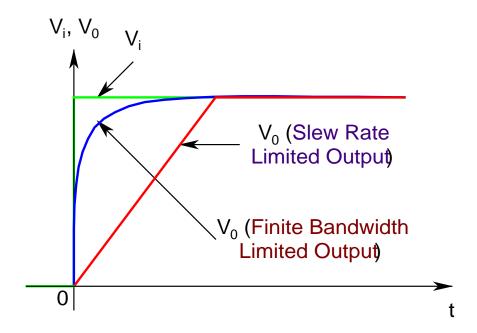
- Similarly, when a *large positive signal* is applied at the *base* of Q<sub>1</sub> (w.r.t. the base of Q<sub>2</sub>), Q<sub>2</sub>-Q<sub>4</sub> branch would *instantly turn off* 
  - $\Rightarrow$  Entire bias current  $I_{C8}$  would flow through the  $Q_1$ - $Q_3$  branch, pushing the same current through  $Q_5$
  - $\Rightarrow$  Q<sub>6</sub> would *carry the same current* (*mirror* with Q<sub>5</sub>)
- This *current* would *flow* from the *output node* through  $C_C$  to  $Q_6$ 
  - $\Rightarrow$  C<sub>C</sub> would *start to discharge*, and V<sub>0</sub> would *start to fall*, going into its *negative swing*
- Since the *same current* (I<sub>C8</sub>) is used to *discharge* the *output node*:

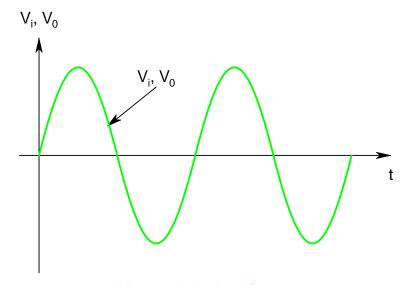
$$SR^{+} = SR^{-} = 1.52 \text{ V/}\mu\text{sec}$$

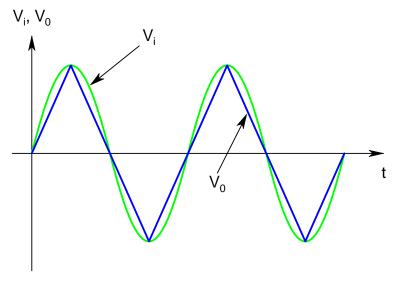


**Slew Rate Limitation of Op-Amps** 

- Situation becomes *more dramatic* if a *sinusoidal signal* is applied at the *non-inverting input* of the op-amp, connected in a *voltage-follower* configuration
- Let *input signal*  $V_i = V_M sin(\omega t)$ , with *large*  $V_M$ 
  - ⇒ *Transistors* in the *differential input stage* act as *switches*
- Under *unity feedback*,  $V_0$  would *follow*  $V_i$  $\Rightarrow dV_0/dt = dV_i/dt = V_M \omega \cos(\omega t)$
- The *maximum value* of this *derivative* occurs when  $\omega t = n\pi$  (n = 0, 1, 2, ...)
  - $\Rightarrow$  It occurs when the *signal crosses zero*
- So long as this rate remains smaller than SR, V<sub>0</sub> would follow V<sub>i</sub> with fidelity

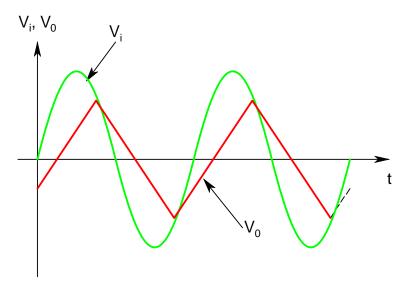
- However, as soon as  $dV_0/dt$  becomes  $\geq SR$ ,  $V_0$  won't be able to follow  $V_i$  anymore rather, it would start to become triangular
- **Note**:  $dV_0/dt$  with an **increase** of **either**  $V_M$  or  $\omega$  or **both** 
  - $\Rightarrow$  What essentially matters is the product  $V_M \omega$
- If this *product* keeps on *increasing* beyond the SR, then V<sub>0</sub> remains *triangular*, however, *two major observations* become apparent:
  - $\diamond$  The zero crossings of  $V_0$  do not quite coincide with those of  $V_i$
  - ❖ The *peak-to-peak swing* of  $V_0$  starts to become *smaller* than that of  $V_i$  due to  $V_0$  *not getting enough time* to *reach its maximum possible value*





**Normal Behavior** 

**Onset of Slew Rate Limitation** 



**Severely Slew Rate Limited** 

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- If  $V_M \omega$  becomes *very large*, then there *may not be any output at all* 
  - $\Rightarrow$  V<sub>0</sub> would *become zero*, implying that the op-amp is *not able to keep up with the variation* of V<sub>i</sub> at all!
- Mathematical Description:
  - ❖ Let the *gain* of the op-amp = A ⇒  $(dV_0/dt)_{max} = AV_Mω$
  - ❖ This must be *less* than the SR of the op-amp to get a *distortion-free output*
  - **\*** *Maximum possible value* of  $AV_M = V_{SAT}$ 
    - $\Rightarrow$  The *maximum allowed value* of  $\omega$  (=  $\omega_{\rm M}$ ) of  $V_{\rm i}$  for  $V_{\rm 0}$  to be *without any distortion* due to *slew rate limitation*:

$$\omega_{\rm M} = {\rm SR/V_{\rm SAT}}$$

- This is an *extremely important relation*, and  $\omega_{\rm M}$  is referred to as the *full-power bandwidth*
- It is a *constant* for a given op-amp
- This *derivation* is for  $V_0$  *swinging* between  $\pm V_{SAT}$
- If the *swing* of  $V_0$  is *less* than this, then  $\omega$  can be *increased* beyond  $\omega_M$ , following the *relation*:

$$SR = \omega_M V_{SAT} = \omega_0 V_0 = \omega_0 A V_i$$

- $\omega_0$ : Frequency till which  $V_0$  won't have any slew rate limited distortion
- $\Rightarrow$  Maximum amplitude of  $V_i$  (of frequency  $\omega_0$ ), beyond which slew rate limited distortion would set in at the output:

$$V_{i,max} = \omega_M V_{SAT} / (\omega_0 A)$$

