FeNN: A RISC-V vector processor for Spiking Neural Network acceleration

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Abstract—Spiking Neural Networks (SNNs) have the potential to drastically reduce the energy requirements of AI systems. However, mainstream accelerators like GPUs and TPUs are designed for the high arithmetic intensity of standard ANNs so are not well-suited to SNN simulation. FPGAs are wellsuited to applications with low arithmetic intensity as they have high off-chip memory bandwidth and large amounts of on-chip memory. Here, we present a novel RISC-V-based soft vector processor (FeNN), tailored to simulating SNNs on FPGAs. Unlike most dedicated neuromorphic hardware, FeNN is fully programmable and designed to be integrated with applications running on standard computers from the edge to the cloud. We demonstrate that, by using stochastic rounding and saturation, FeNN can achieve high numerical precision with low hardware utilisation and that a single FeNN core can simulate an SNN classifier faster than both an embedded GPU and the Loihi neuromorphic system.

Index Terms—SNN, FPGA, RISC-V

I. Introduction

Artificial Neural Networks (ANNs) have demonstrated super-human performance in areas ranging from image classification to playing go [1]. However, training and using ANNs is computationally very expensive and only possible due to modern hardware and software. The primary computational primitive ANNs require is matrix multiplication, which has a very high arithmetic intensity. Numerous accelerators including GPUs and TPUs have been designed to accelerate matrix multiplication, typically by streaming data from high bandwidth external memory into systolic arrays of multiply-accumulate units. However, high arithmetic intensity leads to high energy requirements, representing a growing problem for the AI industry. At the same time, the *biological* neural network of the human brain can still outperform ANNs in many tasks while only consuming 20 W.

While ANNs are *inspired* by the brain, they differ in several key areas, one of which is that biological neurons communicate with sparse binary events called *spikes*. This has been incorporated in artificial Spiking Neural Networks (SNNs) which have already been successful in spatio-temporal processing applications such as audio denoising and autonomous driving [2]. As spikes carry no 'value' to multiply, simulating SNNs does not require matrix multiplication and thus has a

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much lower arithmetic intensity. Bautembach et al. [3], Knight and Nowotny [4] previously showed that SNNs can be simulated effectively on GPUs. However, the low arithmetic intensity of SNNs means that SNN simulations can easily saturate a GPU's memory bandwidth but leave its compute resources under-utilised. Furthermore, while low-precision weights have been widely adopted in GPU designs to reduce memory bandwidth requirements, they are typically handled by the systolic MAC arrays which do not support the operations required for simulating SNNs at low precision [5, 6]. Finally, because simulating SNNs is an iterative process and GPUs have limited capacity for global synchronisation, several kernels typically are launched every simulation timestep causing a significant overhead (order of 10 µs).

Digital neuromorphic systems [7, 8] are tailored to the low arithmetic intensity of SNN simulations, typically only having modest computational resources but large amounts of fast local memory. These systems are very efficient but typically not flexible enough for developing *new* algorithms. Furthermore, neuromorphic systems are not available off the shelf and unlike GPUs cannot easily be integrated with applications running on standard computers as they are typically standalone systems. Field-Programmable Gate Arrays (FPGAs) represent an interesting middle ground as they can be integrated seamlessly with standard computers and not only offer high external memory bandwidth similar to GPUs but also large amounts of on-chip memory that can be accessed in a single cycle (around 40 MiB on a high-end Alveo U55C). Although FPGAs also have DSP cores to enable single-cycle integer multiplication, they run at relatively low clock speeds (typically a few 100 MHz) so cannot compete with GPUs for arithmetically-intensive tasks. This has meant that they have thus far only been used for inference with standard ANNs [9]. However, FPGAs are well-suited to SNN simulations because of their lower arithmetic intensity. Over the years, there has been a plethora of fixed-function FPGA SNN accelerators which demonstrate this [10, 11]. However, such systems have a high entry barrier as they require re-synthesis using FPGA tools and potentially even writing HDL if the emulated model changes. Naylor et al. [12] demonstrated that a fully programmable vector soft-core processor running on an Altera Stratix IV FPGA could simulate SNNs faster than the then state-of-the-art, and more recent programmable systems [13, 14] have shown that this approach continues to be competitive on newer FPGA

