

1. Fourier transform

A linear and time-invariant system can be represented using its response to the unit sample sequence. This response, called the unit impulse response $h(n)$, allows us to compute the system response to any arbitrary input $x(n)$ using the linear convolution:

$$x(n) \longrightarrow \boxed{h(n)} \longrightarrow y(n) = h(n) * x(n)$$

This convolution representation is based on the fact that any signal can be represented by a linear combination of scaled and delayed unit samples.

When the system is linear and time-invariant, only one representation stands out as the most useful. It is based on the complex exponential signal set $\{e^{j\omega n}\}$ and is called the *discrete-time Fourier transform*, which is given by

$$X(e^{j\omega}) \triangleq \mathcal{F}[x(n)] = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n} \quad (3.1)$$

The inverse discrete-time Fourier transform (IDTFT) of $X(e^{j\omega})$ is given by

$$x(n) \triangleq \mathcal{F}^{-1}[X(e^{j\omega})] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega \quad (3.2)$$

The operator transforms a discrete signal $x(n)$ into a complex-valued continuous function $X(e^{j\omega})$ of real variable ω , called a digital frequency, which is measured in radians/sample.

Task 1.

Determine the discrete-time Fourier transform of $x(n) = (0.5)^n u(n)$.

Solution The sequence $x(n)$ is absolutely summable; therefore its discrete-time Fourier transform exists.

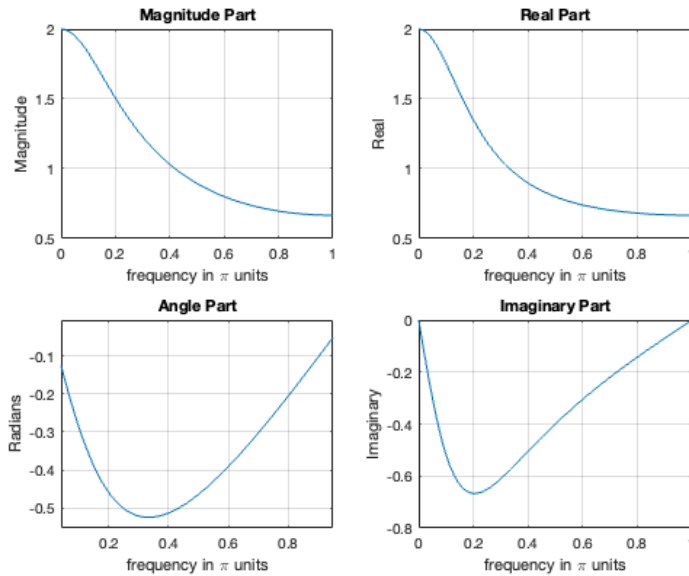
$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n} = \sum_{n=0}^{\infty} (0.5)^n e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} (0.5e^{-j\omega})^n = \frac{1}{1 - 0.5e^{-j\omega}} = \frac{e^{j\omega}}{e^{j\omega} - 0.5} \end{aligned}$$

□

Evaluate $X(e^{j\omega})$ at 501 equispaced points between $[0, \pi]$ and plot its magnitude, angle, real, and imaginary parts, use MATLAB script:

```
w = [0:1:500]*pi/500; % [0, pi] axis divided into 501 points.
X = exp(j*w) ./ (exp(j*w) - 0.5*ones(1,501));
magX = abs(X); angX = angle(X); realX = real(X); imagX = imag(X);
subplot(2,2,1); plot(w/pi,magX); grid
xlabel('frequency in \pi units'); title('Magnitude Part');
ylabel('Magnitude')
subplot(2,2,3); plot(w/pi,angX); grid
xlabel('frequency in \pi units'); title('Angle Part'); ylabel('Radians')
subplot(2,2,2); plot(w/pi,realX); grid
xlabel('frequency in \pi units'); title('Real Part'); ylabel('Real')
subplot(2,2,4); plot(w/pi,imagX); grid
xlabel('frequency in \pi units'); title('Imaginary Part');
ylabel('Imaginary')
```

The resulting plots are shown below. Note that we divided the w array by π before plotting so that the frequency axes are in the units of π and therefore easier to read. This practice is strongly recommended.



