

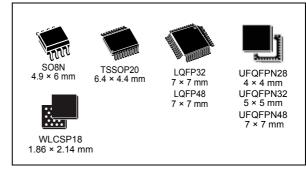
### STM32G031x4/x6/x8

Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ 32-bit MCU, up to 64 KB Flash, 8 KB RAM, 2x USART, timers, ADC, comm. I/Fs, 1.7-3.6V

Datasheet - production data

#### **Features**

- Core: Arm<sup>®</sup> 32-bit Cortex<sup>®</sup>-M0+ CPU, frequency up to 64 MHz
- -40°C to 85°C/105°C/125°C operating temperature
- Memories
  - Up to 64 Kbytes of Flash memory with protection and securable area
  - 8 Kbytes of SRAM with HW parity check
- CRC calculation unit
- Reset and power management
  - Voltage range: 1.7 V to 3.6 V
  - Power-on/Power-down reset (POR/PDR)
  - Programmable Brownout reset (BOR)
  - Programmable voltage detector (PVD)
  - Low-power modes:
     Sleep, Stop, Standby, Shutdown
  - V<sub>BAT</sub> supply for RTC and backup registers
- Clock management
  - 4 to 48 MHz crystal oscillator
  - 32 kHz crystal oscillator with calibration
  - Internal 16 MHz RC with PLL option (±1 %)
  - Internal 32 kHz RC oscillator (±5 %)
- Up to 44 fast I/Os
  - All mappable on external interrupt vectors
  - Multiple 5 V-tolerant I/Os
- 5-channel DMA controller with flexible mapping
- 12-bit, 0.4 µs ADC (up to 16 ext. channels)
  - Up to 16-bit with hardware oversampling
  - Conversion range: 0 to 3.6V
- 11 timers (one 128 MHz capable): 16-bit for advanced motor control, one 32-bit and four 16-bit general-purpose, two low-power 16-bit, two watchdogs, SysTick timer
- Calendar RTC with alarm and periodic wakeup from Stop/Standby/Shutdown



- Communication interfaces
  - Two I<sup>2</sup>C-bus interfaces supporting Fastmode Plus (1 Mbit/s) with extra current sink, one supporting SMBus/PMBus and wakeup from Stop mode
  - Two USARTs with master/slave synchronous SPI; one supporting ISO7816 interface, LIN, IrDA capability, auto baud rate detection and wakeup feature
  - One low-power UART
  - Two SPIs (32 Mbit/s) with 4- to 16-bit programmable bitframe, one multiplexed with I<sup>2</sup>S interface
- Development support: serial wire debug (SWD)
- 96-bit unique ID
- All packages ECOPACK 2 compliant

Table 1. Device summary

Reference	Part number
STM32G031x4	STM32G031C4, STM32G031F4, STM32G031G4, STM32G031K4, STM32G031J4
STM32G031x6	STM32G031C6, STM32G031F6, STM32G031G6, STM32G031K6, STM32G031J6
STM32G031x8	STM32G031C8, STM32G031F8, STM32G031G8, STM32G031K8, STM32G031Y8

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STM32G031x4/x6/x8 Introduction

### 1 Introduction

This document provides information on STM32G031x4/x6/x8 microcontrollers, such as description, functional overview, pin assignment and definition, electrical characteristics, packaging, and ordering codes.

Information on memory mapping and control registers is object of reference manual. Information on Arm<sup>®</sup>(a) Cortex<sup>®</sup>-M0+ core is available from the www.arm.com website.

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Description STM32G031x4/x6/x8

### 2 Description

The STM32G031x4/x6/x8 mainstream microcontrollers are based on high-performance Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ 32-bit RISC core operating at up to 64 MHz frequency. Offering a high level of integration, they are suitable for a wide range of applications in consumer, industrial and appliance domains and ready for the Internet of Things (IoT) solutions.

The devices incorporate a memory protection unit (MPU), high-speed embedded memories (8 Kbytes of SRAM and up to 64 Kbytes of Flash program memory with read protection, write protection, proprietary code protection, and securable area), DMA, an extensive range of system functions, enhanced I/Os, and peripherals. The devices offer standard communication interfaces (two I<sup>2</sup>Cs, two SPIs / one I<sup>2</sup>S, and two USARTs), one 12-bit ADC (2.5 MSps) with up to 19 channels, an internal voltage reference buffer, a low-power RTC, an advanced control PWM timer running at up to double the CPU frequency, four general-purpose 16-bit timers, a 32-bit general-purpose timer, two low-power 16-bit timers, two watchdog timers, and a SysTick timer.

The devices operate within ambient temperatures from -40 to 125°C and with supply voltages from 1.7 V to 3.6 V. Optimized dynamic consumption combined with a comprehensive set of power-saving modes, low-power timers and low-power UART, allows the design of low-power applications.

VBAT direct battery input allows keeping RTC and backup registers powered.

The devices come in packages with 8 to 48 pins.

Table 2. STM32G031x4/x6/x8 family device features and peripheral counts

Peripheral								STN	/32G0	31_						
		_J4	_J6	_Y8	_F4	_F6	_F8	_G4	_G6	_G8	_K4	_K6	_K8	_C4	_C6	_C8
Flas	sh memory (Kbyte)	16	32	64	16	32	64	16	32	64	16	32	64	16	32	64
	SRAM (Kbyte)							8 w	ith pa	rity						
	Advanced control						1 (	16-bit)	high f	reque	псу					
s	General-purpose						4	(16-b	it) + 1	(32-bi	t)					
Timers	Low-power		2 (16-bit)													
⊢	SysTick		1													
	Watchdog		2													
ses	SPI [I2S] <sup>(1)</sup>								2 [1]							
erfa	I2C		2													
r.	USART								2							
Comm. interfaces	LPUART								1							
	RTC		Yes													
	Tamper pins								2							
F	Random number generator								No							

STM32G031x4/x6/x8 Description

Table 2. STM32G031x4/x6/x8 family device features and peripheral counts (continued)

Davimbanal		STM32G031_													
Peripheral	_J4	_J6	_Y8	_F4	_F6	_F8	_G4	_G6	_G8	_K4	_K6	_K8	_C4	_C6	_C8
AES								No							
GPIOs	(	6	16		18		26			30			44		
Wakeup pins	;	3							4						
12-bit ADC channels (external + internal)	6 -	+ 2		14 + 2			15 + 2			16 + 2			16 + 3		
Internal voltage reference buffer						N	lo							Yes	
Max. CPU frequency							6	64 MH	Z						
Operating voltage							1.7	' to 3.6	6 V						
Operating temperature <sup>(2)</sup>		Ambient: -40 to 85 °C / -40 to 105 °C / -40 to 125 °C Junction: -40 to 105 °C / -40 to 125 °C / -40 to 130 °C													
Number of pins	8	3	18		20	•	28				32	•		48	

<sup>1.</sup> The numbers in brackets denote the count of SPI interfaces configurable as  $I^2S$  interface.

<sup>2.</sup> Depends on order code. Refer to Section 7: Ordering information for details.

Description STM32G031x4/x6/x8

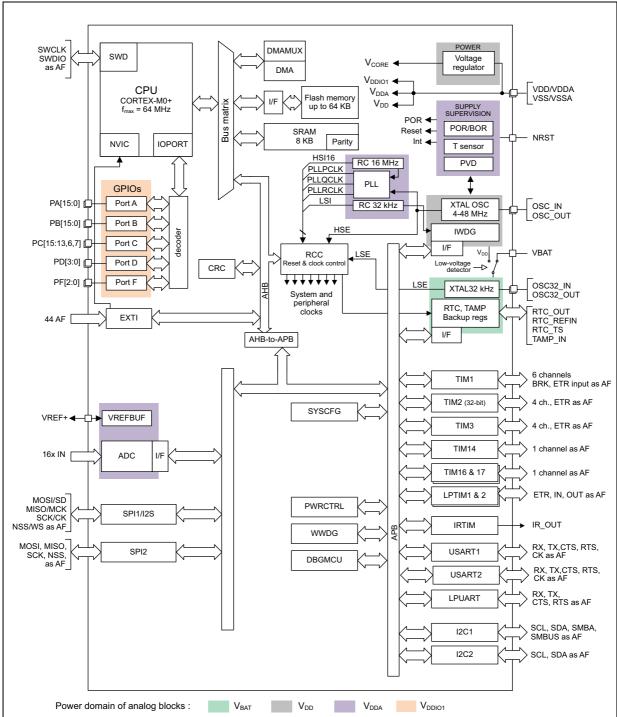


Figure 1. Block diagram



### 3 Functional overview

# 3.1 Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ core with MPU

The Cortex-M0+ is an entry-level 32-bit Arm Cortex processor designed for a broad range of embedded applications. It offers significant benefits to developers, including:

- a simple architecture, easy to learn and program
- ultra-low power, energy-efficient operation
- excellent code density
- deterministic, high-performance interrupt handling
- upward compatibility with Cortex-M processor family
- platform security robustness, with integrated Memory Protection Unit (MPU).

The Cortex-M0+ processor is built on a highly area- and power-optimized 32-bit core, with a 2-stage pipeline Von Neumann architecture. The processor delivers exceptional energy efficiency through a small but powerful instruction set and extensively optimized design, providing high-end processing hardware including a single-cycle multiplier.

The Cortex-M0+ processor provides the exceptional performance expected of a modern 32-bit architecture, with a higher code density than other 8-bit and 16-bit microcontrollers.

Owing to embedded Arm core, the STM32G031x4/x6/x8 devices are compatible with Arm tools and software.

The Cortex-M0+ is tightly coupled with a nested vectored interrupt controller (NVIC) described in Section 3.13.1.

### 3.2 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

### 3.3 Embedded Flash memory

STM32G031x4/x6/x8 devices feature up to 64 Kbytes of embedded Flash memory available for storing code and data.

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Flexible protections can be configured thanks to option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels are available:
  - Level 0: no readout protection
  - Level 1: memory readout protection: the Flash memory cannot be read from or written to if either debug features are connected, boot in RAM or bootloader is selected
  - Level 2: chip readout protection: debug features (Cortex-M0+ serial wire), boot in RAM and bootloader selection are disabled. This selection is irreversible.

Table	J. ACCESS S	tatus vers	us reaudui	protection	level allu t	execution i	iioues			
Area	Protection	U	ser execution	on	Debug, boot from RAM or boot from system memory (loader)					
	level	Read	Write	Erase	Read	Write	Erase			
User	1	Yes	Yes	Yes	No	No	No			
memory	2	Yes	Yes	Yes	N/A	N/A	N/A			
System	1	Yes	No	No	Yes	No	No			
memory	2	Yes	No	No	N/A	N/A	N/A			
Option	1	Yes	Yes	Yes	Yes	Yes	Yes			
bytes	2	Yes	No	No	N/A	N/A	N/A			
Backup	1	Yes	Yes	N/A <sup>(1)</sup>	No	No	N/A <sup>(1)</sup>			
registers	2	Yes	Yes	N/A	N/A	N/A	N/A			

Table 3. Access status versus readout protection level and execution modes

- Write protection (WRP): the protected area is protected against erasing and programming. Two areas per bank can be selected, with 2-Kbyte granularity.
- Proprietary code readout protection (PCROP): a part of the Flash memory can be
  protected against read and write from third parties. The protected area is execute-only:
  it can only be reached by the STM32 CPU as instruction code, while all other accesses
  (DMA, debug and CPU data read, write and erase) are strictly prohibited. An additional
  option bit (PCROP\_RDP) determines whether the PCROP area is erased or not when
  the RDP protection is changed from Level 1 to Level 0.

The whole non-volatile memory embeds the error correction code (ECC) feature supporting:

- single error detection and correction
- double error detection
- readout of the ECC fail address from the ECC register

#### 3.3.1 Securable area

A part of the Flash memory can be hidden from the application once the code it contains is executed. As soon as the write-once SEC\_PROT bit is set, the securable memory cannot be accessed until the system resets. The securable area generally contains the secure boot code to execute only once at boot. This helps to isolate secret code from untrusted application code.

<sup>1.</sup> Erased upon RDP change from Level 1 to Level 0.

### 3.4 Embedded SRAM

STM32G031x4/x6/x8 devices have 8 Kbytes of embedded SRAM with parity. Hardware parity check allows memory data errors to be detected, which contributes to increasing functional safety of applications.

The memory can be read/write-accessed at CPU clock speed, with 0 wait states.

### 3.5 Boot modes

At startup, the boot pin and boot selector option bit are used to select one of the three boot options:

- boot from User Flash memory
- boot from System memory
- boot from embedded SRAM

The boot pin is shared with a standard GPIO and can be enabled through the boot selector option bit. The boot loader is located in System memory. It manages the Flash memory reprogramming through one of the following interfaces:

- USART on pins PA9/PA10 or PA2/PA3
- I<sup>2</sup>C-bus on pins PB6/PB7 or PB10/PB11

### 3.6 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link time and stored at a given memory location.

### 3.7 Power supply management

#### 3.7.1 Power supply schemes

The STM32G031x4/x6/x8 devices require a 1.7 V to 3.6 V operating supply voltage ( $V_{DD}$ ). Several different power supplies are provided to specific peripherals:

V<sub>DD</sub> = 1.7 (1.6) to 3.6 V

 $V_{DD}$  is the external power supply for the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through VDD/VDDA pin.

The minimum voltage of 1.7 V corresponds to power-on reset release threshold  $V_{POR}(max)$ . Once this threshold is crossed and power-on reset is released, the functionality is guaranteed down to power-down reset threshold  $V_{PDR}(min)$ .

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V<sub>DDA</sub> = 1.62 V (ADC) / 2.4 V (VREFBUF) to 3.6 V

 $V_{DDA}$  is the analog power supply for the A/D converter, voltage reference buffer.  $V_{DDA}$  voltage level is identical to  $V_{DD}$  voltage as it is provided externally through VDD/VDDA pin.

•  $V_{DDIO1} = V_{DD}$ 

 $V_{DDIO1}$  is the power supply for the I/Os.  $V_{DDIO1}$  voltage level is identical to  $V_{DD}$  voltage as it is provided externally through VDD/VDDA pin.

- V<sub>BAT</sub> = 1.55 V to 3.6 V. V<sub>BAT</sub> is the power supply (through a power switch) for RTC, TAMP, low-speed external 32.768 kHz oscillator and backup registers when V<sub>DD</sub> is not present. V<sub>BAT</sub> is provided externally through VBAT pin. When this pin is not available on the package, VBAT bonding pad is internally bonded to the VDD/VDDA pin.
- $V_{REF+}$  is the analog peripheral input reference voltage, or the output of the internal voltage reference buffer (when enabled). When  $V_{DDA} < 2 \text{ V}$ ,  $V_{REF+}$  must be equal to  $V_{DDA}$ . When  $V_{DDA} \ge 2 \text{ V}$ ,  $V_{REF+}$  must be between 2 V and  $V_{DDA}$ . It can be grounded when the analog peripherals using  $V_{REF+}$  are not active.

The internal voltage reference buffer supports two output voltages, which is configured with VRS bit of the VREFBUF CSR register:

- V<sub>REF+</sub> around 2.048 V (requiring V<sub>DDA</sub> equal to or higher than 2.4 V)
- V<sub>REF+</sub> around 2.5 V (requiring V<sub>DDA</sub> equal to or higher than 2.8 V)

 $V_{REF+}$  is delivered through VREF+ pin. On packages without VREF+ pin,  $V_{REF+}$  is internally connected with  $V_{DD}$ , and the internal voltage reference buffer must be kept disabled (refer to datasheets for package pinout description).

V<sub>CORE</sub>

An embedded linear voltage regulator is used to supply the  $V_{CORE}$  internal digital power.  $V_{CORE}$  is the power supply for digital peripherals, SRAM and Flash memory. The Flash memory is also supplied with  $V_{DD}$ .

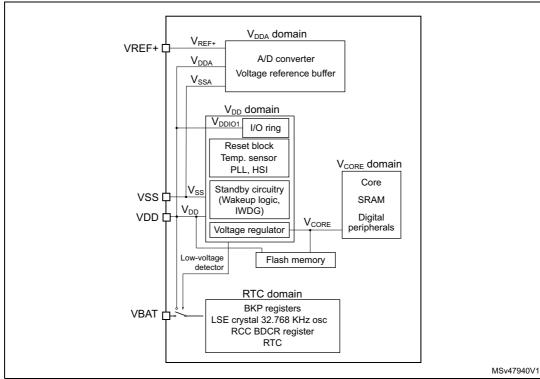


Figure 2. Power supply overview

### 3.7.2 Power supply supervisor

The device has an integrated power-on/power-down (POR/PDR) reset active in all power modes except Shutdown and ensuring proper operation upon power-on and power-down. It maintains the device in reset when the supply voltage is below  $V_{POR/PDR}$  threshold, without the need for an external reset circuit. Brownout reset (BOR) function allows extra flexibility. It can be enabled and configured through option bytes, by selecting one of four thresholds for rising  $V_{DD}$  and other four for falling  $V_{DD}$ .

The device also features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}$  power supply and compares it to  $V_{PVD}$  threshold. It allows generating an interrupt when  $V_{DD}$  level crosses the  $V_{PVD}$  threshold, selectively while falling, while rising, or while falling and rising. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

### 3.7.3 Voltage regulator

Two embedded linear voltage regulators, main regulator (MR) and low-power regulator (LPR), supply most of digital circuitry in the device.

The MR is used in Run and Sleep modes. The LPR is used in Low-power run, Low-power sleep and Stop modes.

In Standby and Shutdown modes, both regulators are powered down and their outputs set in high-impedance state, such as to bring their current consumption close to zero. However, SRAM data retention is possible in Standby mode, in which case the LPR remains active and it only supplies the SRAM.



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#### 3.7.4 Low-power modes

By default, the microcontroller is in Run mode after system or power reset. It is up to the user to select one of the low-power modes described below:

#### Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

#### Low-power run mode

This mode is achieved with  $V_{CORE}$  supplied by the low-power regulator to minimize the regulator's operating current. The code can be executed from SRAM or from Flash, and the CPU frequency is limited to 2 MHz. The peripherals with independent clock can be clocked by HSI16.

#### • Low-power sleep mode

This mode is entered from the low-power run mode. Only the CPU clock is stopped. When wakeup is triggered by an event or an interrupt, the system reverts to the Low-power run mode.

#### Stop 0 and Stop 1 modes

In Stop 0 and Stop 1 modes, the device achieves the lowest power consumption while retaining the SRAM and register contents. All clocks in the  $V_{CORE}$  domain are stopped. The PLL, as well as the HSI16 RC oscillator and the HSE crystal oscillator are disabled. The LSE or LSI keep running. The RTC can remain active (Stop mode with RTC, Stop mode without RTC).

Some peripherals with wakeup capability can enable the HSI16 RC during Stop mode, so as to get clock for processing the wakeup event. The main regulator remains active in Stop 0 mode while it is turned off in Stop 1 mode.

#### Standby mode

The Standby mode is used to achieve the lowest power consumption, with POR/PDR always active in this mode. The main regulator is switched off to power down  $V_{CORE}$  domain. The low-power regulator is either switched off or kept active. In the latter case, it only supplies SRAM to ensure data retention. The PLL, as well as the HSI16 RC oscillator and the HSE crystal oscillator are also powered down. The RTC can remain active (Standby mode with RTC, Standby mode without RTC).

For each I/O, the software can determine whether a pull-up, a pull-down or no resistor shall be applied to that I/O during Standby mode.

Upon entering Standby mode, register contents are lost except for registers in the RTC domain and standby circuitry. The SRAM contents can be retained through register setting.

The device exits Standby mode upon external reset event (NRST pin), IWDG reset event, wakeup event (WKUP pin, configurable rising or falling edge) or RTC event (alarm, periodic wakeup, timestamp, tamper), or when a failure is detected on LSE (CSS on LSE).

### Shutdown mode

The Shutdown mode allows to achieve the lowest power consumption. The internal regulator is switched off to power down the  $V_{CORF}$  domain. The PLL, as well as the



HSI16 and LSI RC-oscillators and HSE crystal oscillator are also powered down. The RTC can remain active (Shutdown mode with RTC, Shutdown mode without RTC).

The BOR is not available in Shutdown mode. No power voltage monitoring is possible in this mode. Therefore, switching to RTC domain is not supported.

SRAM and register contents are lost except for registers in the RTC domain.

The device exits Shutdown mode upon external reset event (NRST pin), IWDG reset event, wakeup event (WKUP pin, configurable rising or falling edge) or RTC event (alarm, periodic wakeup, timestamp, tamper).

#### 3.7.5 Reset mode

During and upon exiting reset, the schmitt triggers of I/Os are disabled so as to reduce power consumption. In addition, when the reset source is internal, the built-in pull-up resistor on NRST pin is deactivated.

### 3.7.6 VBAT operation

The V<sub>BAT</sub> power domain, consuming very little energy, includes RTC, and LSE oscillator and backup registers.

In VBAT mode, the RTC domain is supplied from VBAT pin. The power source can be, for example, an external battery or an external supercapacitor. Two anti-tamper detection pins are available.

The RTC domain can also be supplied from VDD/VDDA pin.

By means of a built-in switch, an internal voltage supervisor allows automatic switching of RTC domain powering between  $V_{DD}$  and voltage from VBAT pin to ensure that the supply voltage of the RTC domain ( $V_{BAT}$ ) remains within valid operating conditions. If both voltages are valid, the RTC domain is supplied from VDD/VDDA pin.

An internal circuit for charging the battery on VBAT pin can be activated if the  $V_{DD}$  voltage is within a valid range.

Note: External interrupts and RTC alarm/events cannot cause the microcontroller to exit the VBAT mode, as in that mode the  $V_{DD}$  is not within a valid range.

### 3.8 Interconnect of peripherals

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources thus power supply consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep and Stop modes.

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**Table 4. Interconnect of peripherals** 

Interconnect source	Interconnect destination	Interconnect action	Run Low-power run	Sleep Low-power sleep	Stop
	TIMx	Timer synchronization or chaining	Υ	Υ	-
TIMx	ADCx	Conversion triggers	Υ	Υ	-
TIIVIX	DMA	Memory-to-memory transfer trigger	Υ	Υ	-
ADCx	TIM1	Υ	Υ	-	
	TIM16	Timer input channel from RTC events	Υ	Υ	-
RTC	LPTIMERx	Low-power timer triggered by RTC alarms or tampers	Y	Υ	Υ
All clock sources (internal and external)	TIM14,16,17	Clock source used as input channel for RC measurement and trimming	Y	Y	-
CSS RAM (parity error) Flash memory (ECC error) PVD	TIM1,16,17	Timer break	Y	Y	-
CPU (hard fault)	TIM1,16,17	Timer break	Y	-	-
O.D.I.O.	TIMx	External trigger	Υ	Υ	-
GPIO	LPTIMERx	External trigger	Υ	Υ	Υ
	ADC	Conversion external trigger	Y	Y	-

### 3.9 Clocks and startup

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

- Clock prescaler: to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- **Safe clock switching:** clock sources can be changed safely on the fly in run mode through a configuration register.
- Clock management: to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.