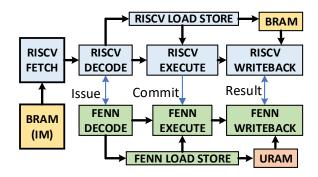


Fig. 1. Block diagram of FeNN, tightly coupled to CV32E40X RISC-V core. Black arrows indicate pipeline stages and key signals. Blue arrows highlight the 'Issue', 'Commit' and 'Result' signals defined by the CV32E40X's XIF extension interface.

architectures like AMD's UltraScale+.

Here, we present our FPGA-Enhanced Neural Network (FeNN) accelerator: a vector soft-processor, tailored to the needs of SNN simulation and the hardware resources available on modern FPGAs. Uniquely, it is designed to operate as an in-memory accelerator, capable of scaling from embedded systems based on Zynq MPSoCs to workstations or cloud nodes equipped with large accelerators such as the Xilinx Alveo U55C. We show that, by using stochastic rounding and saturation, complex spiking neuron models can be simulated accurately using 16 bit fixed-point arithmetic and demonstrate a recurrent SNN that classifies spoken digits [15] on FeNN. This classifier runs significantly faster than a similar model running on Loihi and twice as fast as the same model running on an embedded GPU while using half the energy.

II. METHODS

A. Scalar core

Soft vector processors can either be *tightly coupled*, where individual instructions are offloaded from the scalar processor to the vector processor or *decoupled*, where 'kernels' of vector instructions are assembled separately and copied to the accelerator. While tightly coupled architectures allow finegrained communication between scalar and vector cores, their performance can be restricted by the throughput of the scalar processor and each application must be compiled for the novel architecture [12]. However, decoupled vector processors have to implement the entire control flow for the desired algorithms which would introduce significant complexity.

We have developed a best-of-both-worlds architecture where our FeNN vector processor works as a decoupled, in-memory accelerator for a standard host CPU (a quad-core ARM Cortex-A53 on the Kria KV260) running Linux but is tightly coupled to its own scalar core. The scalar core is a CV32E40X developed by the Open Hardware Group and is a small and efficient, 32-bit, in-order RISC-V core with a 4-stage pipeline. Data and control hazards notwithstanding, this core can issue 1 instruction per cycle and provides a convenient extension interface upon which to implement our co-processor. In order

 $TABLE\ I$ Hardware utilisation. Bracketed values are per lane/PE.

Design	Lanes	Frequency	Utilisation	
		[MHz]	LUT	FF
SNAVA [14]	100	125	148 774 (1488)	97 824 (978)
GaBAN [13]	8	286	28 759 (3595)	9605 (1201)
FeNN	32	166	32 915 (1028)	32 547 (1017)

to make the CV32E40X more suitable for FPGA synthesis, we have simplified its fetch and load-store stages to work with synchronous Block RAMS – removing the need for any instruction prefetch or alignment logic.

B. 3-stage pipelined vector co-processor

While it is common for inference accelerators to use 8 bit or even 4 bit weights, we want to offer more flexibility and, in future, also accelerate SNN training workloads. Hence, our accelerator operates on 16 bit types throughout. Specifically, FeNN operates on vectors of 32 16 bit values using a SIMD architecture and a standard register-register instruction set, backed by a three-port vector register file with 32 512 bit registers, implemented using distributed memory. The majority of on-chip memory on UltraScale+ FPGAs is provided through 'wide' (72 bit) 'UltraRAM' memory which we use to implement local memories, large enough to store all model state in our current prototype. These memories are created using 8 parallel sets of chained UltraRAM blocks, enabling FeNN to load and store entire vectors in a single cycle.

