

Consider all voltage and current sources to be ideal.

- The frequency of the sinusoidal voltage source in the circuit of fig. 3.1 is adjusted until the amplitude of the sinusoidal output voltage (v_o) is maximum. The maximum amplitude of the voltage source (v_i) is 600V. (a) Find ω_s , (b) Amplitude of v_o at ω_s , (c) The bandwidth of the circuit, (d) The Q of the circuit, and the frequencies at which the amplitude of v_o is $0.707\{v_o\}_{\max}$.

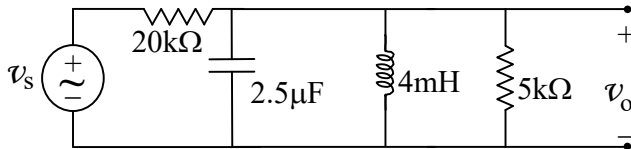


Figure 3.1

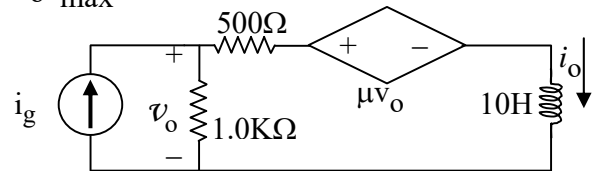


Figure 3.2

- Find the transfer function i_o/i_g as a function of μ for the circuit shown in fig. 3.2. Hence find i_o at $\mu=1$ and $\mu=2.5$.
- Find R, L, and C of a series RLC resonant circuit such that it resonates at $f_0=500\text{MHz}$ and has a bandwidth of $\Delta f=500\text{kHz}$. The constraints are that the maximum current through R is 17.32mA and the maximum power dissipation is 62.84mW.
- The peak amplitude of the sinusoidal voltage source in the circuit shown in fig. 3.3 is $100\sqrt{2}\text{ V}$ and its period is $200\pi\text{ }\mu\text{s}$. The load resistor can be varied from 0 to 300Ω , and the load capacitor can be varied from 1 to $4\mu\text{F}$. Determine the settings of R_L and C_L that will result in the most average power being transferred to R_L . Calculate the average power dissipated in R_L .

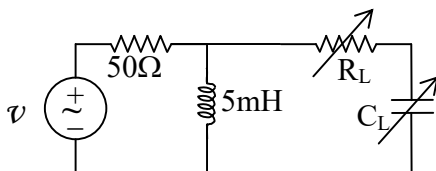


Figure 3.3

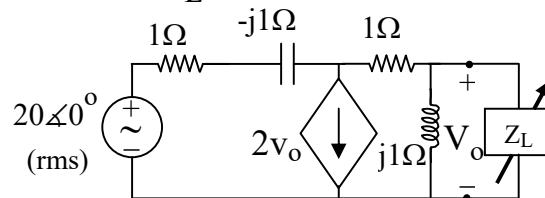


Figure 3.4

- In fig. 3.4 the load impedance is varied till maximum average power is delivered to Z_L . What is this maximum average power? What percentage of the total power developed in the circuit is delivered to Z_L ?
- A machine is running at a load of 1200kW at a power factor (p.f.) of 0.8 lag. This p.f. needs to be improved to a p.f. of 0.96 lagging by adding an extra load of 300kW to the real power load. (a) Find the Reactive Power of the added load. (b) What is its p.f.? (c) If the input voltage to the machine is always maintained at 3kV, then what is the rms current drawn by the machine, before and after addition of the 300kW load?
- Find the numerical expression for the transfer function $H(j\omega) = v_o/v_i$ for the circuit shown in fig. 3.5. Give the numerical value of each pole and zero of $H(j\omega)$.

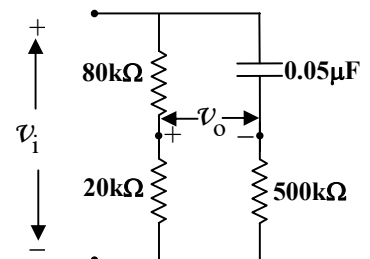


Figure 3.5

8. The numerical expression for a transfer function is $H(j\omega) = \{10^5(j\omega + 5)\} / [(j\omega + 100)(j\omega + 500)]$.

On the basis of a straight-line approximation of $H(j\omega)$ versus ω , estimate (a) the maximum $|H(j\omega)|$ in dB and (b) the value of $\omega > 0$ where the $|H(j\omega)|$ equals unity.

