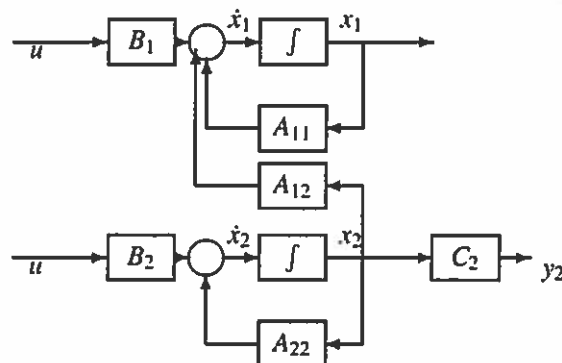


EE4-25

SOLUTIONS: DESIGN OF LINEAR MULTIVARIABLE CONTROL SYSTEMS

1. a) i) The PBH test states that the realisation is observable if and only if  $[(A - \lambda I)^T \ C^T]^T$  has full rank for all complex  $\lambda$ . The matrix loses rank if  $\lambda$  is an eigenvalue of  $A_{11}$  so the realisation is unobservable.
- ii) It follows that the unobservable modes that can be deduced from the structure are the eigenvalues of  $A_{11}$ .
- iii) A realisation is detectable if and only if all the unobservable modes are stable. Since  $A_{22}$  is stable, and the modes of  $A_{11}$  are all unobservable, the realisation is detectable if and only if  $A_{11}$  is stable.
- iv) The diagram is shown below. The subsystem with  $x_1$  is unobservable.

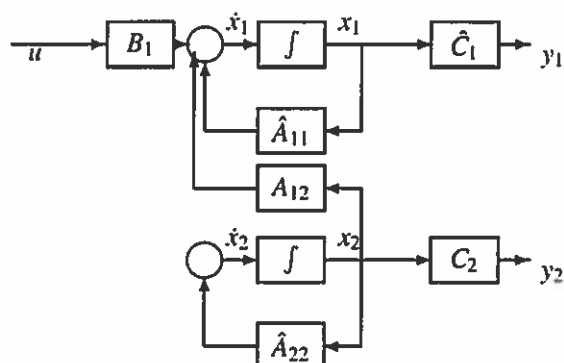


- b) i) Applying the suggested similarity transformation with  $*$  replaced by  $X$  and using the given relations gives

$$G(s) \stackrel{s}{=} \left[ \begin{array}{c|c} T^{-1}AT & T^{-1}B \\ \hline CT & D \end{array} \right] \stackrel{s}{=} \left[ \begin{array}{cc|c} A_{11} + A_{12}X & A_{12} & B_1 \\ 0 & A_{22} - XA_{12} & 0 \\ \hline C_1 + C_2X & C_2 & 0 \end{array} \right].$$

The PBH test now shows that the realisation is uncontrollable.

- ii) These are the modes of  $A_{22} - XA_{12}$ .
- iii) Since  $A_{12}X + A_{11}$  is stable, a necessary and sufficient condition is that  $A_{22} - XA_{12}$  is stable
- iv) The diagram is shown below with  $\hat{A}_{11} = A_{11} + A_{12}X$ ,  $\hat{A}_{22} = A_{22} - XA_{12}$  and  $\hat{C}_1 = C_1 + C_2X$ . The subsystem with  $x_2$  is uncontrollable.



2. a) An inspection of Figure 2 shows that

$$\begin{aligned}\dot{x} - \hat{\dot{x}} &= (A + LC)(x - \hat{x}) + \begin{bmatrix} B_w & -L \end{bmatrix} \begin{bmatrix} w_1 \\ w_2 \end{bmatrix} \\ z &= C_z(x - \hat{x})\end{aligned}$$

It follows that

$$T_{zw}(s) \triangleq \left[ \begin{array}{c|c} A + LC & \begin{bmatrix} B_w & -L \end{bmatrix} \\ \hline C_z & \begin{bmatrix} 0 & 0 \end{bmatrix} \end{array} \right] \triangleq: \left[ \begin{array}{c|c} A_c & B_c \\ \hline C_c & D_c \end{array} \right]$$

