

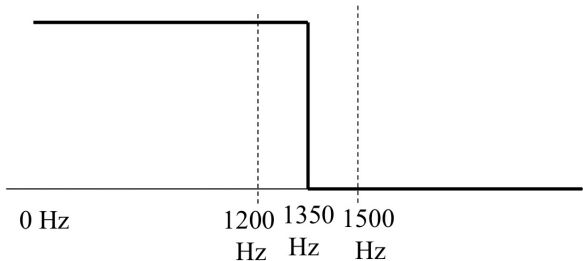
ADSP HW1

R10942152

游家權

(1) Design a Mini-max **lowpass** FIR filter such that (40 scores)

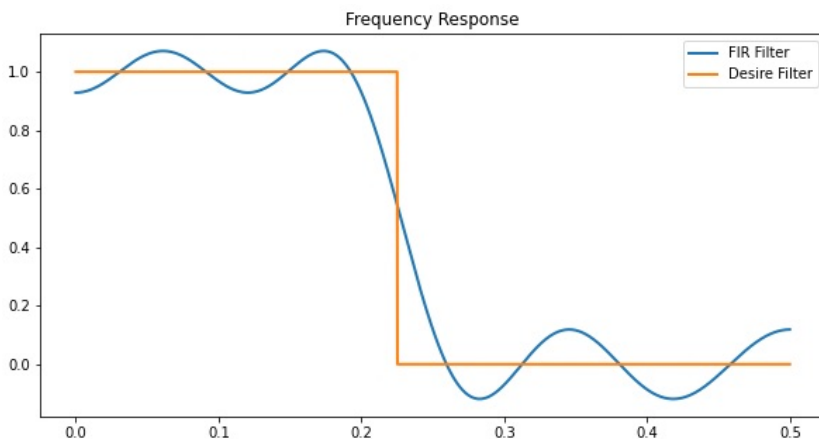
- ① Filter length = 17, ② Sampling frequency $f_s = 6000\text{Hz}$,
- ③ Pass Band 0~1200Hz ④ Transition band: 1200~1500 Hz,
- ⑤ Weighting function: $W(F) = 1$ for passband, $W(F) = 0.6$ for stop band .
- ⑥ Set $\Delta = 0.0001$ in Step 5.



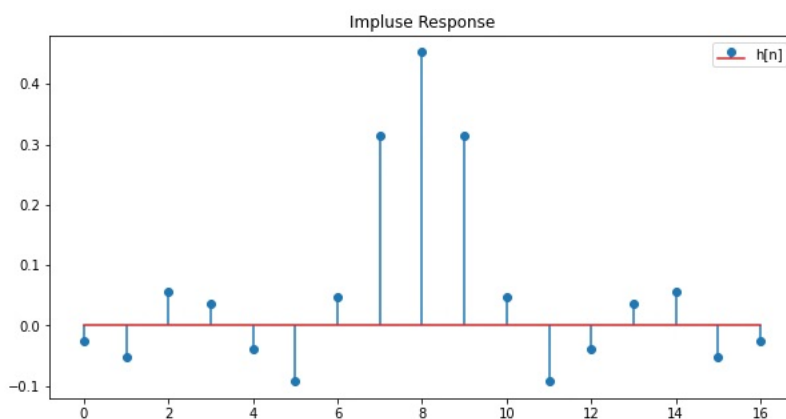
※ The code should be handed out by NTUCool, too.

Show (a) the frequency response, (b) the impulse response $h[n]$, and (c) the maximal error for each iteration.

(a) Frequency Response



(b) Impulse Response



(c) Maximum Error for each iteration

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1st_iteration_MaxErr = 0.12362984401511301
2nd_iteration_MaxErr = 0.07704144356675102
3rd_iteration_MaxErr = 0.07127611617996485
4th_iteration_MaxErr = 0.0712072852346789
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(2) (a) How do we convert convolution into addition?

(b) From the view point of implementation, what are the disadvantages of the discrete Fourier transform? (10 scores)

(a) 先將等號兩邊做 Laplace Transform 再取 \log .

$$\text{e.g. } z(t) = x(t) * y(t) \Rightarrow Z(s) = X(s) \cdot Y(s) \Rightarrow \log(Z(s)) = \log(X(s)) + \log(Y(s))$$

(b) ① 運算上要處理複數乘法, 而一個複數乘法等同於四個實數乘法, 故運算量較大

② 運算上要處理無理數, 因為其值是 π 的倍數, 無理數的數位處理比較麻煩

(3) How do we implement $y[n] = x[n] * (0.8^n u[n] - 0.6^n u[n])$ efficiently where * means convolution and $u[n]$ is the unit step function? (10 scores)

$$\begin{aligned}
 H(z) &= \sum_{n=-\infty}^{\infty} h[n] z^{-n} = \sum_{n=-\infty}^{\infty} [0.8^n u[n] - 0.6^n u[n]] z^{-n} = \sum_{n=0}^{\infty} (0.8^n - 0.6^n) z^{-n} \\
 &= \sum_{n=0}^{\infty} 0.8^n z^{-n} - \sum_{n=0}^{\infty} 0.6^n z^{-n} \\
 &= \frac{1}{1 - 0.8z^{-1}} - \frac{1}{1 - 0.6z^{-1}} = \frac{1 - \frac{0.6}{z} - (1 - \frac{0.8}{z})}{(1 - \frac{0.8}{z})(1 - \frac{0.6}{z})} = \frac{\frac{0.2}{z}}{1 - \frac{1.4}{z} + \frac{0.48}{z^2}}
 \end{aligned}$$

$$Y(z) = X(z) \cdot H(z) = X(z) \cdot \frac{0.2z^{-1}}{1 - 1.4z^{-1} + 0.48z^{-2}}$$

