

Oscillator

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CONTENTS

For the circuit shown in Fig. 1.1, find the loop gain $L(s) = G(s)H(s)$, $L(j\omega)$, the frequency for zero loop phase, and R_2/R_1 for oscillation.

1. Draw the equivalent control system representation for the circuit in Fig. 1.1 as well as the small signal model.

Solution: See Figs. 1.2, 1.3 and 1.4. Oscillators do not include input signal.

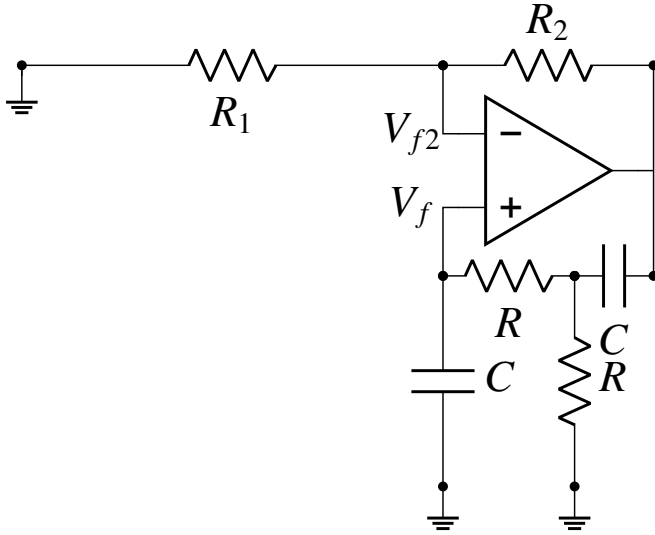


Fig. 1.1

2. Draw the block diagram and circuit diagram for H .

Solution: See Figs. 2.5 and 2.6.

3. Find H .

Solution: In Fig. 2.6, let I_o be the current flowing from V_o . Then

$$I_o = \frac{V_o}{\frac{1}{sC} + R \parallel \left(R + \frac{1}{sC}\right)} \quad (3.1)$$

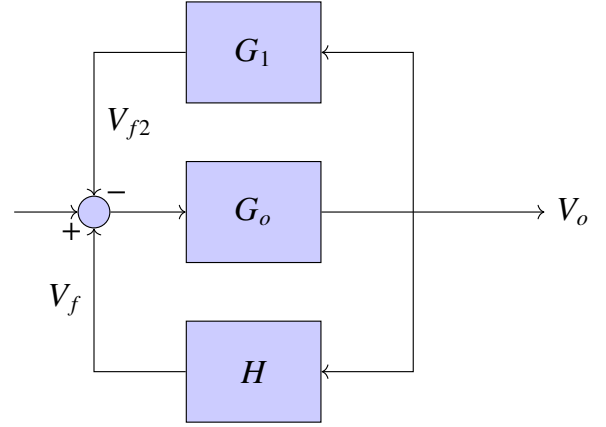


Fig. 1.2: Block diagram

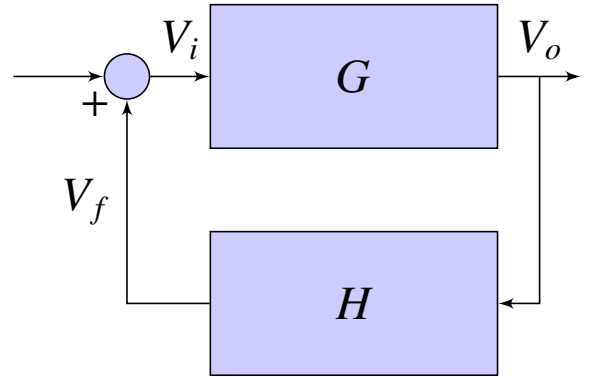


Fig. 1.3: Simplified equivalent block diagram

Using current division,

$$V_f = I_o \frac{R}{R + \left(R + \frac{1}{sC}\right)} \times \frac{1}{sC} \quad (3.2)$$