- b) The Bounded Real Lemma states that  $A_c$  is stable  $\|T_{zw}\|_\infty < \gamma$  if there exists a  $P = P'$  such that

$$\begin{bmatrix} A_c'P + PA_c + C_c'C_c & PB_c + C_c'D_c \\ B_c'P + D_c'C_c & D_c'D_c - \gamma^2 I \end{bmatrix} \prec 0 \\ P = P' \succ 0$$

By substituting the expressions for  $A_c, B_c, C_c$  and  $D_c$ , this becomes

$$\begin{bmatrix} (A + LC)'P + P(A + LC) + C_z'C_z & PB_w & -PL \\ * & -\gamma^2 I & 0 \\ * & * & -\gamma^2 I \end{bmatrix} \prec 0 \\ P = P' \succ 0$$

where  $*$  denotes terms easily inferred from symmetry.

- c) By defining  $Y = PL$ , the matrix inequalities become

$$\begin{bmatrix} PA + A'P + YC + C'Y' + C_z'C_z & PB_w & -Y \\ * & -\gamma^2 I & 0 \\ * & * & -\gamma^2 I \end{bmatrix} \prec 0 \\ P = P' \succ 0$$

which are linear.

- d) Putting the numbers into the LMI:

$$\begin{bmatrix} -2P + 2Y + 2 & P & -Y \\ * & -\gamma^2 I & 0 \\ * & * & -\gamma^2 I \end{bmatrix} \prec 0 \\ P = P' \succ 0$$

effecting a Schur complement, this is equivalent to

$$-2P + 2Y + 2 + \gamma^{-2}Y^2 + \gamma^{-2}P^2 \prec 0, \quad P \succ 0$$

which when completing two squares become

$$(\gamma^{-1}P - \gamma)^2 + (\gamma^{-1}Y + \gamma)^2 + 2 - 2\gamma^2 \prec 0, \quad P \succ 0$$

and so  $2\gamma^2 > 2$  or  $\gamma > 1$ . In the limit when  $\gamma \rightarrow 1$ ,  $P \rightarrow 1$ ,  $Y \rightarrow -1$  and so  $L \rightarrow -1$ .

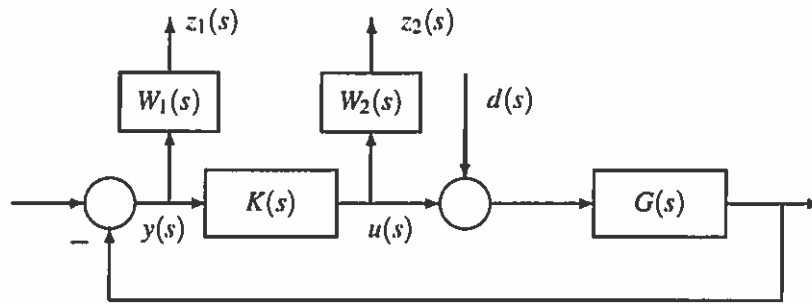
3. a) It is clear that we require  $K(s)$  to be internally stabilising.

A calculation shows that  $y(s) = T_{yd}(s)d(s)$  where  $T_{yd}(s) = -(I + G(s)K(s))^{-1} G(s)$ . It follows that a sufficient condition to achieve the first specification is  $\|T_{yd}(j\omega)\| < |w_1(j\omega)|^{-1} \forall \omega$  or, equivalently,  $\|W_1 T_{yd}\|_\infty < 1$ , where  $W_1(s) = w_1(s)I$ .

A similar calculation shows that  $u(s) = T_{ud}(s)d(s)$  where  $T_{ud}(s) = -K(s)(I + G(s)K(s))^{-1} G(s)$ . It follows that a sufficient condition to achieve the second specification is  $\|T_{ud}(j\omega)\| < |w_2(j\omega)|^{-1} \forall \omega$  or, equivalently,  $\|W_2 T_{ud}\|_\infty < 1$ , where  $W_2(s) = w_2(s)I$ .

