

Discrete source

Ex. 1 (18 points)

There are two binary random variables X and Y with the following **joint probability distribution**:

Y \ X	$x_1=0$	$x_2=1$
$y_1=0$	$1/8$	$3/8$
$y_2=1$	$3/8$	$1/8$

Additionally, define another random variable $Z = XY$ (the general product).

Calculate the following:

(1) $H(X), H(Y), H(Z), H(X, Z), H(Y, Z), H(X, Y, Z)$; (6 points)

(2) $H(X/Y), H(Y/X), H(X/Z), H(Z/X), H(Y/Z), H(Z/Y), H(X/YZ), H(Y/XZ), H(Z/XY)$; (9 points)

(3) $I(X; Y), I(X; Z), I(Y; Z)$ (3 points)

Hint: $P(A) = \sum_{i=1}^n P(A|B_i)P(B_i)$

Solution

(1)

$$p(x_1) = p(x_2) = \frac{1}{2} \quad H(X) = - \sum_i p(x_i) \log_2 p(x_i) = 1 \text{ bit/symbol},$$

$$p(y_1) = p(y_2) = \frac{1}{2} \quad H(Y) = - \sum_j p(y_j) \log_2 p(y_j) = 1 \text{ bit/symbol}$$

$Z = XY$:

$$\begin{bmatrix} Z \\ P(Z) \end{bmatrix} = \begin{bmatrix} z_1=0 & z_2=1 \\ \frac{7}{8} & \frac{1}{8} \end{bmatrix}$$

$$H(Z) = - \sum_k^2 p(z_k) = - \left(\frac{7}{8} \log_2 \frac{7}{8} + \frac{1}{8} \log_2 \frac{1}{8} \right) = 0.544 \text{ bit / symbol}$$

$$p(x_1) = p(x_1 z_1) + p(x_1 z_2)$$

$$p(x_1 z_2) = 0$$

$$p(x_1 z_1) = p(x_1) = 0.5$$

$$p(z_1) = p(x_1 z_1) + p(x_2 z_1)$$

$$p(x_2 z_1) = p(z_1) - p(x_1 z_1) = \frac{7}{8} - 0.5 = \frac{3}{8}$$

$$p(z_2) = p(x_1 z_2) + p(x_2 z_2)$$

$$p(x_2 z_2) = p(z_2) = \frac{1}{8}$$

$$H(XZ) = -\sum_i \sum_k p(x_i z_k) \log_2 p(x_i z_k) = -\left(\frac{1}{2} \log_2 \frac{1}{2} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{1}{8} \log_2 \frac{1}{8}\right) = 1.406 \text{ bit / symbol}$$

$$p(y_1) = p(y_1 z_1) + p(y_1 z_2)$$

$$p(y_1 z_2) = 0$$

$$p(y_1 z_1) = p(y_1) = 0.5$$

$$p(z_1) = p(y_1 z_1) + p(y_2 z_1)$$

$$p(y_2 z_1) = p(z_1) - p(y_1 z_1) = \frac{7}{8} - 0.5 = \frac{3}{8}$$

$$p(z_2) = p(y_1 z_2) + p(y_2 z_2)$$

$$p(y_2 z_2) = p(z_2) = \frac{1}{8}$$

$$H(YZ) = -\sum_j \sum_k p(y_j z_k) \log_2 p(y_j z_k) = -\left(\frac{1}{2} \log_2 \frac{1}{2} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{1}{8} \log_2 \frac{1}{8}\right) = 1.406 \text{ bit / symbol}$$

$$p(x_1 y_1 z_2) = 0$$

$$p(x_1 y_2 z_2) = 0$$

$$p(x_2 y_1 z_2) = 0$$

$$p(x_1 y_1 z_1) + p(x_1 y_1 z_2) = p(x_1 y_1)$$

$$p(x_1 y_1 z_1) = p(x_1 y_1) = 1/8$$

$$p(x_1 y_2 z_1) + p(x_1 y_1 z_1) = p(x_1 z_1)$$

$$p(x_1 y_2 z_1) = p(x_1 z_1) - p(x_1 y_1 z_1) = \frac{1}{2} - \frac{1}{8} = \frac{3}{8}$$

$$p(x_2y_1z_1) + p(x_2y_1z_2) = p(x_2y_1)$$

$$p(x_2y_1z_1) = p(x_2y_1) = \frac{3}{8}$$

$$p(x_2y_2z_1) = 0$$

$$p(x_2y_2z_1) + p(x_2y_2z_2) = p(x_2y_2)$$

$$p(x_2y_2z_2) = p(x_2y_2) = \frac{1}{8}$$

$$H(XYZ) = -\sum_i \sum_j \sum_k p(x_iy_jz_k) \log_2 p(x_iy_jz_k)$$

$$= -\left(\frac{1}{8} \log_2 \frac{1}{8} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{1}{8} \log_2 \frac{1}{8}\right) = 1.811 \text{ bit/symbol}$$

