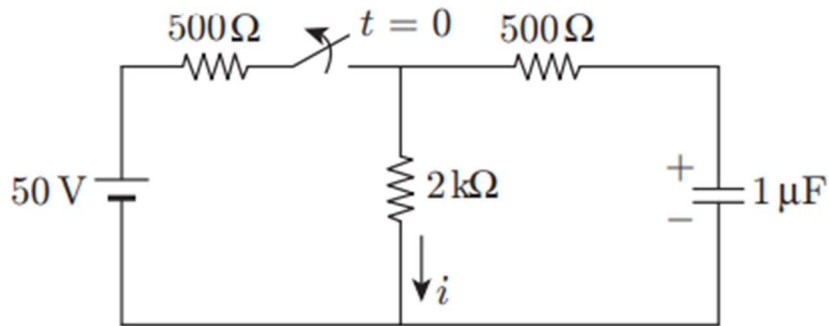


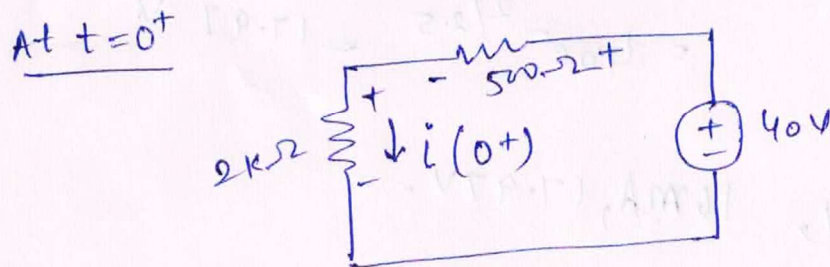
1. For the circuit, shown in the figure below, the switch is closed for a long time and it is opened at $t = 0$. Determine $V_C(0^+)$, $i(0^+)$ and $V_C(2 \text{ ms})$.



Solution -

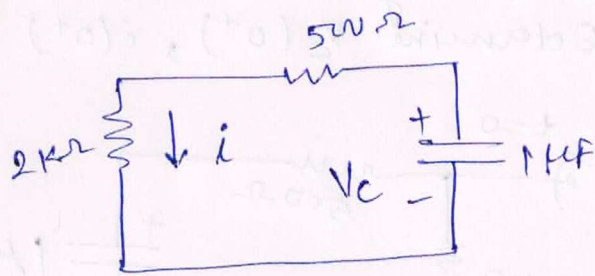
Solⁿ: At $t = 0^-$

$$V_C(0^-) = \frac{50}{2.5 \times 10^3} \times 2 \times 10^3 = 40 \text{ V} = V_C(0^+)$$



$$i(0^+) = \frac{40}{2.5 \times 10^3} = \underline{\underline{16 \text{ mA}}}$$

At $t \geq 0$



$$V_c(t) = 40 e^{(-t / (2.5 \times 10^3 \times 1 \times 10^{-6}))} \text{ V}$$

$$= 40 \times e^{-t / (2.5 \times 10^{-3})} \text{ V}$$

$$V_c(0^-) = V_c(0^+) = 40 \text{ V}$$

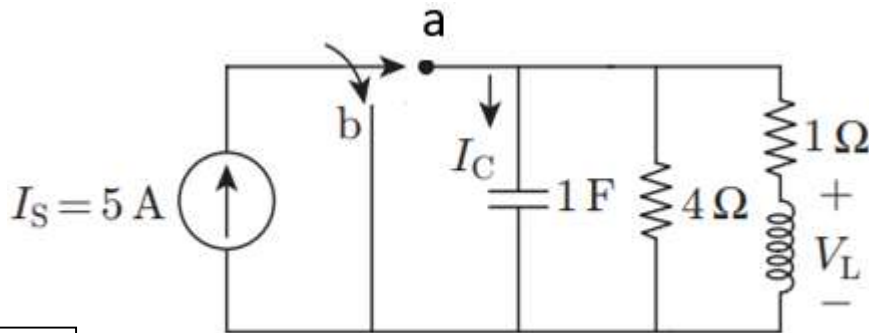
$$t = \underline{2.0 \text{ ms}}$$

$$V_c(2 \text{ ms}) = 40 e^{-\left(\frac{2 \times 10^{-3}}{2.5 \times 10^{-3}}\right)}$$

$$= 40 e^{-2/2.5} = 17.97 \text{ V}$$

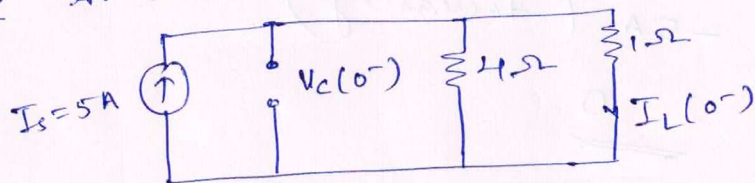
$$\text{Ans} = 40 \text{ V}, 16 \text{ mA}, 17.97 \text{ V}.$$

