



EL2520

Control Theory and Practice

Lecture 2: The closed-loop system

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Goals

After this lecture, you should:

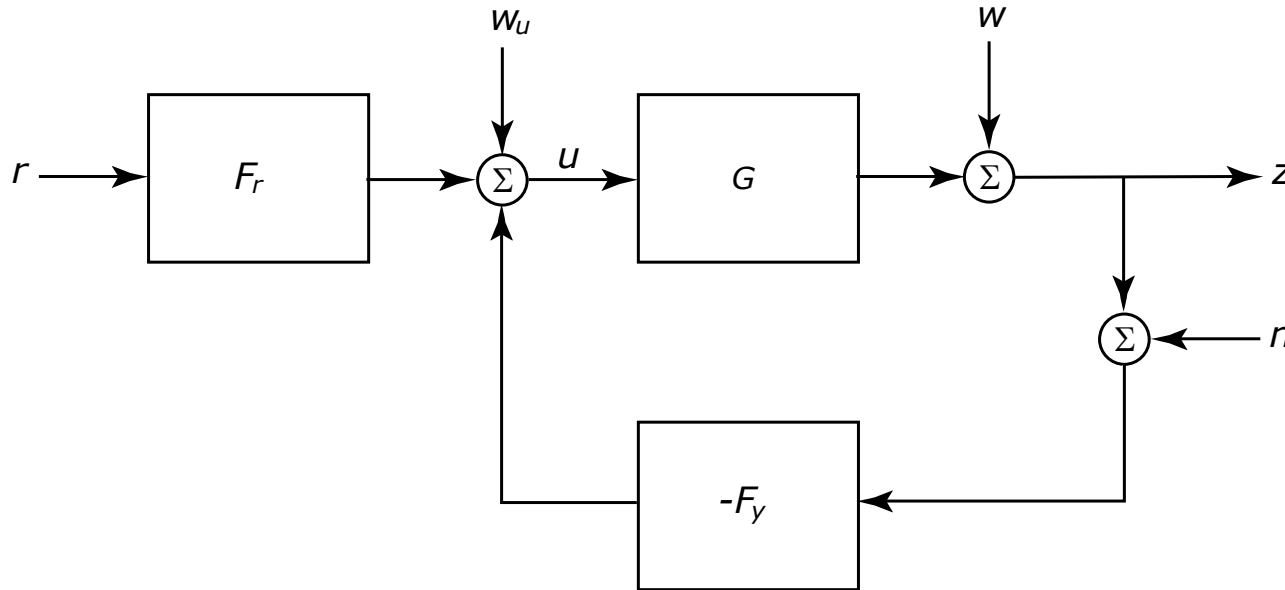
- Know that the closed-loop is characterized by 7 transfer functions
 - several objectives, but trade-offs exist, e.g. $S+T=1$.
 - *internal stability* requires all 7 to be stable
- Be able to determine, analyze and design desired sensitivity functions
 - sensitivity function for disturbance rejection
 - complementary sensitivity function for robust stability and noise
 - control problem formulated as objective of making norm of key transfer-functions small.

Material: course book Chapter 6, lecture notes 2.

Contents

1. The closed-loop system
2. The control problem – and 7 central transfer functions
3. Internal stability
4. The sensitivity function and disturbance rejection
5. The complementary sensitivity, noise and robust stability
6. Design of sensitivity functions using weights

The closed-loop system



Controller: feedforward F_r , feedback F_y
Disturbances: w, w_u drives system from desired state
Measurement noise: corrupts information about z

