

CENG 384 - Signals and Systems for Computer Engineers  
Spring 2023  
Homework 3

Karabacak, Kerem  
e2644417@ceng.metu.edu.tr

I worked individually.  
xxxxxxx@ceng.metu.edu.tr

May 14, 2023

1. We can start with synthesis equation:

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

We can take the integral and calculate  $b_k$  according to that.

$$\int_{-\infty}^t x(s) ds = \int_{-\infty}^t \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 s} ds$$

Now we can use analysis equation on this:

$$b_k = \frac{1}{T} \int_0^T \int_{-\infty}^t \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 s} e^{-jk\omega_0 t} ds dt$$

$$b_k = \frac{1}{T} \sum_{k=-\infty}^{\infty} a_k \int_0^T \int_{-\infty}^t e^{jk\omega_0 s} e^{-jk\omega_0 t} ds dt$$

$$b_k = \frac{1}{T} \sum_{k=-\infty}^{\infty} a_k \int_0^T \frac{1}{jk\omega_0} e^{jk\omega_0 t} e^{-jk\omega_0 t} dt$$

Exponential terms are cancelled.

$$b_k = \frac{1}{T} \sum_{k=-\infty}^{\infty} a_k \int_0^T \frac{1}{jk\omega_0} dt$$

$$b_k = \frac{1}{T} \sum_{k=-\infty}^{\infty} a_k \frac{T}{jk\omega_0}$$

$$b_k = \sum_{k=-\infty}^{\infty} a_k \frac{1}{jk\omega_0}$$

Finally, we know  $\omega_0 = \frac{2\pi}{T}$ .

$$b_k = \sum_{k=-\infty}^{\infty} a_k \frac{1}{jk \frac{2\pi}{T}}$$

2. (a) Let's denote  $x(t)x(t)$ 's coefficients as  $b_k$ .  
Using multiplication property of continuous signals:

$$b_k = \sum_{l=-\infty}^{\infty} a_{l-k} a_l = a_k * a_k$$

- (b) We can rewrite  $\text{Ev}\{x(t)\}$  as

$$\frac{x(t) + x(-t)}{2}$$

Let's denote this representation's coefficients as  $b_k$ .

$$b_k = \frac{a_k + a_{-k}}{2}$$

- (c) Let's denote  $x(t+t_0)$ 's coefficients as  $b_k$ ,  $x(t-t_0)$ 's coefficients as  $c_k$ , all equation's coefficients as  $d_k$ . Using time shifting property of continuous signals:

$$b_k = e^{jk\omega_0 t} a_k$$

$$c_k = e^{-jk\omega_0 t} a_k$$

$$d_k = b_k + c_k = e^{jk\omega_0 t} a_k + e^{-jk\omega_0 t} a_k$$

3. Let's denote period of system as  $T_1$ . Then -2 is  $-T_1$  and 2 is  $T_1$ . According to that  $x(t)$  is:

$$\begin{cases} 2 & 0 < t < \frac{T_1}{2} \\ -2 & -T_1 < t < -\frac{T_1}{2} \\ 0 & \text{otherwise} \end{cases}$$

Using analysis equation for CT signals:

$$a_k = \frac{1}{T} \int_{-T_1}^{T_1} x(t) e^{-jk\omega_0 t} dt = \frac{1}{T} \int_0^{T_1/2} 2e^{-jk\omega_0 t} dt + \frac{1}{T} \int_{-T_1}^{-T_1/2} -2e^{-jk\omega_0 t} dt$$

$$a_k = \frac{-2}{T} \left( \frac{e^{-jk\omega_0 T_1/2} - 1}{jk\omega_0} \right) + \frac{2}{T} \left( \frac{e^{jk\omega_0 T_1/2} + e^{jk\omega_0 T_1}}{jk\omega_0} \right)$$

$$a_k = \frac{2}{jk\omega_0 T} (e^{jk\omega_0 T_1/2} + e^{jk\omega_0 T_1} - e^{-jk\omega_0 T_1/2} + 1)$$

Arranging the equation to use Euler's formula:

$$a_k = \frac{4}{k\omega_0 T} \left( \frac{e^{jk\omega_0 T_1/2} - e^{-jk\omega_0 T_1/2}}{2j} + \frac{e^{jk\omega_0 T_1} + 1}{2j} \right)$$

Also, we can use  $\omega_0 = \frac{2\pi}{T}$ :

$$a_k = \frac{2}{k\pi} (\sin(k\omega_0 T_1/2) + \frac{\cos(k\omega_0 T_1) + j\sin(k\omega_0 T_1) + 1}{2j})$$