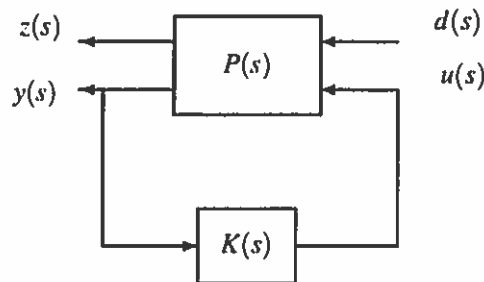
To satisfy both design requirements, it is sufficient that  $\left\| \begin{bmatrix} W_1 T_{yd} \\ W_2 T_{ud} \end{bmatrix} \right\|_\infty < 1$ .

- b) The cost signals are given as  $z_1(s) = W_1(s)y(s)$  and  $z_2(s) = W_2(s)u(s)$ . The block diagram incorporating  $z_1(s)$  and  $z_2(s)$  is shown below.



- c) The corresponding generalised regulator formulation is to find an internally stabilising  $K(s)$  such that  $\|\mathcal{F}_l(P, K)\|_\infty < 1$  where

$$z(s) = \begin{bmatrix} z_1(s) \\ z_2(s) \end{bmatrix}, P(s) = \begin{bmatrix} P_{11}(s) & P_{12}(s) \\ P_{21}(s) & P_{22}(s) \end{bmatrix} = \left[ \begin{array}{c|c} -W_1(s)G(s) & -W_1(s)G(s) \\ \hline 0 & W_2(s) \\ \hline -G(s) & -G(s) \end{array} \right].$$



- d) Suppose that  $\Delta$  and  $S$  are stable. Then the feedback loop is stable if  $\|\Delta S\|_\infty < 1$ .
- e) Let  $K(s)$  be replaced by  $K(s) + \Delta(s)$  in Figure 3 and let  $\epsilon$  be the input and  $\delta$  be the output of  $\Delta$ . Then  $\epsilon = -(I + GK)^{-1} G\delta$ . Using the small gain theorem the maximum stability radius is  $|w_1^{-1}(j\omega)|$ .

4. a) A suitable Lyapunov function for regulating  $x$  is  $V = x'Px$  where  $P = P'$ .  
 b) Set  $u = -Fx$ . Provided that  $P = P' \succ 0$  and  $\dot{V} < 0$  along closed-loop trajectories, we can assume  $\lim_{t \rightarrow \infty} x(t) = 0$ . Then

$$\dot{V} = x'Px + x'P\dot{x} = x'(A'P + PA - F'B'P - PBF)x.$$

Using  $x(\infty) = 0$ ,

$$\int_0^\infty x'(A'P + PA - F'B'P - PBF)x dt = -x_0'Px_0.$$

- c) Adding the last equation to the expression for  $J$  and completing a square:

$$J = x_0'Px_0 + \int_0^\infty \{x'[A'P + PA + C'C - PBB'P]x + \|(F - B'P)x\|^2\} dt.$$

Since the last term is always nonnegative, it follows that the minimizing value of  $F$  is given by  $F = B'P$ . We can set the term in square brackets to zero provided  $P$  satisfies the Riccati equation,

$$A'P + PA + C'C - PBB'P = 0.$$

It follows that the minimum value of  $J$  is  $x_0'Px_0$ .

- d) We need to prove that  $A_c := A - BB'P$  is stable. The Riccati equation can be written as  $A_c'P + PA_c + C'C + PBB'P = 0$ . Let  $\lambda \in \mathcal{C}$  be an eigenvalue of  $A_c$  and  $y \neq 0$  be the corresponding eigenvector. Pre- and post-multiplying the Riccati equation by  $y'$  and  $y$  respectively gives  $(\lambda + \bar{\lambda})y'Py + y'C'Cy + y'PBB'Py = 0$ . Since  $P \succ 0$  and  $y \neq 0$ ,  $y'Py > 0$ ,  $y'y > 0$  and  $y'PBB'Py \geq 0$ . It follows that  $\lambda + \bar{\lambda} < 0$  and the closed loop is stable.  
 e) Since  $\|w\|_2 \leq 1$ , then an upper bound on  $\|z\|_2$  is  $\|T_{zw}\|_\infty$ . Now,

$$\begin{aligned} \dot{x} &= Ax + Bu = Ax + B(w - Fx) \\ &= (A - BF)x + Bw \\ z &= Cx \end{aligned}$$

it follows that  $T_{zw} \stackrel{s}{=} (A - BF, B, C, 0)$ . It follows from the bounded real lemma that  $\|T_{zw}\|_\infty < 1$  if there exists  $P = P' \succ 0$  such that

$$\begin{bmatrix} P(A - BF) + (A - BF)'P + C'C & PB \\ B'P & -I \end{bmatrix} \prec 0$$