In order to provide scope for further expansion, FeNN is implemented within a full 30 bit RISC-V instruction set quadrant (binary prefix 10). Within this quadrant, our initial FeNN design implements a minimal instruction set supporting:

- Addition, subtraction and multiplication of vectors.
- Loading and storing of vectors in UltraRAM.
- Moving data between scalar and vector registers.
- Generating random numbers.
- Computing masks by comparing vectors.
- Selecting between elements of two vectors using masks.

Multiplication is implemented using DSP blocks and, as these can operate at several times our operating frequency, this leaves time in the execute cycle to apply a 16 bit barrel shift to the result in order to implement fixed-point multiplication. Because FeNN has 32 vector lanes, the masks used for non-uniform control flow conveniently fit in a standard 32-bit RISC-V register. Thus, in our tightly-coupled architecture, masks can be processed using vector and scalar instructions, removing hardware requirements for separate mask registers.

1) Pipeline: Figure 1 illustrates how FeNN is pipelined alongside the CV32E40X with synchronisation occurring through the XIF extension interface. All instructions can be executed within a single execute clock cycle and we include additional logic between the writeback and execute stages to enable bypassing of RAW hazards between vector registers. UltraRAM resources are synchronous with a 1-cycle latency so addresses are calculated in the execute cycle and values

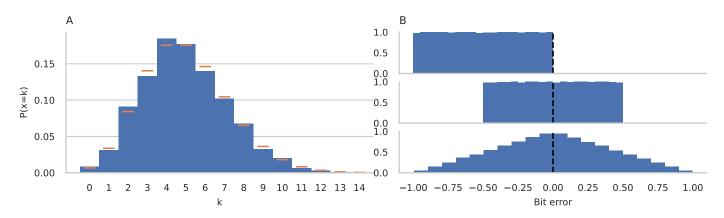


Fig. 2. (A) Histogram of 3200 Poisson variates sampled from distribution with $\lambda = 5$, generated on FeNN. Horizontal lines indicate Probability Mass Function of distribution. (B) Histograms showing the bit error distribution of 21 760 random S0.15 multiplication operations with (top) round-to-zero (middle) round-to-nearest and (bottom) stochastic rounding (after Hopkins et al. [6]).

read from memory are written back to the register file in the writeback cycle.

2) Random number generation: Efficient pseudorandom number generation is important in many SNN models as well as for stochastic rounding [6] so an efficient Random Number Generator (RNG) is a key requirement for FeNN. Because FeNN operates on 16 bit numbers, we require an RNG with a small state size that can generate 16 bit of randomness per cycle in each vector lane. Inspired by its use in the Propeller 2 microcontroller, we chose a Xoroshiro32++ generator [16] which is relatively high quality and extremely hardware friendly - requiring only a handful of addition, XOR, rotation and shift operations. However, if it were to be implemented as a standard RISC-V instruction - which reads at most two operands and writes back another - the instruction to generate a random number would need to read 32 bit and write 48 bit of state per cycle requiring extra register file ports. Furthermore, incorporating the RNG into a stochastic multiplication instruction would require reading two more operands and writing back yet another. Instead, we add two special two-port registers dedicated to holding the RNG state. These can be populated from vector memory using special variants of the vector load instruction.

3) Stochastic rounding: A standard fixed point multiplication is performed by calculating (A * B) >> N where A and B are two fixed-point encoded operands and N is the number of fractional bits in the representation. The right shift by N truncates the result leading to round-to-zero behaviour. However, Mikaitis [17] demonstrated that rounding can be implemented simply by turning the multiplication into a MAC (((A * B) + R) >> N) which is supported in hardware by the DSP blocks. Round-to-nearest can be implemented simply by setting R to 0.5 (in fixed-point with N fractional bits) and stochastic rounding by adding N bits of randomness (the *mean* of which will be 0.5) from the RNG.