From (3.1) and (3.2),

$$\frac{V_f}{V_o} = \frac{R}{R + \left(R + \frac{1}{sC}\right)} \times \frac{1}{sC} \times \frac{1}{\frac{1}{sC} + R \parallel \left(R + \frac{1}{sC}\right)} \quad (3.3)$$

$$\Rightarrow H = \frac{1}{\left(3 + sRC + \frac{1}{sRC}\right)} \quad (3.4)$$

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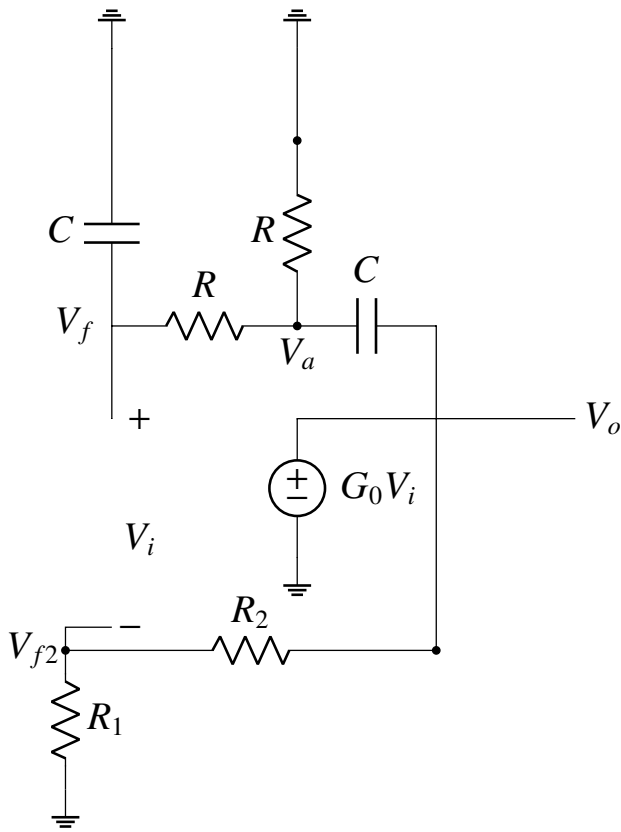


Fig. 1.4

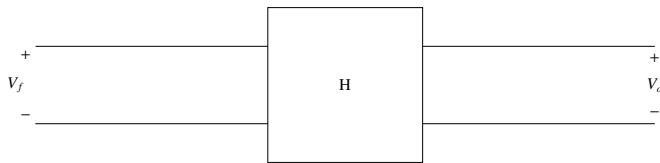


Fig. 2.5: Feedback block diagram

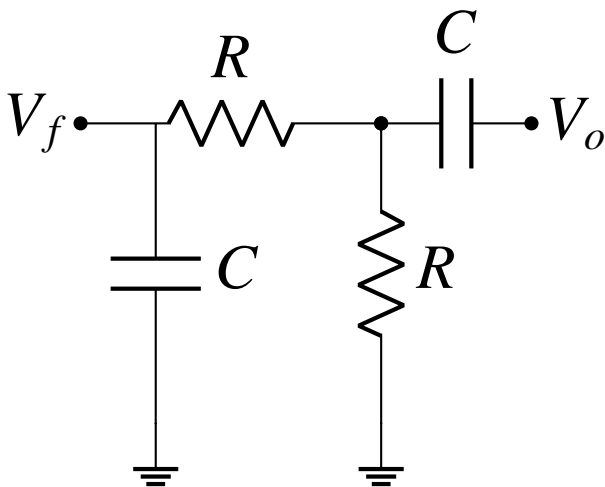


Fig. 2.6: Feedback circuit

after simplification.

4. Find R_{11} and R_{22} from Fig. 2.6.

Solution: Shorting V_o to ground,

$$R_{11} = \frac{1}{sC} \parallel \left(R + R \parallel \frac{1}{sC} \right) \quad (4.1)$$

Shorting V_f to ground,

$$R_{22} = \frac{1}{sC} + \frac{R}{2} \quad (4.2)$$

5. Draw the block diagram and circuit diagram for G .

Solution: See Figs. 5.1 and 5.2.

