Control Systems

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	Contents			7	Compensators		2	
1	Signal	Flow Graph	2		7.1	Phase Lead	2	
1	1.1	Mason's Gain Formula	2		7.2	Lag Lead	2	
	1.2	Matrix Formula	2		7.3	Example	2	
	1.3	Example	2			•		
	1			8	Gain Margin		2	
2	Bode Plot		2		8.1	Introduction	2	
	2.1	Introduction	2					
	2.2	Example	2		8.2	Example	2	
	2.3	Phase	2		8.3	Example	2	
3	Second	andar System	2		8.4	Example	2	
3	Second order System 3.1 Damping							
	3.1	Peak Overshoot	2 2	9	Phase	Margin	4	
	3.3	Example	2		9.1	Intoduction	4	
	3.4	Settling Time	2		9.2	Example	4	
4	Routh Hurwitz Criterion		2	10	Oscille	Oscillator		
	4.1	Routh Array	2	2			4	
	4.2	Marginal Stability	2		10.1	Introduction	4	
	4.3	Stability	2		10.2	Example	4	
	4.4	Example	2					
	4.5	Example	2	11	Root 1	Locus	4	
5	State-Space Model		2		11.1	Introduction	4	
	5.1	Controllability and Observ-			11.2	Example	4	
		ability	2		11.3	Example	4	
	5.2	Second Order System	2			•		
	5.3	Example	2	12	Polar	Plot	4	
	5.4	Example	2		12.1	Introduction	4	
	5.5	Example	2		12.1	introduction	7	
	5.6	Example	2					
	5.7	Example	2			his manual is an introduction to cor		
6	Nyquist Plot		2		systems based on GATE problems.Links to sample Python codes are available in the text.			
	6.1	Introduction	2	coucs	coucs are available in the text.			
	6.2	Example	2	D	ownload	python codes using		

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svn co https://github.com/gadepall/school/trunk/control/codes

1 Signal Flow Graph

- 1.1 Mason's Gain Formula
- 1.2 Matrix Formula
- 1.3 Example

2 Bode Plot

- 2.1 Introduction
- 2.2 Example
- 2.3 Phase

3 Second order System

- 3.1 Damping
- 3.2 Peak Overshoot
- 3.3 Example
- 3.4 Settling Time

4 ROUTH HURWITZ CRITERION

- 4.1 Routh Array
- 4.2 Marginal Stability
- 4.3 Stability
- 4.4 Example
- 4.5 Example

5 STATE-SPACE MODEL

- 5.1 Controllability and Observability
- 5.2 Second Order System
- 5.3 Example
- 5.4 Example
- 5.5 Example
- 5.6 Example
- 5.7 Example

6 Nyquist Plot

- 6.1 Introduction
- 6.2 Example

7 Compensators

- 7.1 Phase Lead
- 7.2 Lag Lead
- 7.3 Example

8 GAIN MARGIN

- 8.1 Introduction
- 8.2 Example
- 8.3 Example
- 8.4 Example
- 8.1. For a unity feedback system shown in Fig 8.1

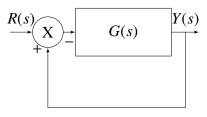


Fig. 8.1

having transfer function

$$G(s) = \frac{K}{(s+3)(s+9)(s+15)}$$
(8.1.1)

design the value of gain(K), for a gain margin of 50 dB.