Using a Schur complement argument, this inequality is equivalent to

$$P(A - BF) + (A - BF)'P + C'C + PBB'P \prec 0.$$

However, it follows from the Riccati equation in Part b above that  $P(A - BF) + (A - BF)'P + C'C + PBB'P = 0$ . This proves that  $\|T_{zw}\|_\infty < 1$  and so  $\|z\|_2 < 1$ .

5. a) i) The (1, 1) block of the inequality gives the inequality  $A'P + PA \prec 0$ . Let  $z \neq 0$  be a right eigenvector of  $A$  and let  $\lambda$  be the corresponding eigenvalue. Then multiplying the inequality from the left by  $z'$  and from the right by  $z$  gives  $(\lambda + \bar{\lambda})z'Pz < 0$ . Since  $P \succ 0$  it follows that  $z'Pz > 0$  and it follows that  $\lambda + \bar{\lambda} < 0$  so that  $A$  is stable.

ii) Let  $x, u$  and  $y$  denote the state, input and output for  $H(s)$ . Since  $A$  is stable,  $\|H\|_\infty < \gamma$  if and only if, with  $x(0) = 0$ ,  $J := \int_0^\infty [y'y - \gamma^2 u'u] dt < 0$ , for all  $u(t)$  such that  $\|u\|_2 < \infty$ . If  $\|u\|_2$  is bounded, then  $\lim_{t \rightarrow \infty} x(t) = 0$ . Now,  $\int_0^\infty \frac{d}{dt} [x'Px] dt = x(\infty)'Px(\infty) - x(0)'Px(0) = 0$ . So,

$$0 = \int_0^\infty (\dot{x}'Px + x'P\dot{x}) dt = \int_0^\infty [x'(A'P + PA)x + x'PBu + u'B'Px] dt.$$

Use  $y = Cx + Du$  and add the last expression to  $J$

$$\begin{aligned} J &= \int_0^\infty [x'(A'P + PA + C'C)x + 2x'(PB + C'D)u + u'(D'D - \gamma^2 I)u] dt \\ &= \int_0^\infty \begin{bmatrix} x' & u' \end{bmatrix} \overbrace{\begin{bmatrix} A'P + PA + C'C & PB + C'D \\ B'P + D'C & D'D - \gamma^2 I \end{bmatrix}}^M \begin{bmatrix} x \\ u \end{bmatrix} dt. \end{aligned}$$

It follows that  $J < 0$ , and so  $\|H\|_\infty < \gamma$ , if  $M \prec 0$ . This proves the result.

b) i) The state equations for Figure 5 give

$$\dot{x} = \underbrace{(A + BFC)}_{A_c} x + \underbrace{BF}_{B_c} r, \quad z = \underbrace{C}_{C_c} x + \underbrace{D}_{D_c} r.$$

It follows that  $T_{zr}(s) = D_c + C_c(sI - A_c)^{-1}B_c$ .

ii) The transfer matrix  $T_{zr}$  is the sensitivity for the feedback-loop and limiting its infinity norm will improve the tracking properties of the loop.

iii) Using the results of part (a), by replacing  $A, B, C$  and  $D$  by  $A_c, B_c, C_c$  and  $D_c$ , we have that there exists a feasible  $F$  if there exists  $P = P'$  such that

$$\begin{bmatrix} (A + BFC)'P + P(A + BFC) + C'C & PBF + C' \\ * & I - \gamma^2 I \end{bmatrix} \prec 0 \\ P \succ 0$$

Noting that the only nonlinearity is due to the product  $PBF$ , and that  $B$  is square and nonsingular, we define  $Z = PBF$  and so there exists a feasible  $F$  if there exists  $P = P'$  and  $Z$  such that

$$\begin{bmatrix} PA + A'P + ZC + C'Z' + C'C & Z + C' \\ * & I - \gamma^2 I \end{bmatrix} \prec 0 \\ P \succ 0$$

in which case  $F = B^{-1}P^{-1}Z$ .

