

EL2520 - Control Theory and Practice - Advanced  
Course  
Solution/Answers – 2017-08-18

1. (a) The determinant of  $G(s)$  is  $-3$  and the LCD of this and the elements are  $(s+1)(s-1)$ , hence poles at  $s = -1$  and  $s = 1$ . The zeros are given by the determinant normalized to have the pole polynomial as denominator, i.e.,  $\det G = -3(s+1)(s-1)/((s+1)(s-1))$  and hence we have zeros at  $s = -1$  and  $s = 1$ . With two poles we need two states in a minimal state space realization.
- (b)  $\|G\|_\infty = \sup_w \bar{\sigma}(G(i\omega))$ . The singular values are given by  $\sqrt{\lambda(G^T G)}$  which gives  $\bar{\sigma}(G) = \frac{1}{\omega^2+1}\sqrt{2}$ . The supremum is at  $\omega = 0$  and hence  $\|G\|_\infty = \sqrt{2}$ .
- (c) Any RHP poles in  $G(s)$  must appear as RHP zeros in  $S(s)$  for internal stability (follows from interpolation constraints). Since  $S(s)$  does not have a zero at  $s = 1$ , the system is not internally stable (a RHP pole has been cancelled).
2. (ai) For a  $2 \times 2$  system, the 1,1-element of the RGA is  $\lambda_{11} = \lambda_{22} = \frac{1}{1 - \frac{g_{12}g_{21}}{g_{11}g_{22}}}$  and  $\lambda_{12} = \lambda_{21}$ . This gives  $\lambda_{11}(i0) = \frac{1}{1-2} = -1$  and  $\lambda_{12} = 2$ . The rule is to never pair on steady-state RGA elements, and hence  $y_1 - u_2, y_2 - u_1$  is the only viable pairing. At bandwidth  $\omega = 0.5$  we get  $\lambda_{12}(i0.5) = 1 - (0.5i + 1)/(0.5i - 1)$  and  $|\lambda_{12}(i0.5)| = 1.79$  which indicates relatively weak interactions and hence decentralized control should be expected to work quite well.
- (aii)  $G(s)$  has a RHP zero at  $s = 1$  and this can not be cancelled by  $W(s)$  due to requirement of internal stability. Hence the proposed decoupler does not provide internal stability.
- (aiii) Must retain zero at  $s = 1$  and with decoupling this means every diagonal element must have this zero. Thus we propose

$$G(s)W(s) = D(s) = \frac{1}{(s+1)^2} \begin{pmatrix} (-s+1) & 0 \\ 0 & (-s+1) \end{pmatrix}$$

which gives

$$W(s) = G^{-1}(s)D(s) = \begin{pmatrix} -1 & \frac{1}{s+1} \\ 2 & -1 \end{pmatrix}$$

- (b) The closed-loop Jacobian is  $A = \begin{bmatrix} 1 & 1 \\ -1 & -1.1 \end{bmatrix}$  and the eigenvalues are  $\lambda = -0.37, 0.27$  which both are inside the unit circle in the complex plane and hence the closed-loop is stable.
3. (a) Disturbance attenuation is needed for frequencies where  $|G_d(i\omega)| > 1$ , and  $|G_d| = 3/\sqrt{100\omega^2 + 1}$  which equals one for  $\omega_d = \sqrt{0.08} = 0.28$ .

The limitations are partly due to the RHP zero in  $G_1$  which gives that disturbance attenuation can only be achieved up to  $\omega = z = 4$  (or  $\omega = z/2 = 2$  if we want to limit the peak of  $S$  to 2), and the delay in  $G_2$  which gives that attenuation can only be achieved approximately up to  $\omega = 2/\theta = 1$ . Neither of these are thus a problem since they exceed the required  $\omega_d$ . However, we also need sufficient input to counteract the disturbance, i.e.,  $|G| > |G_d|\forall\omega < \omega_d$  for perfect disturbance attenuation (for acceptable attenuation it suffices that  $|G| > |G_d| - 1$ ). We see that  $|G_1| > |G_d|\forall\omega$  which  $|G_2| < |G_d| - 1\forall\omega$  and hence we can use  $u_1$  only, but not  $u_2$  only.

