

The Laplace Transform

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Defn

Let $f(t)$ be a given function which is defined for all $t \geq 0$. We multiply $f(t)$ by e^{-st} and integrate w.r.t ' t ' from $0 \rightarrow \infty$. Then if the resulting integral exists, it is a function of s , say $F(s)$.

$$\therefore F(s) = \int_0^{\infty} e^{-st} f(t) dt \dots \dots \quad (1)$$

The function $F(s)$ is called Laplace Transform. The original function $f(t)$ in eqn (1) is called inverse transform.

In this discussion ie for signal, $F(s)$ is denoted by $X(s)$ and $f(t)$ by $x(t)$.

Laplace Transform Evaluations and Theorems

Ex. ①. Find the Laplace transform and abscissa of convergence for the signal $x(t) = e^{-\alpha t} u(t)$.

Ans: \Rightarrow Given $x(t) = e^{-\alpha t} u(t)$

here $u(t)$ is unit step signal.

$$\begin{aligned} \therefore L\{x(t)\} &= X(s) = \int_0^{\infty} e^{-\alpha t} \cdot u(t) e^{-st} dt \\ &= \int_0^{\infty} e^{-(s+\alpha)t} dt = \left[-\frac{e^{-(s+\alpha)t}}{s+\alpha} \right]_0^{\infty} \end{aligned}$$

$$\therefore X(s) = \frac{1}{s+\alpha}$$

Ex-② Find the Laplace transform of $u(t)$.

Ans:⇒

The unit step signal $u(t)$ is the special case of the signal $e^{-at} u(t)$, that occurs when $a=0$. Therefore $x(t) = u(t)$.

$$\therefore X(s) = \frac{1}{s} \text{ as } a=0$$

$$\begin{aligned} X(s) &= \int_0^{\infty} u(t) e^{-st} dt \\ &= \int_0^{\infty} e^{-st} dt \\ &= -\frac{1}{s} [e^{-st}]_0^{\infty} = \frac{1}{s} \end{aligned}$$

Ex-③ J-97 Find the Laplace Transform of the unit ramp signal.

$$\text{Ans:⇒ } x(t) = r(t) = t u(t)$$

$$\therefore L\{r(t)\} = X(s) = \int_0^{\infty} t u(t) e^{-st} dt$$

$$= \int_0^{\infty} t e^{-st} dt = \left[\frac{e^{-st}}{s} (-st - 1) \right]_0^{\infty}$$

$$= \left[0 - \left(-\frac{1}{s^2} \right) \right] = \frac{1}{s^2}$$

$$\begin{aligned} &\int_0^{\infty} t e^{-st} dt \\ &= \frac{-e^{-st}}{s} + \int \frac{e^{-st}}{s} dt \\ &= \left[\frac{t e^{-st}}{-s} - \frac{e^{-st}}{s^2} \right]_0^{\infty} \\ &= -\left(-\frac{1}{s^2} \right) = \frac{1}{s^2} \end{aligned}$$

Theorem① Linearity,

$$\text{if } x(t) \leftrightarrow X(s) \text{ and } y(t) \leftrightarrow Y(s)$$

$$\text{Then } a x(t) + b y(t) \leftrightarrow a X(s) + b Y(s).$$

Problem ④ Find the Laplace Transform of the signal $x(t) = \sin(\omega_0 t) \cdot u(t)$.

$$\text{Ans:⇒ Given } x(t) = \sin(\omega_0 t) u(t)$$

$$\therefore x(t) = \frac{1}{2j} e^{j\omega_0 t} u(t) - \frac{1}{2j} e^{-j\omega_0 t} u(t)$$

Now we know $e^{-\alpha t} x(t) \leftrightarrow \frac{1}{s+\alpha}$ ③

$$\therefore L\{x(t)\} = X(s) = \frac{1}{2j} \left[\frac{1}{s-j\omega_0} \right] - \frac{1}{2j} \left[\frac{1}{s+j\omega_0} \right]$$

$$= \frac{\omega_0}{s^2 + \omega_0^2}$$

Theorem ② Scale change

If $x(t) \leftrightarrow X(s)$ then $x(mt) \leftrightarrow \frac{1}{m} X(\frac{s}{m})$
for $m > 0$; $m \neq 0$ or $m \neq 0$.

