

第7章 频率特性和谐振现象

杨旭强

哈尔滨工业大学电气工程系



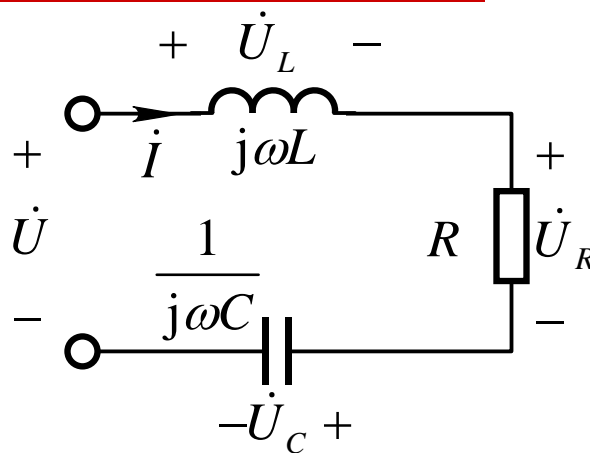
7.3 RLC 串联电路的频率特性

基本要求：了解 RLC 串联电路的网络函数及其频率特性。

主要内容

- 一、以电阻电压为响应的网络函数
- 二、以电容电压为响应的网络函数
- 三、以电感电压为响应的网络函数

7.3 *RLC*串联电路的频率特性



$$H_R(j\omega) = \frac{\dot{U}_R}{\dot{U}} = \frac{R}{R + j[\omega L - 1/(\omega C)]}$$

$$H_C(j\omega) = \frac{\dot{U}_C}{\dot{U}} = \frac{1/(j\omega C)}{R + j\omega L + 1/(j\omega C)}$$

$$H_L(j\omega) = \frac{\dot{U}_L}{\dot{U}} = \frac{j\omega L}{R + j\omega L + 1/(j\omega C)}$$

一、以电阻电压为响应的网络函数

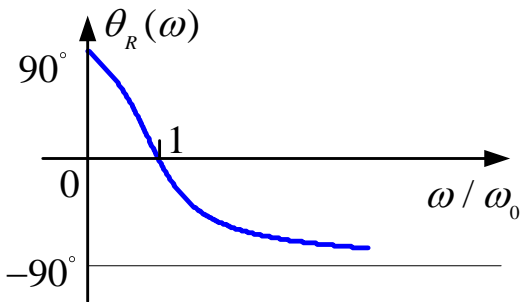
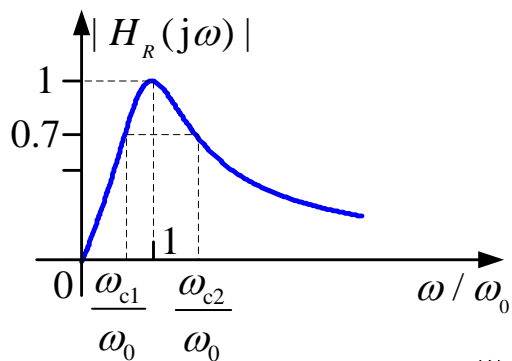
$$H_R(j\omega) = \frac{\dot{U}_R}{\dot{U}} = \frac{R}{R + j[\omega L - 1/(\omega C)]}$$

$$|H_R(j\omega)| = \frac{1}{\sqrt{1 + \frac{1}{R^2} \left(\omega L - \frac{1}{\omega C} \right)^2}} = \frac{1}{\sqrt{1 + Q^2 \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)^2}}$$

$$\theta_R(\omega) = -\arctan Q \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)$$

ω / ω_0	$ H_R(j\omega) $	$\theta_R(\omega)$
0	0	90°
1	1	0°
\vdots	\vdots	\vdots
\vdots	\vdots	\vdots
∞	0	-90°

一、以电阻电压为响应的网络函数



RLC带通电路的频率特性

$$|H_R(j\omega)| = \frac{1}{\sqrt{1 + Q^2 \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)^2}}$$
$$\theta_R(\omega) = -\arctan Q \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)$$