9. In the circuit shown in fig. 3.6 find the transfer function $H(j\omega) = v_o/v_i$. Sketch $\angle H(j\omega)$ versus ω .

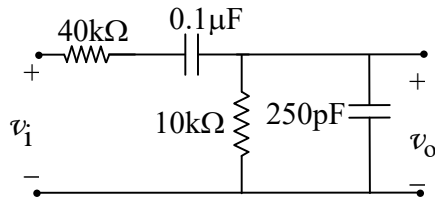


Figure 3.6

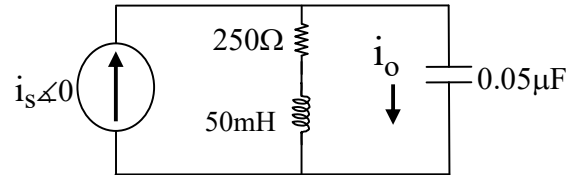


Figure 3.7

10. Derive the transfer function $H(j\omega) = i_o/i_s$ of the circuit shown in fig. 3.7. Sketch the asymptotic $|H(j\omega)|$ as a function of ω and find the bandwidth of the circuit.

1/ (a) This is a parallel resonant circuit.

$$\therefore \omega_0^2 = \frac{1}{LC} = \frac{1}{4 \times 10^{-3} \times 2.5 \times 10^{-6}} = \frac{1}{10 \times 10^{-9}} = 10^8, \omega_0 = 10^4 \text{ rad/s.}$$

$\omega_s = \omega_0$ when V_o is maximum, as the $Z_{eq} = Z_C \parallel Z_L \xrightarrow{\omega_s = \omega_0} \infty$

\therefore All current sent by V_s goes through the $5k\Omega$ resistance.

$$\therefore \omega_s = 10^4 \text{ rad/s.} \quad \& \quad f_s = \frac{10^4}{2\pi} = 1.59 \text{ kHz.}$$

(b) At $\omega_s = \omega_0$ (Resonance of $L \parallel C$) the load to V_s is completely resistive. If $V_s = V_{sm} e^{j\omega t}$ & $V_o = V_{om} e^{j\omega t}$

$$\text{Then } V_{om} = V_{sm} \times \frac{5k}{20k + 5k} \quad \& \quad V_{sm} = 600V \text{ given.}$$

$$\therefore V_{om} = \frac{600 \times 5}{25} = 120V.$$

$$(c) \omega_1 = -\frac{1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 - \left(\frac{1}{LC}\right)^2}, \quad \omega_2 = +\frac{1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 - \left(\frac{1}{LC}\right)^2} \text{ at } 0.707 V_m$$

$\therefore \Delta\omega = \omega_2 - \omega_1 = \frac{1}{RC}$, But, Here there is also a $20k\Omega$ other than the $5k\Omega$ parallel to the $L \parallel C$ combination.

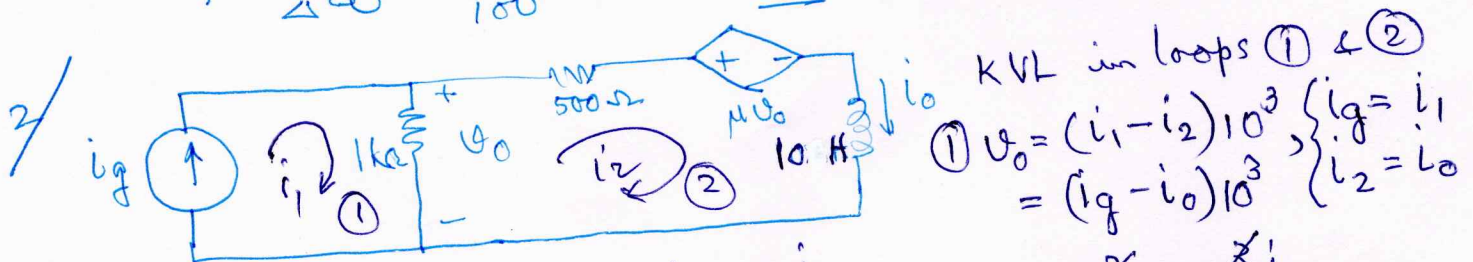
\therefore One need to consider the R_{eq} across $L \parallel C$ combination

$$\therefore R_{eq} = 20k\Omega \parallel 5k\Omega = \frac{20 \times 5}{25} = 4k\Omega$$

$$\therefore \Delta\omega = \frac{1}{4k\Omega \times 2.5\mu} = \frac{1000}{10} = 100 \text{ rad/s} \quad \left\{ \begin{array}{l} \text{Let's call this} \\ \text{the Bandwidth} \\ = \frac{100}{2\pi} = 16 \text{ Hz} \end{array} \right.$$

should give a very high Q.