J-97 7(d) Prove that if the LT of $x(t)$ is $X(s)$, then
LT of $x(mt)$ is $\frac{1}{m} X(\frac{s}{m})$ for $m > 0$. How is the result
modified if $m < 0$?

Ans: \Rightarrow we have $L\{x(t)\} = X(s)$

$$\text{Now } L\{x(mt)\} = \int_0^\infty e^{-st} x(mt) dt \\ = \frac{1}{m} \int_0^\infty e^{-s\gamma/m} x(\gamma) d\gamma; \text{ Putting } mt = \gamma.$$

$$= \frac{1}{m} \cdot \int_0^\infty e^{-s'/\gamma} x(\gamma) d\gamma \text{ where } s' = \frac{s}{m}$$

$$= \frac{1}{m} X(s') = \frac{1}{m} X\left(\frac{s}{m}\right). \text{ Proved.}$$

Now when $m < 0$ i.e. m is -ve, it would result
in a reversal of $x(t)$, meaning $x(mt)$ would be non-
zero only for $t < 0$, since $x(t) = 0$ for $t < 0$. Thus

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the Laplace Transform of $x(mf)$ is $X(s) = 0$ when $m < 0$ and the scaling theorem is not valid.

Ex. 5. Find the Laplace transform of the signal $y(t) = \pi(3t)$.

Solⁿ

$$\text{Let } z(t) = \pi(t)$$

$$z(s) = \frac{1}{s^2}$$

Thus the scale change theorem gives

$$x(s) = \frac{1}{3} z\left(\frac{s}{3}\right) = \frac{1}{3} \left[\frac{1}{\left(\frac{s}{3}\right)^2} \right] = \frac{3}{s^2}$$

$$\text{2nd process} \Rightarrow y(t) = \pi(3t) = 3t u(3t)$$

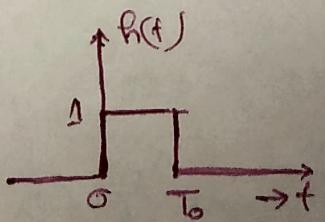
$$= 3t u(t) = 3 \pi(t) \leftrightarrow \frac{3}{s^2}$$

Theorem-3 Time Delay:

If $x(t) \leftrightarrow X(s)$ and $x(t) = 0$ for $t < 0$

then $x(t - t_0) \leftrightarrow X(s) \cdot e^{-st_0}$ for $t_0 > 0$

Ex. 6. A zero-order hold system has the impulse response shown in fig. Find the LT of this impulse response.



Solⁿ From fig.

$$h(t) = u(t) - u(t - T_0)$$

We use the time delay and linearity

theorems and the transform $u(t) \xrightarrow{5} \frac{1}{s}$ to give

$$\begin{aligned} L\{h(t)\} &= H(s) = \frac{1}{s} - \frac{1}{s} \cdot e^{-sT_0} \\ &= \frac{1 - e^{-sT_0}}{s} \end{aligned}$$

This system is called hold system. It holds the value of the strength of an impulse for time interval T_0 .

Theorem 4 s-shift, if $x(t) \leftrightarrow X(s)$. Then

$$e^{-kt} x(t) \leftrightarrow X(s+k); k \text{ is any real or complex number.}$$

Proof \Rightarrow

$$\begin{aligned} L\{e^{-kt} x(t)\} &= \int_0^\infty e^{-st} \cdot e^{-kt} x(t) dt \\ &= \int_0^\infty e^{-(s+k)t} x(t) dt = X(s+k) \end{aligned}$$

Theorem 5 t-multiplication, if $x(t) \leftrightarrow X(s)$

$$\text{then } t x(t) \leftrightarrow - \frac{d X(s)}{ds}$$

Ex.7. Find the Laplace Transform of the signal
 $x(t) = f e^{-at} u(t)$.

