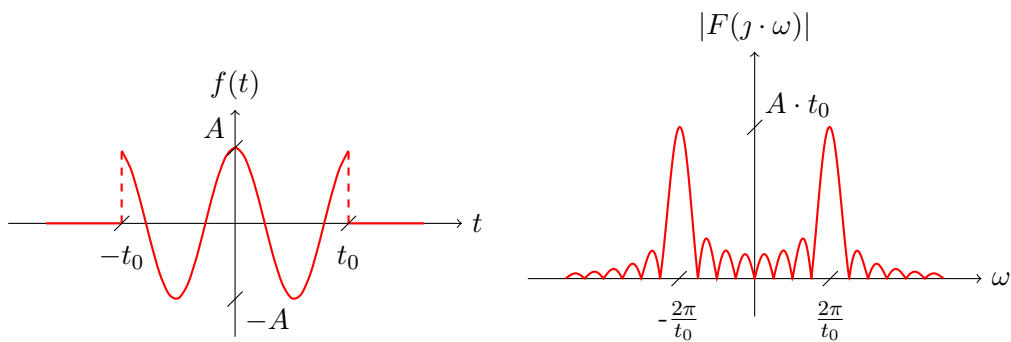


Signal Theory in practise



$$f(t) = A \cdot \Pi\left(\frac{t}{2 \cdot t_0}\right) \cdot \cos\left(\frac{2\pi}{t_0} \cdot t\right)$$

$$F(j\omega) = A \cdot t_0 \cdot [Sa(\omega \cdot t_0 + 2\pi) - Sa(\omega \cdot t_0 - 2\pi)]$$

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May 8, 2020

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ISBN 978-83-939620-3-7
Printed in Poland

Chapter 1

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1.1.1 Mean value of a signal

1.1.2 Energy of a signal

1.1.3 Power and effective value of a signal

Chapter 2

Analysis of periodic signals using orthogonal series

2.1 Trigonometric Fourier series

2.2 Complex exponential Fourier series

2.3 Computing the power of a signal – the Parseval's theorem

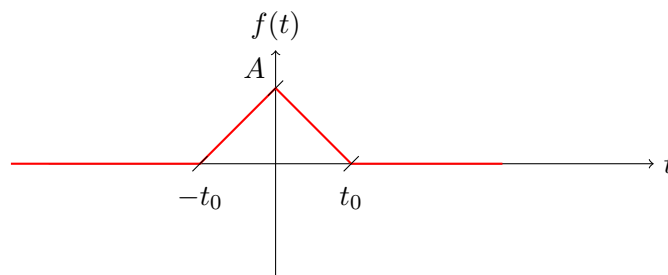
Chapter 3

Analysis of non-periodic signals.

Fourier Transformation and Transform

3.1 Calculation of Fourier Transform by definition

Task 1. Compute the Fourier transform of a triangle impulse shown below.



First of all, describe the $f(t)$ signal using elementary signals:

$$f(t) = A \cdot \Lambda\left(\frac{t}{t_0}\right) \quad (3.1)$$

The Fourier transform is defined as:

$$F(j\omega) = \int_{-\infty}^{\infty} f(t) \cdot e^{-j\omega \cdot t} \cdot dt \quad (3.2)$$

In order to integrate the $f(t)$ signal, we need to describe it as a piecewise signal.

The simplest form of a linear function is:

$$f(t) = m \cdot t + b \quad (3.3)$$

In the first interval (e.g. $t \in (-t_0; 0)$), linear function crosses two points: $(-t_0, 0)$ and $(0, A)$. So, in order to derive m and b , the following system of the equations has to be solved.

$$\begin{cases} 0 = m \cdot (-t_0) + b \\ A = m \cdot 0 + b \end{cases}$$

$$\begin{cases} -b = m \cdot (-t_0) \\ A = b \end{cases}$$

$$\begin{cases} \frac{b}{t_0} = m \\ A = b \end{cases}$$

$$\begin{cases} A = b \\ \frac{A}{t_0} = m \end{cases}$$

As a result we get:

$$f(t) = \frac{A}{t_0} \cdot t + A$$

In the second interval (e.g. $t \in (0; t_0)$), linear function crosses two points: $(0; A)$ and $(t_0, 0)$. So, in order to derive m and b , the following system of the equations has to be solved.

