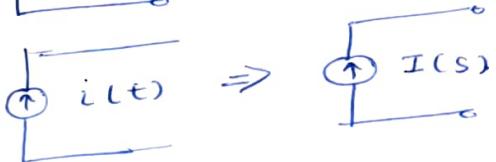


transformed ckt. component representation \rightarrow

Independent Sources \rightarrow

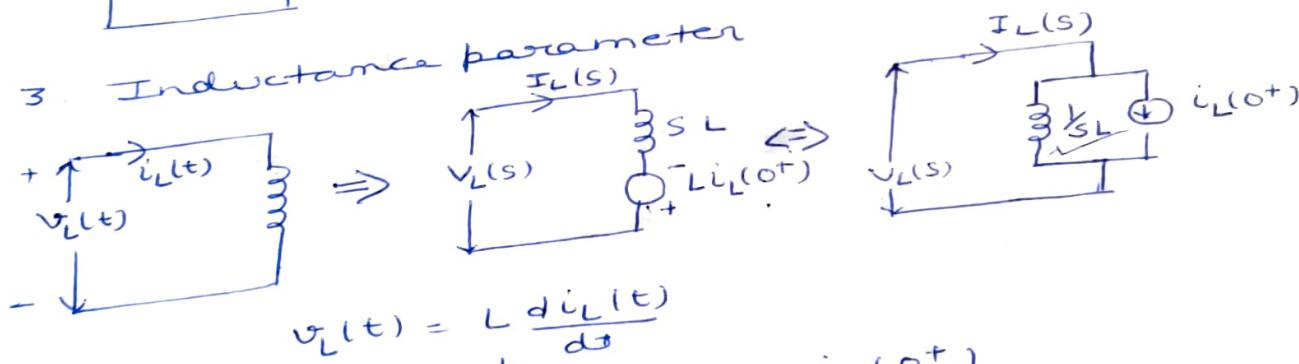


2. Resistance parameter

Diagram illustrating the transformation of a resistor R with voltage $V_R(t)$ and current $i_R(t)$ into a resistor R with voltage $V(s)$ and current $I_R(s)$. The original circuit shows a resistor R with voltage $V_R(t)$ across it and current $i_R(t)$ flowing through it. The transformed circuit shows a resistor R with voltage $V(s)$ across it and current $I_R(s)$ flowing through it.

$$V_R(t) = R i_R(t) \Rightarrow V(s) = R I(s)$$

3. Inductance parameter



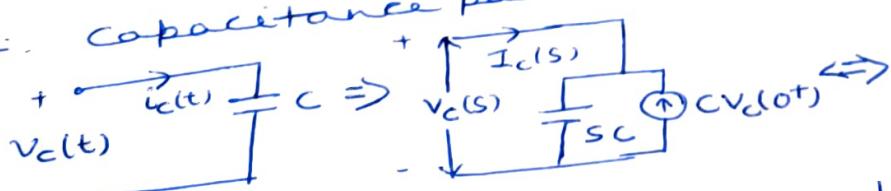
$$V_L(t) = L \frac{di_L(t)}{dt}$$

$$i_L(t) = \frac{1}{L} \int_0^t V_L(t') dt' + i_L(0^+)$$

$$V_L(s) = L s I(s) - L i_L(0^+)$$

$$I_L(s) = \frac{1}{sL} V(s) + \frac{i_L(0^+)}{s}$$

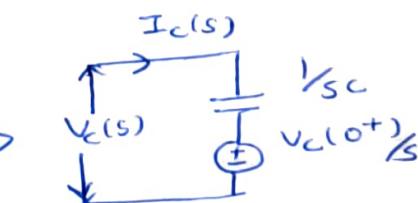
4. Capacitance parameter



$$i_C(t) = C \frac{dV_C(t)}{dt}$$

$$V_C(t) = \frac{1}{C} \int_0^t i_C(t') dt' + V_C(0^+)$$

$$I_C(s) = C V(s) - C V_C(0^+) \quad V_C(s) = \frac{1}{Cs} I_C(s) + \frac{V_C(0^+)}{s}$$



$$I(s) = \frac{k_1}{s+1} + \frac{k_2}{s+2+j^2} + \frac{k_3}{s+2-j^2}$$

$$k_1 = (s+1) I(s) \Big|_{\substack{s=0 \\ s=-1}} = \frac{s^2 + 5s + 9}{s^2 + 4s + 8} = \frac{5}{5} = 1$$

$$k_2 = (s+2+j^2) I(s) \Big|_{s=-2+j^2}$$

$$= \frac{s^2 + 5s + 9}{(s+1)(s+2-j^2)} \Big|_{s=-2+j^2}$$

$$= \frac{-4 + 8j - 10 - 10j + 9}{(-2-j^2+1)(-2-j^2+j^2)} = \frac{-1-2j}{(-1-j^2)(-2j^2)} = \frac{1}{j}$$

$$k_3 = k_2^* = \frac{1}{j^4}$$

$$I(s) = \frac{1}{s+1} + \frac{-1/j^4}{s+2+j^2} + \frac{1/j^4}{s+2-j^2}$$

$$= e^{-t} - \frac{1}{j^4} e^{-(2+j^2)t} + \frac{1}{j^4} e^{-(2-j^2)t}$$

$$= e^{-t} - \frac{1}{j^4} e^{-2t} [e^{-j^2t} - e^{j^2t}]$$

$$= e^{-t} + \frac{1}{j^4} e^{-2t} \sin 2t.$$

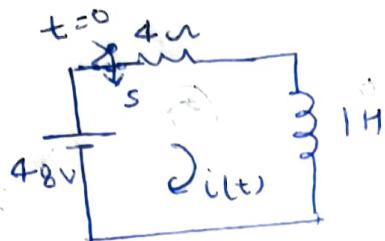
alternatively \rightarrow

$$I(s) = \frac{s^2 + 5s + 9}{(s+1)(s^2 + 4s + 8)}$$

$$= \frac{1}{s+1} + \frac{1}{s^2 + 4s + 8}$$

$$= \frac{1}{s+1} + \frac{1}{2} \left[\frac{2}{(s+2)^2 + 2^2} \right]$$

$$i(t) = e^{-t} + \frac{1}{2} e^{-2t} \sin 2t$$



Assume initial ct. through inductor is 3 A. Using L.T. determine ct. $i(t)$.

By KVL \rightarrow

$$48 = 4i(t) + \frac{di(t)}{dt}$$

Taking L.T.

$$4I(s) + 1 [sI(s) - i(0^+)] = \frac{48}{s}$$

$$4I(s) + sI(s) - 3 = \frac{48}{s}$$

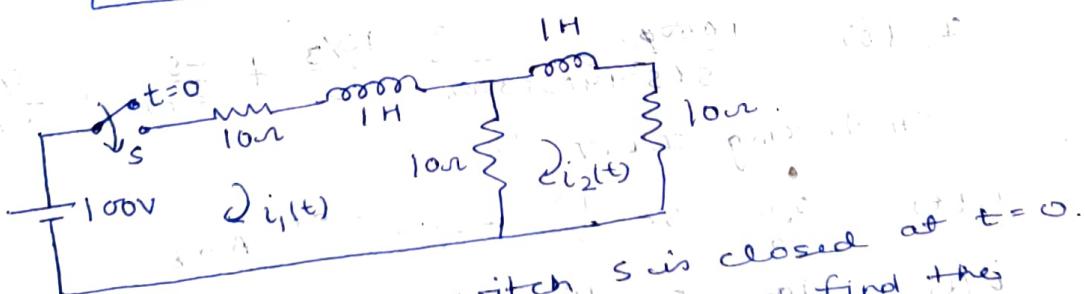
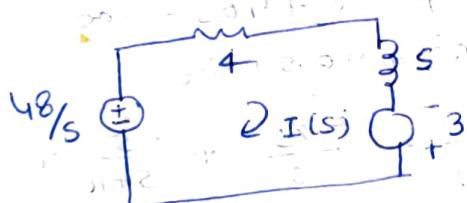
$$I(s) = \frac{3s + 48}{s(s+4)} = \frac{k_1}{s} + \frac{k_2}{s+4}$$

$$k_1 = sI(s)|_{s=0} = \frac{3s + 48}{s+4}|_{s=0} = 12$$

$$k_2 = (s+4)I(s)|_{s=-4} = \frac{3s + 48}{s}|_{s=-4} = -9$$

$$I(s) = \frac{12}{s} - \frac{9}{s+4}$$

$$i(t) = 12 - 9e^{-4t}$$



In the N/w, the switch s is closed at $t=0$. With the N/w parameter value shown, find the expression for $i_1(t) + i_2(t)$, if the N/w is unenergized before the switch s is closed.

