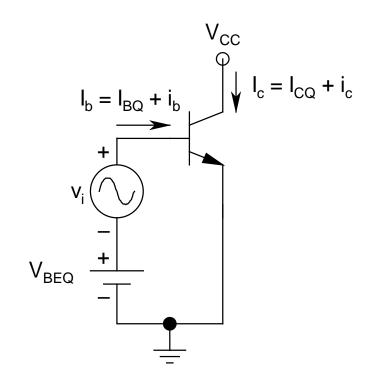
## Validity of the Small-Signal Model

- Basically linearization
  of the operating
  region around the Q point
- This linearization should not contain any higher-order terms



• To start with, assume  $V_A \rightarrow \infty$  $> I_{CO} = I_S \exp(V_{BEO}/V_T)$ 

• Thus:

$$\begin{split} I_{c} &= I_{S} \exp \left( \frac{V_{be}}{V_{T}} \right) = I_{S} \exp \left( \frac{V_{BEQ} + V_{i}}{V_{T}} \right) \\ &= I_{S} \exp \left( \frac{V_{BEQ}}{V_{T}} \right) \exp \left( \frac{V_{i}}{V_{T}} \right) = I_{CQ} \exp \left( \frac{V_{i}}{V_{T}} \right) \end{split}$$

• Expand the *exponential term* in series:

$$I_{c} = I_{CQ} \left[ 1 + \frac{V_{i}}{V_{T}} + \frac{1}{2!} \left( \frac{V_{i}}{V_{T}} \right)^{2} + \frac{1}{3!} \left( \frac{V_{i}}{V_{T}} \right)^{3} + \cdots \right]$$

• Thus:

$$i_{c} = I_{c} - I_{CQ} = I_{CQ} \left[ \frac{v_{i}}{V_{T}} + \frac{1}{2!} \left( \frac{v_{i}}{V_{T}} \right)^{2} + \frac{1}{3!} \left( \frac{v_{i}}{V_{T}} \right)^{3} + \cdots \right]$$

• True linearization of  $i_c$ - $v_i$  relation can be achieved only if all higher-order terms can be neglected  $\Rightarrow v_i$  should be  $<< V_T$ 

## **Small-Signal Model Parameters**

• Incremental Emitter Resistance (r<sub>F</sub>):

$$r_{E} = \frac{v_{i}}{i_{e}} = \frac{\Delta V_{BE}}{\Delta I_{E}} \equiv \frac{dV_{BE}}{dI_{E}} = \frac{V_{T}}{I_{E}}$$

• *Transconductance* (g<sub>m</sub>):

$$g_{\rm m} = \frac{i_{\rm c}}{v_{\rm i}} = \frac{\Delta I_{\rm C}}{\Delta V_{\rm BE}} \equiv \frac{dI_{\rm C}}{dV_{\rm BE}} \bigg|_{V_{\rm CE} \text{ constant}} = \frac{I_{\rm C}}{V_{\rm T}}$$

- ightharpoonup Thus,  $g_m r_E = I_C / I_E = \alpha \approx 1$
- > A frequently used approximation:
  - $g_m = 1/r_E$
- $\triangleright$  For  $I_C = 1$  mA:
  - $r_E = 26 \Omega$  and  $g_m = 1/26 A/V$
- $\rightarrow$  As I<sub>C</sub> $\uparrow$ :
  - $g_m \uparrow$  and  $r_E \downarrow$
  - Also  $P_D \uparrow$
- $\triangleright$  Gain = f(g<sub>m</sub>)
  - ⇒ For *higher gain*, the circuit has to be fed *more* power