
Hardware and software guidelines for use of LPS25H pressure sensor

By Raffaele Di Vaio

Introduction

The LPS25H is an ultra-compact absolute piezoresistive pressure sensor with enhanced digital features in a small package footprint.

Unless specifically noted, all recommendations in this document apply only to the LPS25H.

Please refer to the LPS25H datasheet (available at www.st.com) for device and feature definitions.

Contents

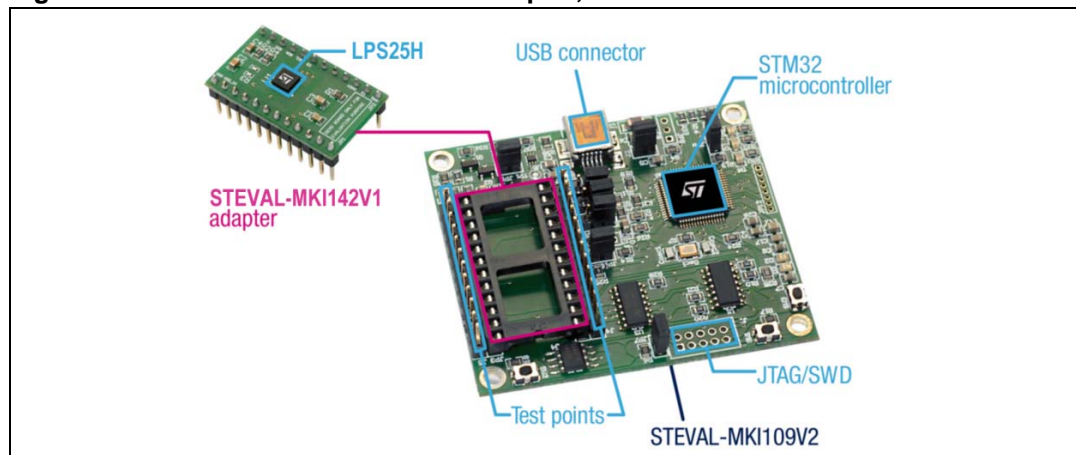
1	Pressure sensor evaluation boards	4
2	Hardware (designing PCB schematics and layout)	6
2.1	LPS25H device package, interconnect and polarization	6
2.1.1	Package drawings	6
2.1.2	Pin mapping	7
2.1.3	Typical application circuit	8
2.2	Pressure sensor PCB layout and soldering recommendations	9
2.2.1	PCB design rules	9
2.2.2	Stencil design and solder paste application	10
2.2.3	Process consideration	10
2.2.4	Manual soldering	11
3	Power supply: consumption estimation and optimization	12
3.1	Common rules for low power consumption	14
4	Using the device step-by-step, from basic to advanced	15
4.1	First time bring-up (I ² C example)	15
4.2	Quick troubleshooting guide	15
4.3	One-shot mode measurement sequence	16
5	Using FIFO modes	17
5.1	Effective use of the FIFO	17
5.1.1	Accessing the FIFO data	17
5.1.2	Bypass mode (F_MODE[2..0]="000" in FIFO_CTRL (0x2E))	18
5.1.3	FIFO mode (F_MODE[2..0]="001" in FIFO_CTRL (0x2E))	18
5.1.4	Stream mode (F_MODE[2..0]="010" in FIFO_CTRL (0x2E))	18
5.1.5	FIFO mean mode (F_MODE[2..0]="110" in FIFO_CTRL (0x2E))	18
5.1.6	Hardware digital filter	19
5.2	Extra FIFO modes	20
5.2.1	Stream to FIFO mode (F_MODE[2..0]="011" in FIFO_CTRL (0x2E))	20
5.2.2	Bypass to Stream mode (F_MODE[2..0]="100" in FIFO_CTRL (0x2E))	20
5.2.3	Bypass to FIFO mode (F_MODE[2..0]="111" in FIFO_CTRL (0x2E))	21

6	Software	22
6.1	Hints	22
6.2	One-shot mode conversion time estimation	22
6.3	Reference SW to get started with LPS25H	22
6.4	Pressure to altitude conversion	22
6.5	SW filtering & internal FIFO	23
6.6	Reset the device to power-on configuration	23
6.7	Absolute accuracy drift due to soldering	24
	6.7.1 Correcting soldering drift (one-point calibration)	24
7	Revision history	25

1 Pressure sensor evaluation boards

UNICO/eMotion evaluation system: [STEVAL-MKI109V2] + [STEVAL-MKI142V1 or V2].

Figure 1. STEVAL-MKI142V1 LPS25H adapter, STEVAL-MKI109V2 MEMS motherboard



The STEVAL-MKI109V2 is a motherboard designed to provide users with a complete, ready-to-use platform for the demonstration of STMicroelectronics MEMS products. The board features a DIL 24 socket to mount all available adapters for both digital and analog output MEMS devices.

The motherboard includes a high-performance 32-bit microcontroller, which functions as a bridge between the sensor and a PC, on which it is possible to use the downloadable graphical user interface (GUI) from ST (Unico SW), or dedicated software routines for customized applications.

A plain terminal (such as MS windows hyper-terminal (C)) can also be used to access the sensor registers by text read/write commands.

Figure 2. STEVAL-MKI142V1 - LPS25H adapter



The STEVAL-MKI142V1 adapter board is designed to facilitate the demonstration of the LPS25H product. The board offers an effective solution for fast system prototyping and device evaluation directly within the user's own application.

The STEVAL-MKI142V1 can be plugged into a standard DIL 24 socket. The adapter provides the complete LPS25H pin-out and comes ready-to-use with the recommended decoupling capacitors on the VDD power supply line. No pull-ups are provided.

The STEVAL-MKI109V2 MEMS (aka eMotion board, shipped with pre-loaded eMotion FW) is completed by a MS Windows™-based SW application called Unico. This tool may be used as a simple real-time demonstrator or to verify device performance.

It also allows easy monitoring of the register status and allows changes to them based on the intended scenario.

Recommended FW/SW revisions fully supporting the STEVAL-MKI142V1 are:

- eMotion V3.0.6
- Unico Rev. 3.0.1.0 beta

or newer.

2 Hardware (designing PCB schematics and layout)

2.1 LPS25H device package, interconnect and polarization

2.1.1 Package drawings

Pressure sensor LPS25H is available in a holed LGA package: HCLGA-10L (1mm thickness).