Sol. apply KVL, loop eqn are

$$\frac{di_1}{dt} + 10i_1(t) + 10[i_1(t) - i_2(t)] = 100$$

$$\frac{di_1}{dt} + 20i_1(t) - 10i_2(t) = 100$$

$$\text{and } \frac{di_2(t)}{dt} + 20i_2(t) - 10i_1(t) = 0 \quad \text{--- (2)}$$

L.T. of eqn (1) & (2)

$$I_1(s) [s + 20] - I_2(s)(10) = 100/s$$

$$\text{and } -10I_1(s) + I_2(s)[20 + s] = 0$$

In Matrix form

$$\begin{bmatrix} s+20 & 10 \\ -10 & 20+s \end{bmatrix} = \begin{bmatrix} 100/s \\ 0 \end{bmatrix}$$

By cramer rule

$$I_1(s) = \frac{\Delta_1}{\Delta}, \quad I_2(s) = \frac{\Delta_2}{\Delta}$$

$$\Delta_1 = \begin{bmatrix} 100/s & -10 \\ 0 & s+20 \end{bmatrix}, \quad \Delta_2 = \begin{bmatrix} s+20 & -100/s \\ -10 & 0 \end{bmatrix}$$

$$= (s+20)\left(\frac{100}{s}\right) + 10 \times 0$$

$$= \frac{(s+20)100}{s}$$

$$\Delta_2 = 10 \times \frac{100}{s}$$

$$= \frac{1000}{s}$$

$$\Delta = \begin{bmatrix} s+20 & -10 \\ -10 & 20+s \end{bmatrix} \Rightarrow s^2 + 400 + 40s - 100$$

$$= s^2 + 40s + 300$$

$$I_1(s) = \frac{(s+20)100}{s(s^2 + 40s + 300)} = \frac{20/3}{s} + \frac{-5}{s+10} + \frac{-5/3}{s+30}$$

$$I_2(s) = \frac{1000}{s(s^2 + 40s + 300)} = \frac{10/3}{s} + \frac{-5}{s+10} + \frac{5/3}{s+30}$$

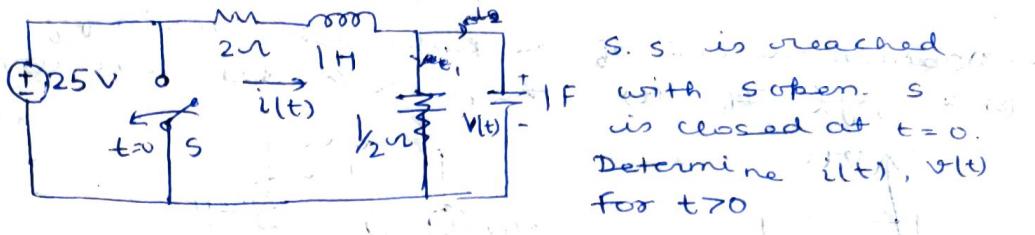
On taking I.L.T.

$$i_1(t) = \frac{20}{3} - 5e^{-10t} + \frac{5}{3}e^{-30t}$$

$$i_2(t) = \frac{10}{3} - 5e^{-10t} + \frac{5}{3}e^{+30t}$$

Amb

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S.S. is reached
with $s \rightarrow \infty$.
If s is closed at $t=0$,
Determine $i(t)$, $v(t)$
for $t > 0$

Sol. at $t = 0^-$ S.S. is reached i.e. L is S.C.
 C is O.C.

$$i(0^-) = \frac{25}{2 + \frac{1}{2}} = 10 \text{ A}$$

$$v(0^-) = 25 - 2i(0^-) - v(0^-) = 0$$

$$v(0^-) = 25 - 2 \times 10 = 5 \text{ V}$$

now s is closed, Apply kVL

$$2i(t) + \frac{di(t)}{dt} + v(t) = 0$$

Taking L.T.

$$2I(s) + LS I(s) - i(0^+) + v(s) = 0$$

$$(2+s)I(s) - 10 + v(s) = 0$$

$$\text{as } i(0^+) = i(0^-) = 10 \text{ A}$$

$$(2+s)I(s) + v(s) = 10 \quad \text{--- (1)}$$

also,

$$i(t) = \frac{v(t)}{\frac{1}{2}} + 1 \cdot \frac{d(v(t))}{dt}$$

Taking L.T.

$$I(s) = 2v(s) + s v(s) - v(0^+)$$

$$I(s) = 2v(s) + sv(s) - 5 \quad \text{--- (2)}$$

$$5 = v(s)(2+s) - I(s) = 0$$

On solving (1) & (2)

$$(2+s)I(s) + v(s) = 10$$

$$- (2+s)I(s) + (2+s)^2 v(s) = 5(2+s)$$

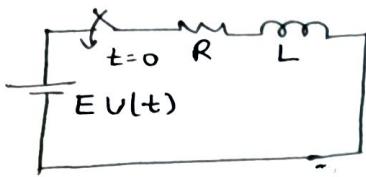
$$v(s)[1 + (2+s)^2] = 10 + [10 + 5s]$$

$$v(s) = \frac{10 + 5s}{s^2 + 4s + 5} = \frac{10s + 50s}{s^2 + 4s + 5 + s + 5}$$

$$= \frac{20 + 5s}{(s+2)^2 + 1^2} = \frac{5(s+2)}{(s+2)^2 + 1} + \frac{10}{(s+2)^2 + 1}$$

$$v(t) = 5e^{-2t} \cos t + 10e^{-2t} \sin t \text{ V}$$

Step Response of R-L Series circuit



at \$t=0\$ switch is closed
so \$i(0^+) = i(0^-) = 0\$

By KVL \$\rightarrow\$

$$E_U(t) = R i(t) + L \frac{di(t)}{dt} \quad \textcircled{1}$$

Input is unit step signal.

On taking Laplace transform of eqn \$\textcircled{1}\$, we have

$$\frac{E}{S} = RI(s) + L [sI(s) - i(0^+)]$$

$$i(0^+) = 0$$

$$\frac{E}{S} = RI(s) + LS I(s)$$

$$\Rightarrow I(s) = \frac{E/L}{S(S + R/L)} = \frac{A}{S} + \frac{B}{S + R/L}$$

$$A = \left. \frac{E/L}{S + R/L} \right|_{S=0} = \frac{E}{R}$$

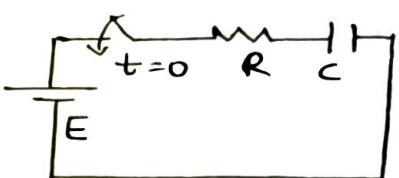
$$B = \left. \frac{E/L}{S} \right|_{S=-R/L} = -\frac{E}{R}$$

$$\text{Thus } I(s) = \frac{E/R}{S} - \frac{E/R}{S + R/L}$$

On taking Inverse Laplace transform

$$i(t) = E/R - \frac{E}{R} e^{-R/L t}$$

Step Response of RC Series Circuit



at $t=0$ switch is closed
By KVL \rightarrow

$$EV(t) = RI(t) + \frac{1}{C} \int i(t) dt \quad \dots \textcircled{1}$$

On taking L.T. of eqn \textcircled{1}

$$\frac{E}{S} = RI(s) + \frac{1}{C} \left[\frac{I(s)}{S} + \frac{i(0^+)}{S} \right]$$

at $t=0$ capacitor acts as s.c. $\rightarrow i(0^+) = 0$

$$\frac{E}{S} = RI(s) + \frac{1}{CS} I(s) + \frac{1}{CS} \cdot \frac{E}{R}$$

$$\Rightarrow \frac{E}{S} - \frac{1}{RCS} E = I(s) \left[\frac{RCs + 1}{CS} \right]$$

$$\Rightarrow \frac{E}{S} \left[\frac{RC - 1}{RC} \right] = I(s) \left[\frac{RCs + 1}{CS} \right]$$

$$\Rightarrow I(s) = \frac{E}{R} \left(\frac{RC - 1}{1 + RCS} \right)$$

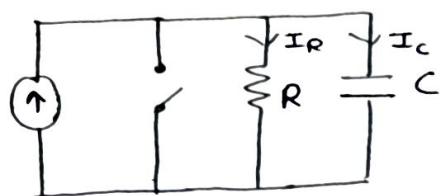
$$= \frac{E}{R} \cdot \frac{RC}{RC} \frac{\left(1 - \frac{1}{RC} \right)}{\left(S + \frac{1}{RC} \right)}$$

$$= \frac{E}{R} \left\{ \left(1 - \frac{1}{RC} \right) / \left(S + \frac{1}{RC} \right) \right\}$$

on taking I.L.T.

$$= \frac{E}{R} \left(1 - \frac{1}{RC} \right) e^{-t/RC}$$

Current Response of RC parallel circuit →
 at $t=0$ switch is opened
 By KVL →



$$I = I_R + I_C \quad \text{--- (1)}$$

also for parallel circuit

$$V_R = V_C$$

$$R I_R = \frac{1}{C} \int I_C dt$$

Multiply eqn (1) both side by R

$$R I = R I_R + R I_C$$

$$R I = \frac{1}{C} \int I_C dt + R I_C$$

On taking Laplace Transform

$$R \frac{I(s)}{s} = \frac{1}{Cs} I_C(s) + R I_C(s)$$

$$= I_C(s) \left[R + \frac{1}{Cs} \right]$$

$$I_C(s) = \frac{I(s) R}{\left(R + \frac{1}{Cs} \right) s}$$

$$= I(s) \left[\frac{1}{s + \frac{1}{RC}} \right]$$

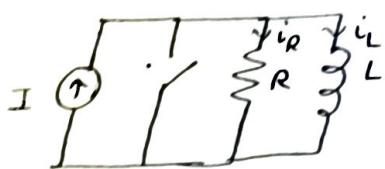
On taking I. L. T.