2. A switch, in the figure below, is in position 'a' for a long time. Switch is moved to position 'b' at $t = 0$. At $t = 0^+$, the values of I_C and V_L are -



Solution -

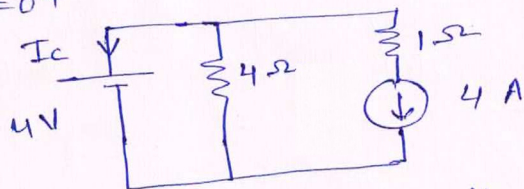
Solⁿ: At $t = 0^-$



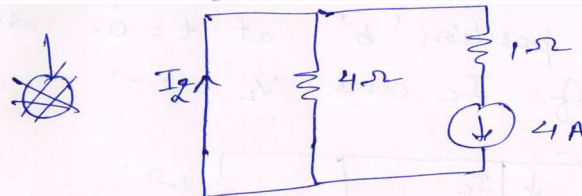
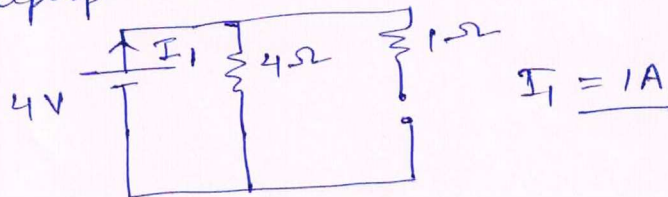
$$\frac{V_C \times 5}{4} = 5 \Rightarrow V_C = 4 \text{ V}$$

$$I_L = 5 \times \frac{4}{4+1} = 4 \text{ A}$$

At $t = 0^+$



By superposition theorem



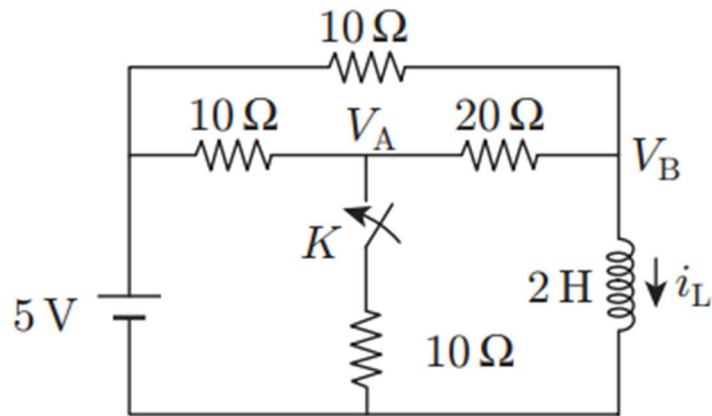
$$I_2 = 4 \text{ A}$$

$$I_C = I_1 + I_2 = 5 \text{ A}$$

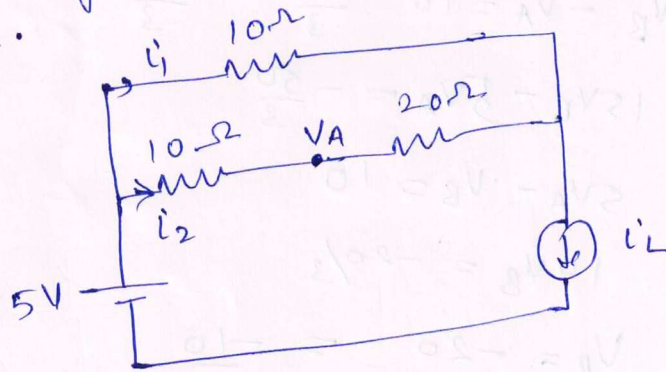
$$V_L = 0$$

(Voltage across inductor is zero as inductor is short circuited)

3. In the network shown in the figure below, a steady state is reached with the switch K open. At $t = 0$, the switch K is closed. For the element values given, determine the values of $V_A(0^-)$ and $V_A(0^+)$.