Figure 3. HCLGA-10L 2.5x2.5x1.0 mm

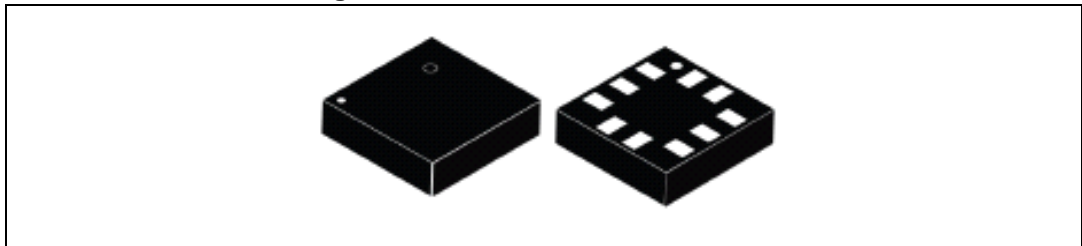
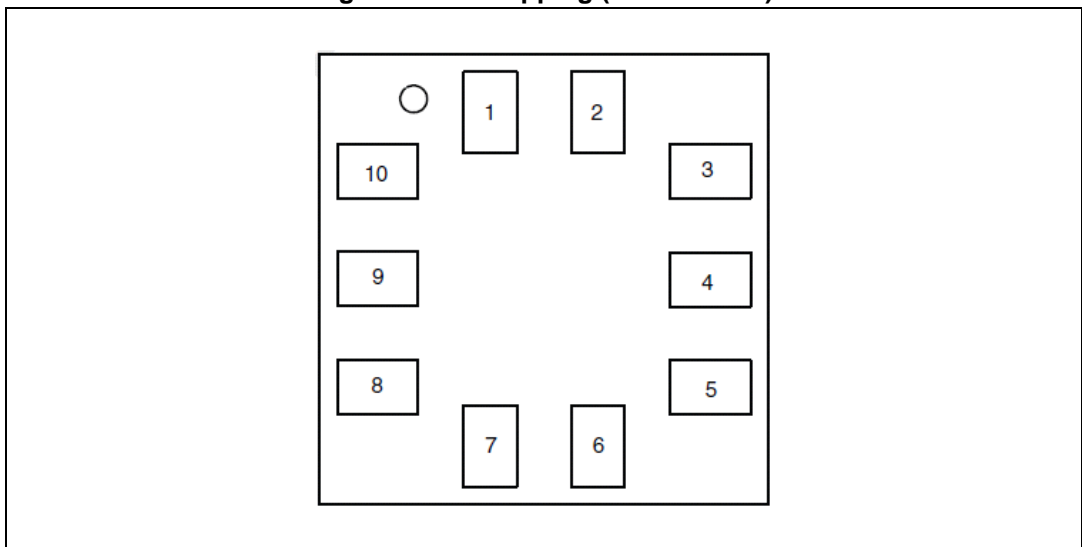


Figure 4. Pin mapping (bottom view)