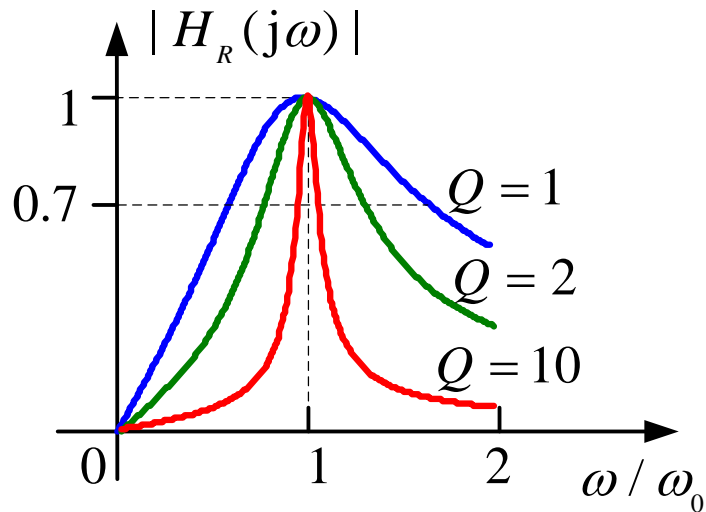
1. $|H_R(j\omega)|$ 具有带通特性

$$\frac{1}{\sqrt{1 + Q^2 \left(\frac{\omega_c}{\omega_0} - \frac{\omega_0}{\omega_c} \right)^2}} = \frac{1}{\sqrt{2}} \quad \xrightarrow{\text{截止频率}} \quad \begin{cases} \omega_{c1} = \omega_0 \left(-\frac{1}{2Q} + \sqrt{\frac{1}{4Q^2} + 1} \right) \\ \omega_{c2} = \omega_0 \left(\frac{1}{2Q} + \sqrt{\frac{1}{4Q^2} + 1} \right) \end{cases}$$

通带宽度 $\Delta\omega = \omega_{c2} - \omega_{c1} = \frac{\omega_0}{Q}$

一、以电阻电压为响应的网络函数

2. 频率特性与品质因数的关系



$$\Delta\omega = \omega_{c2} - \omega_{c1} = \frac{\omega_0}{Q}$$

(1) Q 值越大，截止频率处的曲线越陡，频率选择性越好，带宽越窄。

(2) Q 值越小，带宽越宽，选择性能越差。

RLC 串联电路，频率选择性与带宽存在矛盾。

二、以电容电压为响应的网络函数

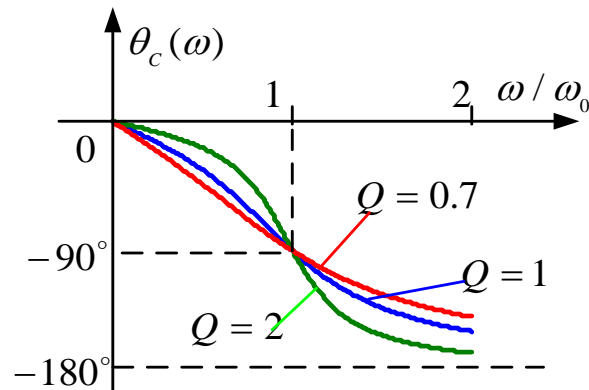
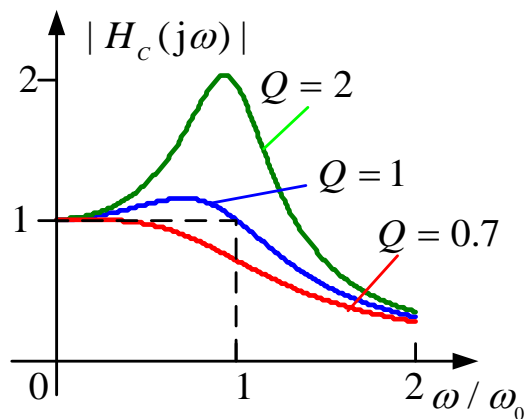
$$H_c(j\omega) = \frac{\dot{U}_c}{\dot{U}} = \frac{1/(j\omega C)}{R + j\omega L + 1/(j\omega C)}$$
$$= \frac{1}{(1 - \omega^2 LC) + j\omega CR} = \frac{1}{[1 - \left(\frac{\omega}{\omega_0}\right)^2] + j\frac{1}{Q}\left(\frac{\omega}{\omega_0}\right)}$$

$$|H_c(j\omega)| = \frac{1}{\sqrt{\left[1 - \left(\frac{\omega}{\omega_0}\right)^2\right]^2 + \frac{1}{Q^2}\left(\frac{\omega}{\omega_0}\right)^2}}$$

$$\theta_c(\omega) = -\arctan \frac{1}{Q\left(\frac{\omega_0}{\omega} - \frac{\omega}{\omega_0}\right)}$$

ω / ω_0	$ H_c(j\omega) $	$\theta_c(\omega)$
0	1	0°
1	Q	-90°
\vdots	\vdots	\vdots
\vdots	\vdots	\vdots
∞	0	-180°

二、以电容电压为响应的网络函数



RLC低通电路的滤波特性

$$|H_c(j\omega)| = \frac{1}{\sqrt{\left[1 - \left(\frac{\omega}{\omega_0}\right)^2\right]^2 + \frac{1}{Q^2} \left(\frac{\omega}{\omega_0}\right)^2}}$$
$$\theta_c(\omega) = -\arctan \frac{1}{Q \left(\frac{\omega_0}{\omega} - \frac{\omega}{\omega_0}\right)}$$

$|H_c(j\omega)|$ 具有低通特性

三、以电感电压为响应的网络函数

$$H_L(j\omega) = \frac{\dot{U}_L}{\dot{U}} = \frac{j\omega L}{R + j\omega L + 1/(j\omega C)}$$