C. System-on-Chip design

In our current prototype we have instantiated a single FeNN core on an AMD Kria KV260 development board. The data

and instruction BRAMs connected to the scalar RISC-V core are exposed to the ARM core via standard AXI BRAM controllers. Similarly, control and status registers are exposed via an AXI GPIO controller. These resources are then memory mapped by user-space applications running on the ARM core. Applications use a C++-based Just-in-Time assembler to generate FeNN instructions and copy them, alongside any initial state, to the memory-mapped BRAMs. FeNN is then reset and simulates multiple SNN timesteps while the host polls, waiting for FeNN to finish.

D. Adaptive Leaky Integrate-and-Fire neuron model

Standard Recurrent SNNs (RSNNs) using Leaky Integrateand-Fire (LIF) neurons have inferior short-term memory capacities compared to recurrent ANN architectures such as LSTMs. Adaptive LIF (ALIF) neurons – which augment a standard LIF neuron with a slowly adapting threshold – are one solution to this problem [18]. The dynamics of an ALIF neuron's membrane voltage V can be calculated using an Exponential Euler scheme,

$$V[t+1] = \alpha V[t] + I^{\text{syn}}[t+1] - S[t]V_{\text{th}}$$
 (1)

where $\alpha=e^{\frac{-1}{\tau_m}}$ controls the neuron's leak $(\tau_m=20)$, $I^{\rm syn}$ is the input current and $V_{\rm th}=0.6$ is the baseline firing threshold which is subtracted from V when a spike occurs. Spikes are triggered when the membrane voltage crosses the adaptive threshold:

$$S[t] = H(V[t] - (V_{th} + \beta A[t]))$$
 (2)

where $\beta=0.0174$ is the adaptation strength and H the Heaviside function. The adaptation A evolves as

$$A[t+1] = \rho A[t] + S[t] \tag{3}$$

where $\rho = e^{\frac{-1}{\tau_a}}$ controls the rate of adaptation ($\tau_a = 2000$).

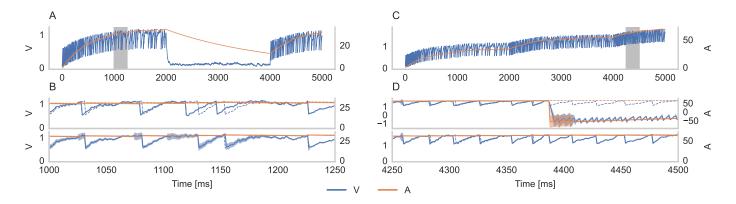


Fig. 3. ALIF neuron simulations (**A**) ALIF stimulated with two periods of Poisson spiking input, separated by a period of low background rate, simulated using 64 bit floating point (**B**) Detail from the shaded area in A) comparing 16 bit fixed point (solid lines) against 64 bit floating point (dashed lines) using standard round-to-zero (top) and stochastic rounding (bottom). (**C**) ALIF stimulated by Poisson spiking input with increasing rates, simulated using 64 bit floating point (**D**) Detail from C) comparing 16 bit fixed point (solid lines) against 64 bit floating point (dashed lines) using stochastic rounding (top) and stochastic rounding with saturation (bottom). Simulations with stochastic rounding were run 32 times and the standard deviations are shown as shaded areas.

III. RESULTS

All code used to generate the results in this section and a bitstream for programming a Kria KV260 with FeNN are available at https://github.com/neworderofjamie/riscv_ise.

A. Synthesis

We synthesised our design using Xilinx Vivado 2023.2. Table I compares hardware utilisation and clock speed with recent programmable FPGA-based SNN accelerators [13, 14]. Due to our focus on 16 bit fixed point, a FeNN core requires significantly fewer LUTs per-lane than other designs. On the other hand, FeNN's deeper pipeline means that it requires more FlipFlops than the other designs, but these are not the scarcest resource and its deeper pipeline means FeNN can process neurons in fewer clock cycles (approximately 20 compared to 85 for GaBAN).