$$(d) Q = \frac{\omega_0}{\Delta\omega} = \frac{10^4}{100} = 10^2 = 100$$



$$\text{② } V_o = 0.5 \times 10^3 i_o + \mu V_o + j\omega \times 10 i_o$$

$$\text{or } 0.5 \times 10^3 i_o + \frac{\mu V_o}{10^3} + \frac{j\omega 10 i_o}{10^2} = i_g \times 10^3 - 10^3 i_o$$

$$\text{or } i_o + 0.5 i_o + 0.01 j\omega i_o + 10^{-3} \mu (i_g - i_o) 10^3 = i_g$$

$$(1.5 - \mu + 0.01 j\omega) i_o = i_g (1 - \mu)$$

$$\therefore \frac{i_o}{i_g} = \frac{1 - \mu}{1.5 - \mu + 0.01 j\omega}$$

$$\therefore i_o|_{\mu=1} = 0, \quad \& \quad i_o|_{\mu=2.5} = \frac{-1.5}{-1 + 0.01 j\omega}$$

$$i_o|_{\mu=2.5} = \frac{1.5}{1 - 0.01 j\omega} = \frac{1.5 / \tan^{-1} \omega}{\sqrt{1 + 10^{-4} \omega^2}}$$

3/ Series Resonant circuit. $\therefore \omega_0 = \frac{1}{\sqrt{LC}}$, $Q = \frac{\omega_0 L}{R} = \frac{\omega_0}{\Delta\omega}$

$f_0 = 500 \text{ MHz} \therefore \omega_0 = 2\pi f_0 = 3.142 \text{ GRad/s}$

$\Delta f = 500 \text{ kHz} \therefore \Delta\omega = 2\pi \Delta f = 3.142 \text{ MRad/s}$, If the voltage across R is supposed to be Max at resonance

$P_{R_{\text{max}}} = I_{R_{\text{max}}}^2 \times R$ or $R = \frac{62.84 \text{ mW}}{(17.32)^2 \mu\text{A}^2} = 209.5 \Omega$ the largest R is used

$Q = \frac{3.142 \times 10^9}{3.142 \times 10^6} = 10^3$ (very high) $\Rightarrow \frac{3.142 \times 10^9 \times L}{209.5}$

$\therefore L = \frac{10^3 \times 209.5}{3.142 \times 10^9} = 66.7 \mu\text{H}$ obvious from Δf being very narrow.

$C = \frac{1}{\omega_0^2 L} = \frac{1}{(3.142)^2 \times 10^{18} \times 66.7 \times 10^{-6}} = 1.5 \times 10^{-15} = 1.5 \text{ fF}$

$\therefore R, L, \& C$ found.

4/ $T = 200 \pi \mu\text{s} \therefore \omega = \frac{2\pi}{T} = \frac{2\pi}{200 \pi \times 10^{-6}} = \frac{1}{10^{-4}} = 10^4 \text{ rad/s}$

Impedance of L is $j\omega L = j \times 10^4 \times 5 \times 10^{-3} = j50 \Omega$

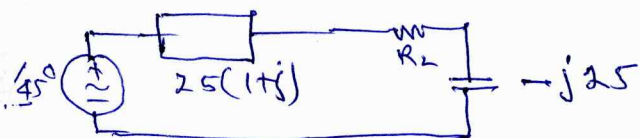
The voltage drop across the $R_L + \frac{1}{j\omega C_L}$ is divided across the two components, but for maximum drop across R_L choose $C_{L_{\text{max}}}$

$\therefore -jX_{C_L} = \frac{1}{j\omega C_L} = \frac{-j}{10^4 \times 4 \times 10^{-6}} = -j25 \Omega$

Compute the Thevenin equivalent of all components seen by R_L, C_L combination. $\therefore Z_{Th} = \frac{50 \times (j50)}{50 + j50} = \frac{j50}{1+j} = \frac{j50(1-j)}{(1+j)(1-j)} = \frac{50(1+j)}{2} = 25(1+j)$