- System clock source: three different sources can deliver SYSCLK system clock:
  - 4-48 MHz high-speed oscillator with external crystal or ceramic resonator (HSE). It can supply clock to system PLL. The HSE can also be configured in bypass mode for an external clock.
  - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software. It can supply clock to system PLL.
  - System PLL with maximum output frequency of 64 MHz. It can be fed with HSE or HSI16 clocks.
- Auxiliary clock source: two ultra-low-power clock sources for the real-time clock (RTC):
  - 32.768 kHz low-speed oscillator with external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for using an external clock.
  - 32 kHz low-speed internal RC oscillator (LSI) with ±5% accuracy, also used to clock an independent watchdog.
- **Peripheral clock sources:** several peripherals (I2S, USARTs, I2Cs, LPTIMs, ADC) have their own clock independent of the system clock.
- Clock security system (CSS): in the event of HSE clock failure, the system clock is automatically switched to HSI16 and, if enabled, a software interrupt is generated. LSE clock failure can also be detected and generate an interrupt. The CCS feature can be enabled by software.
- Clock output:
  - MCO (microcontroller clock output) provides one of the internal clocks for external use by the application
  - LSCO (low speed clock output) provides LSI or LSE in all low-power modes (except in VBAT operation).

Several prescalers allow the application to configure AHB and APB domain clock frequencies, 64 MHz at maximum.

### 3.10 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function (AF). Most of the GPIO pins are shared with special digital or analog functions.

Through a specific sequence, this special function configuration of I/Os can be locked, such as to avoid spurious writing to I/O control registers.

### 3.11 Direct memory access controller (DMA)

The direct memory access (DMA) controller is a bus master and system peripheral with single-AHB architecture.

With 5 channels, it performs data transfers between memory-mapped peripherals and/or memories, to offload the CPU.

Each channel is dedicated to managing memory access requests from one or more peripherals. The unit includes an arbiter for handling the priority between DMA requests.

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Main features of the DMA controller:

- Single-AHB master
- Peripheral-to-memory, memory-to-peripheral, memory-to-memory and peripheral-toperipheral data transfers
- Access, as source and destination, to on-chip memory-mapped devices such as Flash memory, SRAM, and AHB and APB peripherals
- All DMA channels independently configurable:
  - Each channel is associated either with a DMA request signal coming from a peripheral, or with a software trigger in memory-to-memory transfers. This configuration is done by software.
  - Priority between the requests is programmable by software (four levels per channel: very high, high, medium, low) and by hardware in case of equality (such as request to channel 1 has priority over request to channel 2).
  - Transfer size of source and destination are independent (byte, half-word, word), emulating packing and unpacking. Source and destination addresses must be aligned on the data size.
  - Support of transfers from/to peripherals to/from memory with circular buffer management
  - Programmable number of data to be transferred: 0 to 2<sup>16</sup> 1
- Generation of an interrupt request per channel. Each interrupt request originates from any of the three DMA events: transfer complete, half transfer, or transfer error.

#### 3.12 DMA request multiplexer (DMAMUX)

The DMAMUX request multiplexer enables routing a DMA request line between the peripherals and the DMA controller. Each channel selects a unique DMA request line, unconditionally or synchronously with events from its DMAMUX synchronization inputs. DMAMUX may also be used as a DMA request generator from programmable events on its input trigger signals.

#### 3.13 Interrupts and events

The device flexibly manages events causing interrupts of linear program execution, called exceptions. The Cortex-M0+ processor core, a nested vectored interrupt controller (NVIC) and an extended interrupt/event controller (EXTI) are the assets contributing to handling the exceptions. Exceptions include core-internal events such as, for example, a division by zero and, core-external events such as logical level changes on physical lines. Exceptions result in interrupting the program flow, executing an interrupt service routine (ISR) then resuming the original program flow.

The processor context (contents of program pointer and status registers) is stacked upon program interrupt and unstacked upon program resume, by hardware. This avoids context stacking and unstacking in the interrupt service routines (ISRs) by software, thus saving time, code and power. The ability to abandon and restart load-multiple and store-multiple operations significantly increases the device's responsiveness in processing exceptions.

### 3.13.1 Nested vectored interrupt controller (NVIC)

The configurable nested vectored interrupt controller is tightly coupled with the core. It handles physical line events associated with a non-maskable interrupt (NMI) and maskable interrupts, and Cortex-M0+ exceptions. It provides flexible priority management.

The tight coupling of the processor core with NVIC significantly reduces the latency between interrupt events and start of corresponding interrupt service routines (ISRs). The ISR vectors are listed in a vector table, stored in the NVIC at a base address. The vector address of an ISR to execute is hardware-built from the vector table base address and the ISR order number used as offset.

If a higher-priority interrupt event happens while a lower-priority interrupt event occurring just before is waiting for being served, the later-arriving higher-priority interrupt event is served first. Another optimization is called tail-chaining. Upon a return from a higher-priority ISR then start of a pending lower-priority ISR, the unnecessary processor context unstacking and stacking is skipped. This reduces latency and contributes to power efficiency.

#### Features of the NVIC:

- Low-latency interrupt processing
- 4 priority levels
- Handling of a non-maskable interrupt (NMI)
- Handling of 32 maskable interrupt lines
- Handling of 10 Cortex-M0+ exceptions
- Later-arriving higher-priority interrupt processed first
- Tail-chaining
- Interrupt vector retrieval by hardware

### 3.13.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller adds flexibility in handling physical line events and allows identifying wake-up events at processor wakeup from Stop mode.

The EXTI controller has a number of channels, of which some with rising, falling or rising, and falling edge detector capability. Any GPIO and a few peripheral signals can be connected to these channels.

The channels can be independently masked.

The EXTI controller can capture pulses shorter than the internal clock period.

A register in the EXTI controller latches every event even in Stop mode, which allows the software to identify the origin of the processor's wake-up from Stop mode or, to identify the GPIO and the edge event having caused an interrupt.

### 3.14 Analog-to-digital converter (ADC)

A native 12-bit analog-to-digital converter is embedded into STM32G031x4/x6/x8 devices. It can be extended to 16-bit resolution through hardware oversampling. The ADC has up to 16 external channels and 3 internal channels (temperature sensor, voltage reference,  $V_{BAT}$  monitoring). It performs conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.



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The ADC frequency is independent from the CPU frequency, allowing maximum sampling rate of ~2 MSps even with a low CPU speed. An auto-shutdown function guarantees that the ADC is powered off except during the active conversion phase.

The ADC can be served by the DMA controller. It can operate in the whole  $V_{DD}$  supply range.

The ADC features a hardware oversampler up to 256 samples, improving the resolution to 16 bits (refer to AN2668).

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions with timers.

#### 3.14.1 Temperature sensor

The temperature sensor (TS) generates a voltage V<sub>TS</sub> that varies linearly with temperature.

The temperature sensor is internally connected to an ADC input to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor may vary from part to part due to process variation, the uncalibrated internal temperature sensor is suitable only for relative temperature measurements.

To improve the accuracy of the temperature sensor, each part is individually factory-calibrated by ST. The resulting calibration data are stored in the part's engineering bytes, accessible in read-only mode.

Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at a temperature of 130 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75CA - 0x1FFF 75CB

Table 5. Temperature sensor calibration values

### 3.14.2 Internal voltage reference (V<sub>REFINT</sub>)

The internal voltage reference ( $V_{REFINT}$ ) provides a stable (bandgap) voltage output for the ADC.  $V_{REFINT}$  is internally connected to an ADC input. The  $V_{REFINT}$  voltage is individually precisely measured for each part by ST during production test and stored in the part's engineering bytes. It is accessible in read-only mode.

Table 6. Internal voltage reference calibration values

Calibration value name	Description	Memory address
V <sub>DEFINIT</sub>	Raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = V <sub>REF+</sub> = 3.0 V (± 10 mV)	0x1FFF 75AA - 0x1FFF 75AB

### 3.14.3 V<sub>BAT</sub> battery voltage monitoring

This embedded hardware feature allows the application to measure the  $V_{BAT}$  battery voltage using an internal ADC input. As the  $V_{BAT}$  voltage may be higher than  $V_{DDA}$  and thus outside the ADC input range, the VBAT pin is internally connected to a bridge divider by three. As a consequence, the converted digital value is one third the  $V_{BAT}$  voltage.

### 3.15 Voltage reference buffer (VREFBUF)

When enabled, an embedded buffer provides the internal reference voltage to analog blocks (for example ADC) and to VREF+ pin for external components.

The internal voltage reference buffer supports two voltages:

- 2.048 V
- 2.5 V

An external voltage reference can be provided through the VREF+ pin when the internal voltage reference buffer is disabled.

On some packages, the VREF+ pad of the silicon die is double-bonded with supply pad to common VDD/VDDA pin and so the internal voltage reference buffer cannot be used.

### 3.16 Timers and watchdogs

The device includes an advanced-control timer, five general-purpose timers, two low-power timers, two watchdog timers and a SysTick timer. *Table 7* compares features of the advanced-control, general-purpose and basic timers.

Table 7. Timer feature comparison

Timer type	Timer	er Counter Counter operating		Maximum operating frequency	Prescaler factor	DMA request generation	Capture/ compare channels	Comple- mentary outputs
Advanced- control	TIM1	16-bit	Up, down, up/down	128 MHz	Integer from 1 to 2 <sup>16</sup>	Yes	4	3

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Timer type	Timer Counter resolution Counter type Maximum operating factor		Prescaler factor	DMA request generation	Capture/ compare channels	Comple- mentary outputs		
	TIM2	32-bit	Up, down, up/down	64 MHz	Integer from 1 to 2 <sup>16</sup>	Yes	4	-
	TIM3	16-bit	Up, down, up/down	64 MHz	Integer from 1 to 2 <sup>16</sup>	Yes	4	-
General- purpose	TIM14	16-bit	Up	64 MHz	Integer from 1 to 2 <sup>16</sup>	No	1	-
	TIM16 TIM17	16-bit	Up	64 MHz	Integer from 1 to 2 <sup>16</sup>	Yes	1	1
Low-power	LPTIM1 LPTIM2	16-bit	Up	64 MHz	2 <sup>n</sup> where n=0 to 7	No	N/A	-

Table 7. Timer feature comparison (continued)

### 3.16.1 Advanced-control timer (TIM1)

The advanced-control timer can be seen as a three-phase PWM unit multiplexed on 6 channels. It has complementary PWM outputs with programmable inserted dead-times. It can also be seen as a complete general-purpose timer. The four independent channels can be used for:

- input capture
- output compare
- PWM output (edge or center-aligned modes) with full modulation capability (0-100%)
- one-pulse mode output

In debug mode, the advanced-control timer counter can be frozen and the PWM outputs disabled, so as to turn off any power switches driven by these outputs.

Many features are shared with those of the general-purpose TIMx timers (described in Section 3.16.2) using the same architecture, so the advanced-control timers can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

### 3.16.2 General-purpose timers (TIM2, 3, 14, 16, 17)

There are five synchronizable general-purpose timers embedded in the device (refer to *Table 7* for comparison). Each general-purpose timer can be used to generate PWM outputs or act as a simple timebase.

TIM2, TIM3

These are full-featured general-purpose timers:

- TIM2 with 32-bit auto-reload up/downcounter and 16-bit prescaler
- TIM3 with 16-bit auto-reload up/downcounter and 16-bit prescaler

They have four independent channels for input capture/output compare, PWM or onepulse mode output. They can operate in combination with other general-purpose timers via the Timer Link feature for synchronization or event chaining. They can generate

independent DMA request and support quadrature encoders. Their counter can be frozen in debug mode.

TIM14

This timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler. It has one channel for input capture/output compare, PWM output or one-pulse mode output. Its counter can be frozen in debug mode.

TIM16. TIM17

These are general-purpose timers featuring:

- 16-bit auto-reload upcounter and 16-bit prescaler
- 1 channel and 1 complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output. The timers can operate together via the Timer Link feature for synchronization or event chaining. They can generate independent DMA request. Their counters can be frozen in debug mode.

### 3.16.3 Low-power timers (LPTIM1 and LPTIM2)

These timers have an independent clock. When fed with LSE, LSI or external clock, they keep running in Stop mode and they can wake up the system from it.

Features of LPTIM1 and LPTIM2:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output (pulse, PWM)
- Continuous/one-shot mode
- Selectable software/hardware input trigger
- Selectable clock source:
  - Internal: LSE, LSI, HSI16 or APB clocks
  - External: over LPTIM input (working even with no internal clock source running, used by pulse counter application)
- Programmable digital glitch filter
- Encoder mode

### 3.16.4 Independent watchdog (IWDG)

The independent watchdog is based on an 8-bit prescaler and 12-bit downcounter with user-defined refresh window. It is clocked from an independent 32 kHz internal RC (LSI). Independent of the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes. Its counter can be frozen in debug mode.

### 3.16.5 System window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked by the system clock. It has an early-warning interrupt capability. Its counter can be frozen in debug mode.



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### 3.16.6 SysTick timer

This timer is dedicated to real-time operating systems, but it can also be used as a standard down counter.

Features of SysTick timer:

- 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source

### 3.17 Real-time clock (RTC), tamper (TAMP) and backup registers

The device embeds an RTC and five 32-bit backup registers, located in the RTC domain of the silicon die.

The ways of powering the RTC domain are described in Section 3.7.6.

The RTC is an independent BCD timer/counter.

Features of the RTC:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month
- Programmable alarm
- On-the-fly correction from 1 to 32767 RTC clock pulses, usable for synchronization with a master clock
- Reference clock detection a more precise second-source clock (50 or 60 Hz) can be used to improve the calendar precision
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy
- Two anti-tamper detection pins with programmable filter
- Timestamp feature to save a calendar snapshot, triggered by an event on the timestamp pin or a tamper event, or by switching to VBAT mode
- 17-bit auto-reload wakeup timer (WUT) for periodic events, with programmable resolution and period
- Multiple clock sources and references:
  - A 32.768 kHz external crystal (LSE)
  - An external resonator or oscillator (LSE)
  - The internal low-power RC oscillator (LSI, with typical frequency of 32 kHz)
  - The high-speed external clock (HSE) divided by 32

When clocked by LSE, the RTC operates in VBAT mode and in all low-power modes. When clocked by LSI, the RTC does not operate in VBAT mode, but it does in low-power modes except for the Shutdown mode.

All RTC events (Alarm, WakeUp Timer, Timestamp or Tamper) can generate an interrupt and wake the device up from the low-power modes.

The backup registers allow keeping 20 bytes of user application data in the event of V<sub>DD</sub> failure, if a valid backup supply voltage is provided on VBAT pin. They are not affected by



the system reset, power reset, and upon the device's wakeup from Standby or Shutdown modes.

### 3.18 Inter-integrated circuit interface (I2C)

The device embeds two I2C peripherals. Refer to *Table 8* for the features.

The I<sup>2</sup>C-bus interface handles communication between the microcontroller and the serial I<sup>2</sup>C-bus. It controls all I<sup>2</sup>C-bus-specific sequencing, protocol, arbitration and timing.

Features of the I2C peripheral:

- I<sup>2</sup>C-bus specification and user manual rev. 5 compatibility:
  - Slave and master modes, multimaster capability
  - Standard-mode (Sm), with a bitrate up to 100 kbit/s
  - Fast-mode (Fm), with a bitrate up to 400 kbit/s
  - Fast-mode Plus (Fm+), with a bitrate up to 1 Mbit/s and extra output drive I/Os
  - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
  - Programmable setup and hold times
  - Clock stretching
- SMBus specification rev 3.0 compatibility:
  - Hardware PEC (packet error checking) generation and verification with ACK control
  - Command and data acknowledge control
  - Address resolution protocol (ARP) support
  - Host and Device support
  - SMBus alert
  - Timeouts and idle condition detection
- PMBus rev 1.3 standard compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent of the PCLK reprogramming
- · Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

Table 8. I<sup>2</sup>C implementation

I <sup>2</sup> C features <sup>(1)</sup>	I2C1	I2C2
Standard mode (up to 100 kbit/s)	Χ	X
Fast mode (up to 400 kbit/s)	Х	Х
Fast Mode Plus (up to 1 Mbit/s) with extra output drive I/Os	Х	Х
Programmable analog and digital noise filters	Х	Х
SMBus/PMBus hardware support	Х	-
Independent clock	Х	-
Wakeup from Stop mode on address match	Х	-

<sup>1.</sup> X: supported

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# 3.19 Universal synchronous/asynchronous receiver transmitter (USART)

The device embeds universal synchronous/asynchronous receivers/transmitters that communicate at speeds of up to 8 Mbit/s.

They provide hardware management of the CTS, RTS and RS485 DE signals, multiprocessor communication mode, master synchronous communication and single-wire half-duplex communication mode. Some can also support SmartCard communication (ISO 7816), IrDA SIR ENDEC, LIN Master/Slave capability and auto baud rate feature, and have a clock domain independent of the CPU clock, which allows them to wake up the MCU from Stop mode. The wakeup events from Stop mode are programmable and can be:

- start bit detection
- any received data frame
- a specific programmed data frame

All USART interfaces can be served by the DMA controller.

USART modes/features(1) **USART1 USART2** Hardware flow control for modem Х Continuous communication using DMA Χ Χ Multiprocessor communication Х Х Synchronous mode Χ Х Smartcard mode Х Χ Single-wire half-duplex communication Х IrDA SIR ENDEC block Χ LIN mode Χ Dual clock domain and wakeup from Stop mode Χ Х Receiver timeout interrupt Modbus communication Χ Χ Auto baud rate detection Χ **Driver Enable** Χ

**Table 9. USART implementation** 

# 3.20 Low-power universal asynchronous receiver transmitter (LPUART)

The device embeds one LPUART. The peripheral supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

<sup>1.</sup> X: supported

The LPUART has a clock domain independent of the CPU clock, and can wakeup the system from Stop mode. The Stop mode wakeup events are programmable and can be:

- start bit detection
- any received data frame
- a specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

The LPUART interface can be served by the DMA controller.

### 3.21 Serial peripheral interface (SPI)

The device contains two SPIs running at up to 32 Mbits/s in master and slave modes. It supports half-duplex, full-duplex and simplex communications. A 3-bit prescaler gives eight master mode frequencies. The frame size is configurable from 4 bits to 16 bits. The SPI peripherals support NSS pulse mode, TI mode and hardware CRC calculation.

The SPI peripherals can be served by the DMA controller.

The I<sup>2</sup>S interface mode of the SPI peripheral (if supported, see the following table) supports four different audio standards can operate as master or slave, in half-duplex communication mode. It can be configured to transfer 16 and 24 or 32 bits with 16-bit or 32-bit data resolution and synchronized by a specific signal. Audio sampling frequency from 8 kHz up to 192 kHz can be set by an 8-bit programmable linear prescaler. When operating in master mode, it can output a clock for an external audio component at 256 times the sampling frequency.

SPI features <sup>(1)</sup>	SPI1	SPI2
Hardware CRC calculation	Х	Х
Rx/Tx FIFO	Х	Х
NSS pulse mode	X	Х
I <sup>2</sup> S mode	Х	-
TI mode	Х	Х

Table 10. SPI/I2S implementation

### 3.22 Development support

### 3.22.1 Serial wire debug port (SW-DP)

An Arm SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.

<sup>1.</sup> X = supported.

### 4 Pinouts, pin description and alternate functions

Figure 3. STM32G031CxT LQFP48 pinout

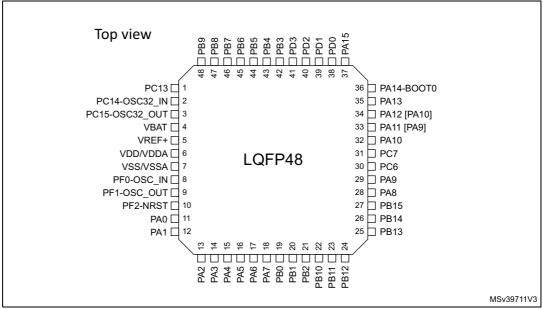
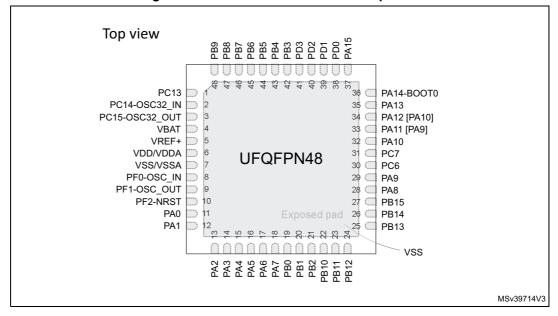


Figure 4. STM32G031CxU UFQFPN48 pinout



Top view

| Compare | Comp

19 PA9
18 PA8
17 PB2

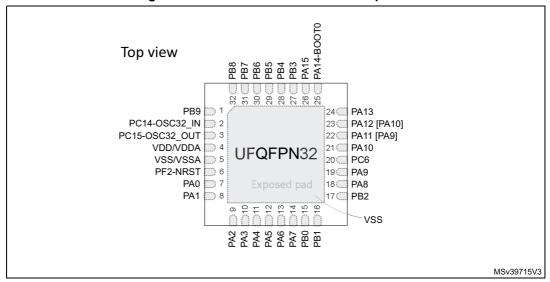
Figure 5. STM32G031KxT LQFP32 pinout



PA2 | 9
PA3 | 10
PA4 | 11
PA5 | 12
PA6 | 13
PA7 | 14
PB0 | 15
PB1 | 16

PF2-NRST 6 PA0 7

PA1



MSv39712V3

Figure 7. STM32G031GxU UFQFPN28 pinout

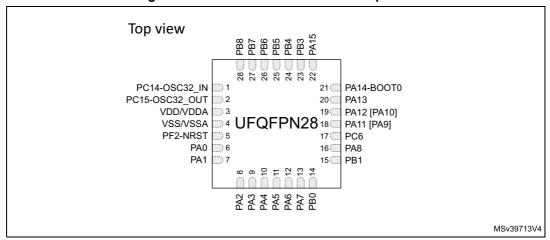


Figure 8. STM32G031Fx TSSOP20 pinout

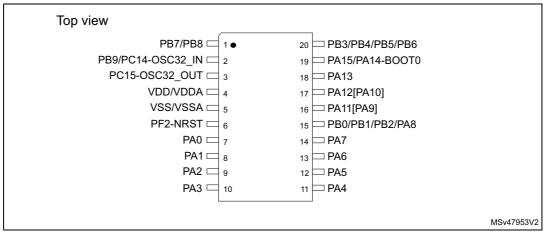
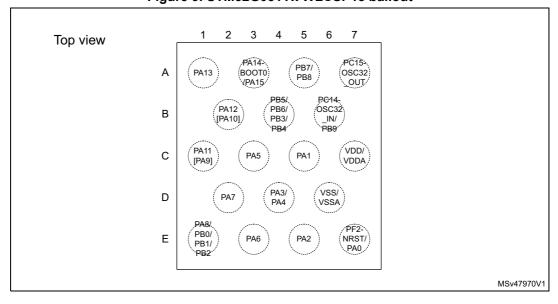


Figure 9. STM32G031Yx WLCSP18 ballout



**47/** 

Figure 10. STM32G031Jx SO8N pinout

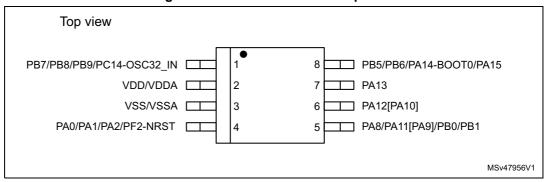


Table 11. Terms and symbols used in *Table 12* 

Col	umn	Symbol Definition				
Pin r	name	Terminal name corresponds t parenthesis under the pin name	o its by-default function at reset, unless otherwise specified in me.			
		S Supply pin				
Pin	type	I	Input only pin			
		I/O	Input / output pin			
		FT	5 V tolerant I/O			
		RST	Bidirectional reset pin with embedded weak pull-up resistor			
		Options for FT I/Os				
I/O str	ructuref		I/O, Fm+ capable			
	_a I/O,		I/O, with analog switch function			
		_e	I/O, with switchable diode to V <sub>DD</sub>			
No	ote	Upon reset, all I/Os are set as analog inputs, unless otherwise specified.				
Pin	Alternate functions	Functions selected through GPIOx_AFR registers				
functions	Additional functions	Functions directly selected/enabled through peripheral registers				