6. a) The generalized regulator formulation is given by

$$\begin{bmatrix} z(s) \\ y(s) \end{bmatrix} = P(s) \begin{bmatrix} w(s) \\ u(s) \end{bmatrix}, u(s) = Fy(s), P(s) = \begin{bmatrix} P_{11}(s) & P_{12}(s) \\ P_{21}(s) & P_{22}(s) \end{bmatrix} \triangleq \left[ \begin{array}{c|c|c} A & B & B \\ \hline C & 0 & 0 \\ \hline 0 & 0 & I \\ \hline I & 0 & 0 \end{array} \right].$$

- b) The requirement  $\|H\|_\infty < \gamma$  is equivalent to  $J := \|z\|_2^2 - \gamma^2 \|w\|_2^2 < 0$ . Let  $V = x^T X x$  and set  $u = Fx$ . Provided that  $X = X^T > 0$  and  $\dot{V} < 0$  along the closed-loop trajectory, we can assume  $\lim_{t \rightarrow \infty} x(t) = 0$ . Then

$$\dot{V} = \dot{x}^T X x + x^T X \dot{x} = x^T (A^T X + XA + F^T B^T X + XBF) x + x^T X B w + w^T B^T X x.$$

Integrating from 0 to  $\infty$  and using  $x(0) = x(\infty) = 0$ ,

$$0 = \int_0^\infty [x^T (A^T X + XA + F^T B^T X + XBF) x + x^T X B w + w^T B^T X x] dt. \quad (6.1)$$

Using the definition of  $J$  and adding the last equation,  $J =$

$$\int_0^\infty \{x^T [A^T X + XA + C^T C + F^T F + F^T B^T X + XBF] x - [\gamma^2 w^T w - x^T X B w - w^T B^T X x]\} dt.$$

Let  $Z = F + B^T X$ . Completing the squares gives

$$J = \int_0^\infty \{x^T [A^T X + XA + C^T C - (1 - \gamma^{-2}) X B B^T X] x + \|Zx\|^2 - \|\gamma v - \gamma^{-1} B^T X x\|^2\} dt.$$

Thus two sufficient conditions for  $J < 0$  are the existence of  $X$  such that

$$A^T X + XA + C^T C - (1 - \gamma^{-2}) X B B^T X = 0, \quad X = X^T > 0.$$

The state feedback gain is  $F = -B^T X$  (ensuring  $Z = 0$ ) and the worst case disturbance is  $w^* = \gamma^{-2} B^T X x$ . The closed-loop with these feedback laws is  $\dot{x} = [A - (1 - \gamma^{-2}) B B^T X] x$  and a third condition is therefore  $\text{Re } \lambda_i [A - (1 - \gamma^{-2}) B B^T X] < 0, \forall i$ .

It remains to prove  $\dot{V} < 0$  along state-trajectory with  $u = Fx$  and  $w = 0$ . But

$$\dot{V} = x^T (A^T X + XA + F^T B^T X + XBF) x = -x^T (C^T C + (1 + \gamma^{-2}) X B B^T X) x < 0$$

for all  $x \neq 0$  (since  $(A, B, C)$  is assumed minimal) proving closed-loop stability.

- c) Setting  $w = 0$  and  $\gamma \rightarrow \infty$  and assuming  $x(0) = x_0 \neq 0$  implies that (6.1) now becomes

$$-x_0^T X x_0 = \int_0^\infty [x^T (A^T X + XA + F^T B^T X + XBF) x + x^T X B w + w^T B^T X x] dt.$$

Adding this to the cost function and proceeding as before gives the Riccati equation as

$$A^T X + XA + C^T C - X B B^T X = 0, \quad X = X^T > 0.$$

and the cost function as

$$J = x_0^T X x_0$$

This may be recognised as the solution of the LQR problem of minimizing  $\|z\|_2$ .