B. Random number generation

In order to test the random number generator described in Section II-B2, we implemented a Poisson generator on FeNN using the direct method first described by Knuth [19] and sampled 3200 variates with $\lambda=5$. As Poisson processes are often used to encode stimuli in SNN models [20], this is a representative test of how random number generation is likely to be used on FeNN. Figure 2A shows that the sampled variates follow the desired distribution and our timing measurements show that, on average, the Poisson generator produces 32 variates every 81 clock cycles (corresponding to around $64 \times 10^6 \, \mathrm{s}^{-1}$).

C. Rounding and saturation

In order to test the correctness of the rounding mechanisms described in Section II-B3, we repeated the experiment described by Hopkins et al. [6] and performed 21 760 multiplication operations with round-to-zero, round-to-nearest and stochastic rounding on random 16 bit operands in S0.15 fixed-point format (S0.15 can only represent numbers with absolute

value < 1 so no overflow will occur) using FeNN. Figure 2B shows that the bit error distributions obtained by comparing the results to 64 bit floating point multiplication match those obtained by Hopkins et al..

Using ALIF neurons has been shown to significantly improve the performance of SNN models [18]. However, accurately simulating their dynamics can prove challenging on digital hardware as fixed-point numbers implemented with standard round-to-zero integer operations prematurely underflow to zero. Figure 3 shows simulations of ALIF neurons driven by challenging input spike trains. Figure 3B highlights a period where the 16 bit fixed point simulation with round-to-zero diverges from the 64 bit floating point reference simulation while the version with stochastic rounding follows closely. Overall, stochastic rounding significantly reduces the Normalised Root-Mean Square Error (NRMSE) compared to the floating point reference (NRMSE of V reduces from 0.13 to 0.045 and A from 0.019 to 0.0066).

When training SNN models or performing inference on live data, stochastic rounding alone is not sufficient for numerically safe simulations. As figure 3D shows, if a neuron's input increases temporarily above the calibrated range of the fixed-point formats, state variables overflow and wrap. This results in catastrophic failure (NRMSE of V of 0.29 and A of 0.39) that can, however, be mitigated by saturating additions and subtractions (NRMSE of V of 0.098 and A of 0.019).

D. Classification

To demonstrate the performance of our FeNN prototype on a real machine learning task, we trained an ALIF-based Recurrent SNN (RSNN) classifier with a single 256 neuron hidden layer on the Spiking Heidelberg Digits (SHD) dataset [15], quantized the trained weights and deployed it onto FeNN. Running on FeNN, this classifier obtains an accuracy of (79.50 ± 0.44) % on the test set – an insignificant difference in performance compared to the 32 bit floating point GeNN implementation which obtains 79.6%. Figure 4A shows the

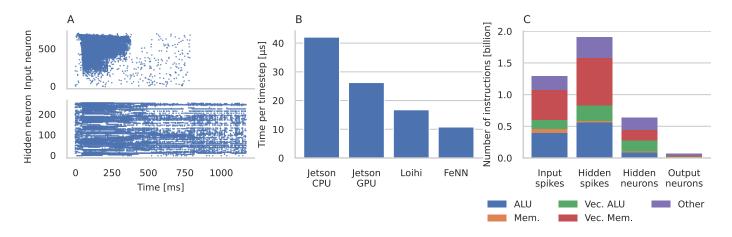


Fig. 4. SHD inference running on FeNN (A) Raster plot of input and hidden neuron activity during one example. (B) Comparison of average time (over the whole test set) taken to simulate a single SNN timestep on FeNN with Jetson Orin Nano using GeNN (batch size=1) [21] and similar RSNN running on Loihi [22]. GPU times were measured using CUDA timing events and CPU and FeNN times using std::chrono:: high_resolution_clock (C) Distribution of different types of instructions in different stages of the simulation.