To do:

- Tape code to Matlab, explain operation in the second line, run the code.
- Plot figures presenting $X(e^{j\omega})$

If $x(n)$ is of finite duration, then MATLAB can be used to compute $X(e^{j\omega})$ numerically at any frequency ω . The approach is to implement (3.1) directly. If, in addition, we evaluate $X(e^{j\omega})$ at equispaced frequencies between $[0, \pi]$, then (3.1) can be implemented as a *matrix-vector multiplication* operation. To understand this, let us assume that the sequence $x(n)$ has N samples between $n_1 \leq n \leq n_N$ (i.e., not necessarily between $[0, N - 1]$) and that we want to evaluate $X(e^{j\omega})$ at

$$\omega_k \triangleq \frac{\pi}{M}k, \quad k = 0, 1, \dots, M$$

which are $(M + 1)$ equispaced frequencies between $[0, \pi]$. Then (3.1) can be written as

$$X(e^{j\omega_k}) = \sum_{\ell=1}^N e^{-j(\pi/M)kn_{\ell}} x(n_{\ell}), \quad k = 0, 1, \dots, M$$



When $\{x(n_\ell)\}$ and $\{X(e^{j\omega_k})\}$ are arranged as *column* vectors \mathbf{x} and \mathbf{X} , respectively, we have

$$\mathbf{X} = \mathbf{W}\mathbf{x} \quad (3.3)$$

where \mathbf{W} is an $(M+1) \times N$ matrix given by

$$\mathbf{W} \triangleq \left\{ e^{-j(\pi/M)kn_\ell}; n_1 \leq n \leq n_N, \quad k = 0, 1, \dots, M \right\}$$

In addition, if we arrange $\{k\}$ and $\{n_\ell\}$ as *row* vectors \mathbf{k} and \mathbf{n} respectively, then

$$\mathbf{W} = \left[\exp \left(-j \frac{\pi}{M} \mathbf{k}^T \mathbf{n} \right) \right]$$

In MATLAB we represent sequences and indices as row vectors; therefore taking the transpose of (3.3), we obtain

$$\mathbf{X}^T = \mathbf{x}^T \left[\exp \left(-j \frac{\pi}{M} \mathbf{n}^T \mathbf{k} \right) \right] \quad (3.4)$$

Note that $\mathbf{n}^T \mathbf{k}$ is an $N \times (M+1)$ matrix. Now (3.4) can be implemented in MATLAB as follows.

```
>> k = [0:M]; n = [n1:n2];
>> X = x * (exp(-j*pi/M)) .^ (n'*k);
```

Task 2.

Numerically compute the discrete-time Fourier transform of the following finite-duration sequence $x(n)$

$$x(n) = \{1 \ 2 \ 3 \ 4 \ 5\}$$

at 501 equispaced frequencies between $[0, \pi]$.


According to the definition 3.1 we get

$$X(e^{j\omega}) = \sum_{-\infty}^{\infty} x(n) e^{-j\omega n} = 1 + 2e^{-j\omega} + 3e^{-j2\omega} + 4e^{-j3\omega} + 5e^{-j4\omega}$$

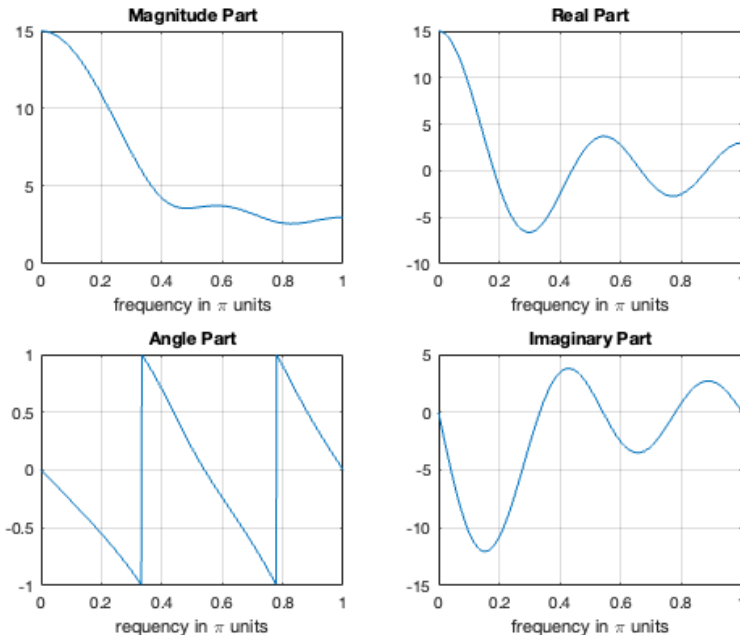
$$n = 0, 1, 2, 3, 4$$

Run the following script

```
clear
n = 0:4; x = 1:5; k = 0:500;
w = (pi/500)*k; % Frequency range 0 to pi
X = x * (exp(-j*pi/500)) .^ (n'*k);
magX = abs(X); angX = angle(X);
realX = real(X); imagX = imag(X);
subplot(2,2,1); plot(k/500,magX);grid
xlabel('frequency in \pi units'); title('Magnitude Part')
subplot(2,2,3); plot(k/500,angX/pi);grid
xlabel('frequency in \pi units'); title('Angle Part')
subplot(2,2,2); plot(k/500,realX);grid
xlabel('frequency in \pi units'); title('Real Part')
subplot(2,2,4); plot(k/500,imagX);grid
xlabel('frequency in \pi units'); title('Imaginary Part')
```

- Tape code to Matlab,  main operation in the third line, run the code.

