

BiliBili Demo

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Abstract—This is abstract.

Index Terms—BiliBili, powerful, template, latex, sublime.

I. INTRODUCTION

Wireless power transfer (WPT) through magnetic coupling has a profound impact on both consumer electronics and industrial applications [1]. Compared with traditional plug-in systems, WPT systems are free of cables, providing users with a more convenient, safe and efficient experience [2]. Currently, most of commercialized WPT systems operate in kHz band, such as at several hundreds kHz [3]. It is mainly because this frequency band provides a richer selection of power electronics components. However, the kHz operation requires large-size coupling coils and ferrite to achieve enough mutual inductance.

$$\begin{cases} X_{\Pi 1} = X_{T1} + X_{T2} + \frac{X_{T1}X_{T2}}{X_{T3}}, \\ X_{\Pi 2} = X_{T2} + X_{T3} + \frac{X_{T2}X_{T3}}{X_{T1}}, \\ X_{\Pi 3} = X_{T3} + X_{T1} + \frac{X_{T3}X_{T1}}{X_{T2}}. \end{cases} \quad (2)$$

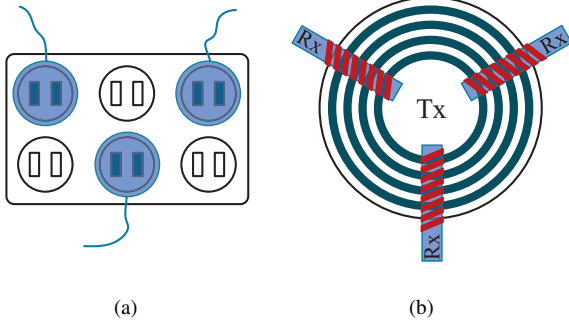


Fig. 1. Caption. (a) Subcaption1. (b) Subcaption2.

Fig.

rm reffig:socketNew shows...

II. A NOVEL METHOD TO DESIGN IMPEDANCE MATCHING NETWORKS FOR MHZ WPT SYSTEMS

Thus the transformed impedance, i.e., the input impedance of the IMN, can be calculated as:

$$\begin{aligned} Z_{\text{net}} &= R_{\text{net}} + jX_{\text{net}} = Z_{T1} + (Z_{\text{load}} + Z_{T2}) // Z_{T3} \\ &= jX_{T1} + \frac{\int_0^\infty jX_{T3} (R_{\text{load}} + jX_{\text{load}} + jX_{T2})}{R_{\text{load}}} \end{aligned} \quad (3)$$

III. PARAMETER DESIGN

A. System Configuration

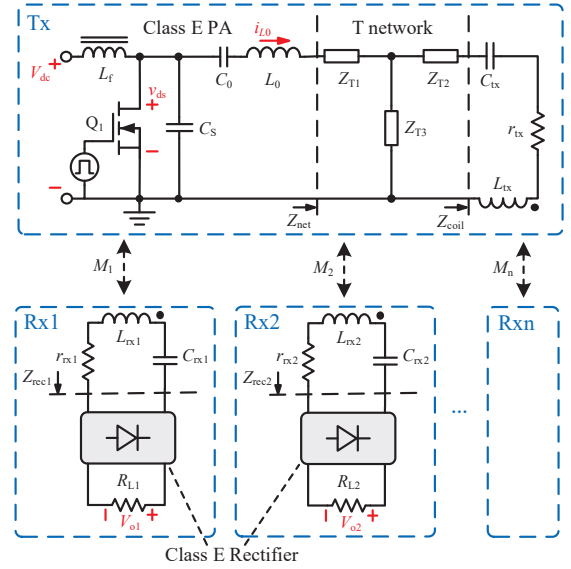


Fig. 2. System configuration of proposed multi-receiver MHz WPT system.