$V_{Th} = \frac{V \times (j50)}{50 + j50} = V \cdot \frac{j(1-j)}{2} = \frac{V}{2}(1+j)$

using max of V , $V_{Th_{\text{max}}} = \frac{100\sqrt{2}}{2}(1+j) = \frac{100\sqrt{2}}{2} \angle 45^\circ = 100 \angle 45^\circ$



Obviously resonance happens at $\omega = 10^4 \text{ rad/s}$ as $j25 + (-j25) = 0$

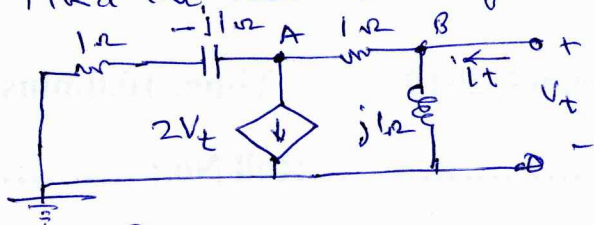
$\therefore I_{R_{L_{\text{max}}}} = \frac{100 \angle 45^\circ}{25 + R_L}$

$= \frac{100 \angle 45^\circ}{25 + 300} = 0.31 \angle 45^\circ$

Again to have bigger drop across R_L as compared to C_L choose the max. $R_L = 300 \Omega$.

$P_{R_{AV}} = R_L \frac{I_{R_{L_{\text{max}}}^2}}{2} = \frac{R_L}{2} I_{R_{L_{\text{max}}} \times I_{R_{L_{\text{max}}}^*} = \frac{300}{2} \times 0.31^2 = 14.2 \text{ W}$

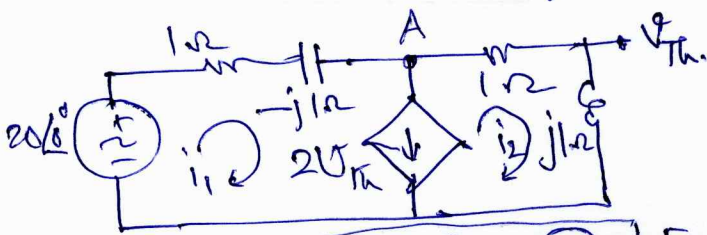
5/ Find the Thevenin equivalent across Z_L . Equ. ckt. is



KCL at (A): $\frac{V_A}{1-j} + 2V_t + \frac{V_A - V_t}{1} = 0$
 $\Rightarrow V_A + 2(1-j)V_t - (1-j)V_t + (1-j)V_A = 0$
 $\Rightarrow (2-j)V_A = -(1-j)V_t$
 $\therefore V_A = -\frac{(1-j)}{2-j} V_t$

$\frac{V_t - V_A}{1} + \frac{V_t}{j} - i_t = 0$ or $V_t + \frac{V_t}{j} + \frac{(1-j)}{2-j} V_t = i_t$

$\therefore V_t \left(1 - j + \frac{(1-j)}{2-j} \right) = i_t \therefore \frac{V_t}{i_t} = \frac{2-j}{-3j+1-j} = \frac{2-j}{1-4j}$



$Z_{Th} = \frac{1}{\frac{1}{\sqrt{4+1}} \angle 36.9^\circ} = 0.4 + j0.3$
 $Z_L = Z_{Th}^* = 0.4 - j0.3$

KVL in the loops (1) & (2) For max Power Transfer $Z_L = Z_{Th}^*$

Loop (2) $\frac{V_{Th}}{j1} = i_2$, $20\angle 0^\circ = 20 = i_1 - j1i_1 + V_A$, $i_1 - i_2 = 2V_{Th}$
 $V_A = V_{Th} + 1\Omega \times i_2 \therefore 20 = (1-j)i_1 + V_{Th} + i_2$

$i_1 - \frac{V_{Th}}{j} = 2V_{Th}$ & $20 = (1-j)i_1 + V_{Th} + \frac{V_{Th}}{j}$
 $\therefore i_1 = (2-j)V_{Th}$ or $20 = (1-j)(2-j)V_{Th} + (1-j)V_{Th}$

$\therefore V_{Th} = \frac{20}{2(1-j)} = \frac{10}{\sqrt{2}} \angle 63.4^\circ$
 $= 1.97 + j4.0 \text{ (rms) V.}$

$\therefore I_{Load} = \frac{10 \angle 63.4^\circ}{\sqrt{5}} / 0.8 = 5.6 \angle 63.4^\circ$