Aim: find controller such that z follows r , with limited use of u

The design problem

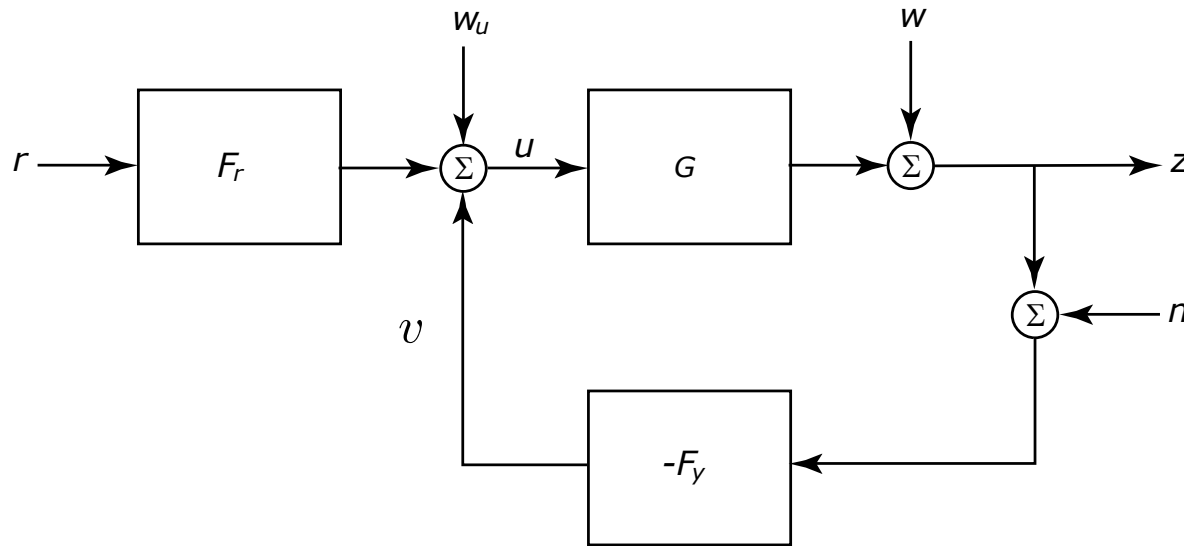
Design problem: find a controller that

- a) Attenuates the effect of disturbances
- b) Does not inject too much measurement noise into the system
- c) Makes the closed loop insensitive to process variations
- d) Makes the output follow command signal

Often convenient with two-degree of freedom controller (can design setpoint tracking independent of disturbance attenuation).

Use feedback to deal with a,b,c; use feedforward to deal with d.

Closed Loop Transfer Functions



$$z = w + G(w_u + F_r r - F_y(z + n)) \Rightarrow$$

$$z = \underbrace{\frac{1}{1 + GF_y}}_S w + \underbrace{\frac{G}{1 + GF_y}}_{SG} w_u + \underbrace{\frac{GF_r}{1 + GF_y}}_{G_c} r - \underbrace{\frac{GF_y}{1 + GF_y}}_T n$$

Similarly, we find

$$u = SF_r r - SF_y(w + n) + Sw_u$$

SISO: Closed-loop characterized by *seven* transfer functions

Transfer functions and observations

$$S = \frac{1}{1 + GF_y} \quad (w \rightarrow z, w_u \rightarrow u) \quad \text{sensitivity function}$$

$$T = \frac{GF_y}{1 + GF_y} \quad (n \rightarrow z) \quad \text{complementary sensitivity}$$

$$G_c = \frac{GF_r}{1 + GF_y} \quad (r \rightarrow z) \quad \text{closed loop system}$$

$$SG = \frac{G}{1 + GF_y} \quad (w_u \rightarrow z)$$

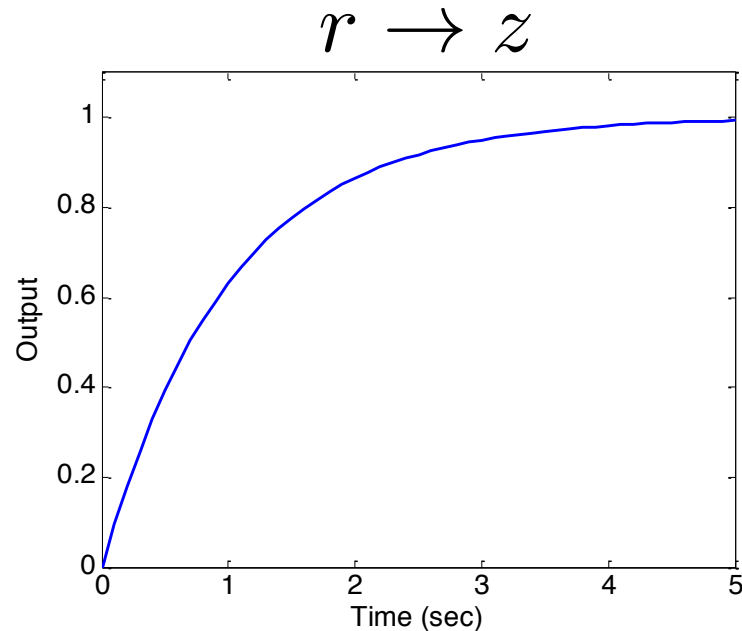
$$SF_y = \frac{F_y}{1 + GF_y} \quad (n \rightarrow u)$$

$$SF_r = \frac{F_r}{1 + GF_y} \quad (r \rightarrow u) \quad TF_r = \frac{F_r GF_y}{1 + GF_y} \quad (r \rightarrow v)$$