$$Y(z) = X(z) + \left[1 - \frac{1 - 1.4z^{-1} + 0.48z^{-2}}{0.2z^{-1}} \right] Y(z)$$

$$= X(z) + \left[\frac{0.2z^{-1} - 1 + 1.4z^{-1} - 0.48z^{-2}}{0.2z^{-1}} \right] Y(z)$$

$$= X(z) + \frac{-0.48z^{-2} + 1.6z^{-1} - 1}{0.2z^{-1}} Y(z)$$

$$\Rightarrow 0.2z^{-1}Y(z) = 0.2z^{-1}X(z) - Y(z) + 1.6z^{-1}Y(z) - 0.48z^{-2}Y(z)$$

$$\Rightarrow Y(z) = 0.2z^{-1}X(z) + 1.4z^{-1}Y(z) - 0.48z^{-2}Y(z)$$

$$\Rightarrow \underline{y[n] = 0.2x[n-1] + 1.4y[n-1] - 0.48y[n-2]}_{\#}$$

(4) Why (a) the step invariance method and (b) the bilinear transform can reduce or avoid the aliasing effect in IIR filter design? (10 scores)

(a) Step invariance 透過積分的方式將高頻的能量壓下來, 故能降低常常在高頻部份出現的 aliasing effect.

(b) Bilinear transform 將整個 $-\infty \sim \infty$ 的頻域 mapping 到 $-\frac{f_s}{2}, \frac{f_s}{2}$ 之間, 使 aliasing effect 完全消失

(5) Suppose that $x[n] = y(0.002n)$ and the length of $x[n]$ is 2000. If $X[m]$ is the FFT of $x[n]$, which frequency do (a) $X[300]$ and (b) $X[1800]$ correspond to? (10 scores)

$$f_s = \frac{1}{0.002} = 500 \text{ Hz}, N = 2000 \quad \text{By: } f = m \cdot \frac{f_s}{N}$$

$$(a) f = 300 \cdot \frac{500}{2000} = 75 \text{ Hz} \#$$

$$(b) \text{ 因為 } 1800 > \frac{N}{2} = 1000, \text{ 故 } f = 1800 \cdot \frac{500}{2000} - 500 = -50 \text{ Hz} \#$$

(6) Suppose that we want to design a 25-point lowpass filter where $F < 0.25$ is the passband. Which one has the least error in the passband? $W(F)$ means the weight function. (10 scores)

(a) transition : $0.23 < F < 0.27$, $W(F) = 0.5$ for $F < 0.23$, $W(F) = 1$ for $F > 0.27$;

(b) transition : $0.2 < F < 0.3$, $W(F) = 2$ for $F < 0.2$, $W(F) = 2$ for $F > 0.3$;

☒ (c) transition : $0.2 < F < 0.3$, $W(F) = 1$ for $F < 0.2$, $W(F) = 0.5$ for $F > 0.3$;

(d) transition : $0.23 < F < 0.27$, $W(F) = 3$ for $F < 0.23$, $W(F) = 2$ for $F > 0.27$.

Ans: C # 因選項 C 有較寬的 transition band, 且在 pass band 的權重也比較高

(7) Use the MSE method to design the 5-point FIR filter that approximates the lowpass filter of $H_d(F) = 1$ for $|F| < 0.3$ and $H_d(F) = 0$ for $0.3 < |F| < 0.5$.

(10 scores)

By ADSP2 P.50

$$S[0] = \int_{-\frac{1}{2}}^{\frac{1}{2}} H_d(F) dF = 0.6$$

$$S[1] = 2 \int_{-\frac{1}{2}}^{\frac{1}{2}} \cos(2\pi n F) H_d(F) dF$$

$$S[1] = 2 \int_{-\frac{1}{2}}^{\frac{1}{2}} \cos(2\pi F) H_d(F) dF$$

$$S[2] = 2 \int_{-0.3}^{0.3} \cos(4\pi F) dF$$

$$= 2 \int_{-0.3}^{0.3} \cos(2\pi F) dF$$

$$= 2 \cdot \left[\sin(4\pi F) \cdot \frac{1}{4\pi} \right]_{-0.3}^{0.3}$$

$$= 2 \cdot \left[\sin(2\pi F) \cdot \frac{1}{2\pi} \right]_{-0.3}^{0.3}$$

$$= \frac{1}{\pi} \cdot \sin(1.2\pi)$$

$$= \frac{1}{\pi} [\sin(0.6\pi) - \sin(-0.6\pi)]$$

$$= \underline{-0.1871}$$

$$= \frac{2}{\pi} \cdot \sin(0.6\pi) = \underline{0.6055}$$

$$\begin{matrix} S[0] \\ 0.6 \end{matrix}$$

$$\begin{matrix} S[1] \\ 0.6055 \end{matrix}$$

$$\begin{matrix} S[2] \\ -0.1871 \end{matrix}$$

$$\frac{S[n]}{2} = r[n]$$

$$\begin{matrix} r[0] \\ 0.6 \end{matrix}$$

$$\begin{matrix} r[1] \\ 0.3025 \end{matrix}$$

$$\begin{matrix} r[2] \\ -0.09355 \end{matrix}$$

$$h[0] = -0.0935$$

$$h[1] = 0.3025$$

$$h[2] = 0.6$$

$$h[3] = 0.3025$$

$$h[4] = -0.0935$$

#

学号尾数2的 Bonus Question:

如果將 pass band 与 stop band 的 weight 都乘上某个常数, 是否會影響結果?

Ans: 不影响. # 這等同於在 A matrix 的最後一行乘上某個常數, 做 A^{-1} 之後, 結果还是一樣