$P_{load} \Big|_{\text{at max Power transfer}} = |I_{Load}|^2 \times 0.4 = 12.54 \text{ W}$

6/ a) Complex Power

$$S = P + jQ$$

$$= \frac{V_m I_m}{2} \cos \theta + j \frac{V_m I_m}{2} \sin \theta$$

$$= 1200k + j \frac{1200k}{0.8} \times \sin 36.87^\circ \text{ VA}$$

$$= 1200k + j900k \text{ VA}$$

After improving P.f.

$$S_i = (1200 + 300)k + j \frac{1500k}{0.96} \times \sin(16.26^\circ)$$

$$= 1500k + j \frac{1500k}{0.96} \times 0.28 \text{ VA} = (1500 + j437.5) \text{ kVA}$$

$$\therefore Q_{\text{load}} = (437.5 - 900) \text{ kVAR} = \underline{\underline{-462.5 \text{ kVAR}}}$$

$$b) S_{\text{load}} = 300k - j462.5k = 551.1k \angle -57^\circ \text{ VA.}$$

$$\text{P.f.}_{\text{load}} = \cos 57^\circ = \underline{\underline{0.545 \text{ leading}}}$$

$$c) 3k \times I_{\text{machine}}^* = 1200k + j900k$$

$$\text{or } I_{\text{machine}}^* = 400 + j300 \text{ or } I_{\text{machine}} = 400 - j300 \text{ A}$$

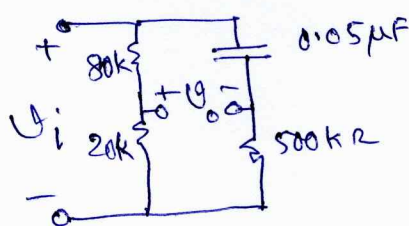
$$= 500 \angle -36.9^\circ$$

$$\therefore I_{\text{machine}}|_{\text{rms}} = 500 \text{ A.}$$

$$d) 3k \times I_{\text{machine}}'^* = 1500k + j437.5k \left\{ \begin{array}{l} I_{\text{ma}}' = 500 - j145.8 \\ I_{\text{rms}}' = 521 \angle -16.26^\circ \end{array} \right.$$

$$I_{\text{machine}}'^* = 500 + j145.8$$

7/ a)



$$V_{o+} = \frac{V_i \times 20}{80 + 20} = 0.2 V_i$$

$$V_{o-} = \frac{V_i \times 500k}{\frac{106}{j\omega \times 0.05} + 500k} = \frac{V_i}{1 + \frac{106}{j\omega \times 0.05 \times 500k}}$$

$$V_o = V_{o+} - V_{o-} = 0.2 V_i - \frac{V_i}{1 - j\frac{40}{\omega}}$$

$$= \frac{V_i}{1 - j\frac{40}{\omega}}$$

$$\frac{V_o}{V_i} = 0.2 - \frac{1}{1 - j\frac{40}{\omega}} = \frac{0.2 - j\frac{8}{\omega} - 1}{1 - j\frac{40}{\omega}} = \frac{-0.8 - j\frac{8}{\omega}}{1 - j\frac{40}{\omega}}$$

$$H(\omega) = \frac{-0.8 \times (1 + j\frac{10}{\omega})}{(1 - j\frac{40}{\omega})} = \frac{-0.8(j\omega - 10)}{(j\omega + 40)} \therefore H(s) = \frac{-0.8(s - 10)}{(s + 40)}$$

$$b) \therefore s = 10 \text{ for } H(s) = 0 \therefore \text{Zero is at } 10 \text{ rad/s.}$$

$$s = -40 \text{ for } H(s) = \infty \therefore \text{Pole is at } -40 \text{ rad/s.}$$

$$\text{where } s = j\omega$$

$$8/ \quad H(j\omega) = \frac{10^5 (j\omega + 5)}{(j\omega + 100)(j\omega + 500)} = \frac{10^5 \times 5}{100 \times 500} \times \frac{(1 + j\frac{\omega}{5})}{(1 + j\frac{\omega}{100})(1 + j\frac{\omega}{500})}$$

$$|H(\omega)|_{dB} = 20 \log 10 + 20 \log |1 + j\frac{\omega}{5}| - 20 \log |1 + j\frac{\omega}{100}| - 20 \log |1 + j\frac{\omega}{500}|$$

$$20 \log 10 = 20 \text{ dB. So for } \omega \ll 5 \ll 100 \ll 500 \text{ then } |H(\omega)|_{dB} = 20 \text{ dB}$$

At $\omega = 5$ there would be a corner frequency and the asymptotic line would increase at $+20 \text{ dB/decade}$.