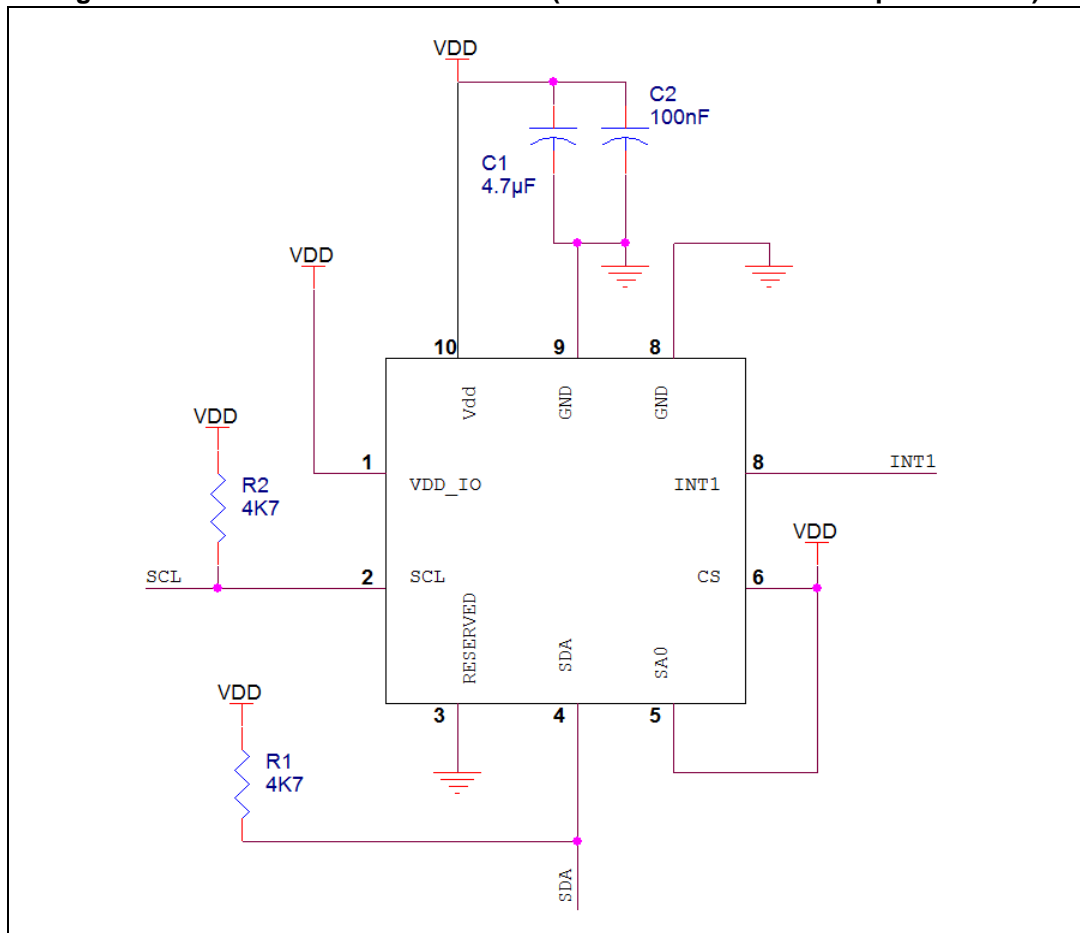
2.1.2 Pin mapping

Table 1. LPS25H pin mapping details

Pin #	Name	Function	What to do
1	VDD_IO	Power supply for I/O pins	I/O supply voltage (1.7V ~ 3.6V)
2	SCL SPC	I ² C serial clock (SCL) SPI serial port clock (SPC)	Optional pull-up resistor, connected to VDD_IO, could be required for I ² C (see I ² C standard for recommended value)
3	Reserved	Reserved	Connect to PCB ground
4	SDA SDI SDIO	I ² C serial data (SDA) 4-wire SPI serial data input (SDI) 3-wire SPI serial data input/output (SDIO)	Optional pull-up resistor, connected to VDD_IO, could be required for I ² C (see I ² C standard for recommended value)
5	SA0 SDO	I ² C slave address select (SA0) 4-wire SPI serial data output (SDO)	SA0=1: 0xBA/BB I ² C slave address + R/W SA0=0: 0xB8/B9 I ² C slave address + R/W
6	CS	I ² C/SPI mode selection SPI chip select	CS=1: I ² C mode CS=ChipSelect in SPI mode
7	INT1	Interrupt 1 (or data ready)	Leave unconnected if unused
8	GND_IO	I/O pins ground	Connect to PCB ground
9	GND	Ground	Connect to PCB ground
10	VDD	Power supply	Supply voltage (1.7V ~ 3.6V)

2.1.3 Typical application circuit

Figure 5. LPS25H electrical connection (I²C communication example w/SA0=1)



Key notes:

The device core is supplied through the VDD line. Power supply decoupling capacitors (100 nF and at least 4.7 µF) must be placed as near as possible to the supply pads of the device.

VDD_IO should match the one requested by the host controller. In any case VDD_IO must be the same or lower than VDD. In the example shown here, VDD_IO is connected to VDD.

Please note that the LPS25H does not have an internal pull-up on the I²C lines (SCL and SDA). They must be added externally according to the I²C bus speed and load, and connected to VDD_IO.

In I²C mode, SA0 = 1 (pin directly connected to VDD_IO) sets the slave address to 0xBA for write and 0xBB for read. A pull-up resistor could be a better choice if there is a need to communicate alternatively in I²C or SPI 4-wires, since pin 5 is SDO for SPI 4-wires and of course it cannot function if hard-connected to VDD_IO.

The I²C interface is compliant with fast mode (400 kHz) I²C standards as well as with normal mode.

In SPI the CS pin is the chip select and it is controlled by SPI master. It goes low at the start of the communication and returns high at the end.

SPI communication can work in 2 different modes: 3-wires and 4-wires. In the first case the pins used are: CS, SPC and SDIO; in the latter: CS, SPC, SDI and SDO. SPI mode selection is done through bit 0 of register CTRL_REG1: 0=4-wires; 1=3-wires.

Both SPI and I²C interfaces are active by default. In case of SPI use only, I²C interface can optionally be disabled by programming the corresponding bit in a control register (CTRL_REG2[3]=1).

2.2 Pressure sensor PCB layout and soldering recommendations

The LPS25H has an aperture on top of the package, so special care is required since sensor performance could be compromised by:

- Mechanical stress coming from the PCB board
 - The whole package surface + air should have minimum temperature gradient
 - Avoid placement in long and narrow PCB area, warp-free area
- Temperature gradients (non-uniform/rapidly changing temperature around sensor)
- Strong electrical field / light source
- Localized air pressure stability (unwanted fast air pressure variation, fans)
- Dust and water exposure/condensation (GORE-TEX[®] protection, etc.)

The HCLGA package is compliant with the ECOPACK[®] standard and is qualified for soldering heat resistance according to JEDEC J-STD-020.

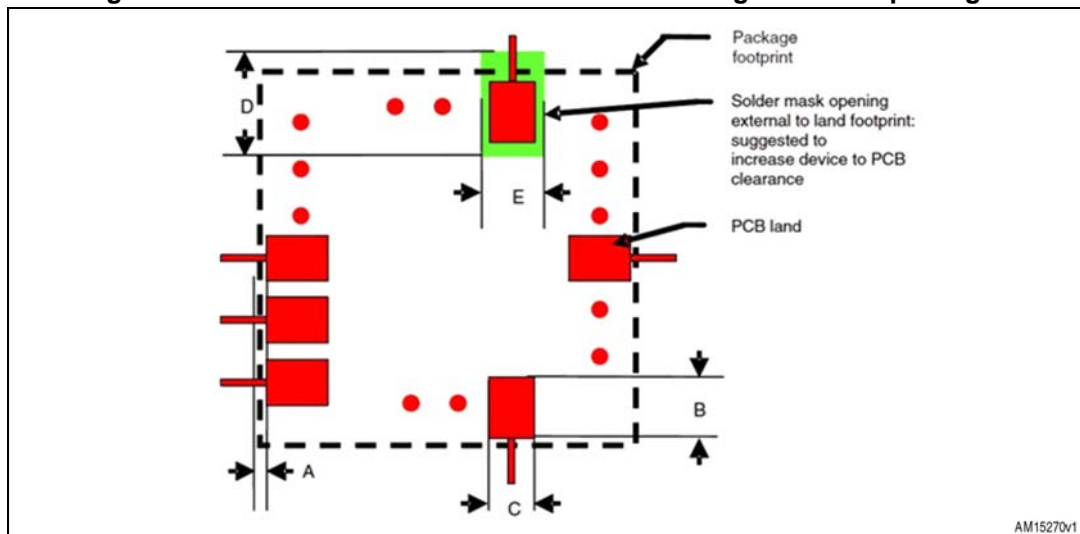
2.2.1 PCB design rules

The pressure sensor is affected by mechanical stress coming from the PCB board, hence it should be minimized. A typical suggestion is to place the pressure sensor at the edge of the PCB where warping is minimal.

PCB land and solder masking general recommendations are shown below. Refer to the LPS25H datasheet for pad count, size and pitch.

- It is recommended to open a solder mask external to PCB land.
- The area below the sensor (on the same side of the board) must be defined as a keep-out area. It is strongly recommended not to place any structure on the top metal layer underneath the sensor.
- This means that it is possible to place the ground plane in the PCB middle layer under the LPS25H, but not in the plane immediately under the LPS25H.
- Traces connected to pads should be as symmetrical as possible. Symmetry and balance for pad connection will help component self-alignment and will lead to better control of solder paste reduction after reflow.
- For better performance over temperature, it is strongly recommended not to place large insertion components like buttons or shielding boxes at distances less than 2 mm from the sensor.
- The pin 1 indicator must be left unconnected to ensure proper device operation.