$$i_C(t) = i(t) \left[e^{-\frac{t}{RC}} \right]$$

~~attribution~~



Step current Response of a R-L parallel C₁₈



By KCL
 $I = I_R + I_L \quad \text{--- (1)}$
 for parallel circuit
 $V_R = V_L$

$$\text{i.e. } RI_R = L \frac{dI_L}{dt} \quad \text{--- (2)}$$

On multiplying eqn (1) by R , we get

$$RI = RI_R + RI_L$$

$$RI = L \frac{dI_L}{dt} + RI_L$$

On taking L.T.

$$R \cdot \frac{I(s)}{s} = (Ls + R) I_L(s)$$

as all initial conditions are zero

$$I_L(s) = I(s) \cdot \frac{R}{s(R+Ls)}$$

$$= I(s) \cdot \frac{R/L}{s(s+R/L)} = I(s) \left[\frac{A}{s} + \frac{B}{s+R/L} \right]$$

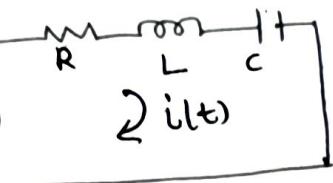
$$A = \left. \frac{R/L}{s+R/L} \right|_{s=0} = 1$$

$$B = \left. \frac{R/L}{s+R/L} \right|_{s=-R/L} = -1$$

$$I_L(s) = I(s) \left[\frac{1}{s} - \frac{1}{s+R/L} \right]$$

$$i_L(t) = i(t) [1 - e^{-R/L t}]$$

Series circuit with step Input \rightarrow



Let all initial conditions are zero.

By KVL

$$V_U(t) = R i(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt$$

On taking Laplace Transform

$$\frac{V}{s} = RI(s) + LS I(s) + \frac{1}{Cs} I(s)$$

$$I(s) = \frac{V/L}{s^2 + R/Ls + 1/LC} \quad \text{--- (1)}$$

Roots of denominator polynomial of eqn (1) is

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

$$s_1 = -\frac{R}{2L} + \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$s_2 = -\frac{R}{2L} - \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$\text{Let } \omega_0 = \frac{1}{\sqrt{LC}} \quad \xi \omega_0 = \frac{R}{2L} \quad \text{i.e. } \xi = \frac{R}{2} \sqrt{\frac{C}{L}}$$

\downarrow natural freq. of oscillation \downarrow damping factor

$$s_1 = -\xi \omega_0 + \omega_0 \sqrt{\xi^2 - 1}$$

$$s_2 = -\xi \omega_0 - \omega_0 \sqrt{\xi^2 - 1}$$

$$I(s) = \frac{V/L}{(s-s_1)(s-s_2)} = \frac{A}{s-s_1} + \frac{B}{s-s_2}$$

$$A = \left. \frac{V/L}{s-s_2} \right|_{s=s_1} = \frac{V}{2 \xi \omega_0 L \sqrt{\xi^2 - 1}}$$

$$B = \left. \frac{V/L}{s-s_1} \right|_{s=s_2} = \frac{-V}{2 \omega_0 L \sqrt{\xi^2 - 1}}$$

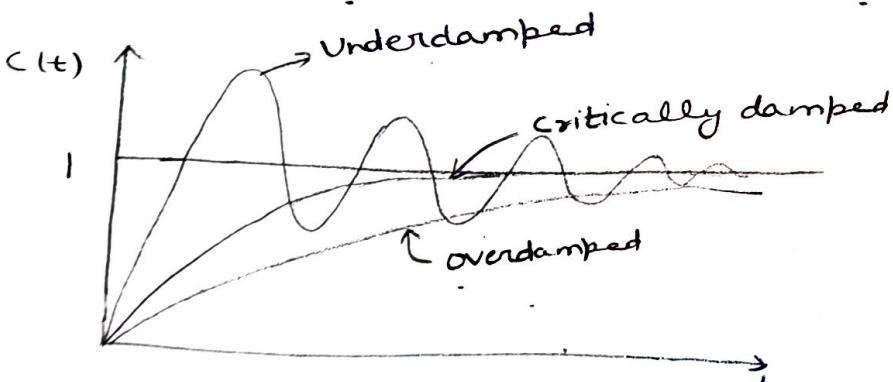
$$I(s) = \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} \left[\frac{1}{s-s_1} - \frac{1}{s-s_2} \right]$$

On taking I.L.T. we get

$$\begin{aligned} i(t) &= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} [e^{s_1 t} - e^{s_2 t}] \\ &= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \left[e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right] \end{aligned}$$

Depending upon the value of R, L, C three cases may appear

- a) $\frac{R}{2L} > \frac{1}{\sqrt{LC}}$ \Rightarrow Overdamped condition
- b) $\frac{R}{2L} < \frac{1}{\sqrt{LC}}$ \Rightarrow Underdamped Condition
- c) $\frac{R}{2L} = \frac{1}{\sqrt{LC}}$ \Rightarrow Critically damped condition



Case I \rightarrow Overdamped condition

$$\frac{R}{2L} > \frac{1}{\sqrt{LC}} \text{ i.e. } \xi > 1$$

$$\begin{aligned} i(t) &= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \left[e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right] \\ &= \frac{V}{\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \sinh(\omega_0 \sqrt{\xi^2 - 1}) t \end{aligned}$$

Critically damped condition ($\xi = 1$)

$$I(s) = \frac{V/L}{s^2 + 2\omega_0 s + \omega_0^2}$$
$$= \frac{V}{L} \left(\frac{1}{(s + \omega_0)^2} \right)$$

On taking I. L. T.

$$i(t) = \frac{V}{L} t e^{-\omega_0 t}$$

III : Underdamped condition ($\xi > 1$)

$$\frac{R}{2L} > \frac{1}{\sqrt{LC}}$$

$$i(t) = \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \left[e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right]$$

$$= \frac{V}{\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \left[\frac{e^{(j\omega_0 \sqrt{1-\xi^2})t} - e^{-(j\omega_0 \sqrt{1-\xi^2})t}}{2j} \right]$$

$$= \frac{V}{\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \sin(\omega_0 \sqrt{1-\xi^2} t)$$

R-L Series Circuit with impulse Input

By KVL

$$V\delta(t) = RI(t) + L \frac{di(t)}{dt}$$

On taking L.T.

$$V = RI(s) + LS I(s)$$

$$I(s) = \frac{V}{R+LS}$$

$$= \frac{V}{L} \left(\frac{1}{s + R/L} \right)$$

On taking I.L.T.

$$i(t) = V_L e^{-R_L t}$$

$$V_R = RI(t) = \frac{VR}{L} e^{-R_L t} \quad (\text{Vg. across Resistor})$$

$$\begin{aligned} V_L &= L \frac{di(t)}{dt} = L \cdot \frac{d}{dt} (V_L e^{-R_L t}) \\ &= L \cdot \frac{V}{L} \cdot (-R_L) e^{-R_L t} \end{aligned}$$

$$V_L = -\frac{VR}{L} e^{-R_L t} \quad (\text{Vg. across inductor})$$

R-C Series Circuit with impulse Input

By KVL \rightarrow

$$V\delta(t) = RI(t) + \frac{1}{C} \int i(t) dt$$

$$V = RI(s) + \frac{1}{Cs} I(s)$$

$$\begin{aligned} I(s) &= \frac{V}{R + 1/Cs} = \frac{V}{R} \left(\frac{s}{s + 1/R_C} \right) \\ &= \frac{V}{R} \left(1 - \frac{1/R_C}{s + 1/R_C} \right) \end{aligned}$$

taking I.L.T.

$$i(t) = \frac{V}{R} \left[\delta(t) - \frac{1}{RC} e^{-t/RC} \right]$$

$$v_R = R i(t) = V \left[\delta(t) - \frac{1}{RC} e^{-t/RC} \right]$$

$$v_C = [V\delta(t) - v_R] = \frac{V}{RC} e^{-t/RC}$$

R-L-C Series Circuit with Impulse Input \rightarrow

By KVL

$$V\delta(t) = R i(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt$$

On taking L.T.