At $\omega = 100$ corner frequency is at 100 and it starts decreasing at -20 dB/decade .

At $\omega = 500$ corner frequency is at 500 and it goes down at -20 dB/decade .

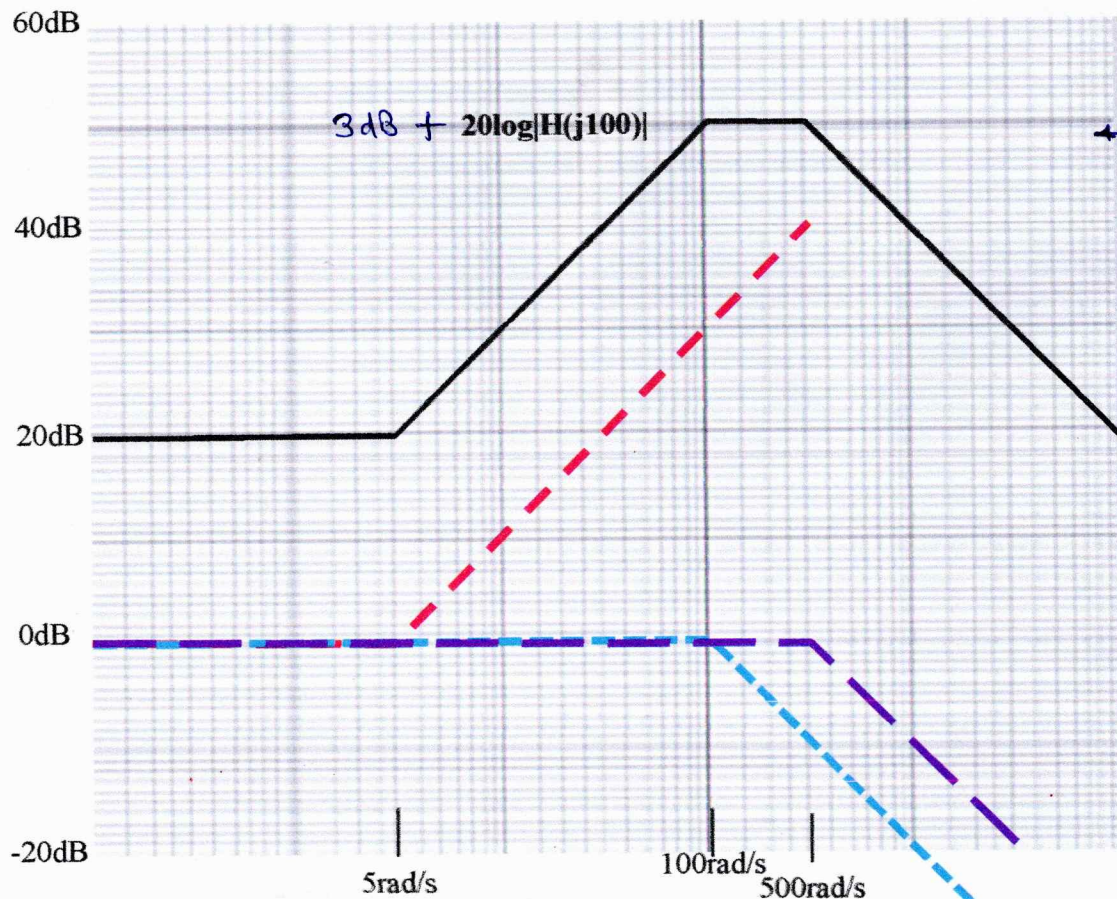
$$H(j100) = \frac{10(1 + j\frac{100}{5})}{(1+j)(1 + j\frac{100}{500})} = \frac{10(1 + j20)}{(1+j)(1 + 0.2j)} \approx \frac{10 \times 20 \angle 87.14^\circ}{\sqrt{2} \angle 45^\circ 1.02 \angle 11.3^\circ}$$