(2)

$$H(XY) = -\sum_i \sum_j p(x_iy_j) \log_2 p(x_iy_j) =$$

$$= -\left(\frac{1}{8} \log_2 \frac{1}{8} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{3}{8} \log_2 \frac{3}{8} + \frac{1}{8} \log_2 \frac{1}{8}\right)$$

$$= 1.811 \text{ bit/symbol}$$

$$H(X/Y) = H(XY) - H(Y) = 1.811 - 1 = 0.811 \text{ bit/symbol}$$

$$H(Y/X) = H(XY) - H(X) = 1.811 - 1 = 0.811 \text{ bit/symbol}$$

$$H(X/Z) = H(XZ) - H(Z) = 1.406 - 0.544 = 0.862 \text{ bit/symbol}$$

$$H(Z/X) = H(XZ) - H(X) = 1.406 - 1 = 0.406 \text{ bit/symbol}$$

$$H(Y/Z) = H(YZ) - H(Z) = 1.406 - 0.544 = 0.862 \text{ bit/symbol}$$

$$H(Z/Y) = H(YZ) - H(Y) = 1.406 - 1 = 0.406 \text{ bit/symbol}$$

$$H(X/YZ) = H(XYZ) - H(YZ) = 1.811 - 1.406 = 0.405 \text{ bit/symbol}$$

$$H(Y/XZ) = H(XYZ) - H(XZ) = 1.811 - 1.406 = 0.405 \text{ bit/symbol}$$

$$H(Z/XY) = H(XYZ) - H(XY) = 1.811 - 1.811 = 0 \text{ bit/symbol}$$

(3)

$$I(X; Y) = H(X) - H(X/Y) = 1 - 0.811 = 0.189 \text{ bit/symbol}$$

$$I(X; Z) = H(X) - H(X/Z) = 1 - 0.862 = 0.138 \text{ bit/symbol}$$

$$I(Y; Z) = H(Y) - H(Y/Z) = 1 - 0.862 = 0.138 \text{ bit/symbol}$$

Multiple-symbol discrete stationary source with memory

Ex. 2 (10 points)

The messages in black and white meteorological facsimiles have only two types: black and white, i.e., the source $X = \{Black, White\}$.

- (1) Let the probability of black appearing be $P(Black) = 0.3$ and the probability of white appearing be $P(White) = 0.7$. Assume that the appearance of black and white messages on the map has **no correlation** before and after, find the entropy $H(X)$; (2 points)
- (2) Now we consider a source that generates a very long sequence. Assume that the messages are correlated, with dependency relationships $P(X_{j+1} = White | X_j = White) = 0.9$, $P(X_{j+1} = Black | X_j = White) = 0.1$, $P(X_{j+1} = White | X_j = Black) = 0.2$, and $P(X_{j+1} = Black | X_j = Black) = 0.8$. Find the entropy $H_\infty(X)$ of this **stationary** first-order Markov source. (5 points)
- (3) Calculate the redundancy of the two sources above, and explain their physical meanings. (3 points)

Solution

(1)

$$H(X) = - \sum_i p(x_i) \log p(x_i) = -(0.3 \log 0.3 + 0.7 \log 0.7) = 0.881 \text{ bit/symbol}$$

(2) We define the black message as e_1 , and white as e_2 .

$$\begin{cases} p(X_{j+1} = e_1) = p(X_j = e_1)p(X_{j+1} = e_1/X_j = e_1) + p(X_j = e_2)p(X_{j+1} = e_1/X_j = e_2) \\ p(X_{j+1} = e_2) = p(X_j = e_2)p(X_{j+1} = e_2/X_j = e_2) + p(X_j = e_1)p(X_{j+1} = e_2/X_j = e_1) \\ p(X_{j+1} = e_1) = 0.8p(X_j = e_1) + 0.1p(X_j = e_2) \\ p(X_{j+1} = e_2) = 0.9p(X_j = e_2) + 0.2p(X_j = e_1) \end{cases}$$

Because it is stationary, we have:

$$\begin{aligned} p(X_j = e_1) &= p(X_{j+1} = e_1) \\ p(X_j = e_2) &= p(X_{j+1} = e_2) \end{aligned}$$