$$V = R I(s) + L s I(s) + \frac{1}{C s} I(s)$$

$$I(s) = \frac{V_L s}{s^2 + R_L s + Y_{LC}}$$

Roots of denominator polynomial are

$$s^2 + \frac{R}{L}s + Y_{LC} = 0$$

$$s_1 = -\frac{R}{2L} + \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$s_2 = -\frac{R}{2L} - \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$\text{Let } \omega_0 = \frac{1}{\sqrt{LC}} \quad \xi \omega_0 = \frac{R}{2L} \quad \text{i.e. } \xi = \frac{R}{2} \sqrt{\frac{C}{L}} = \text{damping ratio}$$

$$s_1 = -\xi \omega_0 + \omega_0 \sqrt{\xi^2 - 1}$$

$$s_2 = -\xi \omega_0 - \omega_0 \sqrt{\xi^2 - 1}$$

$$I(s) = \frac{(V_L)s}{(s-s_1)(s-s_2)} = \frac{A}{s-s_1} + \frac{B}{s-s_2}$$

$$A = \frac{(V_L)s}{s-s_2} \Big|_{s=s_1} = \frac{V s_1}{s_1-s_2} = \frac{V s_1}{2 \omega_0 L \sqrt{\xi^2 - 1}}$$

$$\frac{N_1(s)}{M_2(s)}$$

$$1(s)$$

$$2(s)$$

$$B = \frac{\left(\frac{V}{L}\right)S}{(S-S_1)} \Big|_{S=S_2} = \frac{\frac{V}{L} S_2}{\cancel{2\omega_0 L} (S_2 - S_1)}$$

$$= \frac{V S_2}{2\omega_0 L \sqrt{\xi^2 - 1}}$$

$$I(S) = \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} \left[\frac{S_1}{S - S_1} - \frac{S_2}{S - S_2} \right]$$

$$= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} [S_1 e^{S_1 t} - S_2 e^{S_2 t}]$$

$$= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} [S_1 e^{(\omega_0 \sqrt{\xi^2 - 1})t} - S_2 e^{-(\omega_0 \sqrt{\xi^2 - 1})t}]$$

$$= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} \left[e^{-\xi\omega_0 t} \left\{ (\xi\omega_0 + \omega_0 \sqrt{\xi^2 - 1}) e^{(\omega_0 \sqrt{\xi^2 - 1})t} - (-\xi\omega_0 - \omega_0 \sqrt{\xi^2 - 1}) e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right\} \right]$$

Case I: overdamped condition ($\xi > 1$)

$$i(t) = \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} \left[e^{-\xi\omega_0 t} \left[(-\xi\omega_0 + \omega_0 \sqrt{\xi^2 - 1}) e^{(\omega_0 \sqrt{\xi^2 - 1})t} - (-\xi\omega_0 - \omega_0 \sqrt{\xi^2 - 1}) e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right] \right]$$

$$= \frac{V}{L \sqrt{\xi^2 - 1}} \left[\sqrt{\xi^2 - 1} \cosh(\omega_0 \sqrt{\xi^2 - 1}) t - \xi \sinh(\omega_0 \sqrt{\xi^2 - 1}) t \right]$$

Case II: Critically Damped Condition ($\xi = 1$)

$$I(S) = \frac{V/L S}{S^2 + 2\omega_0 S + \omega_0^2} = \frac{V}{L} \left(\frac{S}{(S + \omega_0)^2} \right)$$

$$= \frac{V}{L} \left[\frac{A}{(S + \omega_0)^2} + \frac{B}{(S + \omega_0)^2} \right]$$

$$\frac{S}{(S + \omega_0)^2} \Big|_{S=-\omega_0} = -\omega_0$$

$$\frac{d}{ds} \left[\frac{S}{(S + \omega_0)^2} \right]_{S=-\omega_0} = 1$$

$$I(s) = \frac{V}{L} \left[\frac{1}{s + \omega_0} - \frac{\omega_0}{(s + \omega_0)^2} \right]$$

on taking I.L.T.

$$i(t) = \frac{V}{L} (1 - \omega_0 t) e^{-\omega_0 t}$$

case III: Underdamped Condition $\xi < 1$

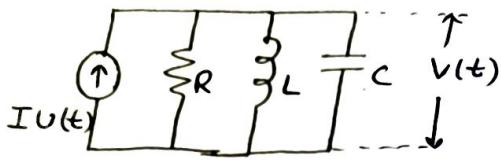
$$\begin{aligned} i(t) &= \frac{V}{2\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi \omega_0 t} \left[\left(\omega_0 \sqrt{\xi^2 - 1} \right) \left\{ e^{(\omega_0 \sqrt{\xi^2 - 1})t} \right. \right. \\ &\quad \left. \left. + e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right\} - \xi \omega_0 \left\{ e^{(\omega_0 \sqrt{\xi^2 - 1})t} \right. \right. \\ &\quad \left. \left. - e^{-(\omega_0 \sqrt{\xi^2 - 1})t} \right\} \right] \\ &= \frac{V}{2\omega_0 L j \sqrt{1 - \xi^2}} e^{-\xi \omega_0 t} \left[\left(j \omega_0 \sqrt{1 - \xi^2} \right) \left\{ e^{(j \omega_0 (1 - \xi^2)t)} \right. \right. \\ &\quad \left. \left. + e^{-(j \omega_0 \sqrt{1 - \xi^2})t} \right\} - \xi \omega_0 \left\{ e^{(j \omega_0 \sqrt{1 - \xi^2})t} \right. \right. \\ &\quad \left. \left. - e^{-(j \omega_0 \sqrt{1 - \xi^2})t} \right\} \right] \\ &= \frac{V}{L \sqrt{1 - \xi^2}} e^{-\xi \omega_0 t} \left[\sqrt{1 - \xi^2} \cos(\omega_0 \sqrt{1 - \xi^2} t) \right. \\ &\quad \left. - \xi \sin(\omega_0 \sqrt{1 - \xi^2} t) \right] \\ &= \frac{V}{L \sqrt{1 - \xi^2}} e^{-\xi \omega_0 t} \cos \left\{ (\omega_0 \sqrt{1 - \xi^2} t + \theta) \right\} \end{aligned}$$

where

$$\theta = \tan^{-1} \left(\frac{\sqrt{1 - \xi^2}}{\xi} \right)$$

$$\begin{aligned} \frac{N_1(s)}{M_2(s)} \\ \frac{I_1(s)}{V_2(s)} \end{aligned}$$

R-L-C parallel circuit with step input



By KCL

$$I(t) = \frac{V(t)}{R} + C \frac{dV(t)}{dt} + \frac{1}{L} \int V(t) dt$$

On taking Laplace Transform

$$\frac{V(s)}{R} + SC V(s) + \frac{1}{SL} V(s) = \frac{I}{s}$$

$$V(s) = \frac{\frac{I}{s}}{s^2 + \frac{1}{RC}s + \frac{1}{LC}}$$

Roots of denominator equation are

$$s^2 + \frac{1}{RC}s + \frac{1}{LC} = 0$$

$$s_1 = -\frac{1}{2RC} + \sqrt{\frac{1}{4R^2C^2} - \frac{1}{LC}}$$

$$s_2 = -\frac{1}{2RC} - \sqrt{\frac{1}{4R^2C^2} - \frac{1}{LC}}$$

$$\text{Let } \omega_0 = \frac{1}{\sqrt{LC}} \quad \xi \omega_0 = \frac{1}{2RC}$$

$$\text{i.e. } \xi = \frac{1}{2R} \sqrt{\frac{1}{LC}} = \text{damping ratio}$$

$$s_1 = -\xi \omega_0 + \omega_0 \sqrt{\xi^2 - 1}$$

$$s_2 = -\xi \omega_0 - \omega_0 \sqrt{\xi^2 - 1}$$

$$V(s) = \frac{\frac{I}{s}}{(s-s_1)(s-s_2)} = \frac{A}{s-s_1} + \frac{B}{s-s_2}$$

$$A = \left. \frac{\frac{I}{s}}{s-s_2} \right|_{s=s_1} = \left. \frac{\frac{I}{s}}{s_1-s_2} \right|_{s=s_1} = \frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}}$$

$$B = \left. \frac{\frac{I}{s}}{s-s_1} \right|_{s=s_2} = \left. \frac{\frac{I}{s}}{s_2-s_1} \right|_{s=s_2} = \frac{-I}{2\omega_0 C \sqrt{\xi^2 - 1}}$$

$$\frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}} \left[\frac{1}{s - s_1} - \frac{1}{s - s_2} \right]$$

taking I. L.T.

$$= \frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}} [e^{s_1 t} - e^{s_2 t}]$$

$$\frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} [e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t}]$$

depending upon the value of R, L, C three cases may appear

Case I Overdamped Condition ($\xi > 1$)

$$v(t) = \frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} [e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t}]$$

$$= \frac{I}{\omega_0 C \sqrt{\xi^2 - 1}} (\sinh(\omega_0 \sqrt{\xi^2 - 1}) t)$$

Case II Critically damped Condition ($\xi = 1$)

$$V(s) = \frac{I/c}{s^2 + 2\omega_0 s + \omega_0^2}$$

$$= \frac{I/c}{(s + \omega_0)^2}$$

$$v(t) = \frac{I}{c} t e^{-\omega_0 t}$$

Case III Underdamped condition ($\xi < 1$)

$$v(t) = \frac{I}{2\omega_0 C \sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} [e^{(\omega_0 \sqrt{\xi^2 - 1})t} - e^{-(\omega_0 \sqrt{\xi^2 - 1})t}]$$

$$= \frac{I}{\omega_0 C \sqrt{1 - \xi^2}} e^{-\xi\omega_0 t} \left[\frac{e^{(j\omega_0 \sqrt{1 - \xi^2})t} - e^{-(j\omega_0 \sqrt{1 - \xi^2})t}}{2j} \right]$$

$$= \frac{I}{\omega_0 C \sqrt{1 - \xi^2}} e^{-\xi\omega_0 t} \sin(\omega_0 \sqrt{1 - \xi^2} t)$$

