Control Systems

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svn co https://github.com/gadepall/school/trunk/control/codes

1 Signal Flow Graph

- 1.1 Mason's Gain Formula
- 1.2 Matrix Formula
- 1.3 Example

2 Bode Plot

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3 Second order System

- 3.1 Damping
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4 ROUTH HURWITZ CRITERION

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5 STATE-SPACE MODEL

- 5.1 Controllability and Observability
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6 Nyquist Plot

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- 6.2 Example

7 Compensators

- 7.1 Phase Lead
- 7.2 Lag Lead
- 7.3 Example

8 GAIN MARGIN

- 8.1 Introduction
- 8.2 Example
- 8.3 Example
- 8.4 Example
- 8.1. For a unity feedback system shown in Fig 8.1

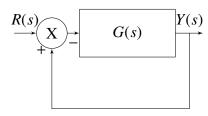


Fig. 8.1

having transfer function

$$G(s) = \frac{K}{(s+3)(s+9)(s+15)}$$
 (8.1.1)

design the value of gain(K), for a gain margin of 50 dB.

8.2. **Solution:**

Gain Margin:

$$GM = -20\log|G(j\omega_{pc})| \tag{8.2.1}$$

where, ω_{pc} is the phase cross-over frequency, at which

$$\angle G(j\omega_{pc}) = -180^{\circ} \tag{8.2.2}$$

First substitute,

$$s = j\omega \tag{8.2.3}$$

$$\implies G(j\omega) = \frac{K}{(-27\omega^2 + 405) + j(-\omega^3 + 207\omega)}$$
(8.2.4)

Now the phase will be

$$\angle G(j\omega) = -\tan^{-1}(\frac{-\omega^3 + 207\omega}{-27\omega^2 + 405}) \qquad (8.2.5)$$

Solving for $\angle G(j\omega) = -180^{\circ}$ gives

$$\omega_{pc} = 14.3875 \tag{8.2.6}$$

Magnitude:

$$|G(j\omega)| = \frac{K}{\sqrt{(\omega^2 + 9)}\sqrt{(\omega^2 + 81)}\sqrt{(\omega^2 + 225)}}$$
(8.2.7)

Substituting value of ω_{pc} in (8.2.1) gives

$$K = 16.406$$
 (8.2.8)

This can be verified from fig 8.2 The following code generates Fig. 8.2

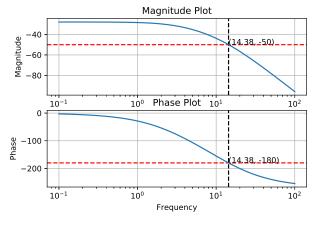


Fig. 8.2

8.3. Design the value gain (K) for a phase margin of 40°.

8.4. Solution:

Phase Margin:

$$PM = 180^{\circ} + \phi_{gc} \tag{8.4.1}$$

where ϕ_{gc} is the phase angle at the gain cross over frequency ω_{gc} . At gain cross over frequency,

$$|G(j\omega_{gc})| = 1 \tag{8.4.2}$$

$$\implies -20\log|G(j\omega_{gc})| = 0 \tag{8.4.3}$$

Given,

$$PM = 40^{\circ} = 180^{\circ} + \phi_{gc}$$
 (8.4.4)

$$\implies \phi_{gc} = -140^{\circ} = \angle G(j\omega_{gc})$$
 (8.4.5)

From (8.2.5)

$$\angle G(j\omega_{gc}) = -\tan^{-1}(\frac{-\omega_{gc}^3 + 207\omega_{gc}}{-27\omega_{gc}^2 + 405}) \quad (8.4.6)$$

$$\implies \omega_{gc} = 8.09623$$
 (8.4.7)

Substituting this value in (8.4.3), we get

$$20 \log K = 65.016 \tag{8.4.8}$$

$$\implies K = 1781.56$$
 (8.4.9)

This again can be verified from fig 8.4. The following code generates Fig. 8.4

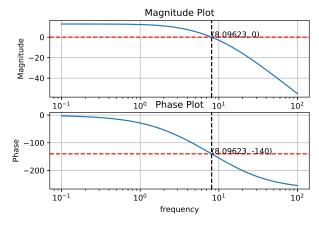


Fig. 8.4

- 8.5. Design the value of gain (K) to yield maximum peak overshoot of 20% for a step input.
- 8.6. **Solution:** Closed loop transfer function:

$$T(s) = \frac{G(s)}{1 + G(s)H(s)}$$
(8.6.1)

where H(s) = 1

$$\implies T(s) = \frac{K}{(s+3)(s+6)(s+15) + K}$$
(8.6.2)

Output will be:

$$\implies Y(s) = \frac{1}{s} \frac{K}{(s+3)(s+6)(s+15) + K}$$
(8.6.3)

Maximum peak overshoot:

$$M_p = \frac{y(t_p) - y(\infty)}{y(\infty)} \tag{8.6.4}$$

which is given as 20%. Here, t_p is the peak time. Solving this, we get

$$\implies \frac{y(t_p)}{y(\infty)} = 1.2 \tag{8.6.5}$$

Plotting y(t) for different values of K, we choose the value of K, which gives the above ratio, which is verified from fig 8.6. Thus, we get

$$t_p = 0.505$$
 (8.6.6)

$$\implies K = 928.035$$
 (8.6.7)

The following code generates fig 8.6

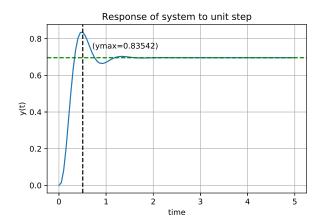


Fig. 8.6

9 Phase Margin

- 9.1 Intoduction
- 9.2 Example
- 10 OSCILLATOR
- 10.1 Introduction
- 10.2 Example
- 11 Root Locus
- 11.1 Introduction
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- 12 Polar Plot
- 12.1 Introduction