Then we can get:

$$\begin{cases} p(X_{j+1} = e_2) = 2p(X_{j+1} = e_1) \\ p(X_{j+1} = e_1) + p(X_{j+1} = e_2) = 1 \\ p(X_{j+1} = e_1) = 1/3 \\ p(X_{j+1} = e_2) = 2/3 \end{cases}$$

$$\begin{aligned} H_\infty(X) &= \lim_{N \rightarrow \infty} H_N(X^N) = \lim_{N \rightarrow \infty} H(X_N | X^{N-1}) \\ &= - \sum_i \sum_j p(X_j = e_i) p(X_{j+1} = e_j / X_j = e_i) \log p(X_{j+1} = e_j / X_j = e_i) \\ &= - \left(\frac{1}{3} \times 0.8 \log 0.8 + \frac{1}{3} \times 0.2 \log 0.2 + \frac{2}{3} \times 0.1 \log 0.1 + \frac{2}{3} \times 0.9 \log 0.9 \right) \\ &= 0.553 \text{ bit/symbol} \end{aligned}$$

(3)

$$\gamma_1 = \frac{H_0 - H(X)}{H_0} = \frac{\log_2 2 - 0.881}{\log_2 2} = 11.9\%$$

$$\gamma_2 = \frac{H_0 - H_\infty(X)}{H_0} = \frac{\log_2 2 - 0.553}{\log_2 2} = 44.7\%$$

The uncertainty of a memoryless source is greater than that of a source with memory. A source with memory contains more structured information, allowing for a greater degree of compression.

Continuous source

Ex. 3 (36 points)

Suppose that the joint probability distribution of random variables X and Y is

$$p(xy) = \frac{1}{2\pi\sqrt{SN}} \exp \left\{ -\frac{1}{2N} \left[x^2 \left(1 + \frac{N}{S} \right) - 2xy + y^2 \right] \right\}$$

with $N > 0$ and $S > 0$.

Noting that $\int_{-\infty}^{\infty} \exp(-at^2) dt = \sqrt{\frac{\pi}{a}}$ and $\int_{-\infty}^{\infty} t^2 \exp(-at^2) dt = \frac{1}{2} \sqrt{\frac{\pi}{a^3}}$, $a > 0$.

Compute:

- 1) $p(x)$ (9 points)
- 2) $H(X)$ (9 points)
- 3) $p(y)$ (9 points)
- 4) $H(Y)$ (9 points)

Solution

1)

$$\begin{aligned} p(x) &= \int_{-\infty}^{\infty} p(xy) dy \\ &= \int_{-\infty}^{\infty} \frac{1}{2\pi\sqrt{SN}} \exp \left\{ -\frac{1}{2N} \left[x^2 \left(1 + \frac{N}{S} \right) - 2xy + y^2 \right] \right\} dy \\ &= \frac{1}{2\pi\sqrt{SN}} \exp \left(-\frac{x^2}{2S} \right) \int_{-\infty}^{\infty} \exp \left[-\frac{1}{2N} (y-x)^2 \right] dy \\ &= \frac{1}{2\pi\sqrt{SN}} \exp \left(-\frac{x^2}{2S} \right) \sqrt{2N\pi} \\ &= \frac{1}{\sqrt{2S\pi}} \exp \left(-\frac{x^2}{2S} \right) \end{aligned}$$

2)

$$\begin{aligned}
H(X) &= - \int_{-\infty}^{\infty} p(x) \log p(x) dx \\
&= - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2S\pi}} \exp\left(-\frac{x^2}{2S}\right) \log \left[\frac{1}{\sqrt{2S\pi}} \exp\left(-\frac{x^2}{2S}\right) \right] dx \\
&= - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2S\pi}} \exp\left(-\frac{x^2}{2S}\right) \log\left(\frac{1}{\sqrt{2S\pi}}\right) dx - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2S\pi}} \exp\left(-\frac{x^2}{2S}\right) \log \left[\exp\left(-\frac{x^2}{2S}\right) \right] dx \\
&= \frac{1}{\sqrt{2S\pi}} \log(\sqrt{2S\pi}) \sqrt{2S\pi} - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2S\pi}} \exp\left(-\frac{x^2}{2S}\right) \left(-\frac{x^2}{2S}\right) \log(e) dx \\
&= \log(\sqrt{2S\pi}) + \frac{\log(e)}{2} \\
&= \frac{1}{2} \log 2\pi e S
\end{aligned}$$