Table 12. Pin assignment and description

		Р	in						it and description		
N8OS	WLCSP18	TSSOP20	UFQFPN28	LQFP32 / UFQFPN32	LQFP48 / UFQFPN48	Pin name (function upon reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	-	-	-	-	1	PC13	I/O	FT	(1)(2)	TIM1_BK	TAMP_IN1, RTC_TS, RTC_OUT1, WKUP2
-	-	ı	-	-	2	PC14- OSC32_IN (PC14)	I/O	FT	(1)(2)	TIM1_BK2	OSC32_IN
1	В6	2	1	2	-	PC14- OSC32_IN (PC14)	I/O	FT	(1)(2)	TIM1_BK2	OSC32_IN, OSC_IN
-	A7	3	2	3	3	PC15- OSC32_OUT (PC15)	I/O	FT	(1)(2)	OSC32_EN, OSC_EN	OSC32_OUT
-	-	-	-	-	4	VBAT	S	-	-	-	VBAT
-	-	-	-	-	5	VREF+	S	-	-	-	VREF_OUT
2	C7	4	3	4	6	VDD/VDDA	S	-	-	-	-
3	D6	5	4	5	7	VSS/VSSA	S	-	-	-	-
-	-	1	-	-	8	PF0-OSC_IN (PF0)	I/O	FT	-	TIM14_CH1	OSC_IN
-	1	ı	-	-	9	PF1- OSC_OUT (PF1)	I/O	FT	-	OSC_EN	OSC_OUT
4	E7	6	5	6	10	PF2-NRST	I/O	ı	(3)	MCO	NRST
4	E7	7	6	7	11	PA0	I/O	FT_a	(3)	SPI2_SCK, USART2_CTS, TIM2_CH1_ETR, LPTIM1_OUT	ADC_IN0, TAMP_IN2,WKUP1
4	C5	8	7	8	12	PA1	I/O	FT_ea	(3)	SPI1_SCK/I2S1_CK, USART2_RTS_DE_CK, TIM2_CH2, I2C1_SMBA, EVENTOUT	ADC_IN1
4	E5	9	8	9	13	PA2	I/O	FT_a	(3)	SPI1_MOSI/I2S1_SD, USART2_TX, TIM2_CH3, LPUART1_TX	ADC_IN2, WKUP4,LSCO
-	D4	10	9	10	14	PA3	I/O	FT_ea	-	SPI2_MISO, USART2_RX, TIM2_CH4, LPUART1_RX, EVENTOUT	ADC_IN3
-	-	-	-	-	15	PA4	I/O	FT_a	-	SPI1_NSS/I2S1_WS, SPI2_MOSI, TIM14_CH1, LPTIM2_OUT, EVENTOUT	ADC_IN4, RTC_OUT2

Table 12. Pin assignment and description (continued)

		Р	in								
N8OS	WLCSP18	TSSOP20	UFQFPN28	LQFP32 / UFQFPN32	LQFP48 / UFQFPN48	Pin name (function upon reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	D4	11	10	11	ı	PA4	I/O	FT_a	-	SPI1_NSS/I2S1_WS, SPI2_MOSI, TIM14_CH1, LPTIM2_OUT, EVENTOUT	ADC_IN4, TAMP_IN1, RTC_TS, RTC_OUT1, WKUP2
-	СЗ	12	11	12	16	PA5	I/O	FT_ea	-	SPI1_SCK/I2S1_CK, TIM2_CH1_ETR, LPTIM2_ETR, EVENTOUT	ADC_IN5
-	E3	13	12	13	17	PA6	I/O	FT_ea	-	SPI1_MISO/I2S1_MCK, TIM3_CH1, TIM1_BK, TIM16_CH1, LPUART1_CTS	ADC_IN6
-	D2	14	13	14	18	PA7	I/O	FT_a	-	SPI1_MOSI/I2S1_SD, TIM3_CH2, TIM1_CH1N, TIM14_CH1, TIM17_CH1	ADC_IN7
5	E1	15	14	15	19	PB0	I/O	FT_ea	-	SPI1_NSS/I2S1_WS, TIM3_CH3, TIM1_CH2N, LPTIM1_OUT	ADC_IN8
5	E1	15	15	16	20	PB1	I/O	FT_ea	-	TIM14_CH1, TIM3_CH4, TIM1_CH3N, LPTIM2_IN1, LPUART1_RTS_DE, EVENTOUT	ADC_IN9
-	E1	15	1	17	21	PB2	I/O	FT_ea	-	SPI2_MISO, LPTIM1_OUT, EVENTOUT	ADC_IN10
-	-	1	ı	ı	22	PB10	I/O	FT_fa	-	LPUART1_RX, TIM2_CH3, SPI2_SCK, I2C2_SCL	ADC_IN11
-	-	1	ı	ı	23	PB11	1/0	FT_fa	-	SPI2_MOSI, LPUART1_TX, TIM2_CH4, I2C2_SDA	ADC_IN15
-	-	1	1	1	24	PB12	I/O	FT_a	-	SPI2_NSS, LPUART1_RTS_DE, TIM1_BK, EVENTOUT	ADC_IN16
-	-	-	-	-	25	PB13	I/O	FT_f	-	SPI2_SCK, LPUART1_CTS, TIM1_CH1N, I2C2_SCL, EVENTOUT	-
-	-	-	-	-	26	PB14	I/O	FT_f	-	SPI2_MISO, TIM1_CH2N, I2C2_SDA, EVENTOUT	-
-	-	-	-	-	27	PB15	I/O	FT	-	SPI2_MOSI, TIM1_CH3N, EVENTOUT	RTC_REFIN
5	E1	15	16	18	28	PA8	I/O	FT	-	MCO, SPI2_NSS, TIM1_CH1, LPTIM2_OUT, EVENTOUT	-



Table 12. Pin assignment and description (continued)

		Р	in							description (continued)	
SO8N	WLCSP18	TSSOP20	UFQFPN28	LQFP32 / UFQFPN32	LQFP48 / UFQFPN48	Pin name (function upon reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	1	-	1	19	29	PA9	I/O	FT_f	-	MCO, USART1_TX, TIM1_CH2, SPI2_MISO, I2C1_SCL, EVENTOUT	-
-	-	-	17	20	30	PC6	I/O	FT	-	TIM3_CH1, TIM2_CH3	-
-	-	-	-	-	31	PC7	I/O	FT	-	TIM3_CH2, TIM2_CH4	-
-	1	1	ı	21	32	PA10	I/O	FT_f	1	SPI2_MOSI, USART1_RX, TIM1_CH3, TIM17_BK, I2C1_SDA, EVENTOUT	-
-	1	1	1	1	33	PA11 [PA9]	I/O	FT_f	(4)	SPI1_MISO/I2S1_MCK, USART1_CTS, TIM1_CH4, TIM1_BK2, I2C2_SCL	-
5	C1	16	18	22	-	PA11 [PA9]	I/O	FT_fa	(4)	SPI1_MISO/I2S1_MCK, USART1_CTS, TIM1_CH4, TIM1_BK2, I2C2_SCL	ADC_IN15
ı	1	1	ı	1	34	PA12 [PA10]	I/O	FT_f	(4)	SPI1_MOSI/I2S1_SD, USART1_RTS_DE_CK, TIM1_ETR, I2S_CKIN, I2C2_SDA	-
6	B2	17	19	23	-	PA12 [PA10]	I/O	FT_fa	(4)	SPI1_MOSI/I2S1_SD, USART1_RTS_DE_CK, TIM1_ETR, I2S_CKIN, I2C2_SDA	ADC_IN16
7	A1	18	20	24	35	PA13	I/O	FT_ea	(5)	SWDIO, IR_OUT, EVENTOUT	ADC_IN17
8	А3	19	21	25	36	PA14-BOOT0	I/O	FT_a	(5)	SWCLK, USART2_TX, EVENTOUT	ADC_IN18, BOOT0
8	А3	19	22	26	37	PA15	1/0	FT	-	SPI1_NSS/I2S1_WS, USART2_RX, TIM2_CH1_ETR, EVENTOUT	-
-	-	-	-	-	38	PD0	I/O	FT	-	EVENTOUT, SPI2_NSS, TIM16_CH1	-
-	-	-	-	-	39	PD1	I/O	FT	-	EVENTOUT, SPI2_SCK, TIM17_CH1	-
-	-	-	1	-	40	PD2	I/O	FT	-	TIM3_ETR, TIM1_CH1N	-
-	1	-	-	-	41	PD3	I/O	FT	ı	USART2_CTS, SPI2_MISO, TIM1_CH2N	-

Table 12. Pin assignment and description (continued)

		Р	in								
N8OS	WLCSP18	TSSOP20	UFQFPN28	LQFP32 / UFQFPN32	LQFP48 / UFQFPN48	Pin name (function upon reset)	Pin type	ON Alternate functions		Additional functions	
-	B4	20	23	27	42	PB3	I/O	FT	-	SPI1_SCK/I2S1_CK, TIM1_CH2, TIM2_CH2, USART1_RTS_DE_CK, EVENTOUT	-
-	В4	20	24	28	43	PB4	I/O	FT	-	SPI1_MISO/I2S1_MCK, TIM3_CH1, USART1_CTS, TIM17_BK, EVENTOUT	-
8	В4	20	25	29	44	PB5	I/O	FT	-	SPI1_MOSI/I2S1_SD, TIM3_CH2, TIM16_BK, LPTIM1_IN1, I2C1_SMBA	WKUP6
8	В4	20	26	30	45	PB6	I/O	FT_f	1	USART1_TX, TIM1_CH3, TIM16_CH1N, SPI2_MISO, LPTIM1_ETR, I2C1_SCL, EVENTOUT	-
-	-	1	-	-	46	PB7	I/O	FT_f	-	USART1_RX, SPI2_MOSI, TIM17_CH1N, LPTIM1_IN2, I2C1_SDA, EVENTOUT	PVD_IN
1	A5	1	27	31	-	PB7	I/O	FT_fa	-	USART1_RX, SPI2_MOSI, TIM17_CH1N, LPTIM1_IN2, I2C1_SDA, EVENTOUT	ADC_IN11, PVD_IN
1	A5	1	28	32	47	PB8	I/O	FT_f	-	SPI2_SCK, TIM16_CH1, I2C1_SCL, EVENTOUT	-
1	В6	2	1	1	48	PB9	I/O	FT_f	1	IR_OUT, TIM17_CH1, SPI2_NSS, I2C1_SDA, EVENTOUT	-

PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited:

 The speed should not exceed 2 MHz with a maximum load of 30 pF
 These GPIOs must not be used as current sources (for example to drive a LED).

- 4. Pins PA9 and PA10 can be remapped in place of pins PA11 and PA12 (default mapping), using SYSCFG\_CFGR1 register.
- Upon reset, these pins are configured as SWD alternate functions, and the internal pull-up on PA13 pin and the internal pull-down on PA14 pin are activated.



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<sup>2.</sup> After an RTC domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the RTC registers. The RTC registers are not reset upon system reset. For details on how to manage these GPIOs, refer to the RTC domain and RTC register descriptions in the RM0444 reference manual.

<sup>3.</sup> PF2-NRST default functionnality is NRST, if used in this mode on SO8 and WLCSP, take particular care to configuration of other IOs connected to this pin.

	Table 13. Port A alternate function mapping										
Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7			
PA0	SPI2_SCK	USART2_CTS	TIM2_CH1_ETR	-	-	LPTIM1_OUT	-	-			
PA1	SPI1_SCK/ I2S1_CK	USART2_RTS _DE_CK	TIM2_CH2	-	-	-	I2C1_SMBA	EVENTOUT			
PA2	SPI1_MOSI/ I2S1_SD	USART2_TX	TIM2_CH3	-	-	-	LPUART1_TX	-			
PA3	SPI2_MISO	USART2_RX	TIM2_CH4	-	-	-	LPUART1_RX	EVENTOUT			
PA4	SPI1_NSS/ I2S1_WS	SPI2_MOSI	-	-	TIM14_CH1	LPTIM2_OUT	-	EVENTOUT			
PA5	SPI1_SCK/ I2S1_CK	-	TIM2_CH1_ETR	-	-	LPTIM2_ETR	-	EVENTOUT			
PA6	SPI1_MISO/ I2S1_MCK	TIM3_CH1	TIM1_BKIN	-	-	TIM16_CH1	LPUART1_CTS	-			
PA7	SPI1_MOSI/ I2S1_SD	TIM3_CH2	TIM1_CH1N	-	TIM14_CH1	TIM17_CH1	-	-			
PA8	MCO	SPI2_NSS	TIM1_CH1	-	-	LPTIM2_OUT	-	EVENTOUT			
PA9	MCO	USART1_TX	TIM1_CH2	-	SPI2_MISO	-	I2C1_SCL	EVENTOUT			
PA10	SPI2_MOSI	USART1_RX	TIM1_CH3	-	-	TIM17_BKIN	I2C1_SDA	EVENTOUT			
PA11	SPI1_MISO/ I2S1_MCK	USART1_CTS	TIM1_CH4	-	-	TIM1_BKIN2	I2C2_SCL	-			
PA12	SPI1_MOSI/ I2S1_SD	USART1_RTS _DE_CK	TIM1_ETR	-	-	I2S_CKIN	I2C2_SDA	-			
PA13	SWDIO	IR_OUT	-	-	-	-	-	EVENTOUT			
PA14	SWCLK	USART2_TX	-	-	-	-	-	EVENTOUT			
PA15	SPI1_NSS/ I2S1_WS	USART2_RX	TIM2_CH1_ETR	-	-	-	-	EVENTOUT			

Table 14. Port B alternate function mapping

Doort	450	A.F.4			A F 4		A.F.O.	A = 7
Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PB0	SPI1_NSS/ I2S1_WS	ТІМЗ_СНЗ	TIM1_CH2N	-	-	LPTIM1_OUT	-	-
PB1	TIM14_CH1	TIM3_CH4	TIM1_CH3N	-	-	LPTIM2_IN1	LPUART1_RTS _DE	EVENTOUT
PB2	-	SPI2_MISO	-	-	-	LPTIM1_OUT	-	EVENTOUT
PB3	SPI1_SCK/ I2S1_CK	TIM1_CH2	TIM2_CH2	-	USART1_RTS _DE_CK	-	-	EVENTOUT
PB4	SPI1_MISO/ I2S1_MCK	TIM3_CH1	-	-	USART1_CTS	TIM17_BKIN	-	EVENTOUT
PB5	SPI1_MOSI/ I2S1_SD	TIM3_CH2	TIM16_BKIN	-	-	LPTIM1_IN1	I2C1_SMBA	-
PB6	USART1_TX	TIM1_CH3	TIM16_CH1N	-	SPI2_MISO	LPTIM1_ETR	I2C1_SCL	EVENTOUT
PB7	USART1_RX	SPI2_MOSI	TIM17_CH1N	-	-	LPTIM1_IN2	I2C1_SDA	EVENTOUT
PB8	-	SPI2_SCK	TIM16_CH1	-	-	-	I2C1_SCL	EVENTOUT
PB9	IR_OUT	-	TIM17_CH1	-	-	SPI2_NSS	I2C1_SDA	EVENTOUT
PB10	-	LPUART1_RX	TIM2_CH3	-	-	SPI2_SCK	I2C2_SCL	-
PB11	SPI2_MOSI	LPUART1_TX	TIM2_CH4	-	-	-	I2C2_SDA	-
PB12	SPI2_NSS	LPUART1_RTS _DE	TIM1_BKIN	-	-	-	-	EVENTOUT
PB13	SPI2_SCK	LPUART1_CTS	TIM1_CH1N	-	-	-	I2C2_SCL	EVENTOUT
PB14	SPI2_MISO	-	TIM1_CH2N	-	-	-	I2C2_SDA	EVENTOUT
PB15	SPI2_MOSI	-	TIM1_CH3N	-	-	-	-	EVENTOUT

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PC6	-	TIM3_CH1	TIM2_CH3	-	-	-	-	-
PC7	-	TIM3_CH2	TIM2_CH4	-	-	-	-	-
PC13	-	-	TIM1_BKIN	-	-	-	-	-
PC14	-	-	TIM1_BKIN2	-	-	-	-	-
PC15	OSC32_EN	OSC_EN	-	-	-	-	-	-

# Table 16. Port D alternate function mapping

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PD0	EVENTOUT	SPI2_NSS	TIM16_CH1	-	-	-	-	-
PD1	EVENTOUT	SPI2_SCK	TIM17_CH1	-	-	-	-	-
PD2	-	TIM3_ETR	TIM1_CH1N	-	-	-	-	-
PD3	USART2_CTS	SPI2_MISO	TIM1_CH2N	-	-	-	-	-

#### Table 17. Port F alternate function mapping

	Table 1111 etc. alternate random mapping							
Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PF0	-	-	TIM14_CH1	-	-	-	-	-
PF1	OSC_EN	-	-	-	-	-	-	-
PF2	МСО	-	-	-	-	-	-	-

# 5 Electrical characteristics

#### 5.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

Parameter values defined at temperatures or in temperature ranges out of the ordering information scope are to be ignored.

Packages used for characterizing certain electrical parameters may differ from the commercial packages as per the ordering information.

## 5.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A(max)$  (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean  $\pm 3\sigma$ ).

# 5.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A$  = 25 °C,  $V_{DD}$  =  $V_{DDA}$  = 3 V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean  $\pm 2\sigma$ ).

#### 5.1.3 Typical curves

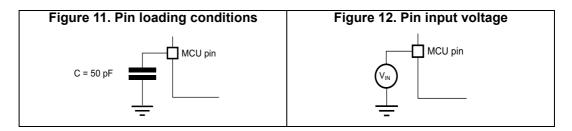
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 5.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 11.

#### 5.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 12.



# 5.1.6 Power supply scheme

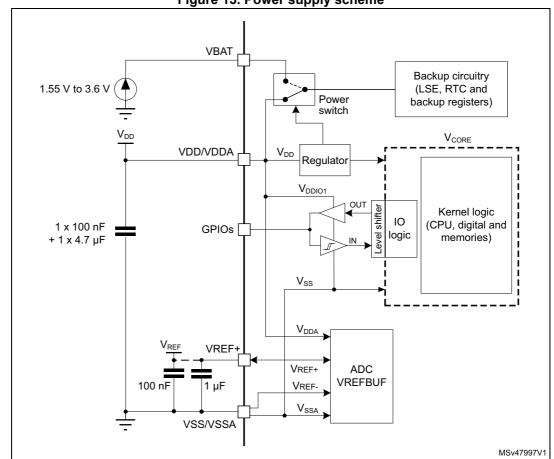


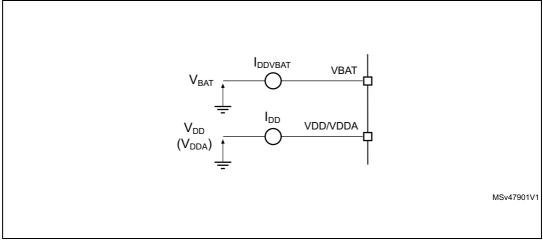
Figure 13. Power supply scheme

Caution:

Power supply pin pair (VDD/VDDA and VSS/VSSA) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

#### 5.1.7 **Current consumption measurement**

Figure 14. Current consumption measurement scheme



#### 5.2 **Absolute maximum ratings**

Stresses above the absolute maximum ratings listed in Table 18, Table 19 and Table 20 may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

All voltages are defined with respect to V<sub>SS</sub>.

**Table 18. Voltage characteristics** 

Symbol	Ratings	Min	Max	Unit
$V_{DD}$	External supply voltage	- 0.3	4.0	
V <sub>BAT</sub>	External supply voltage on VBAT pin	- 0.3	4.0	
V <sub>REF+</sub>	External voltage on VREF+ pin	- 0.3	Min(V <sub>DD</sub> + 0.4, 4.0)	
	Input voltage on FT_xx	- 0.3	V <sub>DD</sub> + 4.0 <sup>(2)</sup>	V
$V_{IN}^{(1)}$	Input voltage on any other pin	- 0.3	4.0	

<sup>1.</sup> Refer to *Table 19* for the maximum allowed injected current values.

<sup>2.</sup> To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.

**Table 19. Current characteristics** 

Symbol	Ratings	Max	Unit
I <sub>VDD/VDDA</sub>	Current into VDD/VDDA power pin (source) <sup>(1)</sup>	100	
I <sub>VSS/VSSA</sub>	Current out of VSS/VSSA ground pin (sink) <sup>(1)</sup>	100	
	Output current sunk by any I/O and control pin except FT_f	15	
I <sub>IO(PIN)</sub>	Output current sunk by any FT_f pin	20	
	Output current sourced by any I/O and control pin	15	
71	Total output current sunk by sum of all I/Os and control pins	80	mA
$\sum I_{IO(PIN)}$	Total output current sourced by sum of all I/Os and control pins	80	
I <sub>INJ(PIN)</sub> <sup>(2)</sup>	Injected current on a FT_xx pin	-5 / NA <sup>(3)</sup>	
Σ I <sub>INJ(PIN)</sub>	Total injected current (sum of all I/Os and control pins) <sup>(4)</sup>	25	

All main power (VDD/VDDA, VBAT) and ground (VSS/VSSA) pins must always be connected to the external power supplies, in the permitted range.

Table 20. Thermal characteristics

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C

# 5.3 Operating conditions

## 5.3.1 General operating conditions

Table 21. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	64	MHz
f <sub>PCLK</sub>	Internal APB clock frequency	-	0	64	IVII IZ
$V_{DD}$	Standard operating voltage	-	1.7 <sup>(1)</sup>	3.6	V
V	Analog supply voltage	For ADC operation	1.62	3.6	V
V <sub>DDA</sub>	Analog supply voltage	For VREFBUF operation	2.4	3.6	
V <sub>BAT</sub>	Backup operating voltage	-	1.55	3.6	V
V <sub>IN</sub>	I/O input voltage	-	-0.3	Min(V <sub>DD</sub> + 3.6, 5.5) <sup>(2)</sup>	V

A positive injection is induced by V<sub>IN</sub> > V<sub>DDIOX</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer also to *Table 18: Voltage characteristics* for the maximum allowed input voltage values.

Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.

When several inputs are submitted to a current injection, the maximum ∑|I<sub>INJ(PIN)</sub>| is the absolute sum of the negative injected currents (instantaneous values).

Symbol	Parameter	Conditions	Min	Max	Unit
		Suffix 6 <sup>(4)</sup>	-40	85	
T <sub>A</sub>	Ambient temperature <sup>(3)</sup>	Suffix 7 <sup>(4)</sup>	-40	105	°C
		Suffix 3 <sup>(4)</sup>	-40	125	
		Suffix 6 <sup>(4)</sup>	-40	105	
$T_J$	Junction temperature	Suffix 7 <sup>(4)</sup>	-40	125	°C
		Suffix 3 <sup>(4)</sup>	-40	130	

Table 21. General operating conditions (continued)

- 1. When RESET is released functionality is guaranteed down to  $V_{\mbox{\scriptsize PDR}}$  min.
- 2. For operation with voltage higher than  $V_{DD}$  +0.3 V, the internal pull-up and pull-down resistors must be disabled.
- 3. The  $T_A(max)$  applies to  $P_D(max)$ . At  $P_D < P_D(max)$  the ambient temperature is allowed to go higher than  $T_A(max)$  provided that the junction temperature  $T_J$  does not exceed  $T_J(max)$ . Refer to Section 6.9: Thermal characteristics.
- 4. Temperature range digit in the order code. See Section 7: Ordering information.