4. (a) We can rewrite  $x(t)$  with the help of Euler's formula as follows:

$$x(t) = 1 + \frac{1}{2j} e^{j\omega_0 t} - \frac{1}{2j} e^{-j\omega_0 t} + e^{j\omega_0 t} + e^{-j\omega_0 t} + \frac{1}{2} e^{j(2\omega_0 t + \frac{\pi}{4})} + \frac{1}{2} e^{-j(2\omega_0 t + \frac{\pi}{4})}$$

If we group them by  $k$  (for instance  $e^{j\omega_0 t}$  corresponds to  $a_1$  since  $k$  is 1 in formula  $e^{jk\omega_0 t}$ ), coefficients are:

$$a_0 = 1 \quad a_1 = 1 - \frac{j}{2} \quad a_{-1} = 1 + \frac{j}{2} \quad a_2 = \frac{1}{2\sqrt{2}} + \frac{1}{2\sqrt{2}}j \quad a_{-2} = \frac{1}{2\sqrt{2}} - \frac{1}{2\sqrt{2}}j$$

All other coefficients are 0.

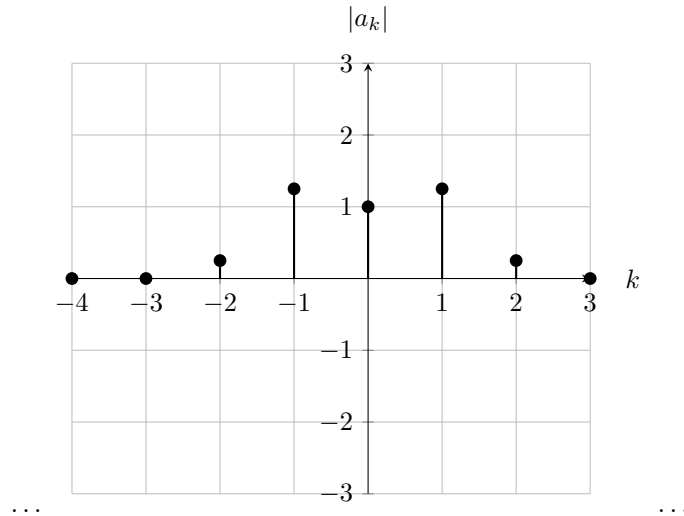


Figure 1:  $k$  vs.  $|a_k|$ .

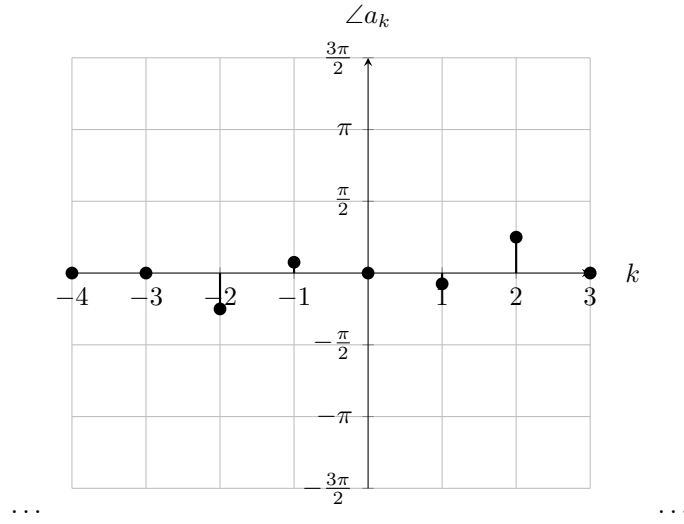


Figure 2:  $k$  vs.  $\angle a_k$ .

(b) We can set  $x(t) = e^{jw_0 t}$  for simplicity. Also, we know  $y(t) = H(jw_0)x(t)$ . Let's put those onto the formula given:

$$H(jw_0)x'(t) + H(jw_0)x(t) = x(t)$$

$$H(jw_0)e^{jw_0 t}jw_0 + H(jw_0)e^{jw_0 t} = e^{jw_0 t}$$

$e^{jw_0 t}$ 's are cancelled.