Figure 6. Recommended land and solder mask design for *LGA packages



- A = Clearance from PCB land edge to solder mask opening ≥ 0.25 mm to ensure that solder mask is opened externally to device area
- B = PCB land length = *LGA solder pad length + 0.1 mm
- C = land width = *LGA solder pad width + 0.1 mm
- D = Solder mask opening length = PCB land length + 0.3 mm: design 0.05 mm inside and 0.25 mm outside
- E = Solder mask opening width = PCB land width + 0.1 mm

2.2.2 Stencil design and solder paste application

The soldering paste thickness and pattern are important for a proper pressure sensor mounting process.

- Stainless steel stencils are recommended
- Stencil thickness of 90 - 150 μm (3.5 - 6 mils) is recommended for screen printing
- The final soldering paste thickness should allow proper cleaning of flux residues and clearance between sensor package and PCB
- Stencil aperture should have a rectangular shape with dimensions up to 25 μm (1 mil) smaller than PCB land
- The openings of the stencil for the signal pads should be between 70 - 80% of the PCB pad area
- Optionally, for better solder paste release, the aperture walls should be trapezoidal and the corners rounded
- The fine IC leads pitch requires accurate alignment of the stencil to the PCB. The stencil and printed circuit assembly should be aligned to within 25 μm (1 mil) prior to application of the solder paste

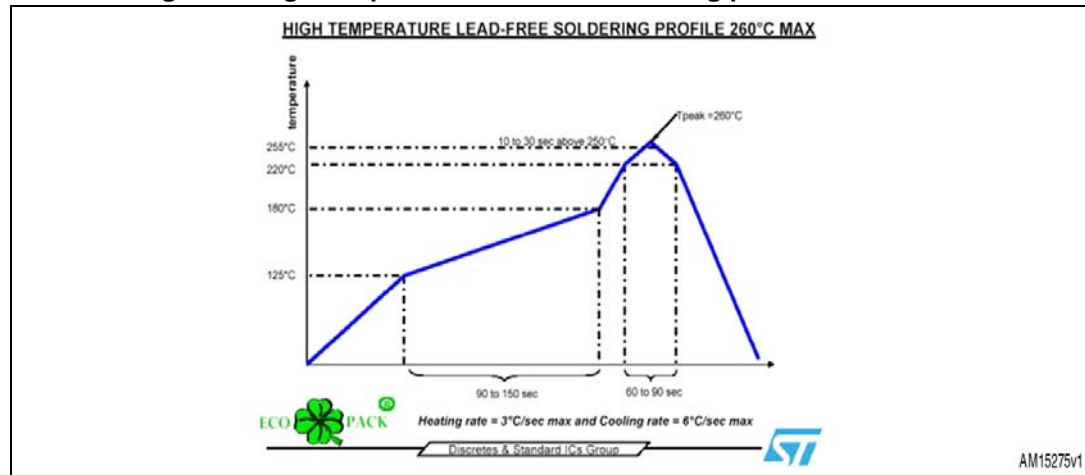
2.2.3 Process consideration

In using non self-cleaning solder paste, proper board washing after soldering must be carried out to remove any possible sources of leakage between pads due to flux residues. However, take care not to perform the cleaning process on top of the pressure sensor.

The PCB soldering profile depends on the number, size and placement of components on the board. The soldering profile should be defined by experience, rather than the pressure sensor soldering profile only.

The JEDEC reflow profile is provided below.

Figure 7. High temperature lead-free soldering profile 260 °C max



No solder material reflow on the side of the package is allowed since *LGA packages show metal trace out of package side.

No need to use any tape to cover the hole on the CAP of the LPS25H during reflow. The cleaning of the PCB is not a general task, as it should be cleaned only when the PCB is contaminated during reflow. ST does not ship the LPS25H covered.

2.2.4 Manual soldering

Apply solder paste (e.g. ALPHA® OM-338-T) on the PCB solder pads using a syringe (air, bubble free), and position the *LGA accordingly. Then press it to ensure it is flat on the PCB, and heat it in an oven using the 260 °C JEDEC profile (the one shown in [Figure 7](#)).

Take care that hand soldering is within spec. In any case, in an emergency situation, coat the solder pad with tin-lead, and use solder wick to flatten the solder ball. Then put the soldering iron on the *LGA land marks to deposit solder on the *LGA pads. Then position it, and use a hot air gun to melt both sides and obtain eutectic bonding (350 °C). However, this is not recommended. Hot air from the gun must be spread around the package dynamically; keep moving the air gun around the *LGA.

Solder heat resistance and environmental specification

The second level interconnect category on ST ECOPACK® lead-free package is marked on the inner box label, in compliance with JEDEC standard JESD97. Soldering condition maximum ratings are also marked on the same label.

HCLGA/HLGA packages for pressure sensors are qualified for soldering heat resistance according to JEDEC J-STD-020, in MSL3 condition.

3 Power supply: consumption estimation and optimization

There are 2 supply voltages in the LPS25H: VDD and VDD_IO.

VDD is the core voltage used to supply to the internal circuits, power on reset, and the sensor.

VDD_IO is the supply for the I²C / SPI blocks and interface signals.

In order to prevent possible leakage in the operational condition, it is necessary to ensure $VDD_IO \leq VDD$.

The operating voltage for both VDD_IO and VDD is 1.7 V to 3.6 V.

The LPS25H can also be powered directly by a small lithium coin battery like the CR2032. This type of battery has a current capability of 220 mAh, enough to sustain the sensor peak current (less than 1 mA).

Provided that the battery fluctuation, which could be in the range of 0.4 V, is within the “normal operating voltage conditions” (even with an aged battery) and the recommended decoupling capacitors are placed near the sensor, our tests with batteries and switched resistive loads show that the sensor functions correctly even with fast slew rates.

During the main power-on transition, the system will initialize the internal registers according to the factory-stored information (such as sensor calibration data).

This is triggered by VDD increasing above 1.6 V.

Current consumption during the device initialization peaks around 200 μ A (for a duration of about 600 μ s).

After these steps, the LPS25H will enter a power-down state. The LPS25H is ready in power-down from power-off in about 2.5 ms.

