c_k = x(k) - a x(k-1) \quad (\text{from (a)}).$$

Here, as we need $\delta(n)$,

$$c_k = \underline{\delta(k) - a \delta(k-1)}$$

Now substituting this into the $y(n)$ expression we got for $T[x(n)]$

$$y(n) = \sum_{k=-\infty}^{\infty} c_k g(n-k)$$

$$\Rightarrow h(n) = \sum_{k=-\infty}^{\infty} (\delta(k) - a \delta(k-1)) g(n-k)$$

4.23.

$$(a) y_1(n) = \begin{cases} x(n) & ; n \text{ even} \\ 0 & ; n \text{ odd} \end{cases}$$

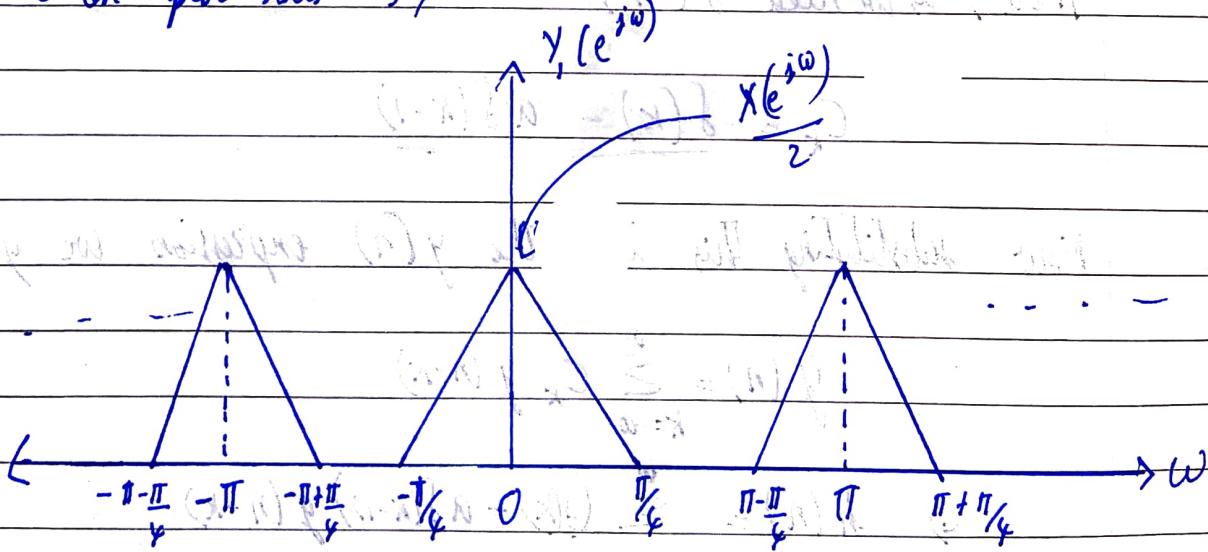
$\therefore y_1(n)$ can be interpreted as all even samples of $x(n)$, and so, we can represent it as,

$$\begin{aligned} y_1(n) &= \frac{x(n) + (-1)^n x(n)}{2} \\ &= x(n) + \frac{e^{j\pi n}}{2} x(n) \end{aligned}$$

If we take DTFT, we can use linearity and frequency shift properties to get,

$$Y_1(e^{j\omega}) = \frac{X(e^{j\omega}) + X(e^{j(\omega-\pi)})}{2}$$

\therefore We can plot this as,



$$(b) y_2(n) = x(2n)$$

This will also only have even samples of $x(n)$.

