The Discrete Forrier Transform (OFT), the practical transform.

Suppose you have a finite tensith observation of XInj:

x(n) observed over n=0,1,..., N-1 and we assume it is a exergular else.

The DTFT would be

$$X(e^{j\omega}) = \sum_{n=0}^{N-1} x^{(n)} e^{j\omega n}$$

The DTFT is still not computationally practical since ω is a continuous variable. It is not possible to compute the DTFT for all values of ω .

Suppose we sample the w variable from 0 to 2th.

$$W_k = \frac{2\pi k}{N}$$
 $k = 0, 1, ..., N-1$

frequency

Samples

Now we have

Note: This will repeat periodically for values of k outside the o to NM vange.

This sampled form of the DTFT is practical in that it can be computed for k=9,2..., N-1.

However, it is a sampled form of the DTFT, but not actually the DIFT any more.

If it is a sampled form of the DTFT, why should we care if it is not exactly the DTFT?

The reason is the inverse transform.

 $X[k] = X(e^{jw_k}) = \sum_{N=0}^{N-1} \times [n] e^{j\frac{2\pi}{N!}k_N}$ $k=0,\ldots,N-1$ The inverse DFT is $X[h] = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}kn} & n=0,1,...,N-1 \\ \frac{1}{N} \sum_{k=0}^{N-1} X[k] & n=0,1,...,N-1 \end{cases}$ else

Notice that the inverse transform equation

I Not X(k) e j N kn

N k=0

is only valid over N=0,1...,N-1.

We have to force the result to be zero outside this range.

If we do not force the vesult to be zero and just blindly use the inverse equation, it will periodically repeat XCN with a period of N.

This periodicity is a result of. Sampling in the frequency domain.

So what can we observe?

- or analyze the frequency domain information about a signal, the DFT as a sampled form of the DTFT is fine and you really don't need to take any special care in doing this.
- the frequency domain information and then inverse transform this back to the time domain then it may be important to consider the fact that the inverse equation implicitly events a periodic time-domain signal.

Another difference between the DFT and the DFFT is a computational tool. We don't need to learn any transform tables or use partial fraction expansion. We compute the DFT and also can compute the inverse DFT (IDFT).

ex X [n]= {1, 1} N=2 N=0

 $X(k) = \sum_{n=0}^{l} x(n) e^{-j\frac{2\pi}{N}kn}$ k=0,1

note: $\frac{2T}{N} = \frac{2T}{2} = T$ $= \frac{2T}{N} = \frac{-j\pi kn}{e^{j\pi}} = -1$

$$Y(k) = \sum_{N=0}^{l} x(N) e^{-j\pi l} kn$$

$$= \sum_{N=0}^{l} x(N) (-1)^{kn} k^{-0,1}$$

$$= \sum_{N=0}^{l} x(N) (-1)^{kn} k^{-0,1}$$

$$X[a] = 1+1 = 2$$

 $X[i] = 1-1 = 0$

What toes this tell us?

First the k doesn't mean much by itself. It is a pointer or index of a frequency.

$$\omega_{k} = \frac{2\pi i}{N} k \qquad k = 0,1,...,N-1$$

$$k \text{ as } \alpha \text{ frequency index}$$

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So for our example N=2 $k=0 \longrightarrow W_k=0$ $k=1 \longrightarrow W_k=T$

So we only have samples at W=0 and w=T. That's a very sparse sampling of the frequencies. What about the vest of the frequencies?

There is a nice trick called zero-padding that gives us a lot more frequencies.

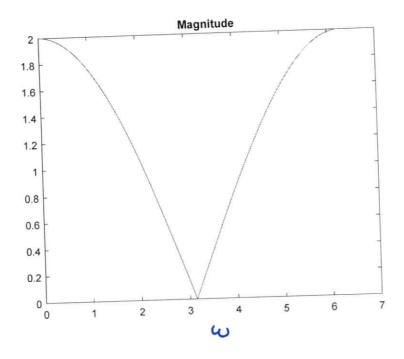
You add a bunch of zeroes onto the end of the signal, this increasing the size of N. This sives us more samples of W.

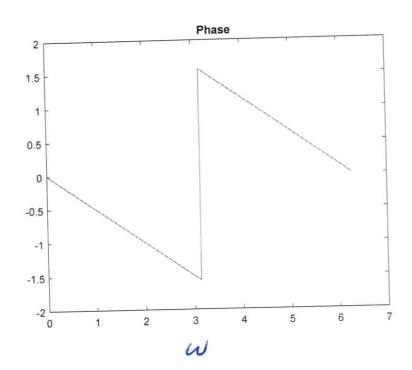
In matlab we use the command, fft, for computing the DFT. The fast formier transform (FFT) is a very fast algorithm for computing the DFT. Additionally, the FFT revolution ized computational signal processing.

For our example:

Set NFFT = 1024 to zero-pad so that the length of the padded signal is 1024. We use 1024 = 21° because it is a power of two, which is optimum for the FFT. x=[1 1]; X= fft (x, NFFT);

K=0: NFFT-1; Wk=2*pi/NFFT *k) figure(i), plot(wk, abs(*)) figure(2), plot(wk, ansle(X))





Note: The DFT/FFT gives results
over the O=WZZT interval.

The second half (right half) corresponds
to a replication of the negative frequencies