$$\begin{cases} 0 = m \cdot t_0 + b \\ A = m \cdot 0 + b \end{cases}$$

$$\begin{cases} -b = m \cdot t_0 \\ A = b \end{cases}$$

$$\begin{cases} -\frac{b}{t_0} = m \\ A = b \end{cases}$$

$$\begin{cases} A = b \\ -\frac{A}{t_0} = m \end{cases}$$

As a result we get:

$$f(t) = -\frac{A}{t_0} \cdot t + A$$

As a result the piecewise linear function is given by:

$$f(t) = A \cdot \Lambda\left(\frac{t}{t_0}\right) = \begin{cases} 0 & \text{for } t \in (-\infty; -t_0) \\ \frac{A}{t_0} \cdot t + A & \text{for } t \in (-t_0; 0) \\ -\frac{A}{t_0} \cdot t + A & \text{for } t \in (0; t_0) \\ 0 & \text{for } t \in (t_0; \infty) \end{cases} \quad (3.4)$$

For the given $f(t)$ signal we get:

$$\begin{aligned} F(j\omega) &= \int_{-\infty}^{\infty} f(t) \cdot e^{-j\omega \cdot t} \cdot dt = \\ &= \int_{-\infty}^{-t_0} 0 \cdot e^{-j\omega \cdot t} \cdot dt + \int_{-t_0}^0 \left(\frac{A}{t_0} \cdot t + A \right) \cdot e^{-j\omega \cdot t} \cdot dt = \end{aligned}$$