Taking DTFT, we get,

$$Y_2(e^{j\omega}) = \sum_{k=-\infty}^{\infty} x[2k] e^{-j\omega k}$$

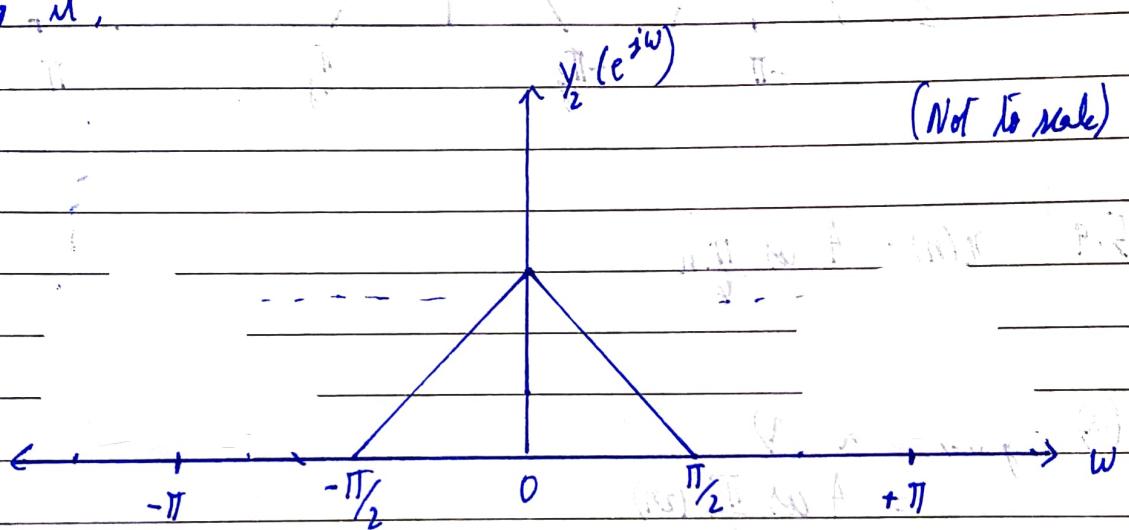
(assuming $x[p] = 0$ for odd p)

Substituting $2k = p$, we get,

$$Y_2(e^{j\omega}) = \sum_{p=-\infty}^{\infty} x[p] e^{-j\omega \frac{p}{2}}$$

$$= X\left(e^{j\frac{\omega}{2}}\right)$$

Plotting it,



$$(c) y_3(n) = \begin{cases} x(n_2) & ; n \text{ even} \\ 0 & ; n \text{ odd} \end{cases}$$

As we can see, this definition of $y_3(n)$ will give us ALL samples of $x(n)$.

Taking DTFT,

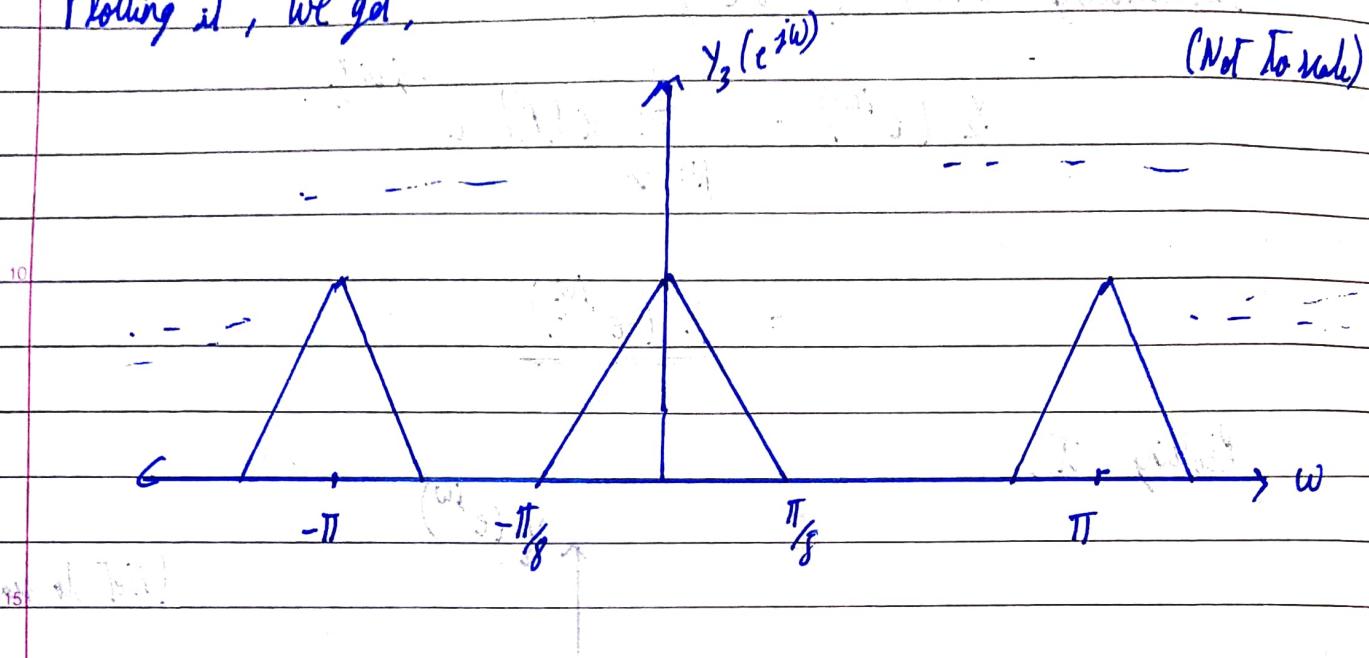
$$Y_3(e^{j\omega}) = \sum_{k=-\infty}^{\infty} x\left(\frac{k}{2}\right) e^{-j\omega k} ; k = \text{even}$$