Observation: need to look at all! Many tradeoffs (e.g. $S+T=1$)

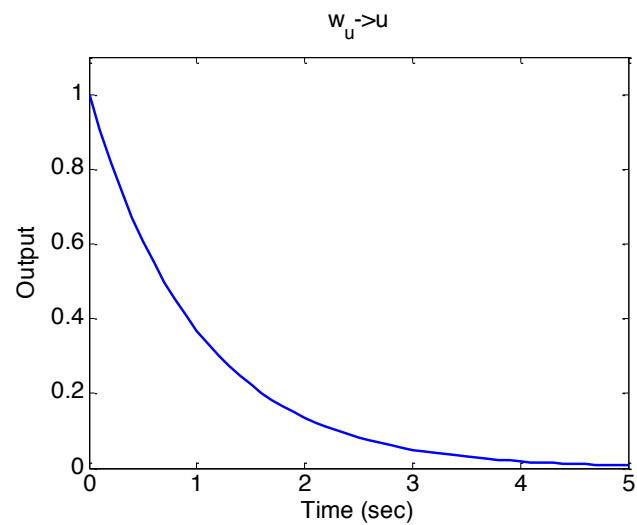
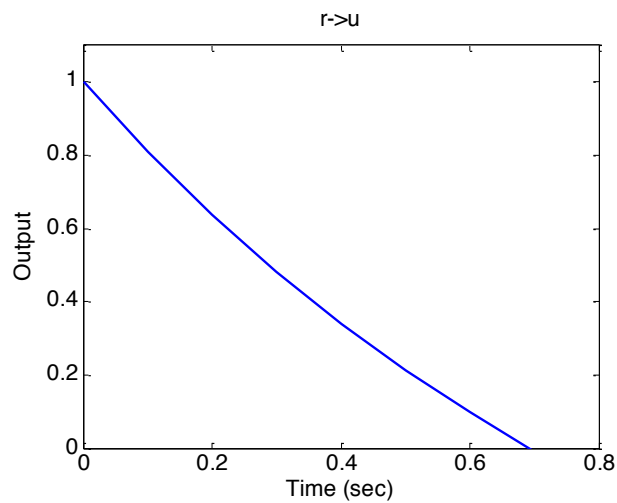
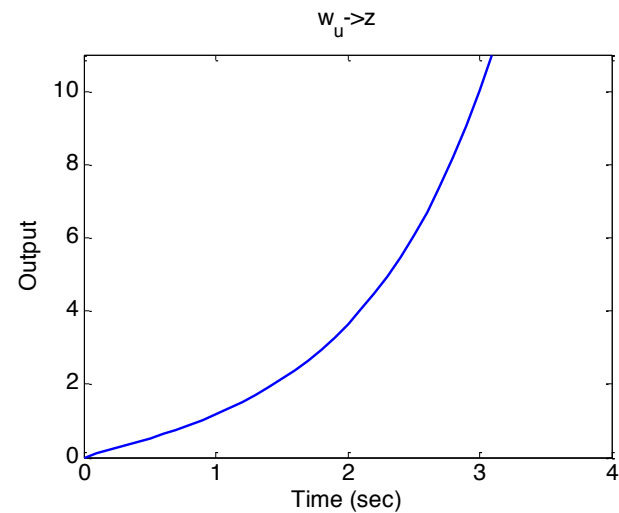
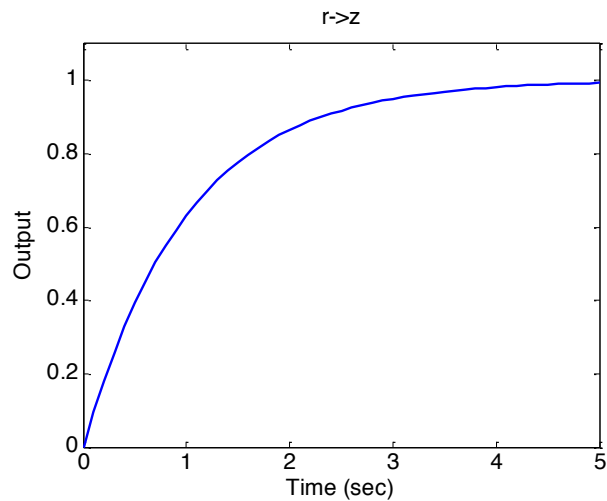
A warning!