$$H(jw_0) = \frac{1}{1 + jw_0}$$

(c) Let's denote  $y(t)$ 's coefficients as  $b_k$ . Also we know:

$$b_k = a_k H(jw_0)$$

I took  $T$  as  $2\pi$  since  $x(t)$  only contains sine and cosine terms i.e.  $w_0 = 1$ . According to that:

$$b_{-2} = \left(\frac{1}{2\sqrt{2}} - \frac{1}{2\sqrt{2}}j\right)\left(\frac{1}{1+j}\right) = \frac{1-j}{2\sqrt{2} + 2\sqrt{2}j}$$

$$b_{-1} = \left(1 + \frac{j}{2}\right)\left(\frac{1}{1+j}\right) = \frac{2+j}{2+2j}$$

$$b_0 = \frac{1}{1+j}$$

$$b_1 = \left(1 - \frac{j}{2}\right)\left(\frac{1}{1+j}\right) = \frac{2-j}{2+2j}$$

$$b_{-2} = \left(\frac{1}{2\sqrt{2}} + \frac{1}{2\sqrt{2}}j\right)\left(\frac{1}{1+j}\right) = \frac{1+j}{2\sqrt{2} + 2\sqrt{2}j} = \frac{1}{2\sqrt{2}}$$

For all other values of  $k$ ,  $b_k$  is 0.

(d)

$$y(t) = \sum_{k=-\infty}^{\infty} b_k e^{jk w_0 t}$$

5. (a) For periodicity  $\frac{\pi}{2}n$  must be equal to  $2\pi$ . So we have period  $N = 4$ .

We can rewrite  $x[n]$  as follows:

$$x[n] = \frac{1}{2j}e^{j\pi n/2} - \frac{1}{2j}e^{-j\pi n/2}$$

As you can see first part is  $a_1$  and second part is  $a_{-1}$ . So

$$a_1 = \frac{1}{2j} \quad a_{-1} = \frac{-1}{2j}$$

Since  $x[n]$  is discrete signal,  $a_n = a_{n+N}$ . Therefore

$$a_{-3} = a_1 = a_5 = \frac{1}{2j} \quad a_{-5} = a_{-1} = a_3 = \frac{-1}{2j}$$

(b) For periodicity  $\frac{\pi}{2}n$  must be equal to  $2\pi$ . So we have period  $N = 4$ . We can rewrite  $y[n]$  as follows:

$$y[n] = 1 + \frac{1}{2}e^{j\frac{\pi}{2}n} + \frac{1}{2}e^{-j\frac{\pi}{2}n}$$

First part is  $b_0$ , second part is  $b_1$  and third part is  $b_{-1}$ .

$$b_0 = 1 \quad b_1 = \frac{1}{2} \quad b_{-1} = \frac{1}{2}$$

Since  $y[n]$  is discrete signal  $b_n = b_{n+N}$ . Therefore

$$b_{-4,0,4..} = 1 \quad b_{-3,1-5..} = \frac{1}{2} \quad b_{-5,-1,3..} = \frac{1}{2}$$

(c) We can directly use multiplication property:

$$c_k = \sum_{l=0}^3 a_l b_{k-l}$$

$$c_0 = 0 + \frac{1}{4j} + 0 - \frac{1}{4j} = 0$$

$$c_1 = 0 + \frac{1}{2j} + 0 + 0 = \frac{1}{2j}$$

$$c_2 = 0 + \frac{1}{4j} + 0 - \frac{1}{4j} = 0$$

$$c_3 = 0 + 0 + 0 - \frac{1}{2j} = \frac{-1}{2j}$$

And also periodicity rule is valid for this one as well.

(d) Multiplying both signals in analysis equation:

$$\frac{1}{4} \sum_{n=0}^3 \sin(\frac{\pi}{2}n) + \sin(\frac{\pi}{2}n)\cos(\frac{\pi}{2}n)e^{-jk\frac{\pi}{2}n} = \frac{1}{4} \sum_{n=0}^3 \sin(\frac{\pi}{2}n) + \frac{1}{2}\sin(\pi n)e^{-jk\frac{\pi}{2}n}$$

$$c_0 = \frac{1}{4}(0 + 1 + 0 - 1) = 0$$

$$c_1 = \frac{1}{4}(0 + e^{-j\frac{\pi}{2}} + 0 + e^{-j\frac{3\pi}{2}}) = \frac{-j}{2} = \frac{1}{2j}$$

$$c_2 = \frac{1}{4}(0 + e^{-j\pi} + 0 - e^{-j3\pi}) = 0$$

$$c_3 = \frac{1}{4}(0 + e^{-j3\pi/2} + 0 - e^{-j\pi/2}) = \frac{j}{2} = \frac{-1}{2j}$$