3)

$$\begin{aligned}
p(y) &= \int_{-\infty}^{\infty} p(xy) dx \\
&= \int_{-\infty}^{\infty} \frac{1}{2\pi\sqrt{SN}} \exp\left\{-\frac{1}{2N} \left[x^2 \left(1 + \frac{N}{S}\right) - 2xy + y^2 \right]\right\} dx \\
&= \int_{-\infty}^{\infty} \frac{1}{2\pi\sqrt{SN}} \exp\left\{-\frac{1}{2N} \left[\left(x - \frac{Sy}{S+N}\right)^2 \left(1 + \frac{N}{S}\right) + \frac{N}{S+N} y^2 \right]\right\} dx \\
&= \frac{1}{\sqrt{2\pi(S+N)}} \exp\left(-\frac{1}{2S+2N} y^2\right)
\end{aligned}$$

4)

$$\begin{aligned}
H(Y) &= - \int_{-\infty}^{\infty} p(y) \log p(y) dy \\
&= - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi(S+N)}} \exp\left(-\frac{1}{2S+2N} y^2\right) \log \left[\frac{1}{\sqrt{2\pi(S+N)}} \exp\left(-\frac{1}{2S+2N} y^2\right) \right] dy \\
&= - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi(S+N)}} \exp\left(-\frac{1}{2S+2N} y^2\right) \log\left(\frac{1}{\sqrt{2\pi(S+N)}}\right) dy \\
&\quad - \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi(S+N)}} \exp\left(-\frac{1}{2S+2N} y^2\right) \log \left[\exp\left(-\frac{1}{2S+2N} y^2\right) \right] dy \\
&= \log\left(\sqrt{2\pi(S+N)}\right) + \frac{\log(e)}{2} \\
&= \frac{1}{2} \log 2\pi e (S+N)
\end{aligned}$$

Channel capacity

Ex. 4 (9 points)

Indicate the kinds of the following four channels (symmetric channel, strongly symmetric channel, quasi symmetrical channel, or general channel), and calculate the channel capacity in bit/symbol.

$$1) \begin{bmatrix} \frac{1}{2} & \frac{1}{2} & 0 & 0 \\ 0 & \frac{1}{2} & \frac{1}{2} & 0 \\ 0 & 0 & \frac{1}{2} & \frac{1}{2} \\ \frac{1}{2} & 0 & 0 & \frac{1}{2} \end{bmatrix} \text{ (3 points)}$$

$$2) \begin{bmatrix} \frac{3}{11} & \frac{2}{11} & \frac{2}{11} & \frac{2}{11} & \frac{2}{11} \\ \frac{2}{11} & \frac{3}{11} & \frac{2}{11} & \frac{2}{11} & \frac{2}{11} \\ \frac{2}{11} & \frac{2}{11} & \frac{3}{11} & \frac{2}{11} & \frac{2}{11} \\ \frac{2}{11} & \frac{2}{11} & \frac{2}{11} & \frac{3}{11} & \frac{2}{11} \\ \frac{2}{11} & \frac{2}{11} & \frac{2}{11} & \frac{2}{11} & \frac{3}{11} \end{bmatrix} \text{ (3 points)}$$

$$3) \begin{bmatrix} 0.5 & 0.2 & 0.3 \\ 0.3 & 0.2 & 0.5 \end{bmatrix} \text{ (3 points)}$$

Solution

- 1) It is a symmetric channel.

$$C = \log_2 m - H(p_{i_1}, p_{i_2}, \dots, p_{i_m}) = \log_2 4 - 2 \times \frac{1}{2} \log_2 2 = 1 \text{ bit/symbol}$$

- 2) It is a strongly symmetric channel, with $m = 5$, $p = \frac{8}{11}$.

$$C = \log_2 m - H(1 - p, p) - p \log_2 (m - 1) = \log_2 5 + \frac{3}{11} \log_2 \frac{3}{11} +$$

$$\frac{8}{11} \log_2 \frac{8}{11} - \frac{8}{11} \log_2 4 = 0.022 \text{ bit/symbol}$$

- 3) It is a quasi-symmetrical channel.

$$C = \log_2 2 - H(0.5, 0.2, 0.3) - 0.8 \log_2 (0.8) - 0.2 \log_2 (0.4) = 0.0365 \text{ bit/symbol}$$

Ex. 5 (20 points)

Calculate the channel capacity in bit/symbol of the following channel with the

transition probability matrix of the channel is $P = \begin{bmatrix} 0 & 0 & 1 & 0 & 0 \\ 0.1 & 0 & 0.2 & 0.6 & 0.1 \\ 0.4 & 0 & 0.3 & 0.2 & 0.1 \\ 0 & 1 & 0 & 0 & 0 \\ 0.1 & 0.2 & 0 & 0.2 & 0.5 \end{bmatrix}$,

and the corresponding probability distribution of the input of the channel if the average mutual information arrives at the channel capacity.