Individual time responses may look good



but you need to verify that *all* transfer functions are as desired!

Four responses, same controller



What is going on?

Process: $G = \frac{1}{s - 1}$

Controller: $F_y = F_r = \frac{s - 1}{s}$

Transfer functions:

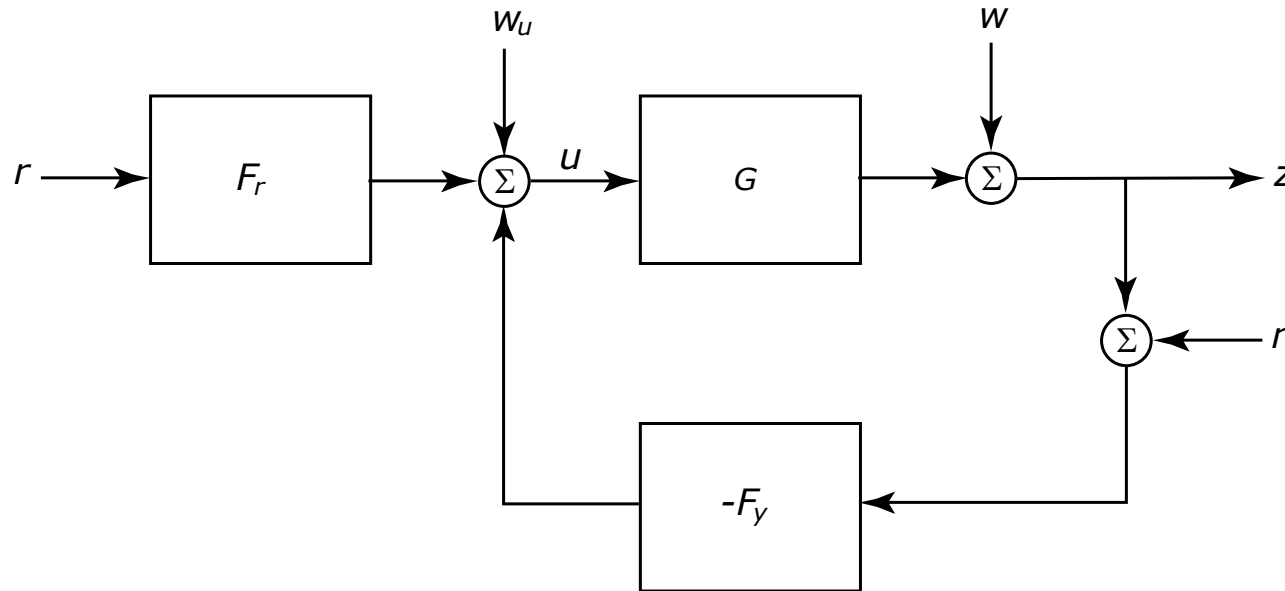
$$T = \frac{1/s}{1 + 1/s} = \frac{1}{s + 1} \quad (\text{stable})$$

$$SG = \frac{s}{(s + 1)(s - 1)} \quad (\text{unstable!})$$

$$SF_y = \frac{(s - 1)}{(s + 1)} \quad (\text{stable})$$

$$S = \frac{s}{(s + 1)} \quad (\text{stable})$$

Internal stability



Definition. The closed loop system above is *internally stable* if it is input-output stable.

Here inputs are all external signals; outputs are all internal signals

Theorem. If G is SISO, the closed-loop system is internally stable if and only if S , SG , SF_y , F_r are all stable

Sensitivity functions

Sensitivity and complementary sensitivity are particularly important:

- S determines attenuation of disturbances and sensitivity to model uncertainty
- T determines sensitivity to noise and robustness to unmodelled dynamics

Both connected to classical stability margins (gain, phase margin) in SISO case

First trade-off: **$S+T=1$** - cannot make both zero at the same time.