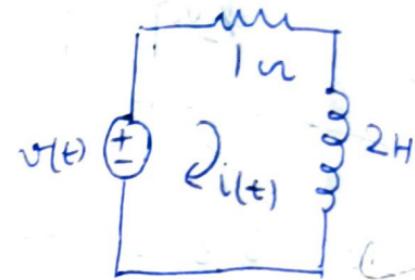
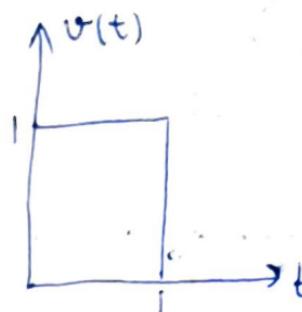
$$\frac{1}{2(s)}$$

s)

$$\begin{aligned} s) &= \frac{N_1(s)}{M_2(s)} \\ &= \frac{N_1(s)}{N_2(s)} \end{aligned}$$

by
"

- (a) Obtain the L.T. of the pulse shown
 (b) In fig (b) If $v(t)$ is the pulse (a) determine



$$v(t) = g_{0,1}(t) = v(t) - v(t-1)$$

$$V(s) = \frac{1}{s} - \frac{1}{s} e^{-s} = \frac{1}{s} (1 - e^{-s})$$

In fig (b) \rightarrow

$$V(t) = 1, i(t) + 2 \frac{di(t)}{dt}$$

$$V(s) = (1 + 2s) I(s)$$

$$I(s) = \frac{V(s)}{1 + 2s}$$

If $v(t)$ is $v(t)$ then

$$I_1(s) = \frac{1}{s \cdot 2(s + \frac{1}{2})} = \frac{1}{s} - \frac{1}{s + \frac{1}{2}}$$

$$i_1(t) = \left(1 - e^{-\frac{1}{2}t}\right) v(t)$$

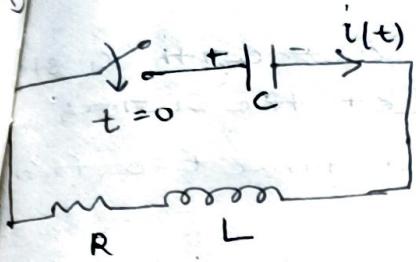
If $v(t)$ is $v(t-1)$ then

$$I_2(s) = \frac{1}{s} e^{-s} \left(\frac{1}{2(s + \frac{1}{2})} \right)$$

$$= e^{-s} \left[\frac{1}{s} - \frac{1}{s + \frac{1}{2}} \right]$$

$$i_2(t) = \left(1 - e^{-(t-\frac{1}{2})}\right) v(t-1)$$

$$i(t) = i_1(t) - i_2(t)$$



$L = 2H, R = 12\Omega, C = 62.5 \text{ mF}$.
Initial condition are $v_C(0^+) = 100V$
 $i_L(0^+) = 1A$. The switch is closed
at $t=0$. find $i(t)$

Sol. Apply KVL

$$L \frac{di(t)}{dt} + R i(t) + \frac{1}{C} \int_{0^+}^t i(t) dt + v_C(0^+) = 0$$

On taking L.T.

$$L [SI(s) - i_L(0^+)] + RI(s) + \frac{1}{Cs} I(s) + \frac{v_C(0^+)}{s} = 0$$

$$2 [SI(s) - 1] + 12I(s) + \frac{1}{(62.5 \times 10^{-3})s} I(s) + \frac{100}{s} = 0$$

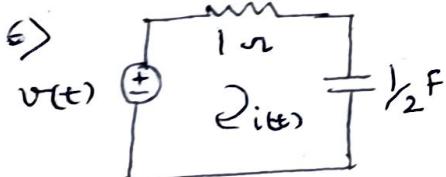
$$I(s) \left[2s + 12 + \frac{16}{s} \right] = -\frac{100}{s} + 2$$

$$I(s) = \frac{-100 + 2s}{2s^2 + 12 + 16} = \frac{-50 + s}{s^2 + 6s + 9} = \frac{s - 50}{(s + 4)(s + 2)}$$

$$I(s) = \frac{27}{s+4} - \frac{26}{s+2}$$

$$i(t) = 27e^{-4t} - 26e^{-2t}$$

If $v(t) = 2e^{-5t}v(t)$, $v_C(0^+) = 0$
find $i(t)$



$$2e^{-5t}v(t) = R i(t) + \frac{1}{C} \int_{-\infty}^t i(t) dt$$

$$2e^{-5t}v(t) = i(t) + 2 \int_0^t i(t) dt + v_C(0^+)$$

$$\frac{2}{s+5} = I(s) + \frac{2}{s} I(s) \neq 0$$

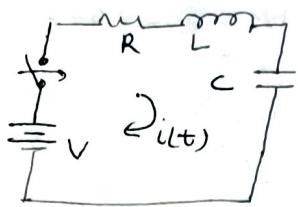
$$I(s) = \frac{2s}{(s+2)(s+5)} = \frac{8/3}{s+2} - \frac{2/3}{s+5}$$

$$i(t) = \frac{8}{3} e^{-2t} - \frac{2}{3} e^{-5t}$$

$$i(t) = \frac{2}{3} (4e^{-2t} - e^{-5t}) v(t) A$$

Q1) (a) Define step response.

7)



series RLC circ. with
Input V. Let the switch $I(s) =$
be closed at time $t=0$

$$Ri + \frac{Ldi}{dt} + \frac{1}{C} \int_0^t idt + v_c(0^+) = V_i(t)$$

$$RI(s) + L[sI(s) - i_L(0^+)] + \frac{1}{Cs} I(s) + \frac{v_c(0^+)}{s} = V_i(t)$$

at $t=0$ there is no stored energy in the circ.
hence $i_L(0^+) = 0$ $v_c(0^+) = 0$. eq'n (2) reduces to

$$(R + sL + \frac{1}{Cs}) I(s) = \frac{V_i(t)}{s}$$

$$I(s) = \frac{\frac{V_i(t)}{s}}{s^2 + R/Ls + 1/LC}$$

The roots of denominator on R.H.S.

$$s_2, s_1 = -\frac{R}{2L} \pm \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$\text{Let } \omega_0 = \frac{1}{\sqrt{LC}}$$

$$\xi \omega_0 = \frac{R}{2L}$$

$$s_1 = -\xi \omega_0 \pm \omega_0 \sqrt{\xi^2 - 1}$$

$$s_2 = -\xi \omega_0 - \omega_0 \sqrt{\xi^2 - 1}$$

$$I(s) = \frac{\frac{V_i(t)}{s}}{(s-s_1)(s-s_2)}$$

$$A = \frac{\frac{V_i(t)}{s}}{s_1 - s_2} = \frac{A}{s-s_1} + \frac{B}{s-s_2}$$

$$B = \frac{\frac{V_i(t)}{s}}{s_2 - s_1}$$

$$A = \frac{\frac{V_i(t)}{s}}{2\omega_0 \sqrt{\xi^2 - 1}} = -B$$

A is found by multiplying
among $s-s_1$ and putting $s=s_1$

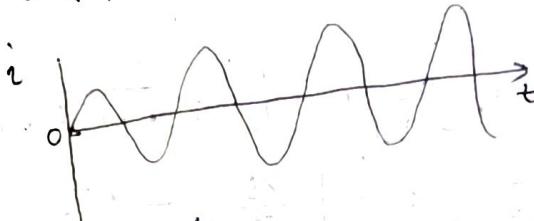
$$I(s) = \frac{V}{2L\omega_0\sqrt{\xi^2 - 1}} \left[\frac{1}{s-s_1} + \frac{1}{s-s_2} \right]$$

$$i(t) = \frac{V}{2L\omega_0\sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} \left[e^{\omega_0\sqrt{\xi^2 - 1}t} - e^{-\omega_0\sqrt{\xi^2 - 1}t} \right]$$

There may be 3 cases depending on value of ξ

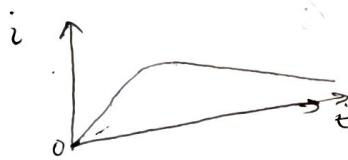
case 1 when $\xi > 1 \Rightarrow \frac{R^2}{4L^2} > \frac{1}{LC} \Rightarrow$ Overdamped case

$$i(t) = \frac{V}{\omega_0 L \sqrt{\xi^2 - 1}} e^{-\xi\omega_0 t} \sinh \omega_0 \sqrt{\xi^2 - 1} t$$