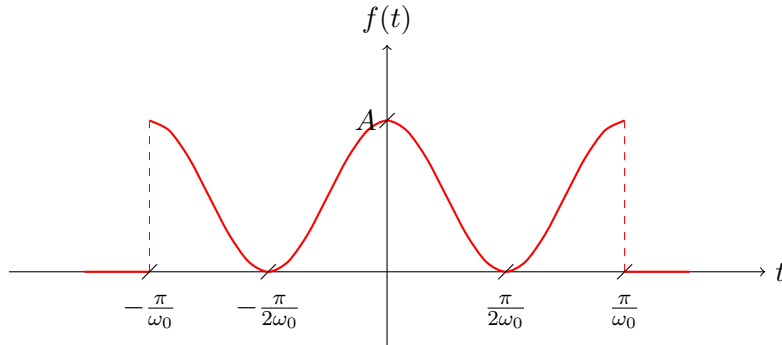
$$\begin{aligned}
& + \int_0^{t_0} \left(-\frac{A}{t_0} \cdot t + A \right) \cdot e^{-j\omega \cdot t} \cdot dt + \int_{t_0}^{\infty} 0 \cdot e^{-j\omega \cdot t} \cdot dt = \\
& = \int_{-\infty}^{-t_0} 0 \cdot dt + \int_{-t_0}^0 \frac{A}{t_0} \cdot t \cdot e^{-j\omega \cdot t} \cdot dt + \int_{-t_0}^0 A \cdot e^{-j\omega \cdot t} \cdot dt = \\
& + \int_0^{t_0} -\frac{A}{t_0} \cdot t \cdot e^{-j\omega \cdot t} \cdot dt + \int_0^{t_0} A \cdot e^{-j\omega \cdot t} \cdot dt + \int_{t_0}^{\infty} 0 \cdot dt = \\
& = 0 + \frac{A}{t_0} \cdot \int_{-t_0}^0 t \cdot e^{-j\omega \cdot t} \cdot dt + A \cdot \int_{-t_0}^0 e^{-j\omega \cdot t} \cdot dt = \\
& - \frac{A}{t_0} \cdot \int_0^{t_0} t \cdot e^{-j\omega \cdot t} \cdot dt + A \cdot \int_0^{t_0} e^{-j\omega \cdot t} \cdot dt + 0 = \\
& = \left\{ \begin{array}{l} u = t \quad dv = e^{-j\omega \cdot t} \cdot dt \\ du = dt \quad v = \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \end{array} \right\} = \\
& = \frac{A}{t_0} \cdot \left(t \cdot \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_{-t_0}^0 - \int_{-t_0}^0 \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \cdot dt \right) = \\
& + A \cdot \left(\frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_{-t_0}^0 \right) = \\
& - \frac{A}{t_0} \cdot \left(t \cdot \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_0^{t_0} - \int_0^{t_0} \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \cdot dt \right) = \\
& + A \cdot \left(\frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_0^{t_0} \right) = \\
& = \frac{A}{t_0} \cdot \left(0 \cdot e^{-j\omega \cdot 0} - (-t_0) \cdot \frac{1}{-j\omega} \cdot e^{-j\omega \cdot (-t_0)} + \frac{1}{j\omega} \left(\frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_{-t_0}^0 \right) \right) = \\
& + \frac{A}{-j\omega} \cdot \left(e^{-j\omega \cdot 0} - e^{-j\omega \cdot (-t_0)} \right) = \\
& - \frac{A}{t_0} \cdot \left(t_0 \cdot \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t_0} - 0 \cdot e^{-j\omega \cdot 0} + \frac{1}{j\omega} \left(\frac{1}{-j\omega} \cdot e^{-j\omega \cdot t} \Big|_0^{t_0} \right) \right) = \\
& + \frac{A}{-j\omega} \cdot \left(e^{-j\omega \cdot t_0} - e^{-j\omega \cdot 0} \right) = \\
& = \frac{A}{t_0} \cdot \left(0 - t_0 \cdot \frac{1}{j\omega} \cdot e^{j\omega \cdot t_0} - \frac{1}{j^2 \cdot \omega^2} \left(e^{-j\omega \cdot 0} - e^{-j\omega \cdot (-t_0)} \right) \right) = \\
& - \frac{A}{j\omega} \cdot \left(1 - e^{j\omega \cdot t_0} \right) = \\
& - \frac{A}{t_0} \cdot \left(t_0 \cdot \frac{1}{-j\omega} \cdot e^{-j\omega \cdot t_0} - 0 - \frac{1}{j^2 \cdot \omega^2} \left(e^{-j\omega \cdot t_0} - e^{-j\omega \cdot 0} \right) \right) = \\
& - \frac{A}{j\omega} \cdot \left(e^{-j\omega \cdot t_0} - 1 \right) = \\
& = -\frac{A}{j\omega} \cdot e^{j\omega \cdot t_0} - \frac{A}{t_0 \cdot j^2 \cdot \omega^2} + \frac{A}{t_0 \cdot j^2 \cdot \omega^2} \cdot e^{j\omega \cdot t_0} - \frac{A}{j\omega} + \frac{A}{j\omega} \cdot e^{j\omega \cdot t_0} = \\
& + \frac{A}{j\omega} \cdot e^{-j\omega \cdot t_0} + \frac{A}{t_0 \cdot j^2 \cdot \omega^2} \cdot e^{-j\omega \cdot t_0} - \frac{A}{t_0 \cdot j^2 \cdot \omega^2} - \frac{A}{j\omega} \cdot e^{-j\omega \cdot t_0} + \frac{A}{j\omega} = \\
& = -\frac{2 \cdot A}{t_0 \cdot j^2 \cdot \omega^2} + \frac{A}{t_0 \cdot j^2 \cdot \omega^2} \cdot \left(e^{j\omega \cdot t_0} + e^{-j\omega \cdot t_0} \right) = \\
& = \frac{2 \cdot A}{t_0 \cdot \omega^2} - \frac{A}{t_0 \cdot \omega^2} \cdot \left(e^{j\omega \cdot t_0} + e^{-j\omega \cdot t_0} \right) = \\
& = \left\{ \cos(x) = \frac{e^{jx} + e^{-jx}}{2} \right\} =
\end{aligned}$$