- (b) Requirement is  $|SG_d| < 1\forall\omega$ . We try with a simple P-controller  $F_y = K_c$ , which gives  $S = 1/(1 + G_1K_c) = (10s + 1)/(10s + 1 + K_c(s - 4))$  and  $SG_d = 3e^{-s}/((10 + K_c)s + (1 - 4K_c))$  which is stable and with magnitude  $< 1\forall\omega$  if  $-10 < K_c < -0.5$ .
- (c) The requirement for  $|y| < 1$  with  $|d| < 1$  translates to  $\|SG_d\|_\infty < 1$ , the requirement for  $|u| < 1$  with  $|d| < 1$  translates into  $\|F_ySG_d\|_\infty < 1$  and the robust stability requirement  $\|0.2T\|_\infty > 1$ . Stacked, this gives the optimization problem

$$\min_u \left\| \begin{array}{c} SG_d \\ F_ySG_d \\ 0.2T \end{array} \right\|_\infty$$

We have  $y = SGdd$  and  $u = F_ySG_d$  and hence we should have  $d$  as an input and  $z = [y \ u]^T$  as an output of the extended system for the first two criteria. Unfortunately, it is difficult to find an output which has the transfer function  $T$  from the input  $d$ . Thus, we have to add another input, e.g., measurement noise  $n$  with gain 0.2 since then  $y = 0.2Tn$ . The disadvantage of adding the second input  $n$  is that we then also include the transfer function from  $n$  to  $u$  in the objective.

4. (a)  $G(s)$  has a RHP zero at  $s = 1$  and  $y_z^H G(z) = 0$  gives the zero direction  $y_z^H = 1/\sqrt{5} [-1 \ 2]$ . Then, the requirement acceptable disturbance attenuation being feasible is  $|y_z^H G_d(z)| < 1$ . For disturbance  $d_1$  we get  $y_z^H G_{d1}(z) = 2.01 > 1$ , and for  $d_2$   $y_z^H G_{d2}(z) = 0 < 1$ . Hence, it is not possible to attenuate disturbance  $d_1$  but there are no given limitations that hinders acceptable attenuation of  $d_2$ .
- (b) Any RHP pole in the open-loop should appear as a RHP zero in  $S(s)$  for internal stability. Here the open-loop has RHP poles, but  $S(s)$  has no RHP zeros and hence the system is not internally stable.
- (c) This is a standard LQ-problem with  $M = I, Q_1 = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, Q_2 = 2$ .

The optimal state feedback is given by  $L = Q_2^{-1}B^T S$  where  $S \geq 0$  is the solution to

$$A^T S + SA + M^T Q_1 M - SBQ_2^{-1}B^T S = 0$$

which gives  $S = \sqrt{2} \begin{pmatrix} \sqrt{\sqrt{2}} & 1 \\ 1 & \sqrt{\sqrt{2}} \end{pmatrix}$  and  $L = [1/\sqrt{2}\sqrt{\sqrt{2}}]$ .

5. (a)  $G = G_0(I + \Delta_G)$  gives

$$\Delta_G = GG_0^{-1} - I = \begin{pmatrix} 0 & \frac{2}{3}\delta_1 \\ \frac{1}{2}\delta_2 & 0 \end{pmatrix}$$

(b) Two poles in  $s = -2$  implies on pole in  $s = -2$  for each loop, i.e.,  $T_{ii} = 2/(s+2)$ . Since  $GF_y = T/(1-T) = 2/s$  we get  $K_1 = 23, \tau_{I1} = 23$  and  $K_2 = 92/3, \tau_{I2} = 23$ .

(c) Derivation not shown here gives the robustness criterion  $\|T_I \Delta_G\|_\infty < 1$  and  $T_I = F_y G_0(I + F_y G_0)^{-1}$  stable. The criterion gives that  $|\delta_1| < \frac{3}{2}$  and  $|\delta_2| < 2$ .

(d) We get

$$FG = \begin{pmatrix} \frac{2}{s} & \frac{\delta_1}{s} \\ \frac{4\delta_2}{3s} & \frac{2}{s} \end{pmatrix}$$

and

$$S = (I + FG)^{-1} = \frac{s}{(s+2)^2 - \frac{4}{3}\delta_1\delta_2} \begin{pmatrix} s+2 & -\delta_1 \\ -\frac{4}{3}\delta_2 & s+2 \end{pmatrix}$$

for which we get stability for  $\delta_1\delta_2 < 3$ . It suffices to check the sensitivity for stability since there are no cancellations between  $F$  and  $G$  in the RHP. This is a less conservative criterion than obtained from the robustness criterion in (c) since we get a bound on the product of  $\delta_1$  and  $\delta_2$  rather than individual bounds. For instance, with  $\delta_1 \leq 0$  we allow  $\delta_2 = \infty$ , but the result in (c) says  $|\delta_2| < 2$  for stability. The conservativeness in (c) comes from the use of the small gain theorem which is sufficient but not necessary for stability.