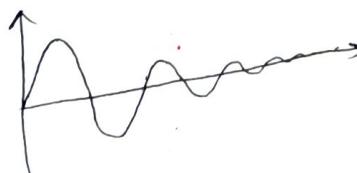
case 2 when $\xi = 1 \Rightarrow \frac{R^2}{4L^2} = \frac{1}{LC}$ = critically damped

$$i(t) = \frac{V}{L} t e^{-\omega_0 t}$$



Case 3 when $\xi < 1 \Rightarrow \frac{R^2}{4L^2} < \frac{1}{LC}$, Underdamped case

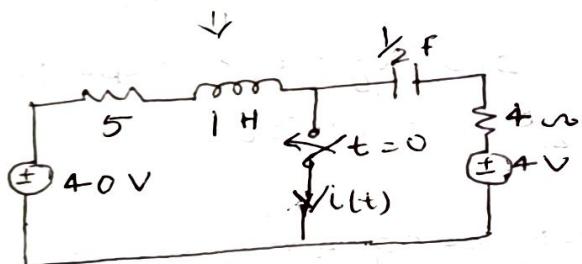
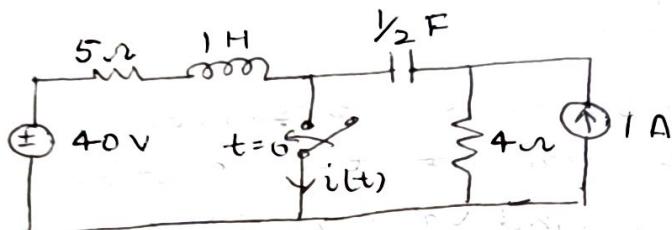
$$i(t) = \frac{V}{\omega_0 L \sqrt{1-\xi^2}} e^{-\xi\omega_0 t} \sin \omega_0 \sqrt{1-\xi^2} t$$



Transf transfer functions \Rightarrow

$$H(s) = \frac{Y(s)}{X(s)} \quad | \text{ all initial conditions are zero.}$$

9>



at s.s. inductor behave as a s.c. and capacitor as a o.c. so $i_1(0^-) = 0$ $v_c(0^-) = 40 - 4 = 36$

$i_1(t) \rightarrow$ ct. due to 40V source
 $i_2(t) \rightarrow$ " " " 4V source

$$i(t) = i_1(t) + i_2(t)$$

apply KVL

$$40 = 5i_1(t) + \frac{di_1(t)}{dt}$$

$$\frac{40}{s} = (5 + s) I_1(s)$$

$$I_1(s) = \frac{\frac{40}{s}}{s(s+5)} = 8 \left[\frac{1}{s} - \frac{1}{s+5} \right]$$

$$i_1(t) = 8(1 - e^{-5t})$$

$$4 = 4i_2(t) + 2 \int_0^t i_2(t) dt - 36$$

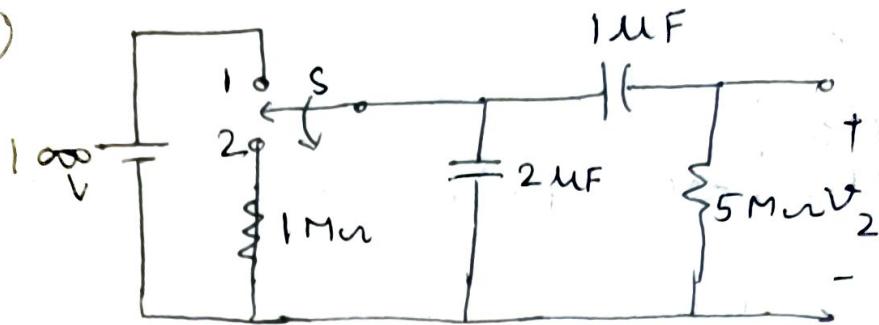
$$\frac{40}{s} = \left(4 + \frac{2}{s}\right) I_2(s)$$

$$I_2(s) = \frac{\frac{40}{s}}{4s+2} = \frac{10}{s+5}$$

$$i_2(t) = 10e^{-5t}$$

$$i(t) = 8(1 - e^{-5t}) + 10e^{-5t}$$

(13)

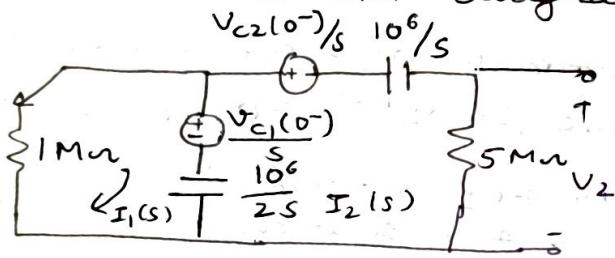


When switch S is in position 1, circuit attains equilibrium at $t=0$. At $t=0$ switch is moved to position 2. Find V_2 a/c $5\text{ M}\Omega$ resistor.

Sol. at position 1 s.s. V_2 a/c capacitor

$$V_{C1}(0^-) = V_{C2}(0^-) = 1000\text{V}$$

at position 2 ckt diagram is as follow



$$10^6 I_1(s) + \frac{1000}{s} + \frac{10^6}{2s} [I_1(s) - I_2(s)] = 0 \quad \text{---(1)}$$

$$I_1(s) = (10s + 3)s I_2(s)$$

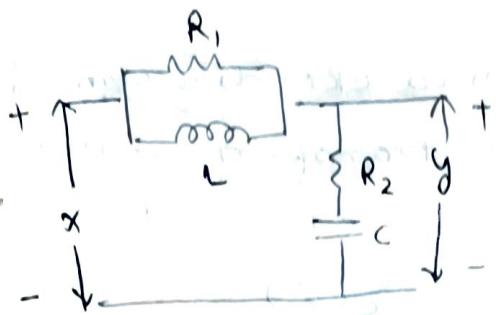
$$\frac{1000}{s} = \frac{1000}{s} + \frac{10^6}{s} I_2(s) + 5 \times 10^6 I_2(s) + \frac{10^6}{2s} [I_2(s) - I_1(s)]$$

from (1) - & (2)

$$I_2(s) = -\frac{1000}{10^6 [10s^2 + 8s + 1]}$$

$$V_2(s) = 5 \times 10^6 I_2(s).$$

$$V_2(t) = 1020 (e^{-645t} - e^{-155t}) \text{V.}$$



Obtain T.F.
 Draw the block diagram
 representation of fig.
 $x \rightarrow$ I/P Variable
 $y \rightarrow$ O/P Variable

$$X(s) - Y(s) = I(s)[R_1 + Ls]$$

$$X(s) - Y(s) = I(s) \cdot \frac{R_1 s}{s + R_1/L}$$

$$Y(s) = I(s) [R_2 + 1/Cs]$$

$$= I(s) \cdot \frac{s + 1/R_2 C}{1/R_2 s}$$

$$H(s) = \frac{Y(s)}{X(s)}$$

$$\text{II. } \frac{C(s)}{R(s)} = \frac{(s+1)(s+3)}{(s+2)(s^2 + 8s + 32)}$$

Determine time response $c(t)$ for unit step I/P

$$\underline{\text{Sol.}} \quad R(s) = 1/s$$

$$C(s) = \frac{(s+1)(s+3)}{s(s+2)(s^2 + 8s + 32)}$$

$$= \frac{A}{s} + \frac{B}{s+2} + \frac{Cs + D}{s^2 + 8s + 32}$$

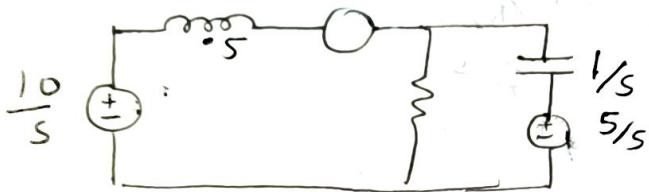
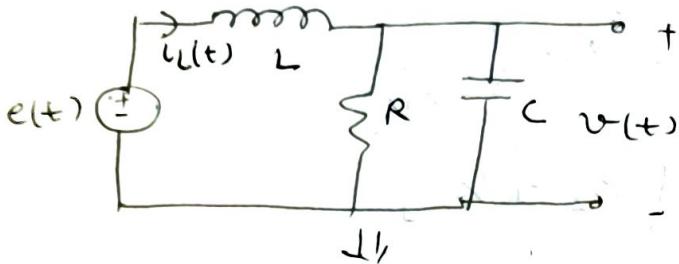
$$A = 3/64 \quad B = 1/40 \quad C = -23/230 \quad D = 19/40$$

$$c(s) = \frac{3}{64} \cdot \frac{1}{s} + \frac{1}{40} \cdot \frac{1}{s+2} - \frac{\frac{23}{320} s + \frac{19}{40}}{(s^2 + 4)^2 + 4^2}$$

$$= \frac{3}{64} + \frac{1}{40} e^{-2t} - \frac{23}{320} e^{-4t} \cos 4t + \frac{61}{320} e^{-4t} \sin 4t$$

12) Draw the transformed ckt. diagram and obtain appropriate transformed nodal equation. Given $R = \frac{1}{3} \Omega$ & $L = 1.5 \text{ H}$ $C = 1 \text{ F}$
 $e(t) = 10 \text{ V}$, $i_L(0^-) = 15 \text{ A}$ $v_C(0^-) = 5 \text{ V}$