The LPS25H runs in normal mode from power-down state (toggling PD bit[7] in CTRL_REG1, 0x20) in a few microseconds. Depending on register RES_CONF (0x10) settings, the first valid measurement is ready after about 2 ms (minimum resolution, with AVGT[1:0]=00; AVGP[1:0]=00, i.e. 8+8 samples) or about 37 ms (maximum resolution, with AVGT[1:0]=11; AVGP[1:0]=11, i.e. 64+512 samples).

The typical conversion time (with +/-3% accuracy at room temperature) is given by the formula:

Equation 1

$$T_{\text{conversion in } \mu\text{s}} = 62 * (T_{\text{avg}} + P_{\text{avg}}) + 975$$

where T_{avg} and P_{avg} are the values, respectively, for temperature and for pressure measurements, selected in register RES_CONF as the number of averaged samples per measurement.

– Example: $T_{\text{avg}} = 64$; $P_{\text{avg}} = 512$; Typ. conversion time 36687 μ s

The LPS25H has been designed to have a very low power consumption. The majority of the internal blocks are switched-on only during the acquisition time. In one-shot mode the device automatically goes into standby when the measurement is completed.

The minimum power consumption occurs when the device is in power-down (PD=0 bit[7] of the CTRL_REG1 (0x20)) AND the interrupt pin is not configured to drain current. In normal

operating mode the power consumption depends on HW-averaging (resolution) and output data rate (auto-refresh frequency) settings. The following formula gives an indication of the current consumption in relation to average and ODR. This can be used to find the best compromise, for a given application, between consumption, update speed and filtering.

Equation 2

$$I_{dd} = [(3\mu A/Hz + 42nA/Hz * P_{avg}) + 30nA/Hz * T_{avg}] * ODR$$

where:

P_{avg} = value set by AVGP[1:0] (8, 32, 128, 512)

T_{avg} = value set by AVGT[1:0] (8, 16, 32, 64)

ODR = rates according to ODR[2:0] (1, 7, 12.5, 25Hz)

Table 2. ODR, AVGT and AVGP correspondence

ODR[6..4]@0x20	ODR	AVGT[3..2]@0x10	Tavg	AVGP[1..0]@0 x10	Pavg
1	1Hz	0	8	0	8
2	7Hz	1	16	1	32
3	12.5Hz	2	32	2	128
4	25Hz	3	64	3	512

- a) Example 1:
 - ODR= 25 Hz
 - P_{avg} = pressure average=512
 - T_{avg} = temperature average=64
 - I_{dd} = 661 μA (V_{dd} independent at first approximation)
- b) Example 2:
 - ODR=1 Hz
 - P_{avg} = pressure average=512
 - T_{avg} = temperature average=64
 - I_{dd} = 26.4 μA
- c) Example 3:
 - ODR= 1 Hz
 - P_{avg} = pressure average=8
 - T_{avg} = temperature average=8
 - I_{dd} = 3.6 μA

In order to improve system power saving, the LPS25H embeds a FIFO buffer capable of storing up to 32 pressure output values. Since the host processor does not need to continuously poll data from the sensor, it can wake up (on programmable interrupt from the sensor) only when requested and burst the significant data out from the buffer.

The FIFO can alternatively be configured as moving average to reduce the noise figure (or to reduce the ODR rate while keeping the same noise figure, traded for delay).

3.1 Common rules for low power consumption

Here are some notes about possible sources of power leakage:

- Do not leave any input pin un-terminated (floating). This may leave the internal circuit in undefined state, creating/introducing extra power consumption. Connect unused input pins (such as CS, and SA0 if a single sensor is used on the I²C bus) to VDD_IO. Use a pull-up resistor where needed.
- Do not connect a push-pull output pin directly to GND or VDD_IO (unless it can be ensured that no change on that pin status will be made by SW).
- Select the proper I²C pull-up resistor (to VDD_IO) according to the bus load and working frequency.

4 Using the device step-by-step, from basic to advanced

4.1 First time bring-up (I²C example)

1. Start by using the same supply for VDD_IO and VDD to check device functionality.
2. Check supply impedances and ensure that all the pins voltages are in static condition.
3. The device is a slave with 1 byte sub-address which MSB should be '1' to enable the multiple data read/write at increasing addresses (bit 7 = 1 enables I²C sub-address multi-byte auto-increment)
4. Use a normal I²C bus speed (<400 kHz)
5. Read the chip ID byte (sub address 0x0F, or 0x8F in case of auto-increment option): 0xBD should be read

4.2 Quick troubleshooting guide

If the device is not communicating properly on the I²C bus:

1. If the I²C slave address is not acknowledged, check:
 - Power supply is present and matches the I²C master VDD_IO
 - I²C line pull-ups are present and with the correct values
 - The slave address is correctly selected
 - There are no other slaves with same address on the I²C bus lines.
 - Make sure the maximum I²C speed is driven by the slowest slave on the same I²C bus.
 - Use an oscilloscope to check that the device acknowledges its slave address.
2. If the device responds and the pressure is always fixed and out of range (e.g. 760 mb), make sure that AVGT and AVGP are in line with the specs and wait for the data ready bit flag before reading the pressure registers.
3. If the device is reporting a value with a small offset compared to the expected one, a single point calibration may be needed in order to compensate the post soldering package stress (see [Section 6.7](#))

4.3 One-shot mode measurement sequence

1. Power down the device (clean start)
 - WriteByte(CTRL_REG1_ADDR = 0x00); // @0x20=0x00
2. Turn on the pressure sensor analog front end in single shot mode
 - WriteByte(CTRL_REG1_ADDR = 0x84); // @0x20=0x84
3. Run one-shot measurement (temperature and pressure), the set bit will be reset by the sensor itself after execution (self-clearing bit)
 - WriteByte(CTRL_REG2_ADDR = 0x01); // @0x21=0x01
4. Wait until the measurement is completed
 - ReadByte(CTRL_REG2_ADDR = 0x00); // @0x21=0x00
5. Read the temperature measurement (2 bytes to read)
 - Read((u8*)pu8, TEMP_OUT_ADDR, 2); // @0x2B(OUT_L)~0x2C(OUT_H)
 - Temp_Reg_s16 = ((u16) pu8[1]<<8) | pu8[0]; // make a SIGNED 16 bit variable
 - Temperature_DegC = 42.5 + Temp_Reg_s16 / 480; // offset and scale
6. Read the pressure measurement
 - Read((u8*)pu8, PRESS_OUT_ADDR, 3); // @0x28(OUT_XL)~0x29(OUT_L)~0x2A(OUT_H)
 - Pressure_Reg_s32 = ((u32)pu8[2]<<16)|((u32)pu8[1]<<8)|pu8[0]; // make a SIGNED 32 bit
 - Pressure_mb = Pressure_Reg_s32 / 4096; // scale
7. Check the temperature and pressure values make sense
 - Reading fixed 760 hPa, means the sensing element is damaged.