Soln. Let $y(t) = e^{-at} u(t)$

$$\therefore Y(s) = \frac{1}{s+a}$$

$$\therefore L\{x(t)\} = X(s) = - \frac{d}{ds} \left[\frac{1}{s+a} \right] = + \frac{1}{(s+a)^2}$$

Note. Extending theorem 5 by induction gives

$$t^n x(t) \leftrightarrow (-1)^n \cdot \frac{d^n [X(s)]}{ds^n} \quad \checkmark$$

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Theorem-6Time Differentiation:If $x(t) \leftrightarrow X(s)$

then $\frac{d^n x(t)}{dt^n} \leftrightarrow s^n X(s) - \sum_{i=0}^{n-1} s^{n-1-i} x^{(i)}(0^-)$

where $x^{(i)}(0^-) = \left. \frac{d^i x(t)}{dt^i} \right|_{t=0^-}$

Thus $L\{x^{(1)}(t)\} = sX(s) - x(0^-)$

$$L\{x^{(2)}(t)\} = s^2 X(s) - s x(0^-) - x'(0^-)$$

Ex.8. Find the Laplace Transform of the derivative of the signal $x(t) = e^{-t} u(t)$. Invert the transform to find the signal derivative.

Soln. By the time differentiation theorem,

$$L\{x^{(1)}(t)\} = sX(s) - x(0^-)$$

$$\text{Now } x(t) = e^{-t} u(t)$$

$$\therefore X(s) = \frac{1}{s+1}$$

and $x(0^-) = 0$ because $u(t) = 0, t < 0$

Therefore,

$$L\{x^{(1)}(t)\} = \frac{s}{s+1} - 0$$

$$= \frac{s+1-1}{s+1} = 1 - \frac{1}{s+1}$$

By using transform —, we find that

the derivative of the signal

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$$x'(t) = \delta(t) - e^{-t} u(t).$$

Theorem (7) Time integration :-

If $x(t) \leftrightarrow X(s)$

$$\text{then } y(t) = \int_0^t x(\tau) d\tau + y(0) \leftrightarrow \frac{X(s)}{s} + \frac{Y(s)}{s}$$

* Find the Laplace Transform of $y(t) = \int_0^t x(\tau) d\tau + y(0)$

when $x(t) = r(t)$ and $y(0) = 2$.

Soln

$$\text{Given } x(t) = r(t)$$

$$\therefore X(s) = \frac{1}{s^2}$$

thus time integration theorem gives

$$Y(s) = \frac{Ys^2}{s} + \frac{2}{s} = \frac{1}{s^3} + \frac{2}{s} = \frac{2s^2 + 1}{s^3}$$

Inverse Laplace Transform Evaluations :-

Ex.① Find the signal $y(t)$, the Laplace transform of which is $y(s) = \frac{s^3 + 7s^2 + 18s + 20}{s^2 + 5s + 6}$.

Soln

$$\text{we have } Y(s) = \frac{s^3 + 7s^2 + 18s + 20}{s^2 + 5s + 6}$$

$$\therefore Y(s) = s+2 + \frac{2s+8}{(s+3)(s+2)}$$

$$\therefore Y(s) = s+2 + 4\left(\frac{1}{s+2}\right) - 2\left(\frac{1}{s+3}\right)$$

$$\therefore y(t) = \delta'(t) + 2\delta(t) + 4e^{-2t}u(t) - 2e^{-3t}u(t)$$

Different Types :-

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Case I Exponential in the Numerator

Ex. ② Find $x(t)$ that corresponds to

$$X(s) = \frac{s^2 e^{-2s} + e^{-3s}}{s(s^2 + 3s + 2)}$$

Soln First we separate the transform into a sum of rational functions multiplied by exponentials in s

$$\begin{aligned} X(s) &= \left[\frac{s}{s^2 + 3s + 2} \right] e^{-2s} + \left[\frac{1}{s(s^2 + 3s + 2)} \right] e^{-3s} \\ &= \left[2\left(\frac{1}{s+2}\right) - \left(\frac{1}{s+1}\right) \right] e^{-2s} + 0.5\left(\frac{1}{s}\right) e^{-3s} \\ &\quad + 0.5 \left[\frac{1}{s+2} \right] e^{-3s} - \left[\frac{1}{s+1} \right] e^{-3s} \end{aligned}$$

$$\begin{aligned} \therefore x(t) &= 2 e^{-2(t-2)} u(t-2) - e^{-(t-2)} u(t-2) \\ &\quad + 0.5 u(t-3) + 0.5 e^{-2(t-3)} u(t-3) - e^{-(t-3)} u(t-3) \end{aligned}$$

Here we have applied time shift and linearity theorems.