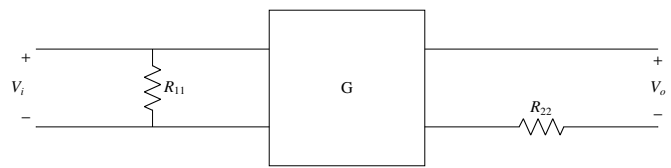


Fig. 5.1: Open loop block diagram

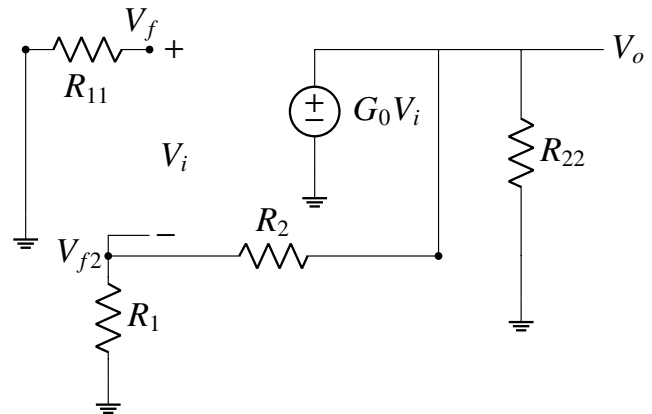


Fig. 5.2: Open loop circuit diagram

6. Find G .

Solution: From Fig. 5.2,

$$V_{f2} = \left(\frac{R_1}{R_1 + R_2} \right) V_o \quad (6.1)$$

From Fig. 1.2,

$$G_1 = \frac{V_{f2}}{V_o} \quad (6.2)$$

$$= \frac{R_1}{R_1 + R_2} \quad (6.3)$$

From Fig. 1.2, G_1 is the negative feedback factor and G_0 is the gain of the op-

amp. Therefore, equivalent G is given by

$$G = \frac{G_0}{1 + G_0 G_1} \quad (6.4)$$

$$= \frac{1}{\frac{1}{G_0} + G_1} \quad (6.5)$$

$$\Rightarrow G \approx \frac{1}{G_1}, \quad G_0 \rightarrow \infty \quad (6.6)$$

$$\text{or, } G = \frac{R_1 + R_2}{R_1} = 1 + \frac{R_2}{R_1} \quad (6.7)$$

using (6.3).

7. Find the loop gain $L(s)$.

Solution: From (6.7) and (3.4),

$$L(s) = G(s)H(s) \quad (7.1)$$

$$\Rightarrow L(s) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + sRC + \frac{1}{sRC}} \right) \quad (7.2)$$

8. Find the closed loop gain $T(s)$.

Solution: From Fig. 1.3,

$$T(s) = \frac{G}{1 - GH(s)} = \frac{G}{1 - L(s)} \quad (8.1)$$

$$= \frac{\left(1 + \frac{R_2}{R_1}\right)}{1 - \left(\frac{1 + \frac{R_2}{R_1}}{3 + sRC + \frac{1}{sRC}}\right)} \quad (8.2)$$

9. Find the conditions for oscillation.

Solution: For oscillations to start,

- $T(s)$ should have imaginary poles.
- $L(0) \geq 1$

For $T(s)$ to have imaginary poles,

$$\text{Im}\{L(j\omega)\} = 0 \quad (9.1)$$

$$\Rightarrow L(j\omega) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + j\left(\omega RC - \frac{1}{\omega RC}\right)} \right) \quad (9.2)$$

From (7.2),

$$L(j\omega) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + j\left(\omega RC - \frac{1}{\omega RC}\right)} \right) \quad (9.3)$$

$$\Rightarrow j\left(\omega RC - \frac{1}{\omega RC}\right) = 0 \quad (9.4)$$

$$\text{or, } \omega = \frac{1}{RC} \quad (9.5)$$

Also, from equation (7.2)

$$L(0) \geq 1 \Rightarrow \left(\frac{1 + \frac{R_2}{R_1}}{3 + j(0)} \right) \geq 1 \quad (9.6)$$

$$\text{or, } \frac{R_2}{R_1} \geq 2 \quad (9.7)$$

10. Find the amplitude and frequency for some arbitrary R,C values given in Table 10.

Parameter	Value
R	250Ω
C	$0.2\mu F$
R_2	$2k\Omega$
R_1	$1k\Omega$

TABLE 10

Solution: Substituting the above values in equation (8.2) results in

$$T(s) = \frac{3(25 \times 10^{-10}s^2 + 15 \times 10^{-5}s + 1)}{25 \times 10^{-10}s^2 + 1} \quad (10.1)$$

The following code gives the step response shown in equation(10.2).

```
codes/ee18btech11047/
ee18btech11047_response.py
```

$$y(t) = 3(3 \sin 20000t + 1)u(t) \quad (10.2)$$

The following code plots the step response of the system. This, in fact is the output of Fig. 1.1.

```
codes/ee18btech11047/ee18btech11047.py
```