Disturbance rejection

The transfer function from w to z in open loop is

$$G_{w \rightarrow z}^{\text{ol}} = 1$$

while the closed-loop counter-part is

$$G_{w \rightarrow z}^{\text{cl}} = \frac{1}{1 + GF_y}$$

Thus

$$\frac{G_{w \rightarrow z}^{\text{CL}}}{G_{w \rightarrow z}^{\text{OL}}} = \frac{1}{1 + GF_y} = S$$

S quantifies disturbance attenuation from feedback

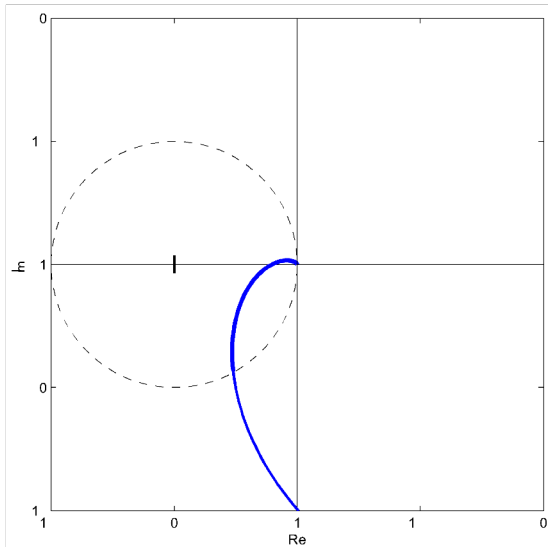
Disturbance at frequencies with

- $|S(i\omega)| < 1$ attenuated by feedback
- $|S(i\omega)| > 1$ amplified by feedback

Nyquist curve interpretation

$|S(i\omega)| = |L(i\omega) + 1|^{-1}$, where $L = GF_y$, is inverse distance from Nyquist curve to the point -1 on the negative real axis

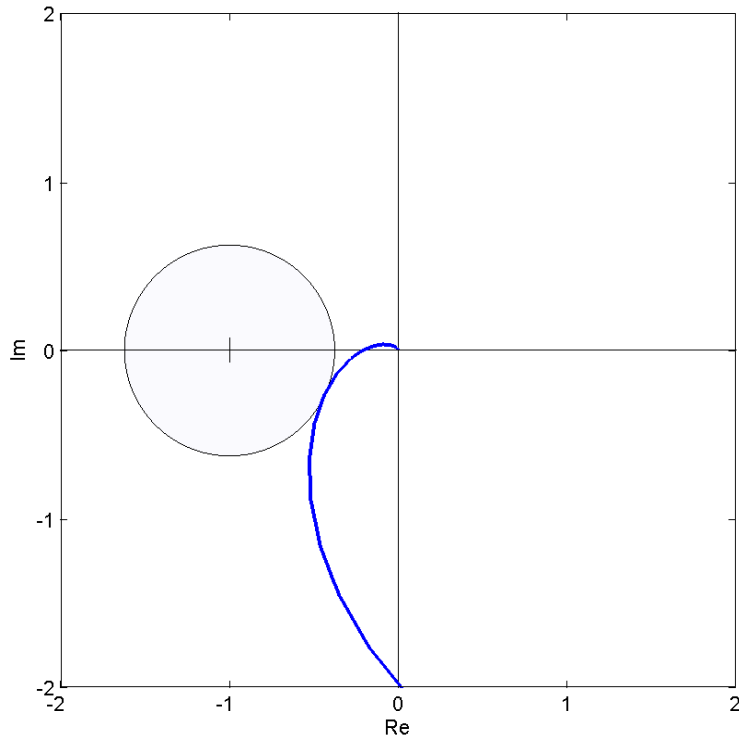
Disturbance amplified at frequencies where Nyquist curve is inside unit circle centered at the -1 point.



Observation: cannot avoid circle with radius one if pole excess is 2 or more, i.e., must amplify disturbances at some frequencies.