The frequency-domain plots are shown below. Note that the **angle plot** is depicted as a **discontinuous function** between $-\pi$ and π . This is because the angle function in MATLAB computes the principal angle.



Note that you can calculate the discrete-time Fourier transform using the `fft` command in Matlab. Note that you have to indicate number of frequencies in the `fft(x,N)` in the **whole frequency range** 0 to 2π .

Run the following script:

```
x = [1:5]; k = 0:500; w = (pi/500)*k;
X = fft(x,1000); % DFT calculated with zero-padding from 0 to 2pi
X=X(1:501); % Plot half spectrum 0 to pi
magX = abs(X); angX = angle(X);
realX = real(X); imagX = imag(X);
subplot(2,2,1); plot(k/500,magX);grid
xlabel('frequency in \pi units'); title('Magnitude Part')
subplot(2,2,3); plot(k/500,angX/pi);grid
xlabel('frequency in \pi units'); title('Angle Part')
subplot(2,2,2); plot(k/500,realX);grid
xlabel('frequency in \pi units'); title('Real Part')
subplot(2,2,4); plot(k/500,imagX);grid
xlabel('frequency in \pi units'); title('Imaginary Part')
```

- Tape code to Matlab, explain operation in the second line, run the code.

To do:

- Plot figures presenting magnitude and angle of $X(e^{j\omega})$
- Plot figure presenting Real and Imaginary part of $X(e^{j\omega})$ in complex plane (*Re vs Im*)
- Compare the results obtained using both scripts.
- Explain the zero-padding operation.

2. Frequency domain representation of LTI systems

Fourier transform representation is the most useful signal representation for LTI systems. An LTI system can be represented in the frequency domain by

$$X(e^{j\omega}) \longrightarrow \boxed{H(e^{j\omega})} \longrightarrow Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega})$$

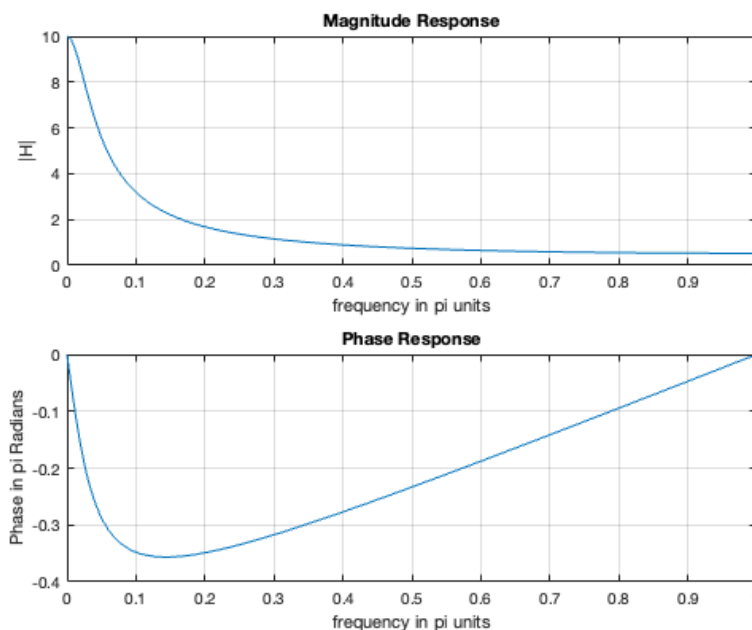
Task 3.

Determine the frequency response $H(e^{j\omega})$ of a system characterized by $h(n) = (0.9)^n u(n)$. Plot the magnitude and the phase responses.

$$\begin{aligned} H(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} h(n) e^{-j\omega n} = \sum_{n=0}^{\infty} (0.9)^n e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} (0.9 e^{-j\omega})^n = \frac{1}{1 - 0.9 e^{-j\omega}} \end{aligned}$$

```
w = [0:1:500]*pi/500; % [0, pi] axis divided into 501 points.
H = exp(j*w) ./ (exp(j*w) - 0.9*ones(1,501));
magH = abs(H); angH = angle(H);
subplot(2,1,1); plot(w/pi,magH); grid;
xlabel('frequency in pi units'); ylabel('|H|');
title('Magnitude Response');
subplot(2,1,2); plot(w/pi,angH/pi); grid;
xlabel('frequency in pi units'); ylabel('Phase in pi Radians');
title('Phase Response');
```

The resulting plots are shown below. Note that we divided the w array by π before plotting so that the frequency axes are in the units of π and therefore easier to read.



To do:

- Tape code to Matlab, explain operation in the second line, run the code.
- Plot figures presenting magnitude and phase responses of $H(j\omega)$