Solⁿ: The steady state is reached with switch 'K' open.

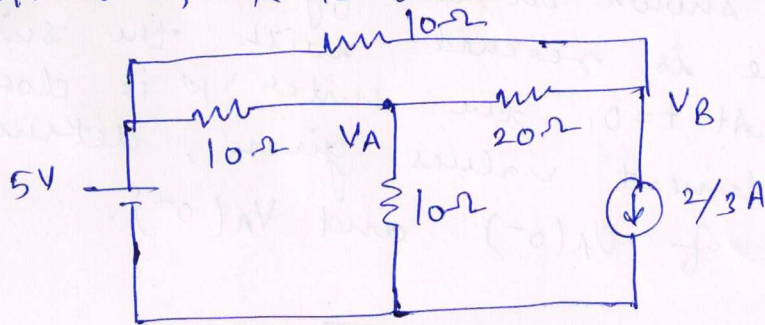


$$i_L = \frac{5}{30} \times 4 = \frac{2}{3} \text{ A}$$

$$\therefore V_A(0^-) = i_2 \times 20 = \frac{10}{40} \times \frac{2}{3} \times 20 = \frac{10}{3} \text{ V}$$

$$i_L(0^-) = i_L(0^+)$$

at $t=0$, K is closed



$$\frac{V_A - 5}{10} + \frac{V_A}{10} + \frac{V_A - V_B}{20} = 0$$

$$\frac{V_B - 5}{10} + \frac{V_B - V_A}{20} + \frac{2}{3} = 0$$

$$2(V_A - 5 + V_A) + V_A - V_B = 0$$

$$5V_A - V_B = 10$$

$$2(V_B - 5) + V_B - V_A = -\frac{2}{3} \times 20$$

$$3V_B - V_A = 10 - \frac{40}{3} = -\frac{10}{3}$$

$$15V_B - 5V_A = -\frac{50}{3}$$

$$5V_A - V_B = 10$$

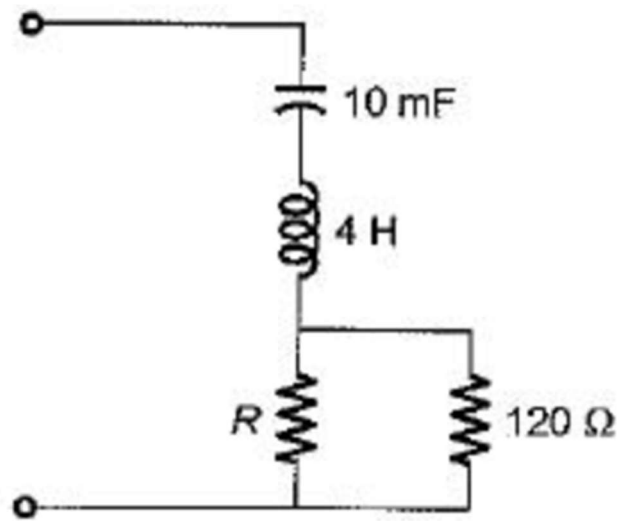
$$14V_B = -\frac{20}{3}$$

$$V_B = \frac{-20}{3 \times 14} = -\frac{10}{21}$$

$$5V_A = 10 - \frac{10}{21} = \frac{200}{21} \Rightarrow V_A(0^+) = \frac{40}{21} \text{ V.}$$

$$\text{Ans.} \left(\frac{10}{3} \text{ V}, \frac{40}{21} \text{ V} \right).$$

4. If the circuit, shown below, is critically damped, find the value of R (in ohms) -



Solution –

Given circuit is series RLC circuit.

For critically damped condition, we have

$$\alpha = \omega_0$$

$$\Rightarrow \alpha = \frac{R_{eq}}{2L} ; \quad \omega_0 = \frac{1}{\sqrt{LC}}$$

$$\Rightarrow \text{Here } R_{eq} = R \parallel 120 = \frac{120 \times R}{120 + R}$$

$$\Rightarrow \frac{R_{eq}}{2L} = \frac{1}{\sqrt{LC}} \Rightarrow R_{eq} = \sqrt{\frac{4L}{C}}$$