Solution

Since $\sum_{j=1}^5 p(y_j|x_i) \log p(y_j|x_i) = \sum_{j=1}^5 p(y_j|x_i) \beta_j$

we have,

$$\begin{cases} \log(1) = \beta_3 \\ 0.1 \log(0.1) + 0.2 \log(0.2) + 0.6 \log(0.6) + 0.1 \log(0.1) = 0.1\beta_1 + 0.2\beta_3 + 0.6\beta_4 + 0.1\beta_5 \\ 0.4 \log(0.4) + 0.3 \log(0.3) + 0.2 \log(0.2) + 0.1 \log(0.1) = 0.4\beta_1 + 0.3\beta_3 + 0.2\beta_4 + 0.1\beta_5 \\ \log(1) = \beta_2 \\ 0.1 \log(0.1) + 0.2 \log(0.2) + 0.2 \log(0.2) + 0.5 \log(0.5) = 0.1\beta_1 + 0.2\beta_2 + 0.2\beta_4 + 0.5\beta_5 \end{cases}$$

We get $\beta_2 = \beta_3 = 0$, $\beta_1 = -3.209$, $\beta_4 = -1.718$, $\beta_5 = -2.193$

So,

$$C = \log_2 \sum_{j=1}^5 2^{\beta_j} = \log_2 [2 + 2^{-3.209} + 2^{-1.718} + 2^{-2.193}] = 1.396 \text{ bits/symbol}$$

(10 points)

$$\begin{aligned} p(y_1) &= 2^{\beta_1 - C} = 0.041 \\ p(y_2) &= p(y_3) = 2^{\beta_2 - C} = 0.38 \\ p(y_4) &= 2^{\beta_4 - C} = 0.116 \\ p(y_5) &= 2^{\beta_5 - C} = 0.083 \end{aligned}$$

Since $\sum_{i=1}^5 p(x_i) p(y_j|x_i) = p(y_j)$,

We have,

$$\begin{cases} 0.1p(x_2) + 0.4p(x_3) + 0.1p(x_5) = p(y_1) \\ p(x_4) + 0.2p(x_5) = p(y_2) \\ p(x_1) + 0.2p(x_2) + 0.3p(x_3) = p(y_3) \\ 0.6p(x_2) + 0.2p(x_3) + 0.2p(x_5) = p(y_4) \\ 0.1p(x_2) + 0.1p(x_3) + 0.5p(x_5) = p(y_5) \end{cases}$$

Finally,

$$\begin{aligned} p(x_1) &= 0.342 \\ p(x_2) &= 0.138 \\ p(x_3) &= 0.035 \\ p(x_4) &= 0.354 \\ p(x_5) &= 0.131 \end{aligned}$$

(10 points)

Ex. 6 (7 points)

A telephone signal has an information rate of 5.6×10^4 bits/second. It is transmitted over a Gaussian channel with a noise **power spectral density** of $N_0 = 5 \times 10^{-6}$ **mW/Hz**, a limited bandwidth W , and a limited power of signal S .

- 1) When $W = 4$ kHz, determine the minimum power of signal S (in watts) required for error-free transmission. (2 points)
- 2) When $W \rightarrow \infty$, determine the minimum power of signal S (in watts) required for error-free transmission. (5 points)

Hint:

$$\ln(1+x) = \sum_{n=1}^{\infty} (-1)^{n-1} \frac{x^n}{n}$$

Solution

1)

$$C_t = W \log \left(1 + \frac{S}{WN_0} \right)$$

$$S = WN_0 \left(2^{\frac{C_t}{W}} - 1 \right) = 4000 \times 5 \times 10^{-9} \times \left(2^{\frac{5.6 \times 10^4}{4000}} - 1 \right) = 0.328 \text{ W}$$

2)

$$W \rightarrow \infty$$

$$C_t = \frac{S}{N_0} \log_2 e$$

$$S = \frac{C_t N_0}{\log_2 e} = \frac{5.6 \times 10^4 \times 5 \times 10^{-9}}{\log_2 2.71828} = 1.94 \times 10^{-4} \text{ W}$$