PERIODIC FUNCTION

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Dec' 97

9(a)

Given a periodic function $f(t)$ with period T and $\int_0^T f(t) e^{-st} dt = F_1(s)$. Show that the Laplace Transform of $f(t)$ is $F(s) = \frac{F_1(s)}{1-e^{-sT}}$.

Soln:

Let $f(t)$ be a periodic function, defined for all +ve 't' and has the period T .

$$\therefore f(t+nT) = f(t) \quad \text{for all } t \geq 0 \\ n = \text{any integer}$$

If $f(t)$ is piecewise continuous on an interval of length T , then its Laplace transform exists, and we can write the integral from zero to infinity as the series of integrals over successive periods.

$$\therefore L\{f(t)\} = \int_0^\infty e^{-st} f(t) dt = \int_0^T e^{-st} f(t) dt + \int_T^{2T} e^{-st} f(t) dt + \int_{2T}^{3T} e^{-st} f(t) dt + \dots$$

If we substitute $t = \gamma + T$ in the second integral, $t = \gamma + 2T$ in the 3rd integral ..., $t = \gamma + (n-1)T$ in the n th integral ..., then the new limits in every integral are zero and T .

Since $f(\gamma+T) = f(\gamma)$, $f(\gamma+2T) = f(\gamma)$ etc.

We can write —

$$F(s) = \int_0^T e^{-sT} f(\gamma) d\gamma + \int_0^T e^{-s(\gamma+T)} f(\gamma) d\gamma + \dots$$

$$+ \int_0^T e^{-s(\tau+2T)} f(\tau) d\tau + \dots$$
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The factors which do not depend on τ can be taken out from under the integral signs; this gives

$$F(s) = [1 + e^{-sT} + e^{-2sT} + e^{-3sT} + \dots] \cdot \int_0^T e^{-sr} f(r) dr$$

$$\therefore F(s) = \left[\frac{1}{1 - e^{-sT}} \right] \int_0^T e^{-st} f(t) dt$$

$$\boxed{\therefore F(s) = \frac{F_i(s)}{1 - e^{-sT}}}$$

Proved.

Laplace Transform Solutions of linear intego-differential equations: —

Ex. ① A second order control system is characterised by the differential equation

$$y''(t) + 5y'(t) + 6y(t) = x(t).$$

Solve for $y(t)$ for $t \geq 0$ when $x(t) = u(t)$ and the initial conditions are $y(0^-) = 2$ and $y'(0^-) = -12$.

Soln:

Using the time differentiation and linearity theorems, we find that the Laplace transform of the differential equation is

$$s^2 Y(s) - s y(0^-) - y'(0^-) + 5[s y(s) - y(0^-)] + 6y(s) = \frac{1}{s}$$

Since $u(t) \leftrightarrow \frac{1}{s}$, substituting the initial conditions and rearranging yields

$$(s^2 + 5s + 6)Y(s) = (2s^2 - 2s + 1)/s.$$

$$\therefore Y(s) = \frac{(2s^2 - 2s + 1)}{s(s+3)(s+2)}$$

$$\therefore Y(s) = \frac{1}{6} \left[\frac{1}{s} \right] - \frac{13}{2} \left[\frac{1}{s+2} \right] + \frac{25}{3} \left[\frac{1}{s+3} \right]$$

∴ The system response is

$$y(t) = \frac{1}{6} u(t) - \frac{13}{2} e^{-2t} u(t) + \frac{25}{3} e^{-3t} u(t)$$