Example of register measurements and conversion:

P = 0x3FF58D means $4191629 / 4096 = 1023.347$ hPa

T = 0xE07C means $42.5 + (-8068/480) = 25.7$ °C

{ as reference: 0x6BD0 = 100°C; 0xB050=0 °C }

The value read from the PRESS_OUT registers is always the difference of the sensor measured pressure (after averaging) and a base value, that can be either the value copied in the RPDS register (reference pressure after soldering, to correct the pressure offset due to mechanical stress, 16-bit value corresponding to the MSB part of the pressure value) by default after power-up, or the value written in the REF_P registers (24-bit value).

To select REF_P as base value, and hence enable differential mode, bit AUTOZERO (bit[1] of CTRL_REG2, (0x21), which is a self-clearing bit) must be set: the current PRESS_OUT value is copied into REF_P registers, and from the next available measurement, the PRESS_OUT value is the difference between the current measure and the “frozen” reference value. In case the application requires to switch back to default mode, the bit RESET_AZ (bit[1] of CTRL_REG1 (0x20), which is a self-clearing bit) must be set. PRESS_OUT is then computed using RPDS registers value as reference.

RPDS_L and RPDS_H calibration values can be written on the volatile memory by the microcontroller after sensor power-on: the value shall be permanently stored in the application memory during manufacturing (see [Section 6.7](#)).

In special cases, there's also the possibility to store the RPDS calibration value inside the LPS25H non-volatile memory.

5 Using FIFO modes

5.1 Effective use of the FIFO

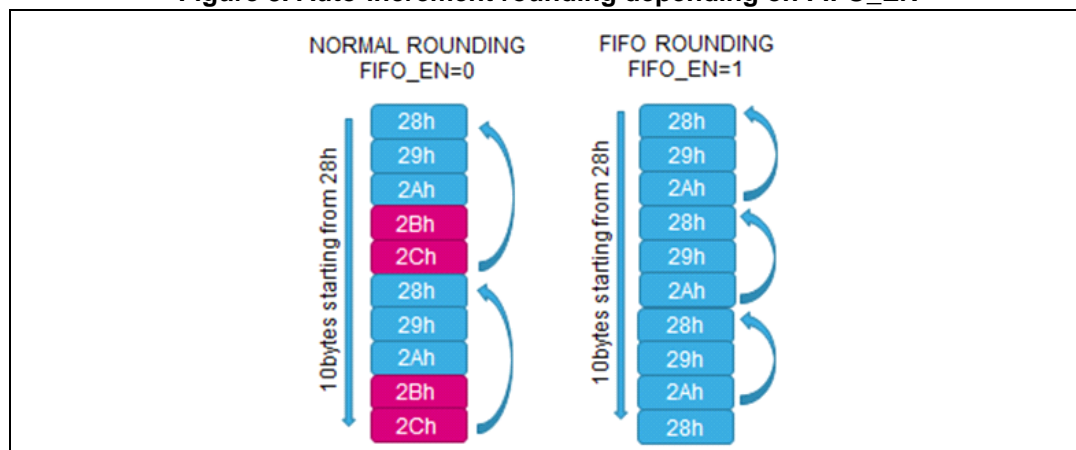
The LPS25H embeds a 32-slot x 24 bit FIFO for pressure data coming from the registers that normally feed PRESS_OUT (@ 0x28/0x29/0x2A). It allows lower frequency of serial bus transactions and provides more time to collect all measurements taken.

The FIFO buffer is enabled by setting FIFO_EN to 1 (bit[6] in CTRL_REG2) and can work in 4 different main modes: Bypass Mode, FIFO Mode, Stream Mode and FIFO Mean Mode. Each mode is selected by the 3 FIFO_MODE bits in FIFO_CTRL register.

Watermark level (P1_WTM), FIFO_empty (P1_EMPTY), FIFO_full (P1_Overrun) or DATA_Ready (P1_DRDY) events can be enabled (in register CTRL_REG4 (0x23)) to generate dedicated interrupts on the INT1 pin (provided that bit[1:0] of register CTRL_REG3 (0x22) are configured "00" to select Data Signal).

When FIFO is enabled temperature data must be read separately from pressure data. The temperature readout is always referred to the last pressure data in the FIFO, as no FIFO is available for temperature data. The register auto-increment "rounding" is then different depending on FIFO_EN status, as shown in the picture below.

Figure 8. Auto-increment rounding depending on FIFO_EN



To switch FIFO on/off, the bypass mode must be used. Switching the FIFO on/off only using the FIFO_en bit of register CTRL_REG2 would cause a wrong FIFO behavior.

5.1.1 Accessing the FIFO data

FIFO data is read through PRESS_OUT registers (0x28/0x29/0x2A). When FIFO is in stream, trigger (i.e. state change on trigger) or FIFO mode, a read operation to the PRESS_OUT registers provides the data stored in the FIFO. When in FIFO mean mode it is not possible to read the data in the buffer, only the final averaged value will be available in PRESS_OUT registers at the end of the process done with the selected speed (see [Section 5.1.5](#)).

Each time data is read from the FIFO, the oldest entry is placed in the PRESS_OUT registers and both single read and multiple read operations can be used.

The whole FIFO content can be read by reading 3x32 bytes from PRESS_OUT_XL location in a single I²C read transaction. Internally the reading address will automatically roll back from 0x2A down to 0x28 when FIFO is active to allow a quick read of its content.

5.1.2 Bypass mode (F_MODE[2..0]="000" in FIFO_CTRL (0x2E))

The FIFO is not operational and for this reason the buffer remains empty. The pressure value is sent directly to PRESS_OUT registers.

5.1.3 FIFO mode (F_MODE[2..0]="001" in FIFO_CTRL (0x2E))

The measurement from the sensor are sent to FIFO buffer, the FIFO content is read using the registers PRESS_OUT_XL (0x28), PRESS_OUT_L (0x29) and PRESS_OUT_H (0x2A).