# 5.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 22* are derived from tests performed under the ambient temperature condition summarized in *Table 21*.

Table 22. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
		V <sub>DD</sub> rising	-	∞	µs/V
t <sub>VDD</sub>	V <sub>DD</sub> slew rate	V <sub>DD</sub> falling; ULPEN = 0	10	8	μ5/ ν
		V <sub>DD</sub> falling; ULPEN = 1	100	8	ms/V

## 5.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 23* are derived from tests performed under the ambient temperature conditions summarized in *Table 21: General operating conditions*.

Table 23. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Тур	Max	Unit
t <sub>RSTTEMPO</sub> <sup>(2)</sup>	POR temporization when $V_{DD}$ crosses $V_{POR}$	V <sub>DD</sub> rising	-	250	400	μs
V <sub>POR</sub> <sup>(2)</sup>	Power-on reset threshold	-	1.62	1.66	1.70	V
V <sub>PDR</sub> <sup>(2)</sup>	Power-down reset threshold	-	1.60	1.64	1.69	V
V	Brownout reset threshold 1	V <sub>DD</sub> rising	2.05	2.10	2.18	V
$V_{BOR1}$	Brownout reset tilleshold 1	V <sub>DD</sub> falling	1.95	2.00	2.08	V
V	Brownout reset threshold 2	V <sub>DD</sub> rising	2.20	2.31	2.38	V
$V_{BOR2}$	brownout reset timeshold 2	V <sub>DD</sub> falling	2.10	2.21	2.28	V
V	Brownout reset threshold 3	V <sub>DD</sub> rising	2.50	2.62	2.68	V
$V_{BOR3}$	Brownout reset tilleshold 3	V <sub>DD</sub> falling	2.40	2.52	2.58	V



Table 23. Embedded reset and power control block characteristics (continued)

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Тур	Max	Unit
V	Brownout reset threshold 4	V <sub>DD</sub> rising	2.80	2.91	3.00	V
$V_{BOR4}$	brownout reset timeshold 4	V <sub>DD</sub> falling	2.70	2.81	2.90	v
V	Programmable voltage detector threshold 0	V <sub>DD</sub> rising	2.05	2.15	2.22	V
$V_{PVD0}$	Programmable voltage detector threshold 0	V <sub>DD</sub> falling	1.95	2.05	2.12	v
V	PVD threshold 1	V <sub>DD</sub> rising	2.20	2.30	2.37	V
$V_{PVD1}$	FVD tillesiloid i	V <sub>DD</sub> falling	2.10	2.20	2.27	v
V	PVD threshold 2	V <sub>DD</sub> rising	2.35	2.46	2.54	V
$V_{PVD2}$	FVD tilleshold 2	V <sub>DD</sub> falling	2.25	2.36	2.44	v
V	D)/D throubold 2	V <sub>DD</sub> rising	2.50	2.62	2.70	V
$V_{PVD3}$	PVD threshold 3	V <sub>DD</sub> falling	2.40	2.52	2.60	V
V	PVD threshold 4	V <sub>DD</sub> rising	2.65	2.74	2.87	V
$V_{PVD4}$	FVD tilleshold 4	V <sub>DD</sub> falling	2.55	2.64	2.77	V
V	PVD threshold 5	V <sub>DD</sub> rising	2.80	2.91	3.03	V
$V_{PVD5}$	FVD tillesiloid 5	V <sub>DD</sub> falling	2.70	2.81	2.93	v
V	PVD threshold 6	V <sub>DD</sub> rising	2.90	3.01	3.14	V
$V_{PVD6}$	FVD tilleshold 6	V <sub>DD</sub> falling	2.80	2.91	3.04	v
V <sub>hyst POR PDR</sub>	Hysteresis of $V_{POR}$ and $V_{PDR}$	Hysteresis in continuous mode	-	20	-	mV
,		Hysteresis in other mode	-	30	-	_
V <sub>hyst_BOR_PVD</sub>	Hysteresis of V <sub>BORx</sub> and V <sub>PVDx</sub>	-	-	100	-	mV
I <sub>DD(BOR_PVD)</sub> <sup>(2)</sup>	BOR and PVD consumption	-	-	1.1	1.6	μA

<sup>1.</sup> Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

# 5.3.4 Embedded voltage reference

The parameters given in *Table 24* are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 24. Embedded internal voltage reference

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>REFINT</sub>	Internal reference voltage	-40°C < T <sub>J</sub> < 130°C	1.182	1.212	1.232	V
t <sub>S_vrefint</sub> (1)	ADC sampling time when reading the internal reference voltage	-	4 <sup>(2)</sup>	-	-	μs
t <sub>start_vrefint</sub>	Start time of reference voltage buffer when ADC is enable	-	-	8	12 <sup>(2)</sup>	μs

<sup>2.</sup> Guaranteed by design.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>DD(VREFINTBUF)</sub>	$V_{REFINT}$ buffer consumption from $V_{DD}$ when converted by ADC	-	-	12.5	20 <sup>(2)</sup>	μΑ
ΔV <sub>REFINT</sub>	Internal reference voltage spread over the temperature range	V <sub>DD</sub> = 3 V		5	7.5 <sup>(2)</sup>	mV
T <sub>Coeff_vrefint</sub>	Temperature coefficient	-	ı	30	50 <sup>(2)</sup>	ppm/°C
A <sub>Coeff</sub>	Long term stability	1000 hours, T = 25 °C	ı	300	1000 <sup>(2)</sup>	ppm
$V_{\mathrm{DDCoeff}}$	Voltage coefficient	3.0 V < V <sub>DD</sub> < 3.6 V	ı	250	1200 <sup>(2)</sup>	ppm/V
V <sub>REFINT_DIV1</sub>	1/4 reference voltage		24	25	26	0/
V <sub>REFINT_DIV2</sub>	1/2 reference voltage	-	49	50	51	$^{ m W}_{ m REFINT}$
V <sub>REFINT_DIV3</sub>	3/4 reference voltage		74	75	76	INCI IIVI

Table 24. Embedded internal voltage reference (continued)

<sup>2.</sup> Guaranteed by design.

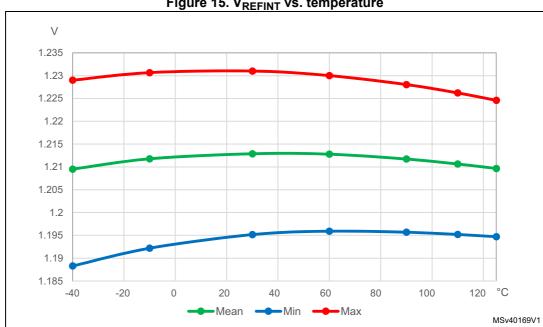


Figure 15. V<sub>REFINT</sub> vs. temperature

#### 5.3.5 **Supply current characteristics**

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in Figure 14: Current consumption measurement scheme

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<sup>1.</sup> The shortest sampling time can be determined in the application by multiple iterations.

#### Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted with the minimum wait states number, depending on the f<sub>HCLK</sub> frequency (refer to the table "Number of wait states according to CPU clock (HCLK) frequency" available in the RM0444 reference manual).
- When the peripherals are enabled f<sub>PCLK</sub> = f<sub>HCLK</sub>
- For Flash memory and shared peripherals f<sub>PCLK</sub> = f<sub>HCLK</sub> = f<sub>HCLKS</sub>

Unless otherwise stated, values given in *Table 25* through *Table 33* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 25. Current consumption in Run and Low-power run modes at different die temperatures

		Cond	litions			Тур			Max <sup>(1)</sup>	)	
I <sub>DD(Run)</sub>	Parameter	General	f <sub>HCLK</sub>	Fetch from <sup>(2)</sup>	25 °C	85 °C	125 °C	25 °C	85 °C	130 °C	Unit
			64 MHz		5.2	5.4	5.6	6.7	6.9	7.6	
			56 MHz		4.7	4.8	4.9	5.9	6.0	6.6	
			48 MHz	Flash	4.2	4.3	4.5	4.8	5.0	5.6	
		Dance 4:	32 MHz	memory	2.9	3.0	3.2	3.8	4.1	4.7	
		Range 1; PLL enabled;	24 MHz		2.3	2.4	2.5	2.9	3.2	3.9	
		f <sub>HCLK</sub> = f <sub>HSE_bypass</sub> (≤16 MHz), f <sub>HCLK</sub> = f <sub>PLLRCLK</sub> (>16 MHz); (3)	16 MHz		1.5	1.5	1.7	2.1	2.4	3.0	
			64 MHz		4.3	4.4	4.6	6.0	6.2	6.6	
			56 MHz		3.8	3.9	4.1	5.4	5.6	5.9	
	Supply current in		48 MHz	SRAM	3.4	3.5	3.7	4.8	5.0	5.3	mA
IDD(Run)	Run mode		32 MHz	OIVAIVI	2.4	2.5	2.6	3.4	3.6	3.9	IIIA
			24 MHz		1.9	1.9	2.0	2.7	2.9	3.3	
			16 MHz		1.2	1.2	1.4	1.9	2.0	2.5	
		D 0:	16 MHz		1.2	1.2	1.4	1.6	1.9	2.5	
		Range 2; PLL enabled;	8 MHz	Flash memory	0.7	0.7	0.8	1.1	1.3	1.7	
		f <sub>HCLK</sub> = f <sub>HSE_bypass</sub>	2 MHz	]	0.2	0.3	0.4	0.5	0.8	1.3	
	(≤16 MHz), f <sub>HCLK</sub> = f <sub>PLLRCLK</sub>	16 MHz		1.0	1.0	1.2	1.5	1.7	2.1		
		'HCLK - 'PLLRCLK	8 MHz	SRAM	0.5	0.6	0.7	1.0	1.1	1.5	
			2 MHz	1	0.2	0.3	0.4	0.5	0.7	1.2	

Table 25. Current consumption in Run and Low-power run modes at different die temperatures (continued)

		Cond	litions			Тур			Max <sup>(1)</sup>	)	
Symbol	Parameter	General	f <sub>HCLK</sub>	Fetch from <sup>(2)</sup>	25 °C	85 °C	125 °C	25 °C	85 °C	130 °C	Unit
			2 MHz		165	205	305	470	660	1190	
			1 MHz		90	120	225	400	580	1140	
			500 kHz	Flash memory	53	81	185	360	530	1060	
	Supply	PLL disabled;	125 kHz	,	23	51	155	310	500	990	
	Supply current in	f <sub>HCLK</sub> = f <sub>HSE</sub> bypass (> 32 kHz),	32 kHz		16	43	145	270	400	990	
IDD(LPRun)	Low-power run mode	f <sub>HCLK</sub> = f <sub>LSE</sub> bypass (= 32 kHz);	2 MHz		146	174	275	460	660	1160	μA
	Turi mode	(3)	1 MHz		83	104	195	390	630	1080	
			500 kHz	SRAM	44	74	155	340	590	1050	
			125 kHz		19	46	135	300	420	920	
			32 kHz		14	34	130	260	340	890	

<sup>1.</sup> Based on characterization results, not tested in production.

<sup>2.</sup> Prefetch and cache enabled when fetching from Flash memory. Code compiled with high optimization for space in SRAM.

<sup>3.</sup> V<sub>DD</sub> = 3.0 V for values in Typ columns and 3.6 V for values in Max columns, all peripherals disabled, cache enabled, prefetch disabled for code and data fetch from Flash and enabled from SRAM

Table 26. Typical current consumption in Run and Low-power run modes, depending on code executed

			onditions		Тур		Тур	
I <sub>DD(Run)</sub> Ct Rt	Parameter	General	Code	Fetch from <sup>(1)</sup>	25 °C	Unit	25 °C	Unit
			Reduced code <sup>(3)</sup>		5.4		84	
			Coremark	Floob	5.2		81	
			Dhrystone 2.1	Flash memory	5.3		82	
		Range 1;	Fibonacci		4.3		67	
		f <sub>HCLK</sub> = f <sub>PLLRCLK</sub> =	While(1) loop		4.0		62	
		64 MHz; (2)	Reduced code <sup>(3)</sup>		4.4		69	
			Coremark		5.6		88	
			Dhrystone 2.1	SRAM	5.6		88	
	0		Fibonacci		4.3		67	
Inn/Pun)	Supply current in		While(1) loop		4.0	mA	63	uA/MHz
DD(IXuII)	Run mode		Reduced code <sup>(3)</sup>		1.6		100	
			Coremark	Flash	1.5		81	
			Dhrystone 2.1	memory	1.5		81	
		Range 2;	Fibonacci		1.1		66	
		f <sub>HCLK</sub> = f <sub>HSl16</sub> = 16 MHz,	While(1) loop		1.0		59	
		PLL disabled,	Reduced code <sup>(3)</sup>		1.3		78	
		(2)	Coremark		1.4		88	
			Dhrystone 2.1	SRAM	1.4		88	
			Fibonacci		1.0		63	
			While(1) loop		1.0		60	
			Reduced code <sup>(3)</sup>	<u> </u>	345		173	
			Coremark	Flash	400		200	
			Dhrystone 2.1	memory	400		200	
	Supply	f <sub>HCLK</sub> = f <sub>HSI16</sub> /8 =	Fibonacci	<u> </u>	365		183	
Inn/Legun)	current in	2 MHz;	While(1) loop		415	uA	208	uA/MHz
DD(Li Null)	Low-power run mode	PLL disabled,	Reduced code <sup>(3)</sup>		225		113	
			Coremark		515		258	
			Dhrystone 2.1	SRAM	515		258	
			Fibonacci		295		148	
			While(1) loop		290		145	

<sup>1.</sup> Prefetch and cache enabled when fetching from Flash. Code compiled with high optimization for space in SRAM.

<sup>2.</sup>  $V_{DD}$  = 3.3 V, all peripherals disabled, cache enabled, prefetch disabled for execution in Flash and enabled in SRAM

<sup>3.</sup> Reduced code used for characterization results provided in *Table 25*.

Table 27. Current consumption in Sleep and Low-power sleep modes

		Condition	ons			Тур			Max <sup>(1</sup>	)	
	Parameter	General	Voltage scaling	f <sub>HCLK</sub>	25 °C	85 °C	125 °C	25 °C	85 °C	130 °C	Unit
				64 MHz	1.3	1.4	1.5	1.8	2.0	2.8	
				56 MHz	1.2	1.3	1.4	1.6	1.8	2.7	
		Flash memory enabled; f <sub>HCLK</sub> = f <sub>HSE</sub> bypass	Dango 1	48 MHz	1.1	1.1	1.3	1.5	1.6	2.6	
	Supply	(≤16 MHz; PLL	Range 1	32 MHz	0.8	0.8	1.0	1.2	1.2	2.2	
I <sub>DD(Sleep)</sub>	current in Sleep	disabled), f <sub>HCLK</sub> = f <sub>PLLRCLK</sub>		24 MHz	0.7	0.7	0.8	0.9	1.0	1.8	mA
(-:	mode	(>16 MHz; PLL		16 MHz	0.4	0.4	0.5	0.6	0.7	1.6	
		enabled); All peripherals disabled		16 MHz	0.3	0.4	0.5	0.5	0.6	1.3	
			Range 2	8 MHz	0.2	0.3	0.4	0.3	0.5	1.2	
				2 MHz	0.1	0.2	0.3	0.2	0.4	1.1	
				2 MHz	40	70	175	145	340	1055	
	Supply	Flash memory disabled; PLL disabled;		1 MHz	27	55	155	125	315	960	
I <sub>DD(LPSleep)</sub>	current in Low-power	f <sub>HCLK</sub> = f <sub>HSE</sub> bypass (> 3	32 kHz),	500 kHz	21	47	150	120	240	825	μΑ
	sleep mode	f <sub>HCLK</sub> = f <sub>LSE</sub> bypass (= 3 All peripherals disabled	32 kHz);	125 kHz	15	42	145	110	225	680	
		, ,		32 kHz	12	40	140	105	215	630	

<sup>1.</sup> Based on characterization results, not tested in production.

Table 28. Current consumption in Stop 0 mode

Symbol	Parameter	Conditions			Тур			Max <sup>(1)</sup>		Unit
Symbol	Farameter	HSI kernel	$V_{DD}$	25°C	85°C	125°C	25°C	85°C	130°C	Oilit
			1.8 V	265	290	375	330	420	745	
		Enabled	2.4 V	265	290	375	330	450	840	
	Supply current in Stop 0	Supply current in	3 V	270	295	380	345	485	940	
			3.6 V	270	295	385	370	495	1000	
I <sub>DD</sub> (Stop 0)			1.8 V	92	130	225	120	180	490	μA
	mode	Disabled	2.4 V	95	130	225	120	220	605	
		Disabled -	3 V	97	135	230	125	235	710	
			3.6 V	99	135	235	125	245	830	

<sup>1.</sup> Based on characterization results, not tested in production.

Table 29. Current consumption in Stop 1 mode

		C	Conditions			Тур		Max <sup>(1)</sup>			
Symbol	Parameter	Flash memory	RTC <sup>(2)</sup>	V <sub>DD</sub>	25°C	85°C	125°C	25°C	85°C	130°C	Unit
				1.8 V	2.9	25	105	9	95	590	
			Disabled	2.4 V	3.1	26	110	11	110	600	
	Supply current in	Not powered	Disabled	3 V	3.3	26	110	16	120	610	
				3.6 V	3.6	26	110	20	130	610	
			Enabled	1.8 V	3.3	25	105	11	95	610	
				2.4 V	3.6	26	110	14	110	615	
I <sub>DD(Stop 1)</sub>	Stop 1 mode			3 V	3.7	26	110	18	120	620	μΑ
	mode			3.6 V	4.2	27	110	23	130	630	1
				1.8 V	7.0	30	110	17	130	630	
		Dowored	Disabled	2.4 V	7.3	30	115	18	135	640	
		Powered	Disabled	3 V	7.5	30	115	19	140	660	
				3.6 V	7.8	31	115	23	140	700	

<sup>1.</sup> Based on characterization results, not tested in production.

Table 30. Current consumption in Standby mode

Symbol	Parameter	Condition	ns		Тур			Max <sup>(1)</sup>	)	Unit
Symbol	Parameter	General	$V_{DD}$	25°C	85°C	125°C	25°C	85°C	130°C	Onn
		RTC disabled	1.8 V	0.1	1.5	5	0.7	9	30	
			2.4 V	0.1	1.8	6	0.7	11	34	
		KTC disabled	3.0 V	0.2	2.1	7	0.9	14	39	
			3.6 V	0.4	2.5	8	1.0	16	45	
			1.8 V	0.3	1.7	5	0.8	14	33	
	F		2.4 V	0.4	2.1	6	1.0	14	39	μΑ
			3.0 V	0.6	2.5	7	1.2	16	46	
	Supply current in Standby		3.6 V	0.8	3.0	8	1.6	16	54	
IDD(Standby)	mode <sup>(2)</sup>	IWDG enabled,	1.8 V	0.2	1.6	5	0.8	10	32	
			2.4 V	0.3	2.0	6	8.0	12	37	
		clocked by LSI	3.0 V	0.4	2.4	7	1.1	14	44	
	ULPEN = 0		3.6 V	0.7	2.8	8	1.7	17	51	
			1.8 V	0.7	1.4	5	1.0	9	32	
		III DEN - 0	2.4 V	0.9	1.7	6	1.0	10	38	
		OLFEN - U	3.0 V	1.1	2.0	7	1.4	10	50	
		-	3.6 V	1.3	2.4	8	1.5	12	55	

<sup>2.</sup> Clocked by LSI

Table 30. Guitent consumption in Standby mode (Continued)										
Symbol	Parameter	Conditions		Тур			Max <sup>(1)</sup>			Unit
Symbol	raiametei	General	$V_{DD}$	25°C	85°C	125°C	25°C	85°C	130°C	Oill
Extra supply current to retain SRAM content <sup>(3)</sup>	current to	SRAM retention	1.8 V	0.3	8.0	3	0.4	7	21	
			2.4 V	0.4	0.9	3	0.9	8	21	μA
	enabled	3.0 V	0.5	1.0	3	1.3	8	23	μΛ	
	content		36 V	0.6	11	4	1.3	9	22	

Table 30. Current consumption in Standby mode (continued)

- 1. Based on characterization results, not tested in production.
- 2. Without SRAM retention and with ULPEN bit set
- 3. To be added to  $I_{DD(Standby)}$  as appropriate

Table 31. Current consumption in Shutdown mode

Symbol	Parameter	Conditions		Тур			Max <sup>(1)</sup>			Unit
- Cymbol		RTC	$V_{DD}$	25 °C	85 °C	125 °C	25 °C	85 °C	130 °C	
			1.8 V	12	370	2950	250	2700	26500	
	Disabled	2.4 V	18	430	3350	450	3000	29500		
	Supply current in Shutdown mode		3.0 V	25	500	3900	800	3700	33500	nA
			3.6 V	42	620	4550	900	4500	38500	
IDD(Shutdown)			1.8 V	185	550	3100	850	2700	26500	ш
		Enabled, clocked by LSE bypass at	2.4 V	265	675	3600	1450	3000	29500	
		32.768 kHz	3.0 V	350	835	4300	2350	3700	33500	
			3.6 V	460	1050	5050	3100	4400	38000	

<sup>1.</sup> Based on characterization results, not tested in production.

Table 32. Current consumption in VBAT mode

Symbol	Parameter	Conditions	5			Unit	
Symbol	Farameter	RTC	$V_{DD}$	25°C	85°C	125°C	Oilit
			1.8 V	180	240	770	
		Enabled, clocked by	2.4 V	250	330	910	
		LSE bypass at 32.768 kHz	3.0 V	330	420	1080	
	Supply current in VBAT mode		3.6 V	430	550	1310	
		Enabled, clocked by LSE crystal at 32.768 kHz	1.8 V	300	360	840	
			2.4 V	380	420	1010	nA
I <sub>DD(VBAT)</sub>			3.0 V	460	480	1190	IIA
			3.6 V	580	700	1410	
			1.8 V	1	59	590	
		Disabled	2.4 V	1	68	660	
		Disabled	3.0 V	2	78	740	
			3.6 V	5	97	860	

#### I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

#### I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 51: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

#### Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

#### I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 33: Current consumption of peripherals*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIO1} \times f_{SW} \times C$$

where

I<sub>SW</sub> is the current sunk by a switching I/O to charge/discharge the capacitive load

V<sub>DDIO1</sub> is the I/O supply voltage

f<sub>SW</sub> is the I/O switching frequency

C is the total capacitance seen by the I/O pin:  $C = C_{INT} + C_{EXT} + C_{S}$ 

C<sub>S</sub> is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

#### On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in the following table. The MCU is placed under the following conditions:

- All I/O pins are in Analog mode
- The given value is calculated by measuring the difference of the current consumptions:
  - when the peripheral is clocked on
  - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in *Table 18: Voltage characteristics*
- The power consumption of the digital part of the on-chip peripherals is given in the following table. The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

Table 33. Current consumption of peripherals

		Con	sumption in μA/	MHz
Peripheral	Bus	Range 1	Range 2	Low-power run and sleep
IOPORT Bus	IOPORT	0.5	0.4	0.3
GPIOA	IOPORT	3.1	2.4	3.0
GPIOB	IOPORT	2.9	2.3	3.0
GPIOC	IOPORT	0.9	0.8	1.0
GPIOD	IOPORT	0.7	0.6	1.0
GPIOF	IOPORT	0.5	0.5	1.0
Bus matrix	AHB	3.2	2.2	2.8
All AHB Peripherals	AHB	9.8	8.2	8.5
DMA1/DMAMUX	AHB	3.4	2.9	3.0
CRC	AHB	0.5	0.4	0.5
FLASH	AHB	4.3	3.6	3.5
All APB peripherals	APB	23.5	20.0	20.5
AHB to APB bridge <sup>(1)</sup>	APB	0.2	0.2	0.1
PWR	APB	0.4	0.3	0.5
SYSCFG/VREFBUF	APB	0.4	0.4	0.5
WWDG	APB	0.2	0.3	0.5
TIM1	APB	7.0	5.9	6.5
TIM2	APB	4.7	3.8	3.5
TIM3	APB	3.6	3.1	3.5
TIM14	APB	1.5	1.3	1.5
TIM16	APB	2.3	2.0	2.5
TIM17	APB	1.0	0.8	0.3
LPTIM1	APB	3.1	2.7	3.0
LPTIM2	APB	2.7	2.3	2.5



Table 33. Current consumption of peripherals (continued)

		Cor	Consumption in µA/MHz					
Peripheral	Bus	Range 1	Range 2	Low-power run and sleep				
I2C1	APB	3.2	2.7	3.0				
I2C2	APB	0.7	0.6	1.0				
SPI1	APB	2.2	1.8	2.0				
SPI2	APB	1.3	1.1	1.5				
USART1	APB	6.6	5.6	6.0				
USART2	APB	1.8	1.5	2.0				
LPUART1	APB	4.3	3.5	4.0				
ADC	APB	1.6	1.5	1.5				

<sup>1.</sup> The AHB to APB Bridge is automatically active when at least one peripheral is ON on the APB.