Maximum sensitivity and M_s -circles

Specification $|S(i\omega)| \leq M_s$: loop gain must stay outside circle with radius M_s^{-1}



Reasonable values: $1.2 < M_s < 2$
(picture shows $M_s=2$)

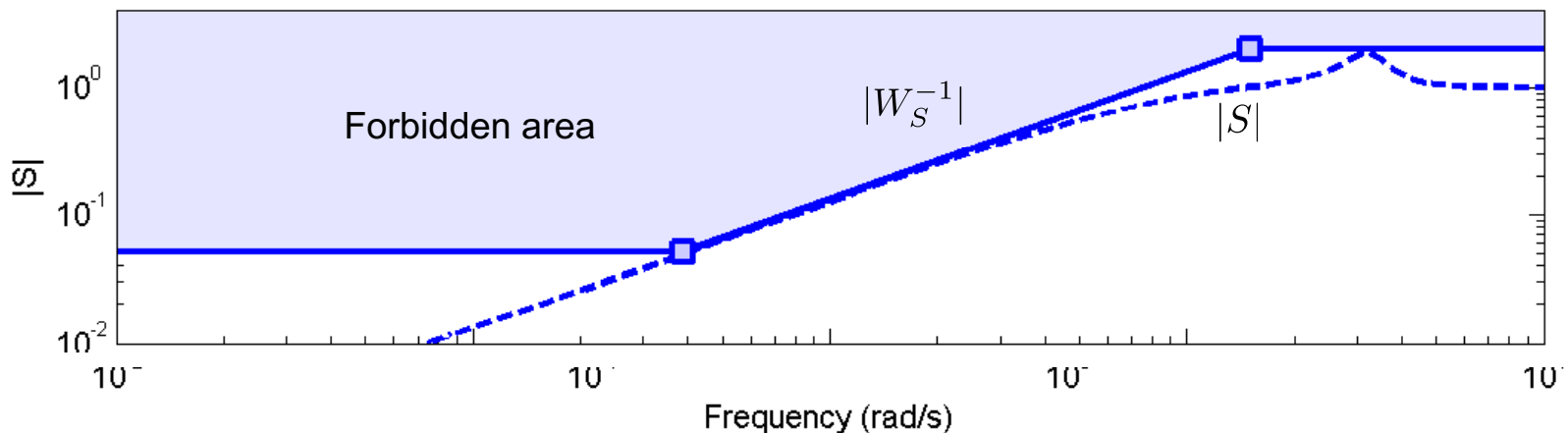
Sensitivity shaping

Observations:

- Can't attenuate disturbances at all frequencies (if pole excess is 2)
- Need to limit $|S(i\omega)|$ at frequencies with significant disturbances

Reasonable design specification

$$|S(i\omega)| \leq |W_S^{-1}(i\omega)| \quad \forall \omega \quad \Leftrightarrow \quad \|W_S S\|_{\infty} \leq 1$$



Sensitivity to uncertainties

The response of z to r is nominally (without uncertainty)

$$z = G_{cl}r = \frac{GF_r}{1 + GF_y}r$$

If there is a model error, so that the true open-loop system is

$$\tilde{G} = (1 + \Delta_G)G$$

then the true response to r is

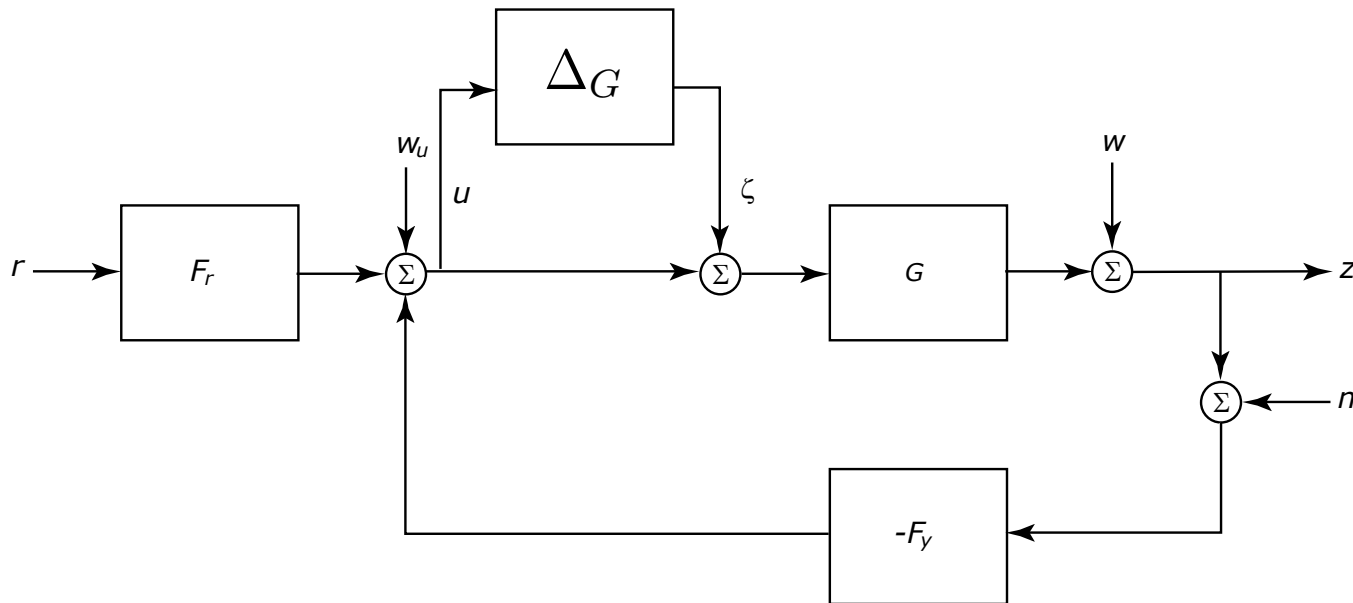
$$\tilde{z} = (1 + \tilde{S}\Delta_G)z$$