Solve for $v(t)$



13) If $R = 1 \Omega$ & $C = 1 \text{ F}$, $v_C(0^-) = 0 \text{ V}$
 find response $v(t)$ when $i(t)$ is (a) Unit impulse fn (b) Unit step fn

$$\text{Circuit diagram: } i(t) \rightarrow \left\{ \begin{array}{l} R \\ \text{---} \\ C \end{array} \right. \quad v(t) \quad \text{V}(s) = I(s) [1 + \frac{1}{s}] = \frac{I(s)}{s+1}$$

$$i(t) = \delta(t)$$

$$I(s) = 1$$

$$\text{V}(s) = \frac{1}{s+1} \quad v(t) = e^{-t} v(t)$$

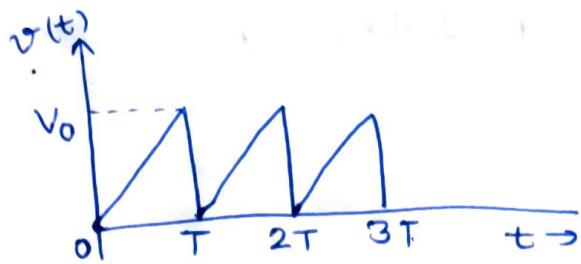
If $i(t) = v(t)$ i.e.

$$I(s) = \frac{1}{s}$$

$$\text{V}(s) = \frac{1}{s(s+1)} = \frac{1}{s} - \frac{1}{s+1}$$

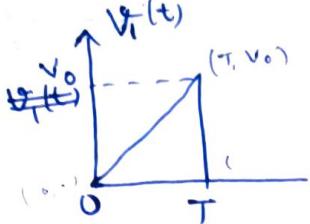
$$v(t) = (1 - e^{-t}) v(t)$$

Practice Questions



find the Laplace Transform

Sol.



Let $v_1(t)$ is the first cycle of given w/f.

$$v_1(t) = \frac{V_0}{T} t + C_{0,T}(t)$$

$$= \frac{V_0}{T} t [u(t) - u(t-T)]$$

$$= \frac{V_0}{T} t u(t) - \frac{V_0}{T} (t-T) u(t-T) - V_0 u(t-T)$$

$$V_1(s) = \frac{\frac{V_0}{T} s^2}{s^2} - \frac{\frac{V_0}{T} s^2}{s^2}$$

$$V_1(s) = \frac{V_0}{T} \cdot \frac{1}{s^2} - \frac{V_0}{T} \cdot \frac{1}{s^2} e^{-Ts} + \frac{V_0}{s} e^{-Ts}$$

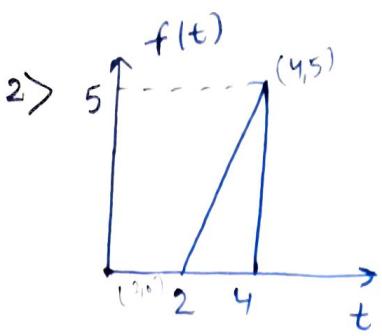
$$= \frac{V_0}{T} \left[\frac{1}{s^2} - \frac{e^{-Ts}}{s^2} \right] - \frac{V_0}{s} e^{-Ts}$$

By using theorem for periodic function

$$V(s) = \frac{1}{1-e^{-Ts}} V_1(s) =$$

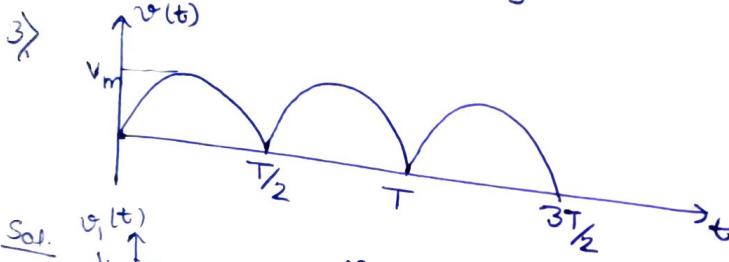
$$= \frac{1}{1-e^{-Ts}} \left[\frac{V_0}{Ts^2} \left[1 - e^{-Ts} \right] - \frac{V_0}{s} e^{-Ts} \right]$$

$$= \frac{V_0}{Ts^2} - \frac{V_0}{s(1-e^{-Ts})} e^{-Ts}$$



find the Laplace transform

$$\begin{aligned}
 \text{Sol. } f(t) &= \frac{5-0}{4-2} (t-2) [u(t-2) - u(t-4)] \\
 &= \frac{5}{2} (t-2) [u(t-2) - u(t-4)] \\
 &= \frac{5}{2} (t-2) u(t-2) - \frac{5}{2} (t-4+2) [u(t-4)] \\
 &= \frac{5}{2} u(t-2) - \frac{5}{2} (t-4) u(t-4) - 5 u(t-4) \\
 &= \frac{5}{2s^2} e^{-2s} - \frac{5}{2s^2} e^{-4s} - \frac{5e^{-4s}}{s}
 \end{aligned}$$

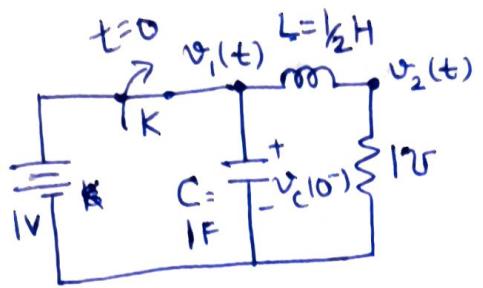


Sol.

$$\begin{aligned}
 v_1(t) &= V_m \sin \omega t [u(t) - u(t - \frac{T}{2})] \\
 &= V_m [\sin \omega t u(t) - \sin \omega t u(t - \frac{T}{2})] \\
 &= V_m [\sin \omega t u(t) - \sin \left\{ \frac{2\pi}{T} (t - \frac{T}{2}) + \pi \right\} u(t - \frac{T}{2})] \\
 &= V_m [\sin \omega t u(t) + \sin \left\{ \frac{2\pi}{T} (t - \frac{T}{2}) \right\} u(t - \frac{T}{2})] \\
 v_1(s) &= V_m \left[\frac{\omega}{s^2 + \omega^2} + e^{-\frac{T}{2}s} \cdot \frac{\omega}{s^2 + \omega^2} \right]
 \end{aligned}$$

By using theorem for periodic function

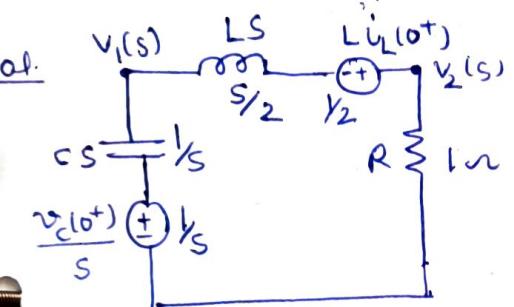
$$V(s) = \frac{1}{1 - e^{-\frac{T}{2}s}} \cdot \frac{V_m \omega}{s^2 + \omega^2} (1 + e^{-\frac{T}{2}s})$$



Switch K is opened at $t = 0$. The circuit was in steady state before the switch is opened.

Find the node voltage $v_1(t)$ and $v_2(t)$. Also draw the transformed circuit.

Initial conditions are $v_C(0^-) = 1V$ $i_L(0^-) = 1A$



$$\text{By KVL}$$

$$\frac{1}{S} + \frac{1}{2} = I(s) \left[\frac{1}{S} + \frac{S}{2} + 1 \right]$$

$$I(s) = \frac{S+2}{S^2+2S+2}$$

$$V_1(s) = \frac{1}{S} - \frac{1}{S} I(s)$$

$$= \frac{1}{S} \left[1 - \frac{S+2}{S^2+2S+2} \right] = \frac{S^2+S}{S(S^2+2S+2)}$$

$$\therefore \frac{S+1}{S^2+2S+2} = \frac{S+1}{(S+1)^2 + (1)^2}$$

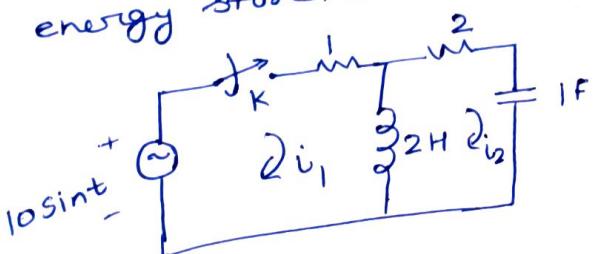
$$v_1(t) = e^{-t} \cos t$$

$$V_2(s) = 1 \cdot I(s) = \frac{S+2}{S^2+2S+2} = \frac{(S+1) + 1}{(S+1)^2 + (1)^2}$$

$$= e^{-t} \cos t + e^{-t} \sin t$$

$$= e^{-t} [\cos t + \sin t]$$

Eg. Determine the steady state mesh current i_1 and i_2 in the given fig. There is no initial energy stored in the circuit.