8.2. Solution:

Gain Margin:

$$GM = -20\log|G(j\omega_{nc})| \qquad (8.2.1)$$

where, ω_{pc} is the phase cross-over frequency. First substitute,

$$s = j\omega \tag{8.2.2}$$

$$\implies G(j\omega) = \frac{K}{(-27\omega^2 + 405) + j(-\omega^3 + 207\omega)}$$
(8.2.3)

Now the phase will be

$$\angle G(j\omega) = -\tan^{-1}(\frac{-\omega^3 + 207\omega}{-27\omega^2 + 405})$$
 (8.2.4)

Solving for $\angle G(j\omega) = -180^{\circ}$ gives

$$\omega_{pc} = 14.3875 \tag{8.2.5}$$

Magnitude:

$$|G(j\omega)| = \frac{K}{\sqrt{(\omega^2 + 9)}\sqrt{(\omega^2 + 81)}\sqrt{(\omega^2 + 225)}}$$
(8.2.6

Substituting value of ω_{pc} in (8.2.1) gives

$$K = 16.406$$
 (8.2.7)

This can be verified from fig 8.2 The following code generates Fig. 8.2

- 8.3. Design the value gain (K) for a phase margin of 40°.
- 8.4. **Solution:**

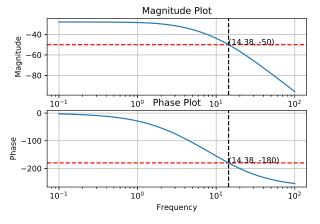


Fig. 8.2

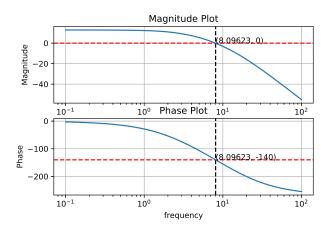


Fig. 8.4

Phase Margin:

$$PM = 180^{\circ} + \phi_{gc} \tag{8.4.1}$$

where ϕ_{gc} is the phase angle at the gain cross over frequency ω_{gc} .

$$-20\log|G(j\omega_{gc})| = 0 (8.4.2)$$

Given,

$$PM = 40^{\circ} = 180^{\circ} + \phi_{gc}$$
 (8.4.3)

$$\implies \phi_{gc} = -140^{\circ} = \angle G(j\omega_{gc})$$
 (8.4.4)

From (8.2.4)

$$\angle G(j\omega_{gc}) = -\tan^{-1}(\frac{-\omega_{gc}^3 + 207\omega_{gc}}{-27\omega_{gc}^2 + 405}) \quad (8.4.5)$$

$$\implies \omega_{gc} = 8.09623$$
 (8.4.6)

Substituting this value in (8.4.2), we get

$$20\log K = 65.016 \tag{8.4.7}$$

$$\implies K = 1781.56$$
 (8.4.8)

This again can be verified from fig 8.4. The following code generates Fig. 8.4

8.5. Design the value of gain (K) to yield maximum peak overshoot of 20% for a step input.

8.6. **Solution:** Closed loop transfer function:

$$T(s) = \frac{K}{(s+3)(s+6)(s+15) + K}$$
 (8.6.1)

Output will be:

$$\implies Y(s) = \frac{1}{s} \frac{K}{(s+3)(s+6)(s+15) + K}$$
(8.6.2)

Maximum peak overshoot:

$$M_p = \frac{y(t_p) - y(\infty)}{y(\infty)} \tag{8.6.3}$$

which is given as 20%. Here, t_p is the peak time.

$$\implies \frac{y(t_p)}{y(\infty)} = 1.2 \tag{8.6.4}$$

Plotting y(t) for different values of K, we choose the value of K, which gives the above ratio, which is verified from fig 8.6. Thus, we get

$$t_p = 0.505 \tag{8.6.5}$$

$$\implies K = 928.035$$
 (8.6.6)

The following code generates fig 8.6

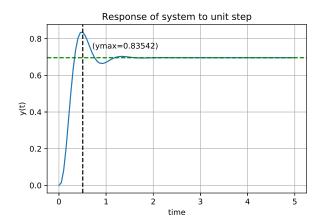


Fig. 8.6

9 Phase Margin

- 9.1 Intoduction
- 9.2 Example
- 10 OSCILLATOR
- 10.1 Introduction
- 10.2 Example
- 11 Root Locus
- 11.1 Introduction
- 11.2 Example
- 11.3 Example
- 12 Polar Plot
- 12.1 Introduction