Amplitude: From Fig. 10 $V(\text{peak-peak})$ is

$$V_{p-p} = 11.99 - (-5.99) = 17.98 \quad (10.3)$$

$$V_{max} = \frac{V_{p-p}}{2} = 8.99 \quad (10.4)$$

Frequency: Theoretical frequency is obtained using equation (9.5)

$$\omega = \frac{1}{RC} = \frac{1}{250 \times (0.2 \times 10^{-6})} \quad (10.5)$$

$$\Rightarrow \omega = 2 \times 10^4 \text{ rad/sec} \quad (10.6)$$

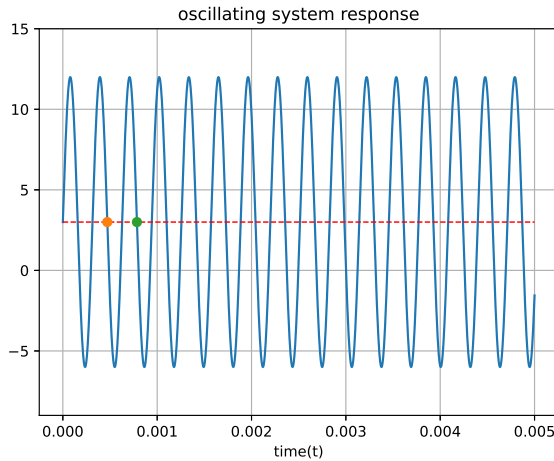


Fig. 10

$$f = \frac{\omega}{2\pi} = 3.183kHz \quad (10.7)$$

From Fig. 10 time period is calculated by any two end points of one cycle.

$$T = 0.000785391 - 0.000471071 \quad (10.8)$$

$$T = 0.00031432sec \quad (10.9)$$

$$f = \frac{1}{T} = 3.181kHz \quad (10.10)$$

11. Verify the amplitude and frequency using spice simulation.

Solution: The following readme file provides necessary instructions to simulate the circuit in spice.

```
codes/ee18btech11047/spice/README
```

The following netlist simulates the given circuit.

```
codes/ee18btech11047/spice/ee18btech11047.
net
```

The following code plots the output from the oscillator spice simulation which is shown in Fig. 11.1.

```
codes/ee18btech11047/spice/
ee18btech11047_spice.py
```

The following code plots a part of the spice output from which we can observe a clear sinusoidal output shown in Fig. 11.2.

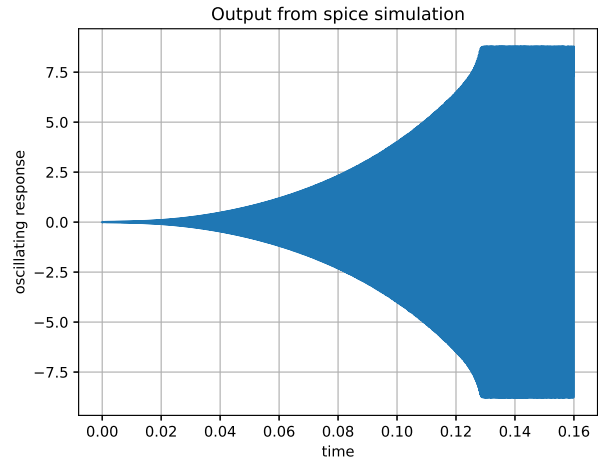


Fig. 11.1

```
codes/ee18btech11047/spice/
ee18btech11047_spice2.py
```

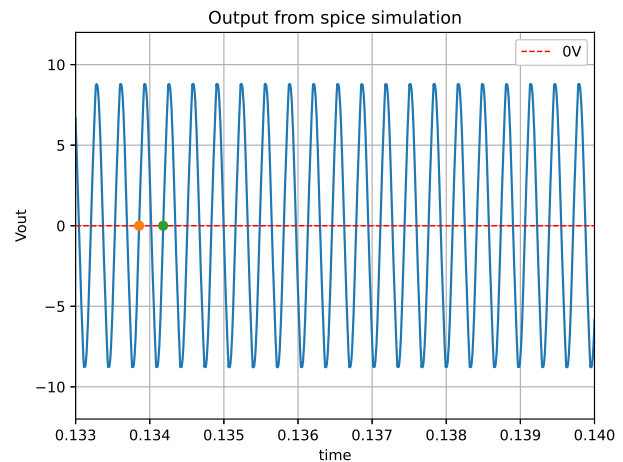


Fig. 11.2

Amplitude: From Fig. 11.2 $V_{(peak-peak)}$ is

$$V_{p-p} = 8.79 - (-8.79) = 17.58 \quad (11.1)$$

$$V_{max} = \frac{V_{p-p}}{2} = 8.79 \quad (11.2)$$

Frequency: From Fig. 11.2 time period is calculated by any two end points of one cycle,

$$T = 0.134181 - 0.133856 = 0.000325sec \quad (11.3)$$

$$f = \frac{1}{T} = 3.076kHz \quad (11.4)$$

Hence, the amplitude and frequency are verified through the spice simulation.