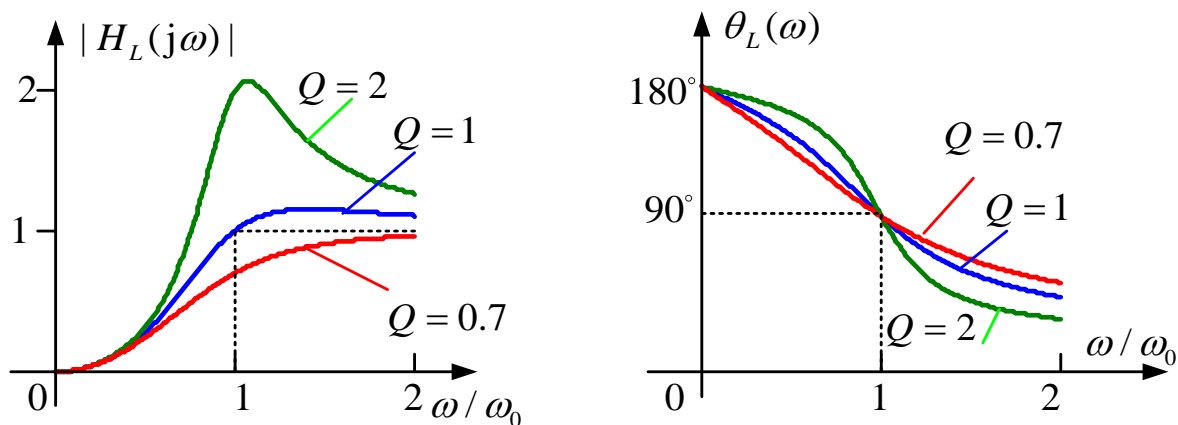
$$= \frac{1}{[1 - 1/(\omega^2 LC)] - jR/(\omega L)} = \frac{1}{\left[1 - \left(\frac{\omega_0}{\omega}\right)^2\right] - j\frac{1}{Q}\left(\frac{\omega_0}{\omega}\right)}$$

$$|H_L(j\omega)| = \frac{1}{\sqrt{\left[1 - \left(\frac{\omega_0}{\omega}\right)^2\right]^2 + \frac{1}{Q^2}\left(\frac{\omega_0}{\omega}\right)^2}}$$

$$\theta_L(\omega) = -\arctan \frac{-1}{Q\left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega}\right)}$$

ω / ω_0	$ H_L(j\omega) $	$\theta_L(\omega)$
0	0	180°
1	Q	90°
\vdots	\vdots	\vdots
\vdots	\vdots	\vdots
∞	1	0°

三、以电感电压为响应的网络函数

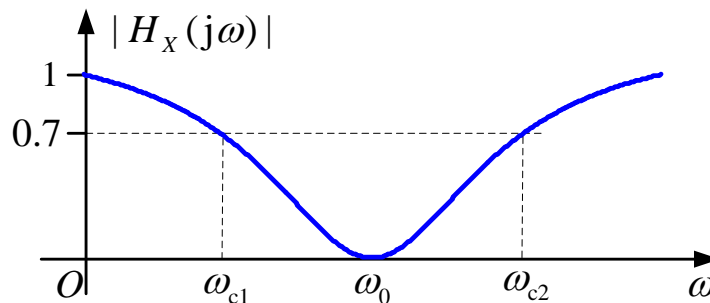
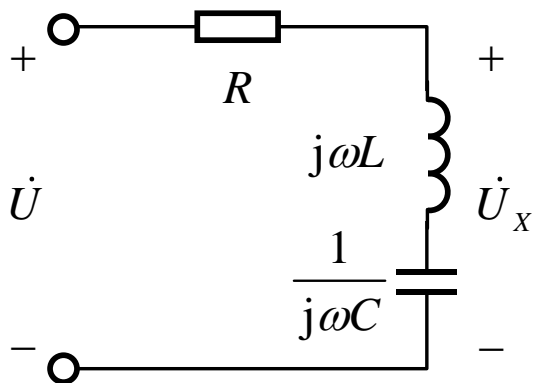


RLC高通电路频率特性

$$|H_L(j\omega)| = \frac{1}{\sqrt{\left[1 - \left(\frac{\omega_0}{\omega}\right)^2\right]^2 + \frac{1}{Q^2} \left(\frac{\omega_0}{\omega}\right)^2}}$$
$$\theta_L(\omega) = -\arctan \frac{-1}{Q \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega}\right)}$$

$|H_L(j\omega)|$ 具有高通特性

思考：RLC串联电路能否实现带阻特性？



$$H_X(j\omega) = \frac{\dot{U}_X}{\dot{U}} = \frac{j[\omega L - 1/(\omega C)]}{R + j[\omega L - 1/(\omega C)]}$$

$$\omega_{c1} = \omega_0 \left(-\frac{1}{2Q} + \sqrt{\frac{1}{4Q^2} + 1} \right)$$

$$\omega_{c2} = \omega_0 \left(\frac{1}{2Q} + \sqrt{\frac{1}{4Q^2} + 1} \right)$$