$$\begin{aligned}
&= \frac{2 \cdot A}{t_0 \cdot \omega^2} - \frac{2 \cdot A}{t_0 \cdot \omega^2} \cdot \cos(\omega \cdot t_0) = \\
&= \frac{2 \cdot A}{t_0 \cdot \omega^2} \cdot (1 - \cos(\omega \cdot t_0)) = \\
&= \left\{ \begin{aligned} \sin^2(x) &= \frac{1}{2} - \frac{1}{2} \cdot \cos(2 \cdot x) \\ \cos(2 \cdot x) &= 1 - 2 \cdot \sin^2(x) \end{aligned} \right\} = \\
&= \frac{2 \cdot A}{t_0 \cdot \omega^2} \cdot (1 - 1 + 2 \cdot \sin^2(\frac{\omega \cdot t_0}{2})) = \\
&= \frac{4 \cdot A}{t_0 \cdot \omega^2} \cdot \sin^2(\frac{\omega \cdot t_0}{2}) = \\
&= \frac{A \cdot t_0}{\frac{t_0^2 \cdot \omega^2}{4}} \cdot \sin^2(\frac{\omega \cdot t_0}{2}) = \\
&= \left\{ \frac{\sin(x)}{x} = \text{Sa}(x) \right\} = \\
&= A \cdot t_0 \cdot \text{Sa}^2(\frac{\omega \cdot t_0}{2})
\end{aligned}$$

The Fourier transform of the $f(t) = A \cdot \Lambda(\frac{t}{t_0})$ is equal to $F(j\omega) = A \cdot t_0 \cdot \text{Sa}^2(\frac{\omega \cdot t_0}{2})$.

3.2 Exploiting properties of the Fourier transformation

Task 1. Oblicz transformatę Fouriera sygnału $f(t)$ przedstawionego na rysunku za pomocą twierdzeń, wiedząc że transformata sygnału prostokątnego $g(t) = \Pi(t)$ jest równa $G(j\omega) = Sa(\frac{\omega}{2})$.



Zacznijmy od napisania wzoru sygnału przedstawionego na rysunku

$$f(t) = \begin{cases} 0 & t \in (-\infty; -\frac{\pi}{\omega_0}) \\ A \cdot \cos^2(\omega_0 \cdot t) & t \in (-\frac{\pi}{\omega_0}; \frac{\pi}{\omega_0}) \\ 0 & t \in (\frac{\pi}{\omega_0}; \infty) \end{cases}$$

Co możemy zapisać za pomocą sygnałów elementarnych jako

$$f(t) = A \cdot \Pi\left(\frac{t \cdot \omega_0}{2\pi}\right) \cdot \cos^2(\omega_0 \cdot t)$$

Nasz sygnał jest iloczynem pewnej funkcji $h(t)$ oraz cosinusów

$$\begin{aligned} f(t) &= A \cdot \Pi\left(\frac{t \cdot \omega_0}{2\pi}\right) \cdot \cos^2(\omega_0 \cdot t) = \\ &= A \cdot h(t) \cdot \cos^2(\omega_0 \cdot t) = \end{aligned}$$

gdzie

$$h(t) = \Pi\left(\frac{t \cdot \omega_0}{2\pi}\right)$$

Wyznamy transformatę sygnału $h(t)$. Z treści zadania wiemy że transformata sygnału $g(t) = \Pi(t)$ jest równa $G(j\omega) = Sa(\frac{\omega}{2})$.