$$R_{eq} = \sqrt{\frac{4 \times 4}{10 \times 10^{-3}}} = 40 \Omega$$

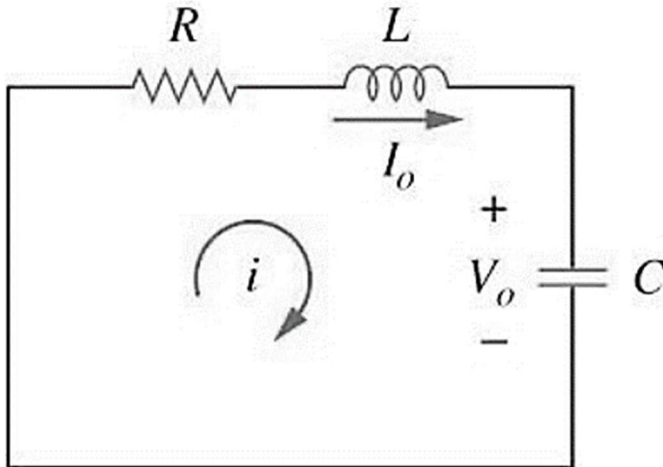
$$\Rightarrow \frac{120R}{120 + R} = 40$$

$$120R = 4800 + 40R$$

$$80R = 4800$$

$$R = 60 \Omega$$

5. In the given circuit, If $L = 2\text{H}$, $C = 1\text{ mF}$, $R = 20\Omega$, the value of loop current $i(t)$ is -



Solution -

Given circuit is series RLC. Also, $R = 20\Omega$, $C = 1\text{ mF}$, $L = 2\text{ H}$

$$\Rightarrow \alpha = \frac{R}{2L} \quad \omega_0 = \frac{1}{\sqrt{LC}}$$

$$= \frac{20}{2 \times 2} = 5 \text{ rad/s} \quad = \frac{1}{\sqrt{2 \times 10^{-3}}} = 22.36 \text{ rad/s}$$

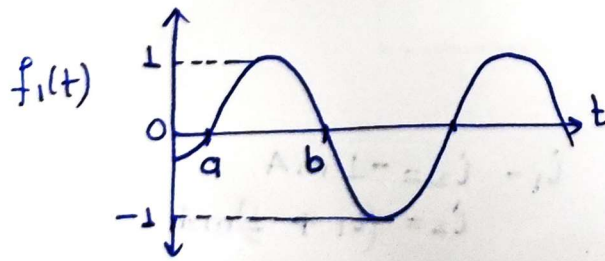
→ Since $\alpha < \omega_0$, $i(t)$ has underdamped response of the form

$$i(t) = e^{-\alpha t} (A_1 \cos \omega_d t + A_2 \sin \omega_d t) ;$$

where $\omega_d = \sqrt{\omega_0^2 - \alpha^2} = 21.8$

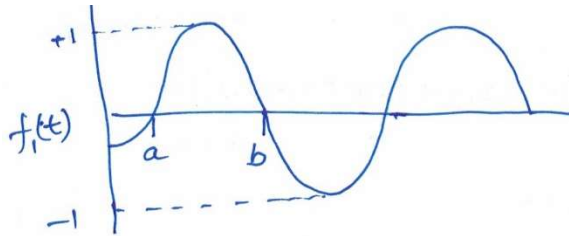
$$\therefore i(t) = e^{-5t} (A_1 \cos 21.8t + A_2 \sin 21.8t)$$

6. Find the Laplace transform of the following function $f_1(t)$ –



$$\omega = \frac{\pi}{b-a}$$

Solution –



$$\omega = \frac{\pi}{b-a}$$

here,

$$\begin{aligned} \mathcal{L}(f_1(t)) &= \int_0^{\infty} f_1(t) e^{-st} dt \\ &= \int_0^{\infty} \sin(\omega t - \omega a) e^{-st} dt \\ &= \int_0^{\infty} (\sin \omega t \cos \omega a - \cos \omega t \sin \omega a) e^{-st} dt \\ &= \int_0^{\infty} \cos \omega a \sin \omega t e^{-st} dt - \int_0^{\infty} \sin \omega a \cos \omega t e^{-st} dt \\ &= \cos \omega a \int_0^{\infty} \sin \omega t e^{-st} dt - \sin \omega a \int_0^{\infty} \cos \omega t e^{-st} dt \\ &= \frac{\cos \omega a \cdot \omega}{s^2 + \omega^2} - \frac{\sin \omega a \cdot s}{s^2 + \omega^2} \\ &= \frac{\omega \cos \omega a - s \sin \omega a}{(s^2 + \omega^2)} \end{aligned}$$