TABLE II
SHD CLASSIFICATION ENERGY. 'TOTAL INFERENCE ENERGY' IS ALL ENERGY SPENT DURING INFERENCE. IDLE POWER AND TIME SPENT COPYING DATA ARE REMOVED TO CALCULATE 'SIMULATION ENERGY' AND THIS IS DIVIDED BY THE TOTAL NUMBER OF SYNAPTIC EVENTS TO GET 'ENERGY PER SYNAPTIC EVENT'.

Device	Total inference energy [J]	Simulation energy [J]	Energy per synaptic event [nJ]
FeNN	347	44	8
Jetson GPU	645	97	18
Jetson CPU	956	113	20

input spikes from an example digit and the hidden layer activity within the network. In Figure 4B, we compare the average time (over the whole SHD test set) taken to simulate a single SNN timestep using FeNN and a Jetson Orin Nano (as this has a similar form factor and power usage to a Kria KV 260) using GeNN [21]. Additionally, we include the time reported by Rao et al. [22] for a similar ALIF-based RSNN classifier on the Loihi [7] neuromorphic system.

E. Energy usage

Neither the Jetson Orin Nano or Kria KV260 provide a sensor for measuring the power used by the GPU or FPGA accelerator. Therefore, we recorded the power usage during SHD inference using a consumer power measurement device at the mains socket, captured the output with a webcam and performed optical character recognition using the ssocr Ubuntu package. We then integrated the power time series using the trapezoidal rule to calculate the energy spent during the whole simulation (we dis-regarded initialisation as it was implemented very differently in the two simulations). As Table II shows, FeNN takes approximately half the energy of the Jetson GPU and a third of the Jetson CPU to evaluate the SHD test set. However, both systems were running desktop Linux operating systems so the majority of this energy goes to driving the display, performing background tasks etc. Furthermore, a

significant amount of time is spent streaming SHD digits to the accelerators which is not currently well-optimised on either system. As neither of these factors are of direct interest to this work, we also show simulation energy with these removed and energy per-synaptic event in Table II. Again, these illustrate that FeNN takes approximately half the energy of the Jetson for the same task.

IV. CONCLUSIONS AND FUTURE WORK

Here we have presented a first prototype of our FeNN accelerator which is already capable of efficiently simulating a moderately-sized RSNN classifier, faster than both embedded GPU and neuromorphic systems. Furthermore, FeNN performs these simulations using around half of the energy of an embedded GPU.

Figure 4C shows the distribution of different instruction types executed by FeNN while evaluating the SHD test. The ratio of 'Vector Memory' and 'Vector ALU' instructions clearly illustrates the low arithmetic intensity of SNN simulation, especially in the spike-processing code where this is 3:1. When processing spikes, a significant number of scalar ALU instructions are spent iterating over spikes and calculating the addresses to load weights from. To reduce this overhead and also allow FeNN to simulate larger models with weights stored in external memory, we plan to develop Direct Memory Access (DMA) hardware which which will calculate addresses from spikes and asynchronously copy weights from external memory to UltraRAM. This DMA hardware will also enable the transfer of data between FeNN cores using an AXI crossbar. As well as these system-on-chip features, we also plan on adding lane-local memories - for efficient processing of sparse connectivity [12] - and instructions to perform reductions across vector lanes to each core. Finally, the SNNs simulated here have all been hand-coded in RISC-V assembly language. However, we are developing a backend for our GeNN library [21] to convert high-level model descriptions into FeNN instructions.

With these developments in place, we intend to scale up our prototype to a 32 core version instantiated on an Alveo U55C accelerator which will be capable of batched Eventprop [23] training in addition to inference. Because Eventprop training has a similar computational complexity as inference, we would expect this system to be able to *train* models at a similar speed and energy cost to the inference task we perform here. For *inference*, the $8\,\mathrm{nJ}$ per-SOP achieved by FeNN is not competitive with the latest neuromorphic systems (at least partly due to the $10\times$ overhead of FPGA vs ASIC logic [24]). However, due to its greater flexibility, including the ability to support inference and Eventprop training in future, FeNN is a valuable intermediate step towards the wider adoption of neuromorphic technologies.

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