Postać funkcji $g(t)$ nie jest identyczna z postacią funkcji $h(t)$, funkcja różni się skalą. Wyznaczamy transformatę funkcji przeskalowanej $h(t) = \Pi(\frac{t \cdot \omega_0}{2\pi})$

Z twierdzenia o zmianie skali mamy

$$g(t) \xrightarrow{\mathcal{F}} G(j\omega)$$

$$h(t) = g(\alpha \cdot t) \xrightarrow{\mathcal{F}} H(j\omega) = \frac{1}{|\alpha|} \cdot G(j\frac{\omega}{\alpha})$$

a więc otrzymujemy

$$h(t) = \Pi\left(\frac{t}{T}\right) =$$

$$= \Pi\left(\frac{t \cdot \omega_0}{2\pi}\right) =$$

$$= \Pi\left(t \cdot \frac{\omega_0}{2\pi}\right) =$$

$$= g\left(t \cdot \frac{\omega_0}{2\pi}\right)$$

gdzie

$$\alpha = \frac{\omega_0}{2\pi}$$

a więc

$$H(j\omega) = \frac{1}{\frac{\omega_0}{2\pi}} \cdot G\left(j\frac{\omega}{\frac{\omega_0}{2\pi}}\right) =$$

$$= \frac{2\pi}{\omega_0} \cdot G\left(\frac{j\omega \cdot 2\pi}{\omega_0}\right) =$$

$$= \frac{2\pi}{\omega_0} \cdot Sa\left(\frac{\frac{\omega \cdot 2\pi}{\omega_0}}{2}\right) =$$

$$= \frac{2\pi}{\omega_0} \cdot Sa\left(\omega \cdot \frac{\pi}{\omega_0}\right)$$

A więc transformata sygnału $h(t)$ jest równa $H(j\omega) = \frac{2\pi}{\omega_0} \cdot Sa\left(\omega \cdot \frac{\pi}{\omega_0}\right)$

Wróćmy do wzoru sygnału i przedstawmy go następująco

$$f(t) = A \cdot \Pi\left(\frac{t \cdot \omega_0}{2\pi}\right) \cdot \cos^2(\omega_0 \cdot t) =$$

$$= A \cdot h(t) \cdot \cos^2(\omega_0 \cdot t) =$$

$$= A \cdot h(t) \cdot \cos(\omega_0 \cdot t) \cdot \cos(\omega_0 \cdot t) =$$

$$= A \cdot k(t) \cdot \cos(\omega_0 \cdot t) =$$

gdzie

$$k(t) = h(t) \cdot \cos(\omega_0 \cdot t) =$$

$$\begin{aligned}
&= \left\{ \cos(x) = \frac{e^{j \cdot x} + e^{-j \cdot x}}{2} \right\} = \\
&= h(t) \cdot \frac{e^{j \omega_0 \cdot t} + e^{-j \omega_0 \cdot t}}{2} = \\
&= \frac{1}{2} \left(h(t) \cdot e^{j \omega_0 \cdot t} + h(t) \cdot e^{-j \omega_0 \cdot t} \right)
\end{aligned}$$

Wyznamy transformatę sygnału $k(t)$. Korzystając z twierdzenia o modulacji mamy

$$\begin{aligned}
h(t) &\xrightarrow{\mathcal{F}} H(j\omega) \\
k(t) &= h(t) \cdot e^{j \omega_0 \cdot t} \xrightarrow{\mathcal{F}} K(j\omega) = H(j(\omega - \omega_0))
\end{aligned}$$

a więc transformata sygnału $k(t)$ wynosi:

$$\begin{aligned}
K(j\omega) &= \frac{1}{2} (H(j(\omega - \omega_0)) + H(j(\omega + \omega_0))) = \\
&= \frac{1}{2} \left(\frac{2\pi}{\omega_0} \cdot \text{Sa} \left((\omega - \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \frac{2\pi}{\omega_0} \cdot \text{Sa} \left((\omega + \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right) = \\
&= \frac{\pi}{\omega_0} \cdot \text{Sa} \left((\omega - \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \frac{\pi}{\omega_0} \cdot \text{Sa} \left((\omega + \omega_0) \cdot \frac{\pi}{\omega_0} \right)
\end{aligned}$$