Sol. Switch k is closed at $t=0$

By KVL

$$10 \sin t = i_1(t) + 2 \left[\frac{di_1}{dt} - \frac{di_2}{dt} \right]$$

$$0 = 2i_2(t) + \frac{1}{1} \int_{-\infty}^t i_2(t) dt + 2 \left[\frac{di_2}{dt} - \frac{di_1}{dt} \right]$$

as there is no initial energy stored in i_1

By E.T.

$$\frac{10}{s^2+1} = I_1(s) + 2sI_1(s) - 2sI_2(s)$$
$$= I_1(s) [1+2s] - 2sI_2(s) \quad \text{---(1)}$$

$$0 = 2I_2(s) + \frac{I_2(s)}{s} + 2sI_2(s) - 2sI_1(s)$$
$$= -2sI_1(s) + I_2(s) \left[2 + \frac{1}{s} + 2s \right] \quad \text{---(2)}$$

On Solving (1) & (2) we have

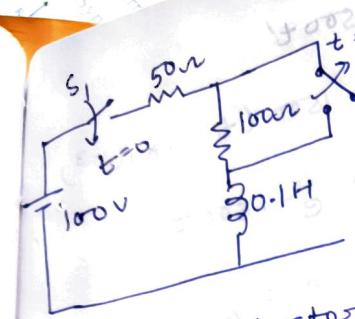
$$I_1(s) = \frac{10(s^2+2s+1)}{(s^2+1)(6s^2+4s+1)}$$

$$I_2(s) = \frac{20s^2}{(s^2+1)(6s^2+4s+1)}$$

S.S. value of i_1 and i_2 is given as

$$i_1(\infty) = \lim_{s \rightarrow 0} sI_1(s) = 0$$

$$i_2(\infty) = \lim_{s \rightarrow 0} sI_2(s) = 0$$



In the given fig. switch S_1 is closed at $t = 0$ and S_2 is opened at $t = 4$ msec. Determine $i(t)$ for $t \geq 0$. (Initially dc-energised)

(Assume inductor is initially de-energised)

$$\text{By KVL} \\ 100 = 50i(t) + 0.1 \frac{di(t)}{dt}$$

$$\text{by L.T.} \\ \frac{100}{s} = 50 I(s) + 0.1 s I(s)$$

$$I(s) = \frac{100}{s(50 + 0.1s)} = \frac{1000}{s(s+500)}$$

$$i(t) = \mathcal{L}^{-1}[I(s)] = \mathcal{L}^{-1}\left[\frac{2}{s} - \frac{2}{s+500}\right] \\ = 2(1 - e^{-500t}) \text{ A}$$

for $t = 4 \times 10^{-3} \text{ sec}$

$$i(4 \times 10^{-3}) = 1.729 \text{ A}$$

for $4 \text{ msec} \leq t < \infty$ If $t' = t - 4 \times 10^{-3}$

then $0 \leq t' < \infty$

$$\text{By KVL} \\ 100 = 150 i(t) + 0.1 \frac{di(t)}{dt}$$

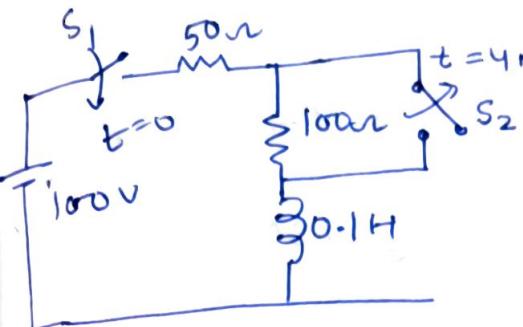
By L.T.

$$\frac{100}{s} = 150 I(s) + 0.1 [s I(s) - i(4 \times 10^{-3})]$$

$$\Rightarrow \frac{100}{s} = (150 + 0.1s) I(s) - 0.1 \times 1.729$$

$$I(s) = \frac{100 + 0.1729s}{s(0.1s + 150)} = \frac{1.729s + 1000}{s(s + 1500)}$$

$$I(s) = \frac{667}{s} + 1.062$$



In the given fig. switch S_1 is closed at $t=0$ and S_2 is opened at $t=4$ msec. Determine $i(t)$ for $t > 0$

(Assume inductor is initially dc-energised)

for $0 \leq t \leq 4$ msec. ; (S_1 is closed & S_2 is closed)

By KVL

$$100 = 50 i(t) + 0.1 \frac{di(t)}{dt}$$

by L.T.

$$\frac{100}{s} = 50 I(s) + 0.1 s I(s)$$

$$i(0+) = 0$$

$$I(s) = \frac{100}{s(50 + 0.1s)} = \frac{1000}{s(s+500)}$$

$$i(t) = \mathcal{L}^{-1}[I(s)] = \mathcal{L}^{-1}\left[\frac{2}{s} - \frac{2}{s+500}\right]$$

$$= 2(1 - e^{-500t}) A$$

for $t = 4 \times 10^{-3}$ sec

$$i(4 \times 10^{-3}) = 1.729 A$$

for $4 \text{ msec} \leq t < \infty$ If $t' = t - 4 \times 10^{-3}$

then $0 \leq t' < \infty$

$$By KVL \rightarrow 100 = 150 i(t) + 0.1 \frac{di(t)}{dt}$$

$$100 = 150 I(s) + 0.1 [s I(s) - i(4 \times 10^{-3})]$$

$$\frac{100}{s} = (150 + 0.1s) I(s) - 0.1 \times 1.729$$

$$\Rightarrow \frac{100}{s} = \frac{100 + 0.1729s}{s(0.1s + 150)} = \frac{1.729s + 1000}{s(s + 1500)}$$

$$I(s) = \frac{100}{s(0.1s + 150)}$$

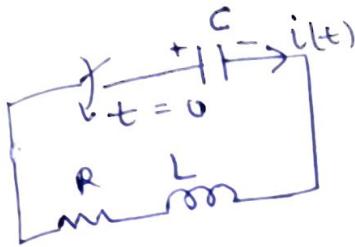
$$I(s) = \frac{667}{s} + \frac{1.062}{s + 1500}$$

$$i(t') = .667 + 1.062 e^{-1500 t'}$$

$$i(t) = .667 + 1.062 e^{-1500 (t - 4 \times 10^{-3})}$$

$$= .667 + 1.062 \cdot e^6 e^{-1500 t}$$

$$= .667 + 428.4 e^{-1500 t} A$$



$$L = 2H \quad R = 12\Omega \quad C = 62.5 \text{ mF}$$

Initial conditions are

$$v_C(0^+) = 100V \quad i_L(0^+) = 1A$$

Switch is closed at $t=0$. find $i(t)$

By KVL

$$Ri(t) + L \frac{di}{dt} + \frac{1}{C} \int i(t) dt + v_C(0^+) = 0$$

on taking L.T.

$$RI(s) + L[sI(s) - i_L(0^+)] + \frac{1}{Cs} I(s) + \frac{v_C(0^+)}{s} = 0$$

$$12I(s) + 2sI(s) - 2 \cdot 1 + \frac{1}{62.5 \times 10^{-3}} s I(s) + \frac{100}{s} = 0$$

$$I(s) \left[2s + 12 + \frac{16}{s} \right] = \frac{-100}{s} + 2$$

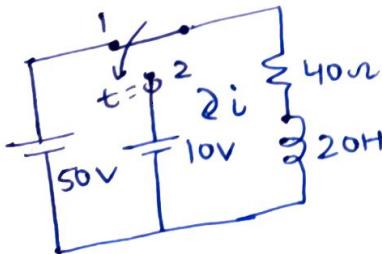
$$I(s) = \frac{-100 + 2s}{2s^2 + 12s + 16}$$

$$= \frac{s - 50}{s^2 + 6s + 8} = \frac{s - 50}{(s+4)(s+2)}$$

$$I(s) = \frac{27}{s+4} - \frac{26}{s+2}$$

$$i(t) = 27e^{-4t} - 26e^{-2t}$$

3)



The switch has been in position 1 for a long time. It is moved to 2 at $t=0$. Obtain expression for i for $t > 0$ using L.T.

When switch was on 1

$$50 = 40i(t) + 20 \frac{di}{dt}$$

$$50 = 40I'(s) + 20 \cdot s I'(s) - i(0^+)$$

Sol:

$$i(0^+) = 0$$

$$\begin{aligned} I'(s) &= \frac{50}{s(40+20s)} \\ &= \frac{2.5}{s(s+2)} \end{aligned}$$

$$i'(t) = 1.25 (1 - e^{-2t})$$

as $t \rightarrow \infty$

$$i'(0^\infty) = 1.25$$

when switch is on 2

$$10 = 40 i(t) + \frac{20 di}{dt}$$

$$\frac{10}{s} = 40 I(s) + 20 [s I(s) - i(0^+)]$$

$$\frac{10}{s} = (40 + 20s) I(s) - 20 \times 1.25$$

$$I(s) = \frac{10 + 25s}{s(40 + 20s)}$$

$$= \frac{1.25 (s+0.4)}{s(s+2)}$$

$$i(t) = (0.25 + e^{-2t}) A$$