An interrupt can be enabled (WTM_EN bit[5] in CTRL_REG2 (0x21) in order to be raised when the FIFO is filled to the level specified by the WTM_POINT[4..0] bits[4..0] in the FIFO_CTRL (0x2E) register. The FIFO continues filling until it is full (32 slots of data for XL, L and H). When full, the FIFO stops collecting incoming pressure measurements.

5.1.4 Stream mode (F_MODE[2..0]="010" in FIFO_CTRL (0x2E))

Like in FIFO mode the measurements are stored in the buffer before being available in PRESS_OUT_XL, PRESS_OUT_L and PRESS_OUT_H. On the contrary of FIFO mode, when full, the FIFO discards the older data as the new arrive. An interrupt can be enabled and set as in FIFO mode though the same register FIFO_CTRL (0x2E).

Stream mode can be used to implement a digital filter averaging the samples stored in the FIFO.

5.1.5 FIFO mean mode (F_MODE[2..0]="110" in FIFO_CTRL (0x2E))

This mode is used in order to enable the "low noise mode". In this mode the pressure data are stored in the FIFO after being averaged depending on AVGP[1..0] bit in register RES_CONF (0x10).

Pressure data stored in the FIFO buffer are further averaged, to implement a moving average, using a number of samples defined by WTM_POINT[3..0] in register FIFO_CTRL (0x2E) and with the speed defined by ODR[2..0] in register CTRL_REG1 (0x20) the result is placed in PRESS_OUT registers.

There are two possible ways to provide data output pressure averaged by FIFO:

1. Same data rate (ODR setting) of data coming from sensor when the FIFO_MEAN_DEC bit[4] of register CTRL_REG2 (0x21) is reset (0=Disable)
2. Decimated output at 1 Hz when the FIFO_MEAN_DEC bit[4] of register CTRL_REG2 (0x21) is set (1=Enable)

In case 1, we will have the output data averaged by the last samples defined by WTM_POINT[4..0] in register FIFO_CTRL (0x2E)

Table 3. Number of averaged samples vs WTM_POINT setting

WTM_POINT[4..0]	Number of averaged samples
00001	2
00011	4
00111	8
01111	16
11111	32

While the speed of new averaged output is the one defined by ODR[2..0] in register CTRL_REG1 (0x20).

25Hz for ODR=4

12.5Hz for ODR=3

7Hz for ODR=2

1Hz for ODR=1

In case 2, the number of samples is not selected by WTM_POINT[4..0] but is automatically set from ODR value.

Table 4. ODR setting in FIFO Mean mode decimated

ODR[2..0]	Data rate from sensor	Number of samples	Output data rate	Data ready every samples
4	25 Hz	32	1 Hz	25
3	12.5 Hz	16	1 Hz	12
2	7 Hz	8	1 Hz	7

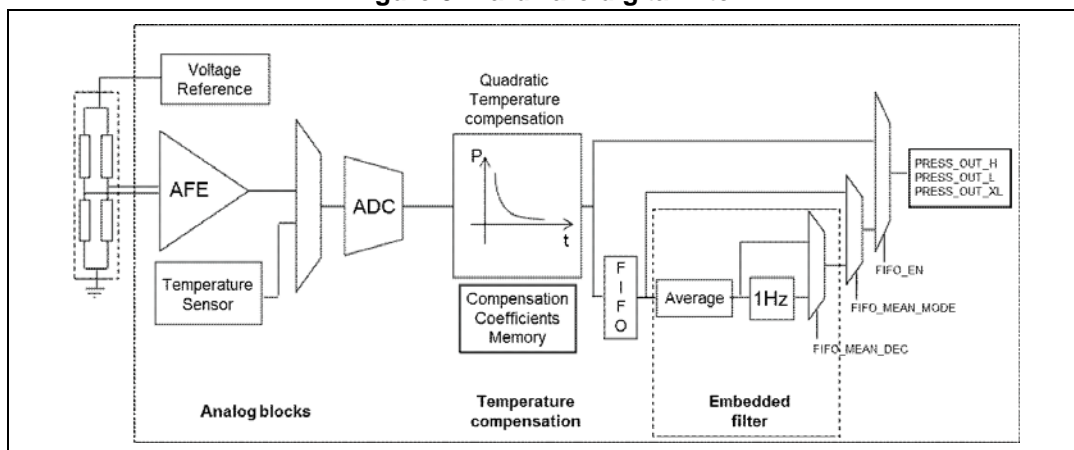
It is important to note that FIFO Mean mode with decimated output is not possible for ODR=1 (data rate at 1 Hz).

As soon as FIFO Mean mode is set, the buffer is cleaned.

5.1.6 Hardware digital filter

To achieve the final averaging, an HW digital filter is used.

Figure 9. Hardware digital filter



The digital filter reduces the pressure noise level to 0.010 hPa rms (1 Pa at 1 sigma) and allows to reduce the internal ADC HW average, reducing power consumption while keeping the same pressure noise level.

Filter enabling and suggested configuration

To reduce power consumption while keeping a low noise figure, the recommendation is to reduce the pressure and temperature averaging, reduce ODR to minimum and enable the digital filter (FIFO).

For example:

- RES_CONF(0x10) = 0x05 (set AVGT=16, AVGP=32, or less)
- FIFO_CTRL (0x2E) = 0xC1 (Set FIFO Mean mode with average on 2 samples or more, up to 0xDF)
- CTRL_REG2 (0x21) = 0x40 (FIFO enabled, decimation disabled)
- CTRL_REG1 (0x20) = 0x90 (ODR = 1 Hz, power-on device)

In this way, the power consumption at 1 Hz is reduced from the typical 26.4 μ A to about 4.5 μ A with a pressure noise of 0.01 hPa rms.

5.2 Extra FIFO modes

For all the following modes the trigger signal is the IA bit[2] of register INT_SOURCE (0x25) which is configured by register INTERRUPT_CFG (0x24).

5.2.1 Stream to FIFO mode (F_MODE[2..0]="011" in FIFO_CTRL (0x2E))

The FIFO works in Stream mode till a trigger event occurs, then it changes to FIFO mode.

5.2.2 Bypass to Stream mode (F_MODE[2..0]="100" in FIFO_CTRL (0x2E))

The FIFO is in Bypass mode, so it stays empty because it is not operational, till a trigger event occurs, then the FIFO enters in Stream mode.