Replacing ~~k~~ with ~~k~~ , $k/2 = p$, (because it still has all samples).

$$Y_3(e^{j\omega}) = \sum_{p=-\infty}^{\infty} x(p) e^{-2j\omega p}$$

$$= \underline{X(e^{j\omega})}$$

Plotting it, we get,



$$5.9 \quad x(n) = A \cos \frac{\pi}{4} n$$

$$(a) \quad y(n) = x(2n)$$

$$= A \cos \frac{\pi}{4} (2n)$$

$$= \underline{A \cos \frac{\pi}{2} n}$$

To find frequency content, we take DTFT.

$$Y(e^{j\omega}) = \sum_{k=-\infty}^{\infty} A \cos \left(\frac{\pi}{2} k \right) e^{jk\omega k}$$

$$= A \sum_{k=-\infty}^{\infty} \cos \left(\frac{\pi}{2} k \right) e^{jk\omega k}$$

$$= A \left(\pi \sum_{l=-\infty}^{\infty} [\delta(\omega - \frac{\pi}{2} - 2\pi l) + \delta(\omega + \frac{\pi}{2} - 2\pi l)] \right)$$

DTFT of $\cos \omega_0 n$
formula.

$$(b) \quad y(n) = x^2(n)$$

$$= A^2 \cos^2 \frac{\pi}{4} n$$

$$= A^2 \left(\frac{1 + \cos \frac{\pi}{2} n}{2} \right)$$

$$= \frac{A^2}{2} + \frac{A^2}{2} \cos \frac{\pi}{2} n$$

\downarrow
DFT

$$\frac{A^2}{2} \left[2\pi \sum_{l=-\infty}^{\infty} \delta(w - 2\pi l) \right] + \frac{A^2}{2} \left[\pi \sum_{l=-\infty}^{\infty} [\delta(w - \frac{\pi}{2} - 2\pi l) + \delta(w + \frac{\pi}{2} - 2\pi l)] \right]$$

$\underbrace{\qquad\qquad\qquad}_{\text{DFFT of } x[n]=1 \text{ formula.}}$ $\underbrace{\qquad\qquad\qquad}_{\text{DFFT of } \cos w_0 n \text{ formula.}}$

$$\therefore Y(e^{jw}) = \frac{A^2}{2} \pi \left(2 \sum_{l=-\infty}^{\infty} \delta(w - 2\pi l) + \sum_{l=-\infty}^{\infty} [\delta(w - \frac{\pi}{2} - 2\pi l) + \delta(w + \frac{\pi}{2} - 2\pi l)] \right)$$

$$(c) \quad y(n) = \cos(\pi n) x(n)$$

$$= A \cos(\pi n) \cos(\frac{\pi}{4} n)$$

$$= \frac{A}{2} (2 \cos(\pi n) \cos(\frac{\pi}{4} n))$$

$$= \frac{A}{2} (\cos((\pi + \frac{\pi}{4})n) + \cos((\pi - \frac{\pi}{4})n))$$

$$(\because \cos C + \cos D = 2 \cos \frac{(C+D)}{2} \cos \frac{(C-D)}{2})$$

\downarrow
DFT

DFFT of $\cos w_0 n$.

$$= \frac{A}{2} \left(\left(\pi \sum_{l=-\infty}^{\infty} \delta(w - \pi - \frac{\pi}{4} - 2\pi l) + \delta(w + \pi + \frac{\pi}{4} - 2\pi l) \right) + \left(\pi \sum_{l=-\infty}^{\infty} \delta(w - \pi + \frac{\pi}{4} - 2\pi l) + \delta(w + \pi - \frac{\pi}{4} - 2\pi l) \right) \right)$$

$$\therefore Y(e^{j\omega}) =$$

$$\frac{A}{2} \pi \left(\left(\sum_{l=-\infty}^{\infty} \delta(\omega - \pi - \frac{\pi}{4} - 2\pi l) + \delta(\omega + \pi + \frac{\pi}{4} - 2\pi l) \right) + \left(\sum_{l=-\infty}^{\infty} \delta(\omega - \pi + \frac{\pi}{4} - 2\pi l) \right. \right.$$

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$$\left. \left. + \delta(\omega + \pi - \frac{\pi}{4} - 2\pi l) \right) \right)$$

As clearly seen, none of these systems are LTI systems, as they are adding frequencies to the input.