6. (a)  $x[n]$ 's period  $N$  is 4. We can directly use analysis equation to compute coefficients:

$$a_k = \frac{1}{4} \sum_{n=0}^3 x[n]e^{-jk\omega_0 n} = \frac{1}{4} \sum_{n=0}^3 x[n]e^{-jk\pi n/2}$$

$$a_0 = (0 + 1 + 2 + 1)/4 = 1$$

$$a_1 = (0 + e^{-j\pi/2} + 2e^{-j\pi} + e^{-j3\pi/2})/4 = -\frac{1}{2}$$

$$a_2 = (0 + e^{-j\pi} + 2e^{-j2\pi} + e^{-j3\pi})/4 = 0$$

$$a_3 = (0 + e^{-j3\pi/2} + 2e^{-j\pi} + e^{-j\pi/2})/4 = \frac{-1}{2}$$

Since  $x[n]$  is discrete signal,  $a_n = a_{n+N}$

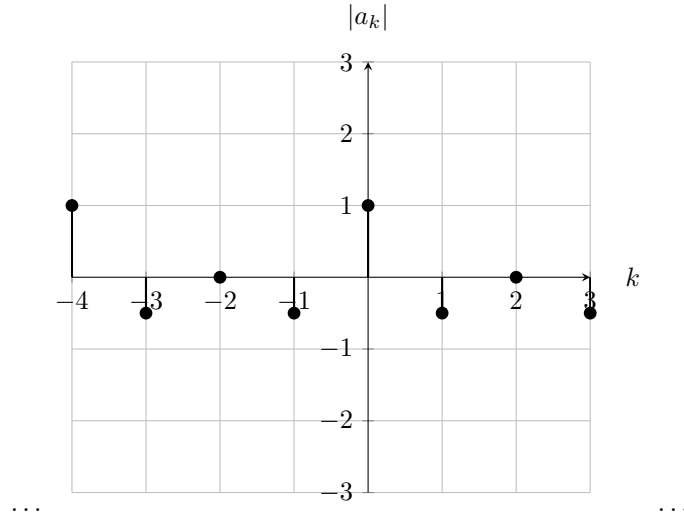


Figure 3:  $k$  vs.  $a_k$ .

(b) Difference between two signal is difference at the  $4k-1$  points, -5, -1, 3. So we can express  $y[n]$  as follows:

$$y[n] = x[n] - \sum_{k=-\infty}^{\infty} x[n] \delta[n - 4k + 1]$$

We can use synthesis equation:

$$b_k = \frac{1}{4} \sum_{n=0}^3 (x[n] - \sum_{k=-\infty}^{\infty} x[n] \delta[n - 4k + 1]) e^{-jkn\pi/2}$$

$$b_0 = \frac{1}{4} \sum_{n=0}^3 x[n] = (0 + 1 + 2 + 1)/4 = 1$$

$$b_1 = \frac{1}{4} \sum_{n=0}^3 x[n] e^{-jn\pi/2} = (0 - j - 2 + j)/4 = \frac{-1}{2}$$

$$b_2 = \frac{1}{4} \sum_{n=0}^3 x[n] e^{-j\pi n} = (-1 + 2 - 1)/4 = 0$$

$$b_3 = \frac{1}{4} \sum_{n=0}^3 e^{-j3\pi n/2} = (1 + j - 1 - j)/4 = 0$$

Since signal is discrete,  $b_n = b_{n+N}$ .

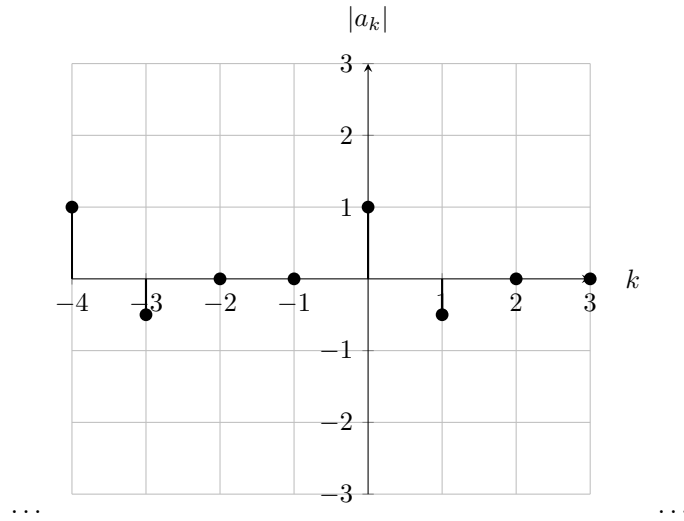


Figure 4:  $k$  vs.  $a_k$ .

7. (a)

$$b_k = \sum_{k=-\infty}^{\infty} a_k H(jw_0)$$

If  $x[n] = y[n]$ , then their coefficients must be equal. Since  $y[n]$ 's coefficients are non-zero at  $[-80, 80]$ ,  $a_k$  is zero when  $|k| > 80$ .

- (b) If  $x(t) \neq y(t)$ , then LTI system has some effect on input and output. Because of this  $a_k$  is non zero at least one point on  $|k| > 80$ .

8.