# 5.3.6 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in *Table 34* are the latency between the event and the execution of the first user instruction.

Table 34. Low-power mode wakeup times<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>WUSLEEP</sub>	Wakeup time from Sleep to Run - mode		11	11	CPU
t <sub>WULPSLEEP</sub>	Wakeup time from Low-power sleep mode	Transiting to Low-power-run-mode execution in Flash memory not powered in Low-power sleep mode; HCLK = HSI16 / 8 = 2 MHz	11	14	cycles
	Wakeup time from	Transiting to Run-mode execution in Flash memory not powered in Stop 0 mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1 or Range 2	5.6	6	110
<sup>T</sup> WUSTOP0	Stop 0	Transiting to Run-mode execution in SRAM or in Flash memory powered in Stop 0 mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1 or Range 2	2	2.4	μs

Table 34. Low-power mode wakeup times<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Тур	Max	Unit
		Transiting to Run-mode execution in Flash memory not powered in Stop 1 mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1 or Range 2		11.2	
+	Wakeup time from	Transiting to Run-mode execution in SRAM or in Flash memory powered in Stop 1 mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1 or Range 2		7.5	110
<sup>t</sup> WUSTOP1	Stop 1	Transiting to Low-power-run-mode execution in Flash memory not powered in Stop 1 mode; HCLK = HSI16/8 = 2 MHz; Regulator in low-power mode (LPR = 1 in PWR_CR1)		25.3	μs
		Transiting to Low-power-run-mode execution in SRAM or in Flash memory powered in Stop 1 mode; HCLK = HSI16 / 8 = 2 MHz; Regulator in low-power mode (LPR = 1 in PWR_CR1)		23.5	
twustby	Wakeup time from Standby mode	Transiting to Run mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1	14.5	30	μs
twushdn	Wakeup time from Shutdown mode	Transiting to Run mode; HCLK = HSI16 = 16 MHz; Regulator in Range 1	258	340	μs
t <sub>WULPRUN</sub>	Wakeup time from Low-power run mode <sup>(2)</sup>	Transiting to Run mode; HSISYS = HSI16/8 = 2 MHz	5	7	μs

<sup>1.</sup> Based on characterization results, not tested in production.

Table 35. Regulator mode transition times<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>VOST</sub>	Transition times between regulator Range 1 and Range 2 <sup>(2)</sup>	HSISYS = HSI16	20	40	μs

<sup>1.</sup> Based on characterization results, not tested in production.

Table 36. Wakeup time using LPUART<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
1 DI IADT haved gate allowing to walke up up from Ctan	Stop mode 0	ı	1.7	μs	
<sup>I</sup> WULPUART	mode when LPUART clock source is HSI16	Stop mode 1	-	8.5	μδ

<sup>1.</sup> Guaranteed by design.



<sup>2.</sup> Time until REGLPF flag is cleared in PWR\_SR2.

<sup>2.</sup> Time until VOSF flag is cleared in PWR\_SR2.

#### 5.3.7 External clock source characteristics

#### High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

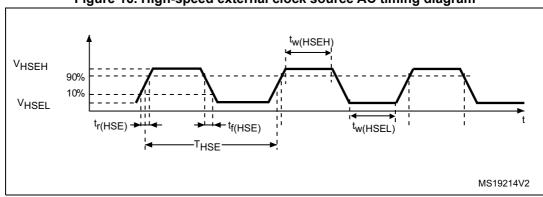
The external clock signal has to respect the I/O characteristics in *Section 5.3.14*. See *Figure 16* for recommended clock input waveform.

Table 37. High-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
f	f <sub>HSE ext</sub> User external clock source frequency	Voltage scaling Range 1	-	8	48	MHz	
f <sub>HSE_ext</sub> User external clock source frequency	Voltage scaling Range 2	-	8	26	IVIITZ		
$V_{HSEH}$	OSC_IN input pin high level voltage	-	0.7 V <sub>DDIO1</sub>	-	V <sub>DDIO1</sub>	V	
V <sub>HSEL</sub>	OSC_IN input pin low level voltage	-	V <sub>SS</sub>	-	0.3 V <sub>DDIO1</sub>	V	
t <sub>w(HSEH)</sub>	OSC_IN high or low time	Voltage scaling Range 1	7	-	-	no	
t <sub>w(HSEL)</sub>	OSC_IN High or low time	Voltage scaling Range 2	18	-	-	ns	

<sup>1.</sup> Guaranteed by design.

Figure 16. High-speed external clock source AC timing diagram



#### Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in *Section 5.3.14*. See *Figure 17* for recommended clock input waveform.

Table 38. Low-speed external user clock characteristics<sup>(1)</sup>

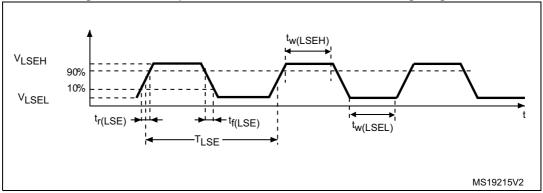
Ī	Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Ī	f <sub>LSE_ext</sub>	User external clock source frequency	-	-	32.768	1000	kHz

	(**************************************									
Symbol	Parameter	Conditions	Min	Тур	Max	Unit				
$V_{LSEH}$	OSC32_IN input pin high level voltage	-	0.7 V <sub>DDIO1</sub>	-	V <sub>DDIO1</sub>	W				
$V_{LSEL}$	OSC32_IN input pin low level voltage	-	$V_{SS}$	-	0.3 V <sub>DDIO1</sub>	V				
t <sub>w(LSEH)</sub>	OSC32_IN high or low time	-	250	-	-	ns				

Table 38. Low-speed external user clock characteristics<sup>(1)</sup> (continued)

<sup>1.</sup> Guaranteed by design.





#### High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 48 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 39*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 39. HSE oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
f <sub>OSC_IN</sub>	Oscillator frequency	-	4	8	48	MHz
$R_{F}$	Feedback resistor	-	-	200	-	kΩ

Table 39. HSE oscillator characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
		During startup <sup>(3)</sup>	-	-	5.5	
		$V_{DD} = 3 \text{ V},$ Rm = 30 $\Omega$ , CL = 10 pF@8 MHz	-	0.58	-	mA/V ms
		V <sub>DD</sub> = 3 V, Rm = 45 Ω, CL = 10 pF@8 MHz	-	0.59	-	
I <sub>DD(HSE)</sub>	HSE current consumption	$V_{DD} = 3 \text{ V},$ Rm = 30 $\Omega$ , CL = 5 pF@48 MHz	-	0.89	-	
		V <sub>DD</sub> = 3 V, Rm = 30 Ω, CL = 10 pF@48 MHz	-	1.14	-	
		V <sub>DD</sub> = 3 V, Rm = 30 Ω, CL = 20 pF@48 MHz	-	1.94	-	
G <sub>m</sub>	Maximum critical crystal transconductance	Startup	-	-	1.5	mA/V
t <sub>SU(HSE)</sub> <sup>(4)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	ms

- 1. Guaranteed by design.
- 2. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
- 3. This consumption level occurs during the first 2/3 of the  $t_{SU(HSE)}$  startup time
- 4. t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 18*).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ .

Note:

For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

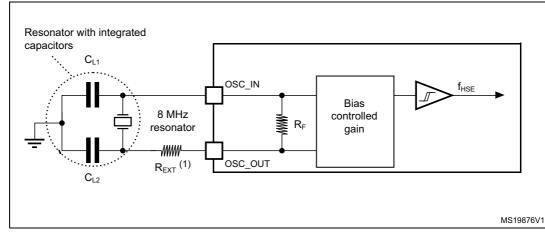


Figure 18. Typical application with an 8 MHz crystal

1.  $R_{\text{EXT}}$  value depends on the crystal characteristics.

#### Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 40*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
		LSEDRV[1:0] = 00 Low drive capability	-	250	-	
1	LSE ourrent consumption	LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	20
I <sub>DD(LSE)</sub>	LSE current consumption	LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	nA
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
	LSEDRV[1:0] = 00 Low drive capability		-	-	0.5	
Gm	Maximum critical crystal	LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	μΑ/V
Gm <sub>critmax</sub>	gm	LSEDRV[1:0] = 10 Medium high drive capability	-	-	1.7	μΑνν
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
t <sub>SU(LSE)</sub> <sup>(3)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	S

Table 40. LSE oscillator characteristics  $(f_{LSE} = 32.768 \text{ kHz})^{(1)}$ 

Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".



<sup>1.</sup> Guaranteed by design.

 t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

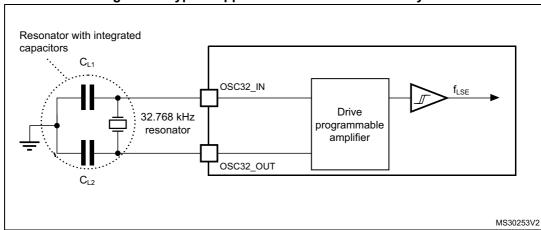


Figure 19. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.

#### 5.3.8 Internal clock source characteristics

The parameters given in *Table 41* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*. The provided curves are characterization results, not tested in production.

# High-speed internal (HSI16) RC oscillator

Table 41. HSI16 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI16</sub>	f <sub>HSI16</sub> HSI16 Frequency		15.88	-	16.08	MHz
_	HSI16 oscillator frequency drift over	T <sub>A</sub> = 0 to 85 °C	-1	ı	1	%
$\Delta$ Temp(HSI16)	temperature	T <sub>A</sub> = -40 to 125 °C	-2	-	1.5	%
$\Delta_{ extsf{VDD}( ext{HSI16})}$	HSI16 oscillator frequency drift over $V_{DD}$	V <sub>DD</sub> =1.62 V to 3.6 V	-0.1	-	0.05	%
		From code 127 to 128	-8	-6	-4	
TRIM	HSI16 frequency user trimming step	From code 63 to 64 From code 191 to 192	-5.8	-3.8	-1.8	%
		For all other code increments	0.2	0.3	0.4	
D <sub>HSI16</sub> <sup>(2)</sup>	Duty Cycle	-	45	-	55	%
t <sub>su(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator start-up time	-	-	0.8	1.2	μs

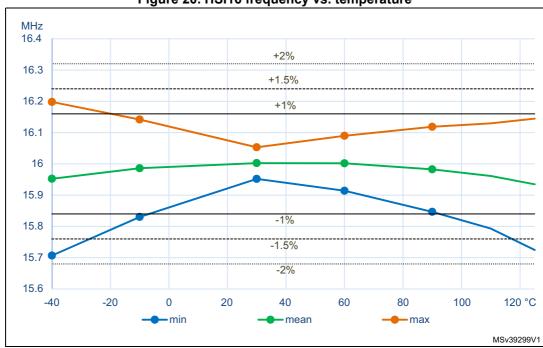
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Table 41. HSI16 oscillator characteristics<sup>(1)</sup> (continued)

Symbol Parameter		Conditions	Min	Тур	Max	Unit
t <sub>stab(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator stabilization time	-	-	3	5	μs
I <sub>DD(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator power consumption	-	-	155	190	μΑ

- 1. Based on characterization results, not tested in production.
- 2. Guaranteed by design.

Figure 20. HSI16 frequency vs. temperature



# Low-speed internal (LSI) RC oscillator

Table 42. LSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		V <sub>DD</sub> = 3.0 V, T <sub>A</sub> = 30 °C	31.04	-	32.96	
f <sub>LSI</sub>	LSI frequency	V <sub>DD</sub> = 1.62 V to 3.6 V, T <sub>A</sub> = -40 to 125 °C	29.5	-	34	kHz
t <sub>SU(LSI)</sub> <sup>(2)</sup>	LSI oscillator start-up time	-	-	80	130	μs
t <sub>STAB(LSI)</sub> (2)	LSI oscillator stabilization time	5% of final frequency	-	125	180	μs
I <sub>DD(LSI)</sub> <sup>(2)</sup>	LSI oscillator power consumption	-	-	110	180	nA

- 1. Based on characterization results, not tested in production.
- 2. Guaranteed by design.

## 5.3.9 PLL characteristics

The parameters given in *Table 43* are derived from tests performed under temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 43. PLL characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>PLL_IN</sub>	PLL input clock frequency <sup>(2)</sup>	-	2.66	-	16	MHz
D <sub>PLL_IN</sub>	PLL input clock duty cycle	-	45	-	55	%
f	PLL multiplier output clock P	Voltage scaling Range 1	3.09	-	122	MHz
f <sub>PLL_P_OUT</sub>	T LE Maitiplier output clock i	Voltage scaling Range 2	3.09	-	40	IVII IZ
f <sub>PLL_Q_OUT</sub>	PLL multiplier output clock Q	Voltage scaling Range 1	12	-	128	MHz
	FEE mainplier output clock Q	Voltage scaling Range 2	12	-	33	IVII IZ
for a see	PLL multiplier output clock R	Voltage scaling Range 1	12	-	64	- MHz
†PLL_R_OUT	FEE mainplier output clock K	Voltage scaling Range 2	12	-	16	IVII IZ
f	PLL VCO output	Voltage scaling Range 1	96	-	344	- MHz
f <sub>VCO_OUT</sub>	FEE VOO output	Voltage scaling Range 2	96	-	128	IVII IZ
t <sub>LOCK</sub>	PLL lock time	-	-	15	40	μs
Jitter	RMS cycle-to-cycle jitter	System clock 56 MHz	-	50	-	±nc.
Jillei	RMS period jitter	- System clock 30 MHz	-	40	-	±ps
		VCO freq = 96 MHz	- 200	200	260	
I <sub>DD(PLL)</sub>	PLL power consumption on V <sub>DD</sub> <sup>(1)</sup>	VCO freq = 192 MHz	-	300	380	μΑ
		VCO freq = 344 MHz	-	520	650	

<sup>1.</sup> Guaranteed by design.

# 5.3.10 Flash memory characteristics

Table 44. Flash memory characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>prog</sub>	64-bit programming time	-	85	125	μs
+	Row (32 double word) programming time	Normal programming	2.7	4.6	
<sup>t</sup> prog_row	(32 double word) programming time	Fast programming	1.7	2.8	
+	Page (2 Kbyte) programming time	Normal programming	21.8	36.6	ms
<sup>l</sup> prog_page	Page (2 Kbyte) programming time	Fast programming	13.7	22.4	
t <sub>ERASE</sub>	Page (2 Kbyte) erase time	-	22.0	40.0	
+	Bank (64 Kbyte <sup>(2)</sup> ) programming time	Normal programming	0.7	1.2	
<sup>T</sup> prog_bank	Bank (04 Kbyte 7) programming time	Fast programming	0.4	0.7	S
t <sub>ME</sub>	Mass erase time	-	22.1	40.1	ms



<sup>2.</sup> Take care of using the appropriate division factor M to obtain the specified PLL input clock values.

Symbol	Parameter	Conditions	Тур	Max	Unit
		Programming	3	-	mA
I <sub>DD(FlashA)</sub>	Average consumption from V <sub>DD</sub>	Page erase	3	-	mA
		Mass erase	5	-	
I <sub>DD(FlashP)</sub>	Maximum current (peak)	Programming, 2 µs peak duration	7	-	mA
( )		Erase, 41 µs peak duration	7	-	

Table 44. Flash memory characteristics<sup>(1)</sup> (continued)

<sup>2.</sup> Values provided also apply to devices with less Flash memory than one 64 Kbyte bank

Tubic 40. Fluori memory endurance and data retention								
Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Unit				
N <sub>END</sub>	Endurance	T <sub>A</sub> = -40 to +105 °C	10	kcycles				
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30					
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	15					
	Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 125 °C	7	Vann				
t <sub>RET</sub>	Data retention	10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 55 °C	30	Years				
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 85 °C	15					
		10 kcycles <sup>(2)</sup> at T <sub>4</sub> = 105 °C	10					

Table 45. Flash memory endurance and data retention

#### 5.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 46*. They are based on the EMS levels and classes defined in application note AN1709.



<sup>1.</sup> Guaranteed by design.

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Cycling performed over the whole temperature range.

Table	16	<b>EMS</b>	characteristics
Table	40.		Characteristics

Symbol	Parameter	Conditions	Level/ Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD}$ = 3.3 V, $T_A$ = +25 °C, $f_{HCLK}$ = 64 MHz, LQFP48, conforming to IEC 61000-4-2	2B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on $V_{DD}$ and $V_{SS}$ pins to induce a functional disturbance	$V_{DD}$ = 3.3 V, $T_{A}$ = +25 °C, $f_{HCLK}$ = 64 MHz, LQFP48, conforming to IEC 61000-4-4	5A

# Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

#### **Software recommendations**

The software flowchart must include the management of runaway conditions such as:

- corrupted program counter
- unexpected reset
- critical data corruption (for example control registers)

#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

#### **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f <sub>HSE</sub> /f <sub>HCLK</sub> ]  8 MHz / 64 MHz	Unit	
S <sub>EMI</sub>	Peak level	V <sub>DD</sub> = 3.6 V, T <sub>A</sub> = 25 °C, LQFP64 package compliant with IEC 61967-2	0.1 MHz to 30 MHz	-4		
			30 MHz to 130 MHz	1	dΒμV	
			130 MHz to 1 GHz	3		
			1 GHz to 2 GHz	8		
			EMI level	2.5	-	

Table 47. EMI characteristics

## 5.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 48. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	$T_A$ = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001	2	2000	V
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C, conforming to ANSI/ESDA/JEDEC JS-002	C2a	500	V

<sup>1.</sup> Based on characterization results, not tested in production.

#### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current is injected to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 49. Electrical sensitivity

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +125 °C conforming to JESD78	II Level A



# 5.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DDIO1}$  (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out-of-range parameter: ADC error above a certain limit (higher than 5 LSB TUE), induced leakage current on adjacent pins out of conventional limits (-5  $\mu$ A/+0  $\mu$ A range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

	iabi	e 30. I/O current injection sus	ceptibility		
Symbol			Functional s		
		Description	Negative injection	Positive injection	Unit
		All except PA1, PA3, PA5, PA6, PA13, PB0, PB1, PB2, and PB8	-5	N/A	
I <sub>INJ</sub>	Injected current on pin	PA1, PA5, PA13, PB1, PB2	0	+5 / N/A <sup>(2)</sup>	mA
		PA3, PA6, PB0	-5	+5 / N/A <sup>(2)</sup>	
	PB8		0	N/A	

Table 50. I/O current injection susceptibility<sup>(1)</sup>

<sup>1.</sup> Based on characterization results, not tested in production.

<sup>2.</sup> The injection current value is applicable when the switchable diode is activated, N/A when not activated.

# 5.3.14 I/O port characteristics

# General input/output characteristics

Unless otherwise specified, the parameters given in *Table 51* are derived from tests performed under the conditions summarized in *Table 21: General operating conditions*. All I/Os are designed as CMOS- and TTL-compliant.

Table 51. I/O static characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IL</sub> <sup>(1)</sup>	I/O input low level voltage	All	1.62 V < V <sub>DDIO1</sub> < 3.6 V	-	-	0.3 x V <sub>DDIO1</sub>	V
VIL.						0.39 x V <sub>DDIO1</sub> - 0.06 <sup>(3)</sup>	
V <sub>IH</sub> <sup>(1)</sup>	I/O input high level voltage		1.62 V < V <sub>DDIO1</sub> < 3.6 V	0.7 x V <sub>DDIO1</sub> (	-	-	V
VIH		All		0.49 x V <sub>DDIO1</sub> + 0.26 <sup>(3)</sup>	-	-	V
V <sub>hys</sub> <sup>(3)</sup>	I/O input hysteresis	FT_xx, NRST	1.62 V < V <sub>DDIO1</sub> < 3.6 V	-	200	-	mV
	Input leakage current <sup>(3)</sup>	All except FT_e	$0 < V_{IN} \leq V_{DDIO1}$	-	-	±70	nA
			$V_{\text{DDIO1}} \le V_{\text{IN}} \le V_{\text{DDIO1}} + 1 \text{ V}$	-	-	600 <sup>(4)</sup>	
I <sub>lkg</sub>			$V_{\rm DDIO1} + 1 \ V < V_{\rm IN} \le 5.5 \ V^{(3)}$	-	-	150 <sup>(4)</sup>	
		FT_e	0 < VIN < VDDIO1	-	-	5	μA
R <sub>PU</sub>	Weak pull-up equivalent resistor (6)	$V_{IN} = V_{SS}$		25	40	55	kΩ
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(6)</sup>	V <sub>IN</sub> = V <sub>DDIO1</sub>		25	40	55	kΩ
C <sub>IO</sub>	I/O pin capacitance		-	-	5	-	pF

<sup>1.</sup> Refer to Figure 21: I/O input characteristics.

<sup>2.</sup> Tested in production.

<sup>3.</sup> Guaranteed by design.

<sup>4.</sup> This value represents the pad leakage of the I/O itself. The total product pad leakage is provided by this formula:  $I_{Total\_Ileak\_max} = 10 \ \mu A + [number of I/Os where V_{IN}]$  is applied on the pad]  $x \ I_{lkg}(Max)$ .

<sup>5.</sup> FT\_e with diode enabled. Input leakage current of FT\_e I/Os with the diode disabled is the same as standard I/Os.

Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters, as shown in *Figure 21*.

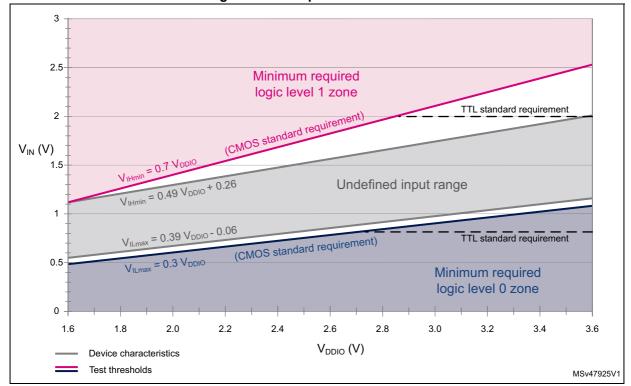


Figure 21. I/O input characteristics

## Characteristics of FT\_e I/Os

The following table and figure specify input characteristics of FT e I/Os.