Thus, sensitivity function also quantifies “attenuation” of model uncertainty

Robust stability

Uncertainty also affect stability of closed-loop system.

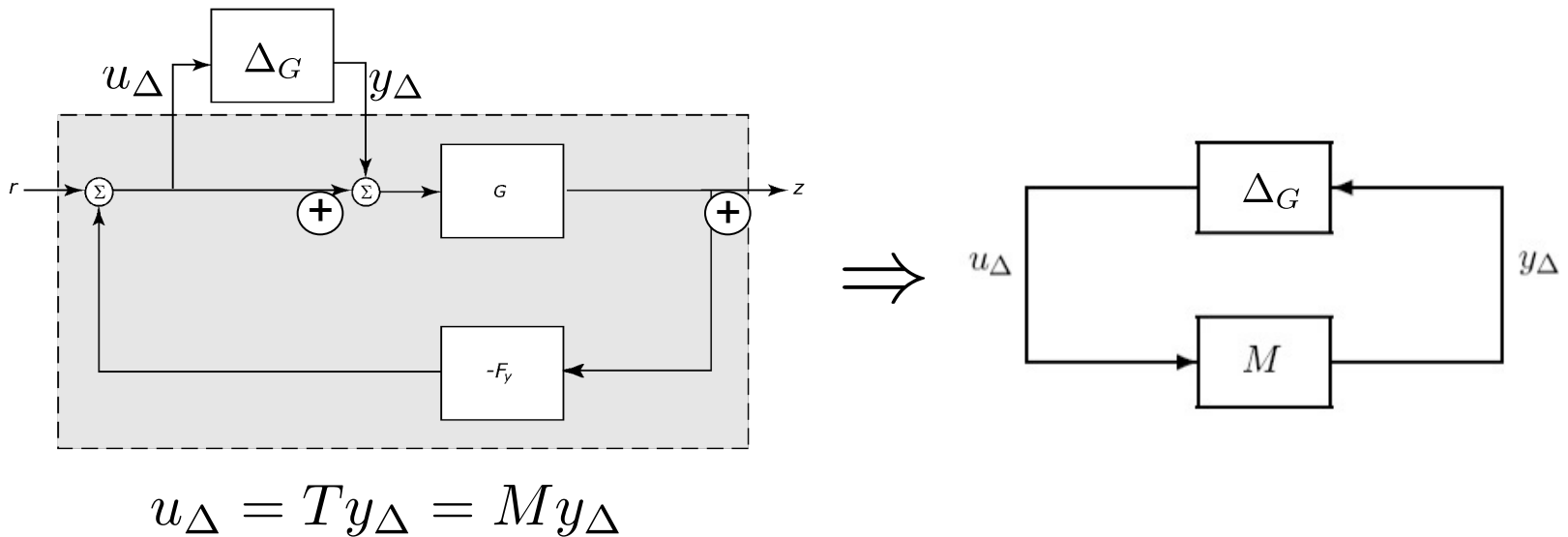
Assume true system is given by $\tilde{G} = (1 + \Delta_G)G$



What linear Δ_G can be tolerated without risking stability?