$$= 138.85 \angle 30.84^\circ$$

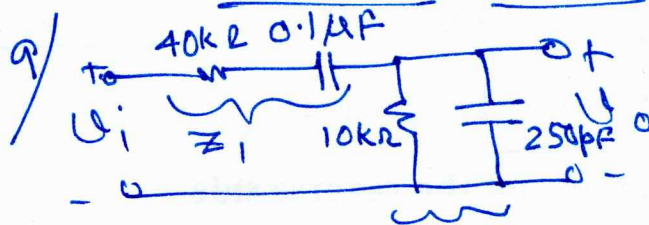
$$20 \log |H(j100)| = 20 \log 138.85 = 42.8 \text{ dB.}$$

$$20 \log |H(j500)| \approx 20 \log \left| \frac{10 \times (100)}{(1+j5)(1+j1)} \right| \approx 20 \log \left| \frac{10 \times (100)}{5 \cdot 1 \times \sqrt{2}} \right| = 42.84 \text{ dB.}$$

\therefore Corners would be $+3 \text{ dB}$ above this value $\approx \underline{46 \text{ dB}}$



should have been 46 dB
But due to bad graph paper, looks more like 30 dB



Looks like the interstage coupling circuit of an amplifier.

$$Z_1 = 40k - \frac{j \times 10^6}{0.1\omega}$$

$$= -j \frac{10^7}{\omega} + 4 \times 10^4$$

$$= 4 \times 10^4 \left(1 - j \frac{250}{\omega} \right)$$

$$(s = j\omega) \Rightarrow 4 \times 10^4 \left(1 + \frac{250}{s} \right)$$

$$Z_2 = \frac{-j10k / 250 \times 10^{-12} \times \omega}{10k + j \frac{1}{250 \times 10^{-12} \omega}}$$

$$= \frac{10^{12} / 250 \omega}{j + \frac{10^{12}}{250 \times 10^4 \omega}}$$

$$= \frac{4 \times 10^9}{4 \times 10^5 + j\omega}$$

$$\Rightarrow \frac{4 \times 10^9}{s \left(1 + \frac{4 \times 10^5}{s} \right)}$$

$$\frac{U_o}{U_i} = H(s) = \frac{Z_2}{Z_1 + Z_2}$$

$$= \frac{\frac{4 \times 10^9}{s \left(1 + \frac{4 \times 10^5}{s} \right)}}{4 \times 10^4 \left(1 + \frac{250}{s} \right) + \frac{4 \times 10^9}{s \left(1 + \frac{4 \times 10^5}{s} \right)}}$$

$$= \frac{\frac{10^5 s}{(s + 4 \times 10^5)}}{(s + 250) + \frac{10^5 s}{(s + 4 \times 10^5)}}$$

$$= \frac{10^5 s}{s^2 + 4 \times 250 \times 10^5 + 4 \times 10^5 s + 250 s + 10^5 s}$$

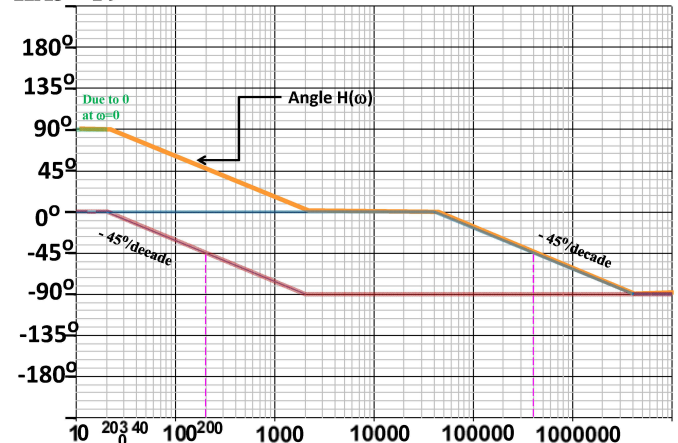
$$= \frac{10^5 s}{s^2 + (5 \times 10^5 + 250) s + 10^8}$$

$$= \frac{10^5 s}{(s + 199.98)(s + 500050)}$$

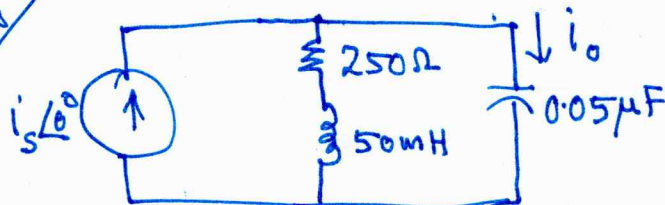
\therefore Poles at $\omega_1 = -199.98 \text{ rad/s}$
and $\omega_2 = -500,050 \text{ rad/s}$