**Symbol Conditions** Unit **Parameter** Min Max Тур Injected current on pin 5  $I_{INJ}$ mΑ ٧  $V_{DDIO1}-V_{IN}$ Voltage over V<sub>DDIO1</sub> 2  $I_{INJ} = 5 \text{ mA}$ Diode dynamic serial resistor  $I_{INJ} = 5 \text{ mA}$ 300  $R_{d}$ Ω

Table 52. Input characteristics of FT\_e I/Os

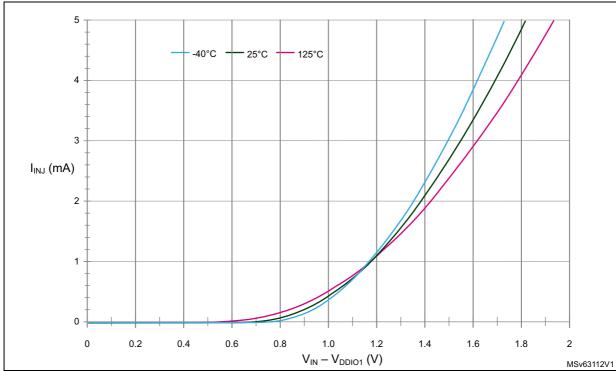


Figure 22. Current injection into FT\_e input with diode active

#### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 6$  mA, and up to  $\pm 15$  mA with relaxed  $V_{OL}/V_{OH}$ .

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 5.2*:

- The sum of the currents sourced by all the I/Os on V<sub>DDIO1</sub>, plus the maximum consumption of the MCU sourced on V<sub>DD</sub>, cannot exceed the absolute maximum rating I<sub>VDD</sub> (see *Table 18: Voltage characteristics*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub>, plus the maximum consumption of the MCU sunk on V<sub>SS</sub>, cannot exceed the absolute maximum rating I<sub>VSS</sub> (see *Table 18: Voltage characteristics*).

#### **Output voltage levels**

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*. All I/Os are CMOS- and TTL-compliant (FT OR TT unless otherwise specified).

	Table 33. Outpu	it voitage characteristics	,		
Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OL}$	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup>	-	0.4	
V <sub>OH</sub>	Output high level voltage for an I/O pin	I <sub>IO</sub>   6 mA V <sub>DDIO1</sub> ≥ 2.7 V	V <sub>DDIO1</sub> - 0.4	-	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	TTL port <sup>(2)</sup>	-	0.4	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 6 mA V <sub>DDIO1</sub> ≥ 2.7 V	2.4	-	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	All I/Os	-	1.3	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 15 mA V <sub>DDIO1</sub> ≥ 2.7 V	V <sub>DDIO1</sub> - 1.3	-	٧
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 3 mA	-	0.4	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	V <sub>DDIO1</sub> ≥ 1.62 V	V <sub>DDIO1</sub> - 0.45	-	
V <sub>OLFM+</sub>	Output low level voltage for an FT I/O	I <sub>IO</sub>   = 20 mA V <sub>DDIO1</sub> ≥ 2.7 V	-	0.4	
(3)	pin in FM+ mode (FT I/O with _f option)	I <sub>IO</sub>   = 9 mA V <sub>DDIO1</sub> ≥ 1.62 V	-	0.4	

Table 53. Output voltage characteristics<sup>(1)</sup>

### Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 23* and *Table 54*, respectively.

Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Speed	Symbol	Parameter	Conditions	Min	Max	Unit
			C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	2	
	Fmax		C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	0.35	MHz
	FILIAX	Maximum frequency	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	3	IVIITZ
00			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	0.45	
00		T. T.	C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	100	
	Tr/Tf		C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	225	no
Tr/Tf	Output rise and fall time	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	75	ns	
			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	_	150	1

Table 54. I/O AC characteristics<sup>(1)(2)</sup>

The I<sub>IO</sub> current sourced or sunk by the device must always respect the absolute maximum rating specified in *Table 18: Voltage characteristics*, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI<sub>IO</sub>.

<sup>2.</sup> TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

<sup>3.</sup> Guaranteed by design.

Table 54. I/O AC characteristics<sup>(1)(2)</sup> (continued)

Speed	Symbol	Parameter	Conditions	Min	Max	Unit
			C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	10	
	Fmax	Maximum frequency	C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	2	MHz
	IIIIax	iviaximum nequency	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	15	IVIIIZ
01			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	2.5	]
01			C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	30	
	Tr/Tf	Output rise and fall time	C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	60	ne
	11/11	Output rise and fair time	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	15	ns
			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	30	1
			C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	30	
	Fmax	Maximum frequency	C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	15	MHz
	Tillax	iviaximum frequency	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	60	IVII IZ
10			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	30	1
10			C=50 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	11	
	Tr/Tf	Output rise and fall time	C=50 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	22	ne
	11/11	Output rise and fair time	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	4	ns
			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	8	1
			C=30 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	60	
	Fmax	Maximum frequency	C=30 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	30	MHz
	Fillax	iviaximum frequency	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	80 <sup>(3)</sup>	IVITIZ
11			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	40	1
11			C=30 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	5.5	
	Tr/Tf	Output rice and fall time	C=30 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	11	]
	11/11	Output rise and fall time	C=10 pF, 2.7 V ≤ V <sub>DDIO1</sub> ≤ 3.6 V	-	2.5	ns
			C=10 pF, 1.6 V ≤ V <sub>DDIO1</sub> ≤ 2.7 V	-	5	
Fm+	Fmax	Maximum frequency	C-50 pF 16 V < V < 22 C V	-	1	MHz
FIII+	Tf	Output fall time <sup>(4)</sup>	C=50 pF, 1.6 V $\leq$ V <sub>DDIO1</sub> $\leq$ 3.6 V	-	5	ns

The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG\_CFGR1 register. Refer to the RM0444 reference manual for a description of GPIO Port configuration register.

<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> This value represents the I/O capability but the maximum system frequency is limited to 64 MHz.

<sup>4.</sup> The fall time is defined between 70% and 30% of the output waveform, according to  $I^2C$  specification.

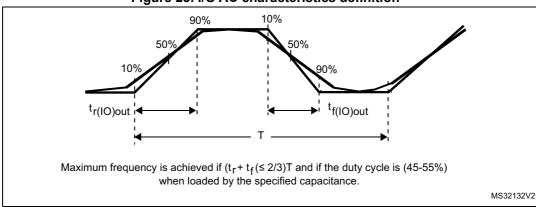


Figure 23. I/O AC characteristics definition<sup>(1)</sup>

1. Refer to Table 54: I/O AC characteristics.

## 5.3.15 NRST input characteristics

The NRST input driver uses CMOS technology. It is connected to a permanent pull-up resistor,  $\mathsf{R}_{\text{PU}}.$ 

Unless otherwise specified, the parameters given in the following table are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Tuble 66: 14161 pin characteristics						
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL(NRST)</sub>	NRST input low level voltage	-	-	-	0.3 x V <sub>DDIO1</sub>	V
V <sub>IH(NRST)</sub>	NRST input high level voltage	-	0.7 x V <sub>DDIO1</sub>	-	-	
V <sub>hys(NRST)</sub>	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(2)</sup>	V <sub>IN</sub> = V <sub>SS</sub>	25	40	55	kΩ
V <sub>F(NRST)</sub>	NRST input filtered pulse	-	-	-	70	ns
V <sub>NF(NRST)</sub>	NRST input not filtered pulse	1.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	350	-	-	ns

Table 55. NRST pin characteristics<sup>(1)</sup>

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).

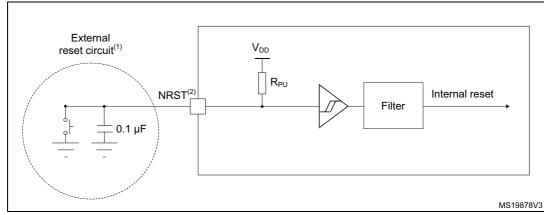


Figure 24. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V<sub>IL(NRST)</sub> max level specified in Table 55: NRST pin characteristics. Otherwise the reset will not be taken into account by the device.
- 3. The external capacitor on NRST must be placed as close as possible to the device.

### 5.3.16 Analog switch booster

Table 56. Analog switch booster characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Тур	Max	Unit
$V_{DD}$	Supply voltage	1.62 V	-	3.6	V
t <sub>SU(BOOST)</sub>	Booster startup time	-	-	240	μs
I <sub>DD(BOOST)</sub>	Booster consumption for 1.62 V ≤ V <sub>DD</sub> ≤ 2.0 V	-	-	250	
	Booster consumption for 2.0 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	-	500	μΑ
	Booster consumption for $2.7 \text{ V} \leq \text{V}_{\text{DD}} \leq 3.6 \text{ V}$	-	-	900	

<sup>1.</sup> Guaranteed by design.

### 5.3.17 Analog-to-digital converter characteristics

Unless otherwise specified, the parameters given in *Table 57* are preliminary values derived from tests performed under ambient temperature, f<sub>PCLK</sub> frequency and V<sub>DDA</sub> supply voltage conditions summarized in *Table 21: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

Table 57. ADC characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage	-	1.62	-	3.6	V
V	Positive reference	V <sub>DDA</sub> ≥ 2 V	2	-	$V_{DDA}$	V
V <sub>REF+</sub>	voltage	V <sub>DDA</sub> < 2 V	$V_{DDA}$			V



Table 57. ADC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
,	100 1 16	Range 1	0.14	-	35	
f <sub>ADC</sub>	ADC clock frequency	Range 2	0.14	-	16	MHz
		12 bits	-	-	2.50	
	O a mana lina na manta	10 bits	-	-	2.92	MOrra
f <sub>s</sub>	Sampling rate	8 bits	-	-	3.50	MSps
		6 bits	-	-	4.38	
£	External trigger	f <sub>ADC</sub> = 35 MHz; 12 bits	-	-	2.33	N.41.1-
f <sub>TRIG</sub>	frequency	12 bits	-	-	f <sub>ADC</sub> /15	MHz
V <sub>AIN</sub> <sup>(3)</sup>	Conversion voltage range	-	V <sub>SSA</sub>	-	V <sub>REF+</sub>	V
R <sub>AIN</sub>	External input impedance	-	-	-	50	kΩ
C <sub>ADC</sub>	Internal sample and hold capacitor	-	-	5	-	pF
t <sub>STAB</sub>	ADC power-up time	-		2		Conversion cycle
	Calibration times	f <sub>ADC</sub> = 35 MHz		2.35		μs
t <sub>CAL</sub>	Calibration time - 82			1/f <sub>ADC</sub>		
		CKMODE = 00	2	-	3	1/f <sub>ADC</sub>
+	Trigger conversion	CKMODE = 01		6.5		
t <sub>LATR</sub>	latency	CKMODE = 10		12.5		
		CKMODE = 11		3.5	_	
		f <sub>ADC</sub> = 35 MHz;	0.043	-	4.59	μs
t <sub>s</sub>	Sampling time	V <sub>DDA</sub> > 2V	1.5	-	160.5	1/f <sub>ADC</sub>
'S		$f_{ADC}$ = 35 MHz;	0.1	-	4.59	μs
		V <sub>DDA</sub> < 2V	3.5		160.5	1/f <sub>ADC</sub>
t <sub>ADCVREG_STUP</sub>	ADC voltage regulator start-up time	-	-	-	20	μs
	Total conversion time	f <sub>ADC</sub> = 35 MHz Resolution = 12 bits	0.40	-	4.95	μs
t <sub>CONV</sub>	(including sampling time)	Resolution = 12 bits	t <sub>s</sub> + 12.5 cycles for successive approximation = 14 to 173		1/f <sub>ADC</sub>	
t <sub>IDLE</sub>	Laps of time allowed between two conversions without rearm	-	-	-	100	μs

				/		
Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Тур	Max	Unit
		f <sub>s</sub> = 2.5 MSps	-	410	-	
I <sub>DDA(ADC)</sub>	ADC consumption from V <sub>DDA</sub>	f <sub>s</sub> = 1 MSps	-	164	-	μΑ
		f <sub>s</sub> = 10 kSps	-	17	-	
		f <sub>s</sub> = 2.5 MSps	-	65	-	
I <sub>DDV(ADC)</sub>	ADC consumption from V <sub>REF+</sub>	f <sub>s</sub> = 1 MSps	-	26	-	μΑ
	- INLI F	f <sub>s</sub> = 10 kSps	-	0.26	-	1

Table 57. ADC characteristics<sup>(1)</sup> (continued)

Table 58. Maximum ADC R<sub>AIN</sub>

Resolution	Sampling cycle at 35 MHz	Sampling time at 35 MHz [ns]	Max. R <sub>AIN</sub> <sup>(1)(2)</sup> (Ω)
	1.5 <sup>(3)</sup>	43	50
	3.5	100	680
	7.5	214	2200
12 bits	12.5	357	4700
12 0115	19.5	557	8200
	39.5	1129	15000
	79.5	2271	33000
	160.5	4586	50000
	1.5 <sup>(3)</sup>	43	68
	3.5	100	820
	7.5	214	3300
10 hita	12.5	357	5600
10 bits	19.5	557	10000
	39.5	1129	22000
	79.5	2271	39000
	160.5	4586	50000

<sup>1.</sup> Guaranteed by design

<sup>2.</sup> I/O analog switch voltage booster must be enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA}$  < 2.4 V and disabled when  $V_{DDA} \ge 2.4$  V.

V<sub>REF+</sub> is internally connected to V<sub>DDA</sub> on some packages.Refer to Section 4: Pinouts, pin description and alternate functions for further details.

Table 58. Maximum ADC  $R_{AIN}$  (continued)

Resolution	Sampling cycle at 35 MHz	Sampling time at 35 MHz [ns]	Max. R <sub>AIN</sub> <sup>(1)(2)</sup> (Ω)
	1.5 <sup>(3)</sup>	43	82
	3.5	100	1500
	7.5	214	3900
Q hita	12.5	357	6800
8 bits	19.5	557	12000
	39.5	1129	27000
	79.5	2271	50000
	160.5	4586	50000
	1.5 <sup>(3)</sup>	43	390
	3.5	100	2200
	7.5	214	5600
C hita	12.5	357	10000
6 bits	19.5	557	15000
	39.5	1129	33000
	79.5	2271	50000
	160.5	4586	50000

<sup>1.</sup> Guaranteed by design.

Table 59. ADC accuracy<sup>(1)(2)(3)</sup>

Symbol	Parameter	Conditions <sup>(4)</sup>	Min	Тур	Max	Unit
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_{S} \le 2.5 \text{ MSps};$ $T_{A} = 25 \text{ °C}$	-	3	4	
ET	Total unadjusted error	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_{s}$ ≤ 2.5 MSps; $T_{A}$ = entire range	-	3	6.5	LSB
		1.65 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $T_A$ = entire range Range 1: $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.2 MSps; Range 2: $f_{ADC}$ = 16 MHz; $f_s$ ≤ 1.1 MSps;	-	3	7.5	

<sup>2.</sup> I/O analog switch voltage booster must be enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA} < 2.4 \text{ V}$  and disabled when  $V_{DDA} \ge 2.4 \text{ V}$ .

<sup>3.</sup> Only allowed with  $V_{DDA} > 2 V$ 

Table 59. ADC accuracy<sup>(1)(2)(3)</sup> (continued)

Symbol	Parameter	Conditions <sup>(4)</sup>	Min	Тур	Max	Unit
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_{S} \le 2.5 \text{ MSps};$ $T_{A} = 25 \text{ °C}$	-	1.5	2	
EO	Offset error	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_{s}$ ≤ 2.5 MSps; $T_{A}$ = entire range	-	1.5	4.5	LSB
		$\begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } f_{\text{ADC}} = 35 \text{ MHz}; f_{\text{s}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } f_{\text{ADC}} = 16 \text{ MHz}; f_{\text{s}} \leq 1.1 \text{ MSps}; \\ \end{array}$	-	1.5	5.5	
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_{S} \le 2.5 \text{ MSps};$ $T_{A} = 25 \text{ °C}$	-	3	3.5	
EG	Gain error	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_{s}$ ≤ 2.5 MSps; $T_{A}$ = entire range	-	3	5	LSB
		$ \begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } f_{\text{ADC}} = 35 \text{ MHz}; f_{\text{S}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } f_{\text{ADC}} = 16 \text{ MHz}; f_{\text{S}} \leq 1.1 \text{ MSps}; \\ \end{array} $	-	3	6.5	
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_{S} \le 2.5 \text{ MSps};$ $T_{A} = 25 ^{\circ}\text{C}$	-	1.2	1.5	
ED	Differential linearity error	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_{s}$ ≤ 2.5 MSps; $T_{A}$ = entire range	-	1.2	1.5	LSB
		$\begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } f_{\text{ADC}} = 35 \text{ MHz}; f_{\text{s}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } f_{\text{ADC}} = 16 \text{ MHz}; f_{\text{s}} \leq 1.1 \text{ MSps}; \\ \end{array}$	-	1.2	1.5	
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_{S} \le 2.5 \text{ MSps};$ $T_{A} = 25 ^{\circ}\text{C}$	-	2.5	3	
EL	Integral linearity error	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_{s}$ ≤ 2.5 MSps; $T_{A}$ = entire range	-	2.5	3	LSB
	oa.ny onoi	$ \begin{array}{l} 1.65~\text{V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < \ 3.6~\text{V}; \\ T_{\text{A}} = \text{entire range} \\ \text{Range 1: } f_{\text{ADC}} = 35~\text{MHz}; \ f_{\text{s}} \leq 2.2~\text{MSps}; \\ \text{Range 2: } f_{\text{ADC}} = 16~\text{MHz}; \ f_{\text{s}} \leq 1.1~\text{MSps}; \\ \end{array} $	-	2.5	3.5	

Table 59. ADC accuracy<sup>(1)(2)(3)</sup> (continued)

Symbol	Parameter	Conditions <sup>(4)</sup>	Min	Тур	Max	Unit
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_s \le 2.5 \text{ MSps};$ $T_A = 25 \text{ °C}$	10.1	10.2	-	
ENOB	ENOB Effective number of bits	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.5 MSps; $T_A$ = entire range	9.6	10.2	-	bit
		$\begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } \text{f}_{\text{ADC}} = 35 \text{ MHz}; \text{f}_{\text{s}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } \text{f}_{\text{ADC}} = 16 \text{ MHz}; \text{f}_{\text{s}} \leq 1.1 \text{ MSps}; \\ \end{array}$	9.5	10.2	1	
	Signal-to-noise SINAD and distortion ratio	$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_s \le 2.5 \text{ MSps};$ $T_A = 25 \text{ °C}$	62.5	63	-	
SINAD		2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.5 MSps; $T_A$ = entire range	59.5	63	ı	dB
		$ \begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < \ 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } \text{f}_{\text{ADC}} = 35 \text{ MHz}; \text{f}_{\text{s}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } \text{f}_{\text{ADC}} = 16 \text{ MHz}; \text{f}_{\text{s}} \leq 1.1 \text{ MSps}; \\ \end{array} $	59	63	-	
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_s \le 2.5 \text{ MSps};$ $T_A = 25 \text{ °C}$	63	64	1	
SNR	Signal-to-noise ratio	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.5 MSps; $T_A$ = entire range	60	64	-	dB
		$\begin{array}{l} 1.65 \text{ V} < \text{V}_{\text{DDA}} = \text{V}_{\text{REF+}} < 3.6 \text{ V}; \\ \text{T}_{\text{A}} = \text{entire range} \\ \text{Range 1: } \text{f}_{\text{ADC}} = 35 \text{ MHz}; \text{f}_{\text{s}} \leq 2.2 \text{ MSps}; \\ \text{Range 2: } \text{f}_{\text{ADC}} = 16 \text{ MHz}; \text{f}_{\text{s}} \leq 1.1 \text{ MSps}; \\ \end{array}$	60	64	1	
		$V_{DDA} = V_{REF+} = 3 \text{ V};$ $f_{ADC} = 35 \text{ MHz}; f_s \le 2.5 \text{ MSps};$ $T_A = 25 ^{\circ}\text{C}$	-	-74	-73	
TUD I	Total harmonic distortion	2 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.5 MSps; $T_A$ = entire range	_	-74	-70	dB
		1.65 V < $V_{DDA}$ = $V_{REF+}$ < 3.6 V; $T_A$ = entire range Range 1: $f_{ADC}$ = 35 MHz; $f_s$ ≤ 2.2 MSps; Range 2: $f_{ADC}$ = 16 MHz; $f_s$ ≤ 1.1 MSps;	-	-74	-70	

<sup>1.</sup> Based on characterization results, not tested in production.

<sup>2.</sup> ADC DC accuracy values are measured after internal calibration.

<sup>3.</sup> Injecting negative current on any analog input pin significantly reduces the accuracy of A-to-D conversion of signal on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins susceptible to receive negative current.

<sup>4.</sup> I/O analog switch voltage booster enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA} < 2.4 \text{ V}$  and disabled when  $V_{DDA} \ge 2.4 \text{ V}$ .

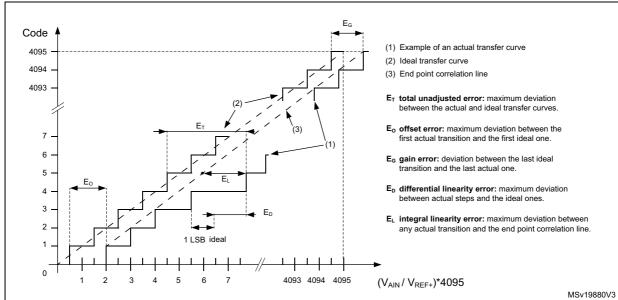


Figure 25. ADC accuracy characteristics





- 1. Refer to Table 57: ADC characteristics for the values of R<sub>AIN</sub> and C<sub>ADC</sub>.
- 2. C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to *Table 51: I/O static characteristics* for the value of the pad capacitance). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.
- 3. Refer to Table 51: I/O static characteristics for the values of Ilkg.

#### General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 13: Power supply scheme*. The 100 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.