Robust stability

Assume all exogenous inputs (r , w , w_u , n) zero, and re-write



Robust stability

Assume Δ_G stable and nominal-system T internally stable.

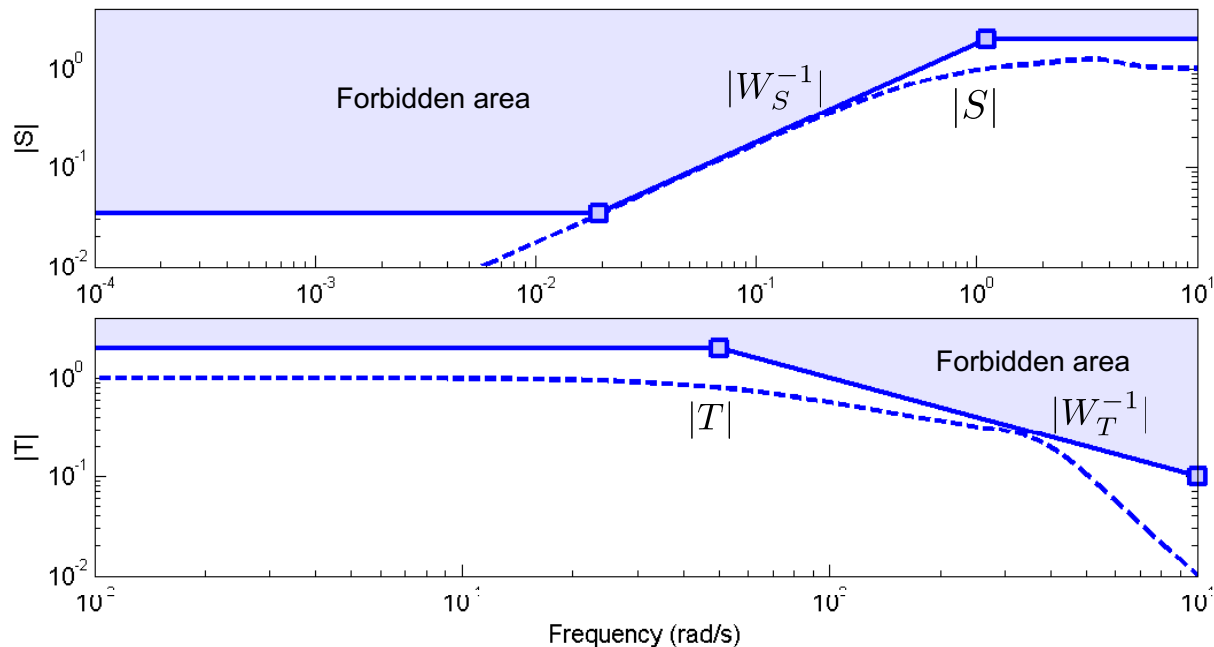
If $\|T\Delta_G\|_\infty < 1$, then the system is input-output stable.

$$|T(i\omega)| < \frac{1}{|\Delta_G(i\omega)|} \quad \forall \omega \quad \Leftrightarrow \quad \|T\Delta_G\|_\infty < 1$$

Proof: Small Gain Theorem.

Sensitivity shaping

Reasonable design criterion: make sure that both the sensitivity S and the complementary sensitivity T avoid “forbidden areas”



$$\begin{aligned}
 |S(i\omega)| &\leq |W_S^{-1}(i\omega)| \\
 |T(i\omega)| &\leq |W_T^{-1}(i\omega)|
 \end{aligned}
 \Leftrightarrow
 \begin{aligned}
 \|SW_S\|_{\infty} &\leq 1 \\
 \|TW_T\|_{\infty} &\leq 1
 \end{aligned}$$

Summary

- Closed-loop system characterized by 7 transfer functions
 - Need to consider all!
 - All must be stable for internal stability.
- Sensitivity and complementary especially important
 - S: disturbance attenuation, “performance sensitivity”
 - T: noise attenuation, robust stability
- Control system design via “sensitivity shaping”
- Conflicts and limitations
 - $S+T=1$
 - $|S(i\omega)| \geq 1$ for some ω (disturbance amplification!)
 - More on this in the next two lectures!