5.2.3 Bypass to FIFO mode (F_MODE[2..0]="111" in FIFO_CTRL (0x2E))

The FIFO is in Bypass mode, so it stays empty because it is not operational, till a trigger event occurs, then the FIFO mode starts.

6 Software

A detailed register list is reported in the LPS25H datasheet. In This section we collect useful information on how to handle possible exceptions on the communication bus and how to properly use the information coming from the device.

6.1 Hints

1. If the I²C line has glitches and slaves are holding SDA low, a simple I²C error recovery would be to send 9 stop bits (which will flush the bus). This is usually done by SW using the I²C GPIOs resources of the host device. A good practice would be to check that SDA is high prior to generating a START bit.
2. The serial bus power consumption can be reduced by reducing the data traffic:
 - Using built-in HW averaging instead of SW averaging
 - Reduce the polling rate when waiting for the one-shot measurement completion ([Section 4.3](#))
 - The INT1 pin can be used as event to minimize serial bus polling

6.2 One-shot mode conversion time estimation

Typical conversion time $\approx 62 \cdot (P_{avg} + T_{avg}) + 975 \mu s$

- ex: $T_{avg} = 64$; $P_{avg} = 512$; Typ. conversation time ≈ 36.7 ms (compatible with ODT=25 Hz)
- ex: $T_{avg} = 32$; $P_{avg} = 128$; Typ. conversation time ≈ 10.9 ms
- The formula is accurate within +/- 3% at room temperature

6.3 Reference SW to get started with LPS25H

There are some examples of C source files that provide the baseline in initializing, configuring and performing measurements with the pressure sensor.

6.4 Pressure to altitude conversion

The simplest and widely used barometer (altitude) formula used in most watches comes from the US Standard Atmosphere, for example the 1976 edition.

Example of the source code is shown below:

```
void From_Pressure_mb_To_Altitude_US_Std_Atmosphere_1976_ft(double*
Pressure_mb, double* Altitude_ft) {
  //=(1-(Pressure/1013.25)^0.190284)*145366.45
  *Altitude_ft = (1-pow(*Pressure_mb/1013.25,0.190284))*145366.45;
}

void From_ft_To_m(double* ft, double* m) {
```

```
//=feet/3.280839895  
*m = *ft/3.280839895;  
}
```

Terminology is very delicate for altitude. There is not one definition of altitude:

1. Absolute pressure sensor measures the air pressure at the sensing point. The base rule of thumb is that the pressure drops by 1 hPa every 8.3 meters.
2. The “sea level pressure average” is 1013.25 hPa (standard altitude measurement done in most digital barometer wrist watches)
3. The “instant sea level pressure” is weather and location dependent and can vary by more than 5 hPa (compare SFO and SIN airports on the web)
4. Also beware that pressure altitude may be different than GPS altitude, which may also be different than density altitude.

Aircraft flying between airports use the ISA altitude which is based on yearly/earth-wide sea level pressure average value (as shown above) to avoid plane collision; this is the altimeter reference.

When aircraft approach an airport, local equivalent sea level and airfield temperature are wirelessly shared from airport to plane to compute local instant airfield altitude. This is known as QNH, using METAR data.

6.5 SW filtering & internal FIFO

If higher precision is required, or if the air pressure/flow is unstable, a SW filter could be implemented on the sensor measurements.

The internal FIFO buffer and HW filter can also be used, as indicated in previous sections; this is strongly suggested since it reduces the level of pressure noise and in the meantime decreases the power consumption.

For indoor navigation when sub meter detection is required, a special recursive filter could be used. Tailoring the SW filter for the sensor characteristic and the application requirement is usually recommended. Sensor fusion is the most advanced filtering which uses all available information and uses them to reduce the positioning noise.

6.6 Reset the device to power-on configuration

If there are any doubts regarding the register content, there is the possibility to reset the configuration of the device at the power-on condition setting both BOOT bit and SWRESET bit in CTRL_REG2 (0x21).

The bit[2] SWRESET resets all the non-trimming registers to their reset values and stops any internal state machines.

The bit[7] BOOT can be used, when the device is enabled, to reinitialize the volatile registers with the content of the internal non-volatile trimming memory. These values are factory trimmed and are different for every device. They permit good behavior of the device and normally they should not be changed. If, for any reason, the content of the trimming registers is modified, it is sufficient to use the BOOT bit to restore the correct values.

Follow this sequence:

1. Turn on BOOT bit (bit[7] = 1 of CTRL_REG2 (0x21))
2. Wait for the BOOT bit to reset (BOOT bit is self-clearing, it is set again to '0' by hardware at the end of the process)
3. Wait an additional 5 ms
4. Configure the registers
5. Power on the device for normal use

6.7 Absolute accuracy drift due to soldering

Reflow soldering may cause a spread of the device population accuracy. The spread is PCB construction, assembly and layout specific.

Beware that board warping may cause additional drift in the accuracy.

If very good absolute accuracy is required, a one-point calibration in the production line could be implemented.

6.7.1 Correcting soldering drift (one-point calibration)

The following is a brief guideline to reduce the impact of soldering.

1. Soldering drift is a complex process and it is not easy to identify the single root cause of soldering stress.
2. We define the soldering drift as the difference between the accuracy of the pressure sensor before and after soldering.
3. Soldering temperature profile is one of the major contributors to soldering shift.
4. A well-controlled temperature soldering profile, that avoids peak temperature over the max JEDEC spec can reduce the accuracy drift.

Here some hints on how to minimize post-soldering drift in SW:

1. Be sure to re-hydrate the device package by keeping it at 70%RH for 12 h or waiting a minimum of 36 h at ambient humidity.
2. After point 1, if a residual drift is still present and above the application intended limits, perform a one-point calibration by following these steps:
 - a) During board final test use a precision barometer to get the reference value and calculate the difference with respect to the sensor measurements (taken at maximum resolution)
 - b) Store the difference (offset) in the application non-volatile memory
 - c) Use the offset value to correct all measurements from the sensor
 - d) Correction can be performed entirely by SW, or the offset value (upper 16-bit only) can be written in dedicated registers of the sensor (RPDSx), at each power-on, so that the values read from the sensor are already corrected
 - e) When the device is calibrated after soldering as proposed above, thanks to the embedded quadratic compensation, the absolute accuracy is within +/-1 hPa
 - f) The SW offset correction could also take into account for sensor aging test results.

7 Revision history

Table 5. Document revision history

Date	Revision	Changes
29-Apr-2014	1	Initial release.

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