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## 5.3.18 Voltage reference buffer characteristics

Table 60. VREFBUF characteristics<sup>(1)</sup>

Symbol	Parameter	Conditio	ns	Min	Тур	Max	Unit
		No was all was alla	V <sub>RS</sub> = 0	2.4	-	3.6	
	Analog supply	Normal mode	V <sub>RS</sub> = 1	2.8	-	3.6	
$V_{DDA}$	voltage	Degraded mode <sup>(2)</sup>	V <sub>RS</sub> = 0	1.65	-	2.4	
		Degraded mode(=)	V <sub>RS</sub> = 1	1.65	-	2.8	v
			V <sub>RS</sub> = 0	2.038	2.042	2.046	V
V <sub>REFBUF</sub> _	Voltage	iload = 100 μA	V <sub>RS</sub> = 1	2.497	2.5	2.503	
OUT	reference output	T = 30 °C	V <sub>RS</sub> = 0	V <sub>DDA</sub> -150 mV	-	$V_{DDA}$	
		V <sub>RS</sub> = 1	V <sub>DDA</sub> -150 mV	ı	$V_{DDA}$		
TRIM	Trim step resolution	-	-	-	±0.05	±0.1	%
CL	Load capacitor	-	-	0.5	1	1.5	μF
esr	Equivalent Serial Resistor of C <sub>load</sub>	-	-	-	-	2	Ω
I <sub>load</sub>	Static load current	-	-	-	-	4	mA
	ine_reg Line regulation	201/21/ / 261/	I <sub>load</sub> = 500 μA	-	200	1000	nnm/\/
Iline_reg		$2.8 \text{ V} \le \text{V}_{\text{DDA}} \le 3.6 \text{ V}$	I <sub>load</sub> = 4 mA	-	100	500	ppm/V
I <sub>load_reg</sub>	Load regulation	500 μA ≤ I <sub>load</sub> ≤4 mA	Normal mode	-	50	500	ppm/mA
T <sub>Coeff_vrefbuf</sub>	Temperature coefficient of VREFBUF <sup>(3)</sup>	-40 °C < T <sub>J</sub> < +125 °C	;	-	-	50	ppm/ °C
DODD	Power supply	DC		40	60	-	40
PSRR	rejection	100 kHz		25	40	-	dB
		$CL = 0.5 \mu F^{(4)}$		-	300	350	
t <sub>START</sub>	Start-up time	CL = 1.1 µF <sup>(4)</sup>		-	500	650	μs
		CL = 1.5 µF <sup>(4)</sup>		-	650	800	
I <sub>INRUSH</sub>	Control of maximum DC current drive on VREFBUF_OUT during start-up phase <sup>(5)</sup>	-		-	8	-	mA
	VREFBUF	I <sub>load</sub> = 0 μA		-	16	25	
I <sub>DDA(VREFB</sub>	consumption	I <sub>load</sub> = 500 μA		-	18	30	μΑ
UF)	from V <sub>DDA</sub>	I <sub>load</sub> = 4 mA		-	35	50	

<sup>1.</sup> Guaranteed by design.



- In degraded mode, the voltage reference buffer can not maintain accurately the output voltage which will follow (V<sub>DDA</sub> drop voltage).
- 3. The temperature coefficient at VREF+ output is the sum of  $T_{Coeff\_vrefint}$  and  $T_{Coeff\_vrefibuf}$ .
- 4. The capacitive load must include a 100 nF capacitor in order to cut-off the high frequency noise.
- 5. To correctly control the VREFBUF inrush current during start-up phase and scaling change, the  $V_{DDA}$  voltage should be in the range [2.4 V to 3.6 V] and [2.8 V to 3.6 V] respectively for  $V_{RS} = 0$  and  $V_{RS} = 1$ .

#### 5.3.19 Temperature sensor characteristics

Table 61. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>TS</sub> linearity with temperature	-	±1	±2	°C
Avg_Slope <sup>(2)</sup>	Average slope	2.3	2.5	2.7	mV/°C
V <sub>30</sub>	Voltage at 30°C (±5 °C) <sup>(3)</sup>	0.742	0.76	0.785	V
t <sub>START(TS_BUF)</sub> (1)	Sensor Buffer Start-up time in continuous mode <sup>(4)</sup>	-	8	15	μs
t <sub>START</sub> (1)	Start-up time when entering in continuous mode <sup>(4)</sup>	-	70	120	μs
t <sub>S_temp</sub> <sup>(1)</sup>	ADC sampling time when reading the temperature	5	-	-	μs
I <sub>DD(TS)</sub> <sup>(1)</sup>	Temperature sensor consumption from $V_{DD}$ , when selected by ADC	-	4.7	7	μΑ

- 1. Guaranteed by design.
- 2. Based on characterization results, not tested in production.
- 3. Measured at  $V_{DDA}$  = 3.0 V ±10 mV. The  $V_{30}$  ADC conversion result is stored in the TS\_CAL1 byte.
- 4. Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

### 5.3.20 V<sub>BAT</sub> monitoring characteristics

Table 62. V<sub>BAT</sub> monitoring characteristics

Symbol	Parameter	Min	Тур	Max	Unit
R	Resistor bridge for V <sub>BAT</sub>	-	39	-	kΩ
Q	Ratio on V <sub>BAT</sub> measurement	-	3	-	-
Er <sup>(1)</sup>	Error on Q	-10	-	10	%
t <sub>S_vbat</sub> <sup>(1)</sup>	ADC sampling time when reading the VBAT	12	-	-	μs

<sup>1.</sup> Guaranteed by design.

Table 63. V<sub>BAT</sub> charging characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R <sub>BC</sub> c	Battery	VBRS = 0	-	5	-	
	charging resistor	VBRS = 1	-	1.5	-	kΩ



#### 5.3.21 Timer characteristics

The parameters given in the following tables are guaranteed by design. Refer to Section 5.3.14: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 64. TIMx<sup>(1)</sup> characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
+	Timer resolution time	-	1	-	t <sub>TIMxCLK</sub>
<sup>t</sup> res(TIM)	Timer resolution time	f <sub>TIMxCLK</sub> = 64 MHz	15.625	-	ns
f	Timer external clock frequency	-	0	f <sub>TIMxCLK</sub> /2	MHz
f <sub>EXT</sub>	on CH1 to CH4	f <sub>TIMxCLK</sub> = 64 MHz	0	40	IVII IZ
Res <sub>TIM</sub>	Timer resolution	TIMx (except TIM2)	-	16	bit
IXESTIM	Timer resolution	TIM2	-	32	
t	16-bit counter clock period	-	1	65536	t <sub>TIMxCLK</sub>
<sup>t</sup> COUNTER	To-bit counter clock period	f <sub>TIMxCLK</sub> = 64 MHz	0.015625	1024	μs
t	Maximum possible count with	-	-	65536 × 65536	t <sub>TIMxCLK</sub>
<sup>t</sup> MAX_COUNT	32-bit counter	f <sub>TIMxCLK</sub> = 64 MHz	-	67.10	s

<sup>1.</sup> TIMx is used as a general term in which x stands for 1, 2, 3, 4, 5, 6, 7, 8, 15, 16 or 17.

Table 65. IWDG min/max timeout period at 32 kHz LSI clock<sup>(1)</sup>

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFF	Unit
/4	0	0.125	512	
/8	1	0.250	1024	
/16	2	0.500	2048	
/32	3	1.0	4096	ms
/64	4	2.0	8192	
/128	5	4.0	16384	
/256	6 or 7	8.0	32768	

The exact timings further depend on the phase of the APB interface clock versus the LSI clock, which causes an
uncertainty of one RC period.

#### 5.3.22 Characteristics of communication interfaces

### I<sup>2</sup>C-bus interface characteristics

The  $I^2C$ -bus interface meets timing requirements of the  $I^2C$ -bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The timings are guaranteed by design as long as the I2C peripheral is properly configured (refer to the reference manual RM0444) and when the I2CCLK frequency is greater than the minimum shown in the following table.

**Symbol Parameter** Condition Typ Unit Standard-mode 2 Analog filter enabled 9 DNF = 0Fast-mode Minimum I2CCLK Analog filter disabled 9 frequency for correct DNF = 1MHz f<sub>I2CCLK(min)</sub> operation of I2C peripheral Analog filter enabled 18 DNF = 0Fast-mode Plus Analog filter disabled 16 DNF = 1

Table 66. Minimum I2CCLK frequency

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and  $V_{\rm DDIO1}$  is disabled, but is still present. Only FT\_f I/O pins support Fm+ low-level output current maximum requirement. Refer to Section 5.3.14: I/O port characteristics for the I2C I/Os characteristics.

All I2C SDA and SCL I/Os embed an analog filter. Refer to the following table for its characteristics:

	Table 07. 120 dilatog litter characteristics							
Symbol	Parameter	Min	Max	Unit				
t <sub>AF</sub>	Limiting duration of spikes suppressed by the filter <sup>(2)</sup>	50	260	ns				

Table 67, I2C analog filter characteristics<sup>(1)</sup>

#### SPI/I<sup>2</sup>S characteristics

Unless otherwise specified, the parameters given in *Table 68* for SPI are derived from tests performed under the ambient temperature, f<sub>PCLKx</sub> frequency and supply voltage conditions summarized in *Table 21: General operating conditions*. The additional general conditions are:

- OSPEEDRy[1:0] set to 11 (output speed)
- capacitive load C = 30 pF
- measurement points at CMOS levels: 0.5 x V<sub>DD</sub>

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).



<sup>1.</sup> Based on characterization results, not tested in production.

<sup>2.</sup> Spikes shorter than the limiting duration are suppressed.

Table 68. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode 1.65 < V <sub>DD</sub> < 3.6 V Range 1			32	
		Master transmitter 1.65 < V <sub>DD</sub> < 3.6 V Range 1			32	
f <sub>SCK</sub>	SPI clock frequency	Slave receiver 1.65 < V <sub>DD</sub> < 3.6 V Range 1		_	32	MHz
1/t <sub>c(SCK)</sub>		Slave transmitter/full duplex 2.7 < V <sub>DD</sub> < 3.6 V Range 1			32	
		Slave transmitter/full duplex 1.65 < $V_{DD}$ < 3.6 V Range 1			25	
		1.65 < V <sub>DD</sub> < 3.6 V Range 2			8	
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI prescaler = 2	4 x T <sub>PCLK</sub>	-	-	ns
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI prescaler = 2	2 x T <sub>PCLK</sub>	-	-	ns
t <sub>w(SCKH)</sub>	SCK high time	Master mode	T <sub>PCLK</sub> - 1.5	T <sub>PCLK</sub>	T <sub>PCLK</sub> + 1	ns
t <sub>w(SCKL)</sub>	SCK low time	Master mode	T <sub>PCLK</sub> - 1.5	T <sub>PCLK</sub>	T <sub>PCLK</sub> + 1	ns
t <sub>su(MI)</sub>	Master data input setup time	-	1	-	-	ns
t <sub>su(SI)</sub>	Slave data input setup time	-	3	-	-	ns
t <sub>h(MI)</sub>	Master data input hold time	-	5	-	-	ns
t <sub>h(SI)</sub>	Slave data input hold time	-	2	-	-	ns
t <sub>a(SO)</sub>	Data output access time	Slave mode	9	-	34	ns
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	9	-	16	ns
		2.7 < V <sub>DD</sub> < 3.6 V Range 1	-	9	12	
t <sub>v(SO)</sub>	Slave data output valid time	1.65 < V <sub>DD</sub> < 3.6 V Range 1	-	9	19.5	ns
		1.65 < V <sub>DD</sub> < 3.6 V Voltage Range 2	-	11	24	
t <sub>v(MO)</sub>	Master data output valid time	-	-	3	5	ns

<b>Table 68. S</b>	SPI characteristics <sup>(1)</sup> (	(continued)
--------------------	--------------------------------------	-------------

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>h(SO)</sub>	Slave data output hold time	-	5	-	-	ns
t <sub>h(MO)</sub>	Master data output hold time	-	1	-	-	ns

<sup>1.</sup> Based on characterization results, not tested in production.

Figure 27. SPI timing diagram - slave mode and CPHA = 0

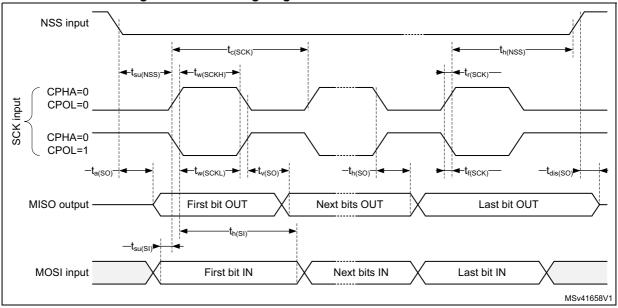
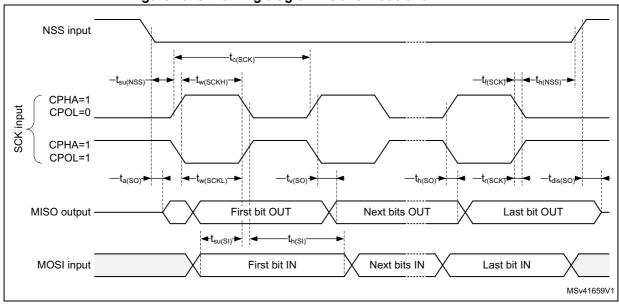


Figure 28. SPI timing diagram - slave mode and CPHA = 1



1. Measurement points are done at CMOS levels:  $0.3 V_{DD}$  and  $0.7 V_{DD}$ .

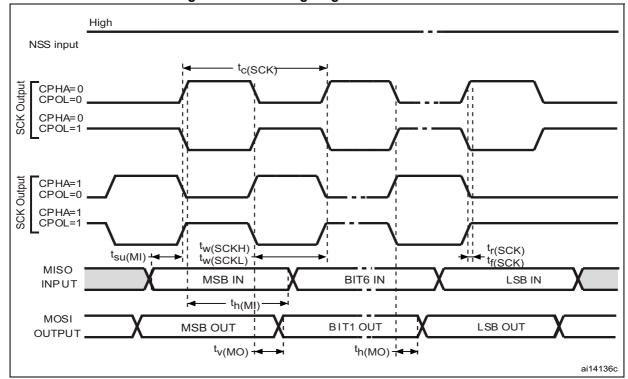


Figure 29. SPI timing diagram - master mode

1. Measurement points are set at CMOS levels: 0.3  $\rm V_{DD}$  and 0.7  $\rm V_{DD}$ .

Table 69. I<sup>2</sup>S characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>MCK</sub>	I2S main clock output	f <sub>MCK</sub> = 256 x Fs; (Fs = audio sampling frequency) Fs <sub>min</sub> = 8 kHz; Fs <sub>max</sub> = 192 kHz;	2.048	49.152	MHz
f	12C alook fraguanay	Master data	-	64xFs	MHz
<sup>I</sup> CK	f <sub>CK</sub> I2S clock frequency	Slave data	-	64xFs	IVITIZ
D <sub>CK</sub>	I2S clock frequency duty cycle	Slave receiver	30	70	%

Symbol	Parameter	Conditions	Min	Max	Unit
t <sub>v(WS)</sub>	WS valid time	Master mode	-	6	
t <sub>h(WS)</sub>	WS hold time	Master mode	3	-	
t <sub>su(WS)</sub>	WS setup time	Slave mode	3	-	
t <sub>h(WS)</sub>	WS hold time	Slave mode	2	-	
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	4	-	
t <sub>su(SD_SR)</sub>	Data input setup time	Slave receiver	5	-	
t <sub>h(SD_MR)</sub>	Data input hold time	Master receiver	4.5	-	
t <sub>h(SD_SR)</sub>	Data input noid time	Slave receiver	2	-	ns
t	Data output valid time -	after enable edge; 2.7 < V <sub>DD</sub> < 3.6V		10	
t <sub>v(SD_ST)</sub>	slave transmitter	after enable edge; 1.65 < V <sub>DD</sub> < 3.6V	_	15	
t <sub>v(SD_MT)</sub>	Data output valid time - master transmitter	after enable edge	-	5.5	
t <sub>h(SD_ST)</sub>	Data output hold time - slave transmitter	after enable edge	7	-	
t <sub>h(SD_MT)</sub>	Data output hold time - master transmitter	after enable edge	1	-	

Table 69. I<sup>2</sup>S characteristics<sup>(1)</sup> (continued)

<sup>1.</sup> Based on characterization results, not tested in production.

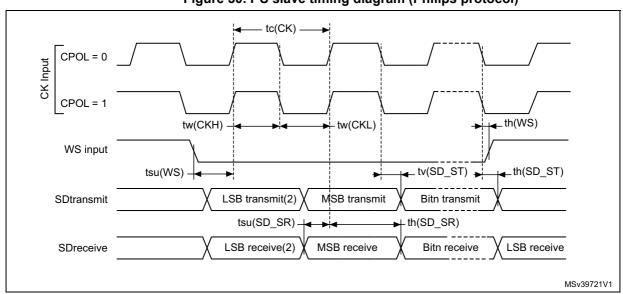


Figure 30. I<sup>2</sup>S slave timing diagram (Philips protocol)

- 1. Measurement points are done at CMOS levels: 0.3  $\rm V_{DDIO1}$  and 0.7  $\rm V_{DDIO1}.$
- 2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

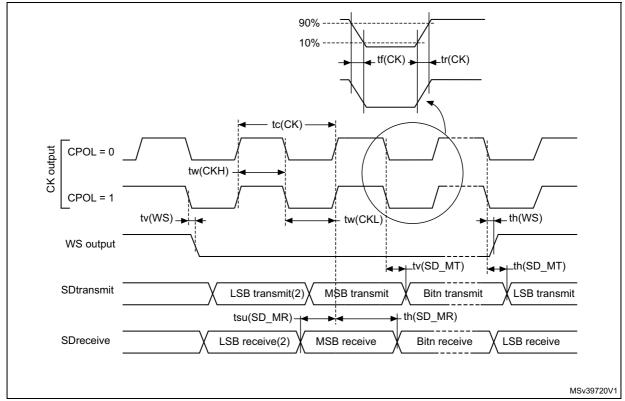


Figure 31. I<sup>2</sup>S master timing diagram (Philips protocol)

- 1. Based on characterization results, not tested in production.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

#### **USART** characteristics

Unless otherwise specified, the parameters given in *Table 70* for USART are derived from tests performed under the ambient temperature, f<sub>PCLKx</sub> frequency and supply voltage conditions summarized in *Table 21: General operating conditions*. The additional general conditions are:

- OSPEEDRy[1:0] set to 10 (output speed)
- capacitive load C = 30 pF
- measurement points at CMOS levels: 0.5 x V<sub>DD</sub>

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, CK, TX, and RX for USART).

Table 70. USART characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	USART clock frequency	Master mode	-	-	8	MHz
ICK	OSANT Clock frequency	Slave mode	-	-	21	IVII IZ

**Table 70. USART characteristics** 

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>su(NSS)</sub>	NSS setup time	Slave mode	t <sub>ker</sub> + 2	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode	2	-	-	
t <sub>w(CKH)</sub>	CK high time	Master mode	1 / f <sub>CK</sub> / 2 - 1	1 / f <sub>CK</sub> / 2	1 / f <sub>CK</sub> / 2	
t <sub>w(CKL)</sub>	CK low time	Master mode	- 1	171CK72	+ 1	
t	Data input setup time	Master mode	t <sub>ker</sub> + 2	-	-	
t <sub>su(RX)</sub>	Data input setup time	Slave mode	3	-	-	ns
+	Data input hold time	Master mode	2	-	-	115
t <sub>h(RX)</sub>	Data input noid time	Slave mode	1	-	-	
+	Data output valid time	Master mode	-	1	2	
t <sub>v(TX)</sub>	Data output valid time	Slave mode	-	10	19	
+	Data output hold time	Master mode	0	-	-	
t <sub>h(TX)</sub>	Data output hold time	Slave mode	7	-	-	

# 6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: <a href="https://www.st.com">www.st.com</a>. ECOPACK is an ST trademark.

## 6.1 SO8N package information

SO8N is an 8-lead 4.9 x 6 mm plastic small-outline package with 150 mils body width.

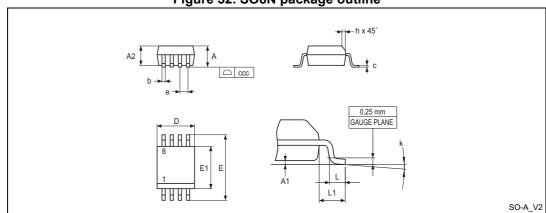


Figure 32. SO8N package outline

1. Drawing is not to scale.

Table 71. SO8N package mechanical data

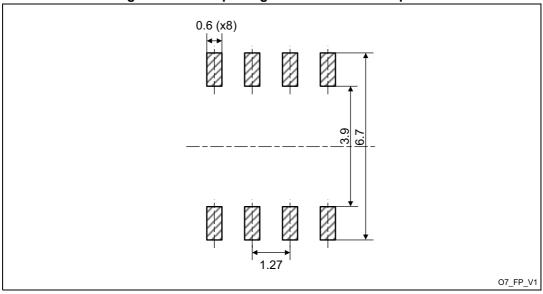
Symbol		millimeters			inches <sup>(1)</sup>	inches <sup>(1)</sup>		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.		
Α	-	-	1.750	-	-	0.0689		
A1	0.100	-	0.250	0.0039	-	0.0098		
A2	1.250	-	-	0.0492	-	-		
b	0.280	-	0.480	0.0110	-	0.0189		
С	0.170	-	0.230	0.0067	-	0.0091		
D	4.800	4.900	5.000	0.1890	0.1929	0.1969		
Е	5.800	6.000	6.200	0.2283	0.2362	0.2441		
E1	3.800	3.900	4.000	0.1496	0.1535	0.1575		
е	-	1.270	-	-	0.0500	-		
h	0.250	-	0.500	0.0098	-	0.0197		
k	0°	-	8°	0°	-	8°		
L	0.400	-	1.270	0.0157	-	0.0500		

Table 71. SO8N package mechanical data (continued)

Symbol millimeters				inches <sup>(1)</sup>		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
L1	-	1.040	-	-	0.0409	-
ccc	-	-	0.100	-	-	0.0039

<sup>1.</sup> Values in inches are converted from mm and rounded to four decimal digits.

Figure 33. SO8N package recommended footprint



1. Dimensions are expressed in millimeters.

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks that identify the parts throughout supply chain operations, are not indicated below.

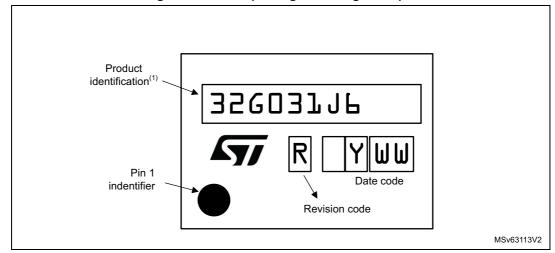


Figure 34. SO8N package marking example

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

## 6.2 WLCSP18 package information

This WLCSP is a 18 ball, 1.86 x 2.14 mm, 0.4 mm pitch, wafer level chip scale package.

// bbb Z - A1 A1 BALL LOCATION 000 В DETAIL A e2 С D  $\circ$ Ε @ O O 🛭 7 6 5 4 3 2 1 D (4x) aaa TOP VIEW **BOTTOM VIEW** SIDE VIEW А3 BUMP FRONT VIEW △ ccc Z Ż Ø b (36x) SEATING PLANE DETAIL A ROTATED 90° B06E\_WLCSP18\_DIE466\_ME\_V1

Figure 35. WLCSP18 - Outline

- 1. Drawing is not to scale.
- 2. Dimension is measured at the maximum bump diameter parallel to primary datum Z.
- 3. Primary datum Z and seating plane are defined by the spherical crowns of the bump.
- 4. Bump position designation per JESD 95-1, SPP-010.