Zero at $\omega = 0 \rightarrow \angle 90^\circ$

HA3 - P9



10



Parallel \therefore use Admittance.

$$Y_s = \frac{1}{R + j\omega L} + j\omega C$$

$$= \frac{R}{R^2 + (\omega L)^2} - \frac{j\omega L}{R^2 + (\omega L)^2} + j\omega C$$

At resonance $\frac{\omega_0 L}{R^2 + (\omega_0 L)^2} = \omega_0 C$

or $L = R^2 C + \omega_0^2 L^2 C$ or $1 = \frac{R^2 C}{L} + \omega_0^2 L C$

$\therefore \omega_0^2 = \frac{1}{LC} - \left(\frac{R}{L}\right)^2$ $\therefore \omega_0 = \sqrt{\frac{1}{LC} - \left(\frac{R}{L}\right)^2} = \sqrt{4 \times 10^8 - (5 \times 10^3)^2} = 1.94 \times 10^4 \frac{\text{rad}}{\text{s}}$

Let $j\omega = s$.

$\therefore \frac{i_o}{i_s} = \frac{R + sL}{R + sL + \frac{1}{sC}}$ or $H(\omega) = \frac{250 + j\omega \times 0.05}{250 + j\omega \times 0.05 - j \frac{106}{\omega \times 0.05}}$

$\therefore |H(\omega)| = \frac{\sqrt{250^2 + \omega^2 \times 0.05^2}}{\sqrt{250^2 + \left[(\omega \times 0.05)^2 - \left(\frac{106}{\omega \times 0.05}\right)^2\right]^2}}$

$\therefore |H(\omega_0)| = \frac{\sqrt{250^2 + 1.94^2 \times 10^8 \times 0.05^2}}{\sqrt{250^2 + \left[(1.94 \times 0.05 \times 10^4)^2 - \left(\frac{106}{1.94 \times 0.05 \times 10^4}\right)^2\right]^2}}$

$$= \frac{\sqrt{250^2 + 940900}}{\sqrt{250^2 + [970 - 1031]^2}} = \frac{1001.7}{257.33} \approx 3.9$$

At ω_1 & ω_2 $|H(\omega_{1,2})| = 0.707 |H(\omega_0)| = 2.76$

$\therefore 2.76 = \left| \frac{250 + j\omega_{1,2} \times 0.05}{250 + j\omega_{1,2} \times 0.05 - j \frac{2 \times 10^7}{\omega_{1,2}}} \right|$

$\therefore 2.76 = \left| \frac{250 \omega_{1,2} + j\omega_{1,2}^2 \times 0.05}{250 \omega_{1,2} + j\omega_{1,2}^2 - j 2 \times 10^7} \right| = \frac{\sqrt{250^2 \omega_{1,2}^2 + \omega_{1,2}^4 \times 0.05^2}}{\sqrt{250^2 \omega_{1,2}^2 + (\omega_{1,2}^2 - 2 \times 10^7)^2}}$

$= \frac{\sqrt{250^2 + \omega_{1,2}^2 \times 0.05^2}}{\sqrt{250^2 + \left(1 - \frac{2 \times 10^7}{\omega_{1,2}^2}\right)^2}}$ or $\frac{250^2 + \omega_{1,2}^2 \times 0.05^2}{250^2 + \left(1 - \frac{2 \times 10^7}{\omega_{1,2}^2}\right)^2}$

$$= 2.76^2 \times 250^2 + 2.76^2 \left(1 - \frac{2 \times 10^7}{\omega_{1,2}^2}\right)^2$$

or $41.36 \times 10^4 = 25 \times 10^{-4} \omega_{1,2}^2 - \frac{30.5 \times 10^{14}}{\omega_{1,2}^4} + \frac{30.5 \times 10^{14}}{\omega_{1,2}^2}$

or $41.36 \omega_{1,2}^4 = 25 \times 10^{-8} \omega_{1,2}^6 - 30.5 \times 10^{10} + 30.5 \times 10^{10} \omega_{1,2}^2$

or $\omega_{1,2}^4 - 7.37 \omega_{1,2}^2 \times 10^9 + 7.37 \times 10^{10} \approx 0$

Too lengthy a problem \rightarrow Solve for $\omega_{1,2}$

That would give the asymptotic plot