Fig.

rm reffig:system illustrates configuration of the proposed multi-receiver MHz WPT system, which is composed of a PA, an IMN of T-network, a transmitting (Tx) coil and several receiving (Rx) coils connected with corresponding rectifiers. In this system, Class E typology is applied in both the PA and the rectifier, due to its zero voltage switching (ZVS) and zero voltage derivative switching (ZVDS) characteristics. In the figure, $M_1 \sim M_n$ are the mutual inductance between the Tx coil and different Rx coils, with the cross coupling between

$$\begin{aligned} a &= \sqrt{(R_{\text{loadA}} + R_{\text{loadB}})^2 + (X_{\text{loadA}} - X_{\text{loadB}})^2 + 4R_{\text{loadA}}R_{\text{loadB}}\tan^2\theta_{\text{ref}}} \cdot \sqrt{(R_{\text{loadA}} - R_{\text{loadB}})^2 + (X_{\text{loadA}} - X_{\text{loadB}})^2} \\ b &= R_{\text{loadB}}^2 + R_{\text{loadA}}^2 (1 + 2\tan^2\theta_{\text{ref}}) + (X_{\text{loadA}} - X_{\text{loadB}})^2 + 2R_{\text{loadA}} [R_{\text{loadB}} (1 + \tan^2\theta_{\text{ref}}) + \tan\theta_{\text{ref}} (X_{\text{loadB}} - X_{\text{loadA}})] \end{aligned} \quad (1)$$

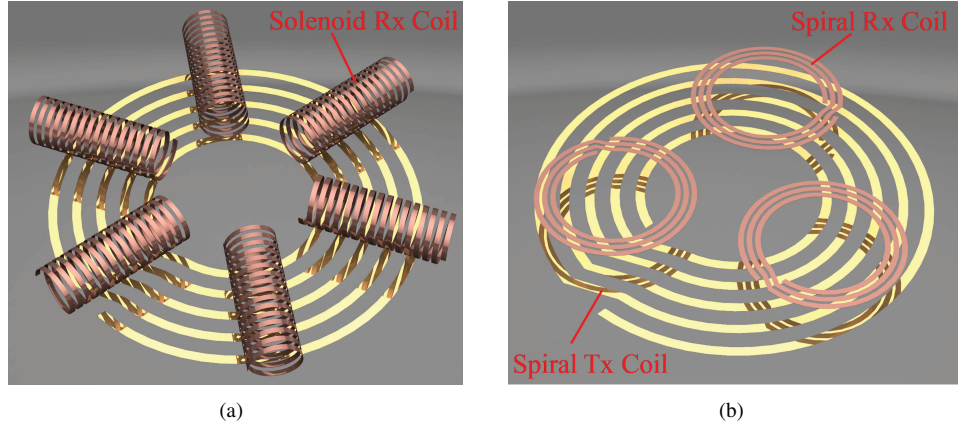


Fig. 3. Coil shapes. (a) Solenoid. (b) Spiral.

TABLE I
TARGET SETTING AND CALCULATED PARAMETERS OF THE IMN

Original Impedances	
$Z_{loadA} (Z_{coilA})$	$27+0j \Omega$
$Z_{loadB} (Z_{coilB})$	$9+0j \Omega$
Target Setting	
Z_{ref}	$14.7+12.3j \Omega$
θ_{ref}	-88°
Calculated T-net	
Z_{T1}	$26.7j \Omega$
Z_{T1}	$7j \Omega$
Z_{T1}	$-23.4j \Omega$

the Rx coils ignored. L_{tx} is inductance of the Tx coil and $L_{rx1} \sim L_{rxn}$ are the inductances of Rx coils. Their parasitic resistors and compensation capacitors are also shown in the figure. sys sys sys

Table

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IV. HYBRID COUPLING COILS

Based on the above factors and the preliminary simulations, the hybrid coupler has the following 2 advantages:

- Higher receiver capacity;
- Suitable for those receivers with special shapes.

[4]

V. EXPERIMENTAL VERIFICATION

VI. CONCLUSIONS

A multi-receiver MHz WPT system with hybrid coupler is proposed.

REFERENCES

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TABLE II
PARAMETERS OF THE EXPERIMENTAL SYSTEM

Parameters	Value	
f	6.78 MHz	
L_f	10 uH	
L_0	2.17 uH	$\Rightarrow Z_{T1} = 26.7j \Omega$
C_0^*	357 pF	
C_s	287 pF	
L_{tx}	6.65 uH	$\Rightarrow Z_{T2} = 7j \Omega$
C_{tx}^*	85 pF	
r_{tx}	1.1 Ω	
C_{T3}	1005 pF	$\Rightarrow Z_{T3} = -23.4j \Omega$
Z_{ref}	$9 \sim 27 \Omega$	
$M_1 \sim M_3$	0.45 uH	
L_r	4.7 uH	
$C_{r1} \sim C_{r3}$	540 pF	
C_L	10 uF	
$R_{L1} \sim R_{L3}$	40 Ω	

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