Table 72. WLCSP18 - Mechanical data

0		millimeters			inches <sup>(1)</sup>	Typ Max  - 0.023  0.007 - 0.015  0.001 - 0.011  0.073 0.074  0.084 0.085  0.016 - 0.047 - 0.055 - 0.013  0.015 - 0.015  0.015 - 0.015  0.015 - 0.015		
Symbol	Min	Тур	Max	Min	Тур	Max		
A <sup>(2)</sup>	-	-	0.59	-	-	0.023		
A1	-	0.18	-	-	0.007	-		
A2	-	0.38	-	-	0.015	-		
A3 <sup>(3)</sup>	-	0.025	-	-	0.001	-		
b <sup>(4)</sup>	0.22	0.25	0.28	0.009	0.010	0.011		
D	1.84	1.86	1.88	0.072	0.073	0.074		
E	2.12	2.14	2.16	0.083	0.084	0.085		
е	-	0.40	-	-	0.016	-		
e1	-	1.20	-	-	0.047	-		
e2	-	1.39	-	-	0.055	-		
F <sup>(5)</sup>	-	0.330	-	-	0.013	-		
G <sup>(5)</sup>	-	0.390	-	-	0.015	-		
aaa	-	-	0.10	-	-	0.004		
bbb	-	-	0.10	-	-	0.004		
ccc	-	-	0.10	-	-	0.004		
ddd	-	-	0.05	-	-	0.002		
eee	-	-	0.05	-	-	0.002		

- 1. Values in inches are converted from mm and rounded to 3 decimal digits.
- 2. The maximum total package height is calculated by the RSS method (Root Sum Square) using nominal and tolerances values of A1 and A2.
- Back side coating. Nominal dimension is rounded to the 3rd decimal place resulting from process capability.
- 4. Dimension is measured at the maximum bump diameter parallel to primary datum Z.
- 5. Calculated dimensions are rounded to the 3rd decimal place

Figure 36. WLCSP18 - Recommended footprint

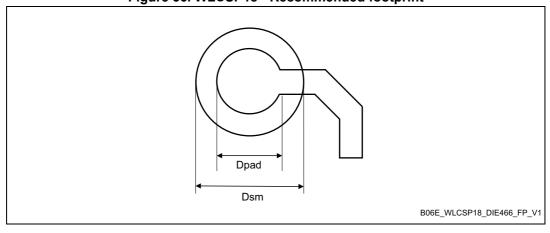


Table 73. WLCSP18 - Recommended PCB design rules

Dimension	Recommended values
Pitch	0.4 mm
Dpad	0,225 mm
Dsm	0.290 mm typ. (depends on soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm

#### **Device marking**

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks that identify the parts throughout supply chain operations, are not indicated below.

Product identification<sup>(1)</sup>

Product identification<sup>(1)</sup>

Revision code

MSv63115V1

Figure 37. WLCSP18 package marking example

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

## 6.3 TSSOP20 package information

TSSOP20 is a 20-lead, 6.5 x 4.4 mm thin small-outline package with 0.65 mm pitch.

Figure 38. TSSOP20 package outline

1. Drawing is not to scale.

Table 74. TSSOP20 package mechanical data

Coursels at		millimeters			inches <sup>(1)</sup>	
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
А	-	-	1.200	-	-	0.0472
A1	0.050	-	0.150	0.0020	-	0.0059
A2	0.800	1.000	1.050	0.0315	0.0394	0.0413
b	0.190	-	0.300	0.0075	-	0.0118
С	0.090	-	0.200	0.0035	-	0.0079
D <sup>(2)</sup>	6.400	6.500	6.600	0.2520	0.2559	0.2598
Е	6.200	6.400	6.600	0.2441	0.2520	0.2598
E1 <sup>(3)</sup>	4.300	4.400	4.500	0.1693	0.1732	0.1772
е	-	0.650	-	-	0.0256	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	-	8°	0°	-	8°
aaa	-	-	0.100	-	-	0.0039

<sup>1.</sup> Values in inches are converted from mm and rounded to four decimal digits.

<sup>2.</sup> Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs shall not exceed 0.15mm per side.

Dimension "E1" does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.

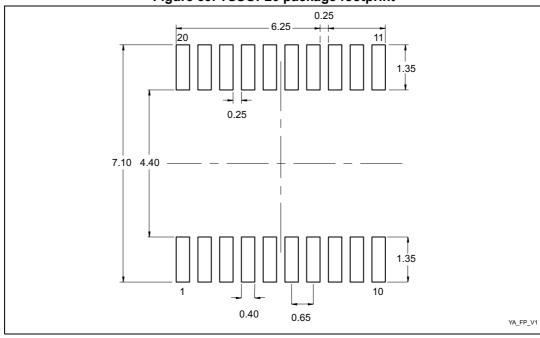


Figure 39. TSSOP20 package footprint

1. Dimensions are expressed in millimeters.

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks that identify the parts throughout supply chain operations, are not indicated below.

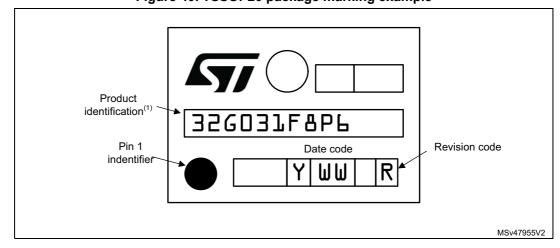


Figure 40. TSSOP20 package marking example

Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified
and therefore not approved for use in production. ST is not responsible for any consequences resulting
from such use. In no event will ST be liable for the customer using any of these engineering samples in
production. ST's Quality department must be contacted prior to any decision to use these engineering

101/120

samples to run a qualification activity.

## 6.4 UFQFPN28 package information

UFQFPN is a 28-lead, 4x4 mm, 0.5 mm pitch, ultra-thin fine-pitch quad flat package.

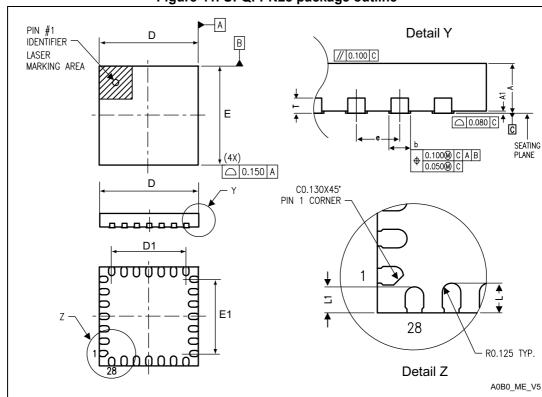


Figure 41. UFQFPN28 package outline

1. Drawing is not to scale.

Table 75. UFQFPN28 package mechanical data<sup>(1)</sup>

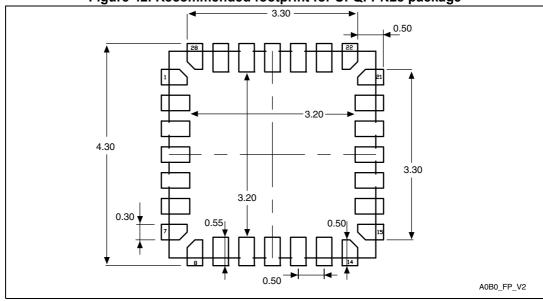
Symbol		millimeters			inches		
Symbol	Min	Тур	Max	Min	Тур	Max	
Α	0.500	0.550	0.600	0.0197	0.0217	0.0236	
A1	-	0.000	0.050	-	0.0000	0.0020	
D	3.900	4.000	4.100	0.1535	0.1575	0.1614	
D1	2.900	3.000	3.100	0.1142	0.1181	0.1220	
Е	3.900	4.000	4.100	0.1535	0.1575	0.1614	
E1	2.900	3.000	3.100	0.1142	0.1181	0.1220	
L	0.300	0.400	0.500	0.0118	0.0157	0.0197	
L1	0.250	0.350	0.450	0.0098	0.0138	0.0177	
Т	-	0.152	-	-	0.0060	-	

Table 75. UFQFPN28 package mechanical data<sup>(1)</sup> (continued)

Symbol	millimeters			inches		
Symbol	Min	Тур	Max	Min	Тур	Max
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 42. Recommended footprint for UFQFPN28 package



1. Dimensions are expressed in millimeters.

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Product identification (1)

GOBLGB

Revision code

Pin 1 identifier

MSv47951V2

Figure 43. UFQFPN28 package marking example

Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified
and therefore not approved for use in production. ST is not responsible for any consequences resulting
from such use. In no event will ST be liable for the customer using any of these engineering samples in
production. ST's Quality department must be contacted prior to any decision to use these engineering
samples to run a qualification activity.

## 6.5 UFQFPN32 package information

UFQFPN32 is a 32-pin, 5x5 mm, 0.5 mm pitch ultra-thin fine-pitch quad flat package.

Figure 44. UFQFPN32 package outline

- 1. Drawing is not to scale.
- 2. There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this backside pad to PCB ground.

inches<sup>(1)</sup> millimeters **Symbol** Min Тур Max Min Тур Max 0.0217 Α 0.500 0.550 0.600 0.0197 0.0236 0.0007 0.0020 Α1 0.000 0.020 0.050 0.000 **A3** 0.0060 0.152 0.180 0.280 0.0071 0.230 0.0091 0.0110  $D^{(2)}$ 4.900 5.000 5.100 0.1929 0.1969 0.2008 0.1417 D1 3.400 3.500 3.600 0.1339 0.1378 3.400 3.500 3.600 0.1339 0.1378 0.1417  $E^{(2)}$ 4.900 5.000 5.100 0.1929 0.1969 0.2008 E1 3.400 3.500 3.600 0.1339 0.1378 0.1417 0.1417 E2 3.400 3.500 3.600 0.1339 0.1378 е 0.500 0.0197 L 0.300 0.400 0.500 0.0118 0.0157 0.0197 ddd 0.080 0.0031

Table 76. UFQFPN32 package mechanical data

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

<sup>2.</sup> Dimensions D and E do not include mold protrusion, not to exceed 0,15mm.

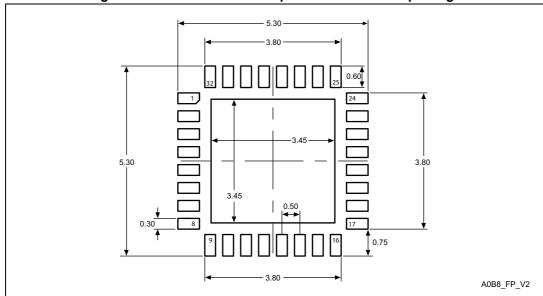


Figure 45. Recommended footprint for UFQFPN32 package

1. Dimensions are expressed in millimeters

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

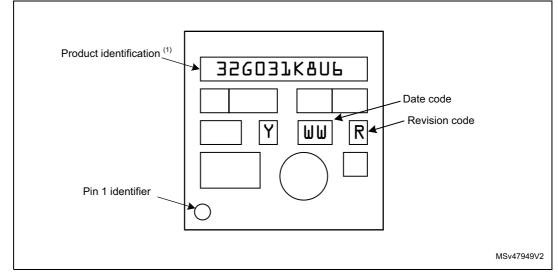


Figure 46. UFQFPN32 package marking example

Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified
and therefore not approved for use in production. ST is not responsible for any consequences resulting
from such use. In no event will ST be liable for the customer using any of these engineering samples in
production. ST's Quality department must be contacted prior to any decision to use these engineering
samples to run a qualification activity.

#### LQFP32 package information 6.6

LQFP32 is a 32-pin, 7 x 7 mm low-profile quad flat package.

SEATING PLANE С 0.25 mm GAUGE PLANE CCC С D A D1 D3 16  $\blacksquare$ ₩-E3 П --₩ ш ╨ <del>------</del>9 <u>e</u> 5V\_ME\_V2

Figure 47. LQFP32 package outline

1. Drawing is not to scale.

Table 77. LQFP32 mechanical data

Symbol		millimeters			inches <sup>(1)</sup>	inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max		
Α	-	-	1.600	-	-	0.0630		
A1	0.050	-	0.150	0.0020	-	0.0059		
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571		

Table 77. LQFP32 mechanical data (continued)

Symbol		millimeters		,	inches <sup>(1)</sup>	
Symbol	Min	Тур	Max	Min	Тур	Max
b	0.300	0.370	0.450	0.0118	0.0146	0.0177
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.600	-	-	0.2205	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.600	-	-	0.2205	-
е	-	0.800	-	-	0.0315	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.100	-	-	0.0039

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

1. Dimensions are expressed in millimeters.

## **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

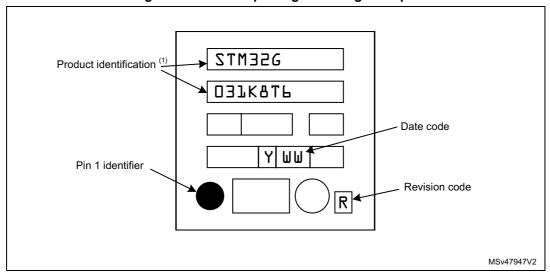


Figure 49. LQFP32 package marking example

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity. Package information STM32G031x4/x6/x8

# 6.7 UFQFPN48 package information

UFQFPN48 is a 48-lead, 7x7 mm, 0.5 mm pitch, ultra-thin fine-pitch quad flat package

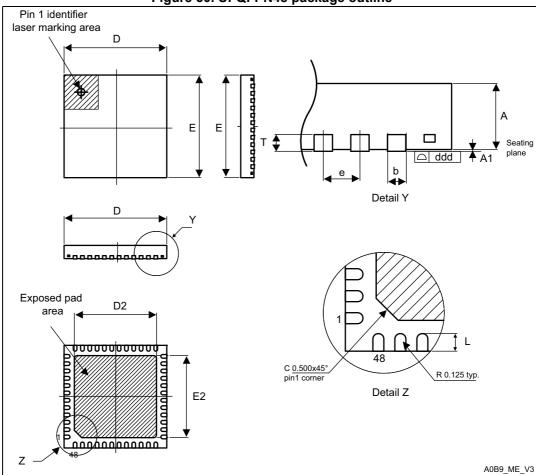


Figure 50. UFQFPN48 package outline

- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this back-side pad to PCB ground.

Table 78. UFQFPN48 package mechanical data

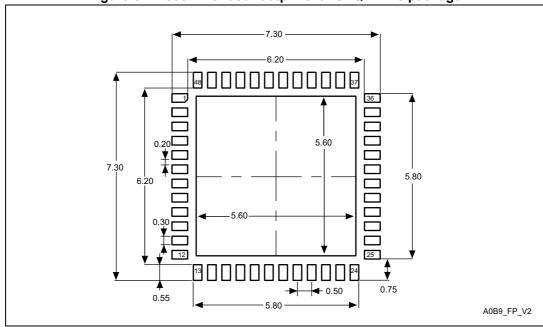
Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
А	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
Е	6.900	7.000	7.100	0.2717	0.2756	0.2795
D2	5.500	5.600	5.700	0.2165	0.2205	0.2244
E2	5.500	5.600	5.700	0.2165	0.2205	0.2244

Table 78. UFQFPN48 package mechanical data (continued)

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
Т	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 51. Recommended footprint for UFQFPN48 package



1. Dimensions are expressed in millimeters.

Package information STM32G031x4/x6/x8

#### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Product identification (1) STM32G

Date code

Pin 1 identifier

MSv47945V2

Figure 52. UFQFPN48 package marking example

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

# 6.8 LQFP48 package information

LQFP48 is a 48-pin, 7 x 7 mm low-profile quad flat package.

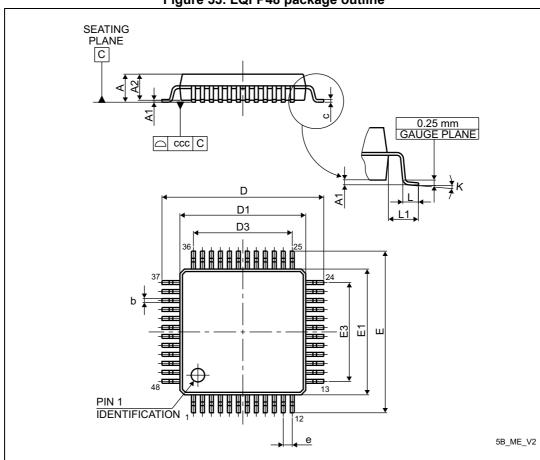


Figure 53. LQFP48 package outline

1. Drawing is not to scale.

Table 79. LQFP48 mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.500	-	-	0.2165	-

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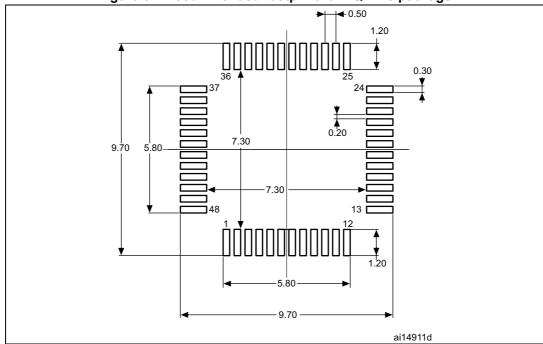
Package information STM32G031x4/x6/x8

Table 79. LQFP48 mechanical data (continued)

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	7.200 0.2677 0.2756		0.2835
E3	-	5.500	-	-	0.2165	-
е	-	0.500	-	-	0.0197 -	
L	0.450	0.600	0.750	0.0177 0.0236		0.0295
L1	-	1.000	-	- 0.0394		-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 54. Recommended footprint for LQFP48 package



1. Dimensions are expressed in millimeters.

## **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

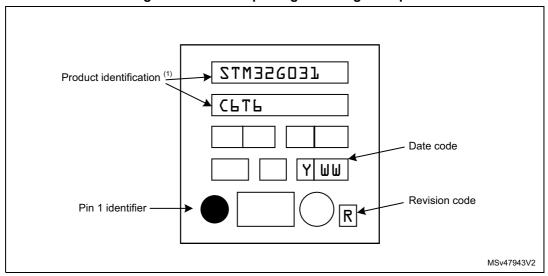


Figure 55. LQFP48 package marking example

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

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## 6.9 Thermal characteristics

The operating junction temperature  $T_J$  must never exceed the maximum given in *Table 21: General operating conditions* 

The maximum junction temperature in °C that the device can reach if respecting the operating conditions, is:

$$T_J(max) = T_A(max) + P_D(max) \times \Theta_{JA}$$

#### where:

- T<sub>A</sub>(max) is the maximum operating ambient temperature in °C,
- Θ<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- $\bullet \qquad \mathsf{P}_\mathsf{D} = \mathsf{P}_\mathsf{INT} + \mathsf{P}_\mathsf{I/O},$ 
  - P<sub>INT</sub> is power dissipation contribution from product of I<sub>DD</sub> and V<sub>DD</sub>
  - P<sub>I/O</sub> is power dissipation contribution from output ports where:

$$\mathsf{P}_\mathsf{I/O} = \Sigma \; (\mathsf{V}_\mathsf{OL} \times \mathsf{I}_\mathsf{OL}) + \Sigma \; ((\mathsf{V}_\mathsf{DDIO1} - \mathsf{V}_\mathsf{OH}) \times \mathsf{I}_\mathsf{OH}),$$

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Symbol	Parameter	Package	Junction- to-ambient	Junction- to-board	Junction- to-case	Unit	
Θ The		LQFP48 7 × 7 mm	66 35 17		17		
		UFQFPN48 7 × 7 mm	33	18	14	ı	
		LQFP32 7 × 7 mm	66	35	17		
	Thermal resistance	UFQFPN32 5 × 5 mm	41	23	20	°C/W	
		UFQFPN28 4 × 4 mm		45	30		
		TSSOP20 6.4 × 4.4 mm		55	26		
		SO8N 4.9 × 6 mm	111	70	38		

Table 80. Package thermal characteristics

#### 6.9.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (still air). Available from www.jedec.org.

## 6.9.2 Selecting the product temperature range

The temperature range is specified in the ordering information scheme shown in *Section 7: Ordering information*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and to a specific maximum junction temperature.

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As applications do not commonly use microcontrollers at their maximum power consumption, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range best suits the application.

The following example shows how to calculate the temperature range needed for a given application.

#### **Example:**

Assuming the following worst application conditions:

- ambient temperature T<sub>A</sub> = 50 °C (measured according to JESD51-2)
- $I_{DD} = 50 \text{ mA}; V_{DD} = 3.6 \text{ V}$
- 20 I/Os simultaneously used as output at low level with I<sub>OL</sub> = 8 mA (V<sub>OL</sub>= 0.4 V), and
- 8 I/Os simultaneously used as output at low level with I<sub>OL</sub> = 20 mA (V<sub>OL</sub>= 1.3 V),

the power consumption from power supply  $P_{\mathsf{INT}}$  is:

$$P_{INT} = 50 \text{ mA} \times 3.6 \text{ V} = 118 \text{ mW},$$

the power loss through I/Os P<sub>IO</sub> is

$$P_{IO} = 20 \times 8 \text{ mA} \times 0.4 \text{ V} + 8 \times 20 \text{ mA} \times 1.3 \text{ V} = 272 \text{ mW},$$

and the total power  $P_{\text{D}}$  to dissipate is:

$$P_D = 180 \text{ mW} + 272 \text{ mW} = 452 \text{ mW}$$

For a package with  $\Theta_{JA}$ = 65 °C/W, the junction temperature stabilizes at:

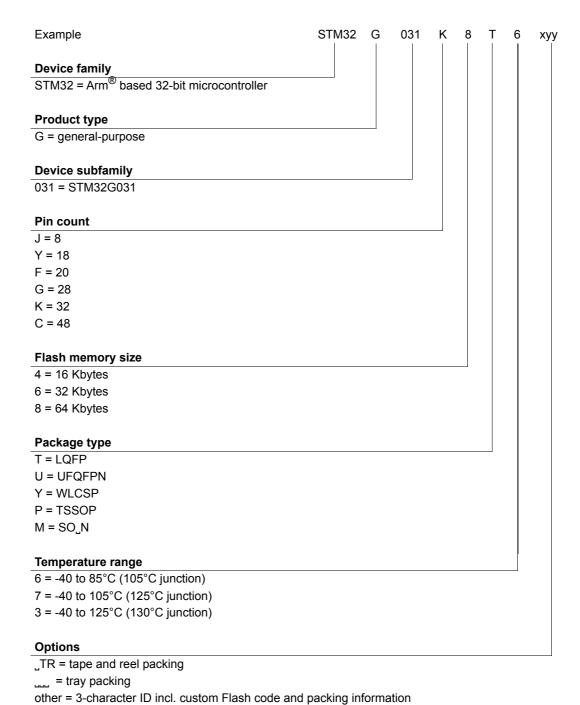
$$T_{.1} = 50^{\circ}C + (65^{\circ}C/W \times 452 \text{ mW}) = 50^{\circ}C + 29.4^{\circ}C = 79.4^{\circ}C$$

As a conclusion, product version with suffix 6 (maximum allowed  $T_J$  = 105° C) is sufficient for this application.

If the same application was used in a hot environment with maximum  $T_A$  greater than 75.5 °C, the junction temperature would exceed 105°C and the product version allowing higher maximum  $T_J$  would have to be ordered.



# 7 Ordering information



For a list of available options (memory, package, and so on) or for further information on any aspect of this device, please contact your nearest ST sales office.

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STM32G031x4/x6/x8 Revision history

# 8 Revision history

Table 81. Document revision history

Date	Revision	Changes
26-Jun-2019	1	Initial release
22-Apr-2020	2	Cover page updated; Section 2: Description updated; Figure 1: Block diagram updated; Table 12: Pin assignment and description updated; Table 18: Voltage characteristics updated; Table 19: Current characteristics: Note 2 removed; Table 21: General operating conditions and Section 7: Ordering information: Added Ta range -40 to 105°C; Table 57: ADC characteristics: major update; Corrected figures with package marking examples.
05-Oct-2021 3		<ul> <li>Table 2: STM32G031x4/x6/x8 family device features and peripheral counts</li> <li>Section 3.7.1: Power supply schemes</li> <li>Section 3.9: Clocks and startup</li> <li>Table 12: Pin assignment and description</li> <li>Table 43: PLL characteristics (footnotes)</li> <li>Table 48: ESD absolute maximum ratings</li> <li>Table 60: VREFBUF characteristics</li> <li>Table 6: Package information: all subsections rearranged</li> <li>Table 72: WLCSP18 - Mechanical data</li> <li>Table 80: Package thermal characteristics</li> <li>Example in Section 6.9.2: Selecting the product temperature range</li> </ul>

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