Wróćmy do wzoru sygnału $f(t)$

$$\begin{aligned}
f(t) &= A \cdot \Pi \left(\frac{t \cdot \omega_0}{2\pi} \right) \cdot \cos^2(\omega_0 \cdot t) = \\
&= A \cdot h(t) \cdot \cos^2(\omega_0 \cdot t) = \\
&= A \cdot h(t) \cdot \cos(\omega_0 \cdot t) \cdot \cos(\omega_0 \cdot t) = \\
&= A \cdot k(t) \cdot \cos(\omega_0 \cdot t) = \\
&= \left\{ \cos(x) = \frac{e^{j \cdot x} + e^{-j \cdot x}}{2} \right\} = \\
&= A \cdot k(t) \cdot \frac{e^{j \omega_0 \cdot t} + e^{-j \omega_0 \cdot t}}{2} = \\
&= \frac{A}{2} \left(k(t) \cdot e^{j \omega_0 \cdot t} + k(t) \cdot e^{-j \omega_0 \cdot t} \right)
\end{aligned}$$

Znając transformatę sygnału $k(t)$ i korzystając z twierdzenia o modulacji możemy wyznaczyć transformatę sygnału $f(t)$.

$$\begin{aligned}
k(t) &\xrightarrow{\mathcal{F}} K(j\omega) \\
f(t) &= k(t) \cdot e^{j \omega_0 \cdot t} \xrightarrow{\mathcal{F}} F(j\omega) = K(j(\omega - \omega_0))
\end{aligned}$$

a więc transformata sygnału $f(t)$ wynosi

$$F(j\omega) = \frac{A}{2} (K(j(\omega - \omega_0)) + K(j(\omega + \omega_0))) =$$

$$\begin{aligned}
&= \frac{A}{2} \left(\frac{\pi}{\omega_0} \cdot Sa \left((\omega - \omega_0 - \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \frac{\pi}{\omega_0} \cdot Sa \left((\omega - \omega_0 + \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \right. \\
&\quad \left. + \frac{\pi}{\omega_0} \cdot Sa \left((\omega + \omega_0 - \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \frac{\pi}{\omega_0} \cdot Sa \left((\omega + \omega_0 + \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right) = \\
&= \frac{A}{2} \left(\frac{\pi}{\omega_0} \cdot Sa \left((\omega - 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) + \frac{\pi}{\omega_0} \cdot Sa \left(\omega \cdot \frac{\pi}{\omega_0} \right) + \right. \\
&\quad \left. + \frac{\pi}{\omega_0} \cdot Sa \left(\omega \cdot \frac{\pi}{\omega_0} \right) + \frac{\pi}{\omega_0} \cdot Sa \left((\omega + 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right) = \\
&= \frac{A}{2} \left(\frac{\pi}{\omega_0} \cdot Sa \left((\omega - 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) + 2 \cdot \frac{\pi}{\omega_0} \cdot Sa \left(\omega \cdot \frac{\pi}{\omega_0} \right) + \right. \\
&\quad \left. + \frac{\pi}{\omega_0} \cdot Sa \left((\omega + 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right) = \\
&= \frac{A \cdot \pi}{2 \cdot \omega_0} \left(Sa \left((\omega - 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) + 2 \cdot Sa \left(\omega \cdot \frac{\pi}{\omega_0} \right) + Sa \left((\omega + 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right)
\end{aligned}$$

Ostatecznie transformata sygnału $f(t)$ wynosi $F(j\omega) = \frac{A \cdot \pi}{2 \cdot \omega_0} \left(Sa \left((\omega - 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) + 2 \cdot Sa \left(\omega \cdot \frac{\pi}{\omega_0} \right) + Sa \left((\omega + 2 \cdot \omega_0) \cdot \frac{\pi}{\omega_0} \right) \right)$

3.3 Calculating energy of the signal from its Fourier transform. The Parseval's theorem

Chapter 4

Processing of signals by linear and time invariant (LTI) systems

4.1 Linear convolution

4.2 Filters

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ISBN 978-83-939620-3-7

