

IMPERIAL COLLEGE LONDON

DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING  
EXAMINATIONS 2013

EEE/EIE PART II: MEng, Beng and ACGI

**COMMUNICATION SYSTEMS**

Friday, 7 June 2:00 pm

Time allowed: 2:00 hours

**There are THREE questions on this paper.**

**Answer ALL questions.**

**Q1 carries 40% of the marks. Questions 2 and 3 carry equal marks (30% each).**

**Any special instructions for invigilators and information for candidates are on page 1.**

Examiners responsible	First Marker(s) :	C. Ling
	Second Marker(s) :	J.A. Barria

## EXAM QUESTIONS

### Information for Students

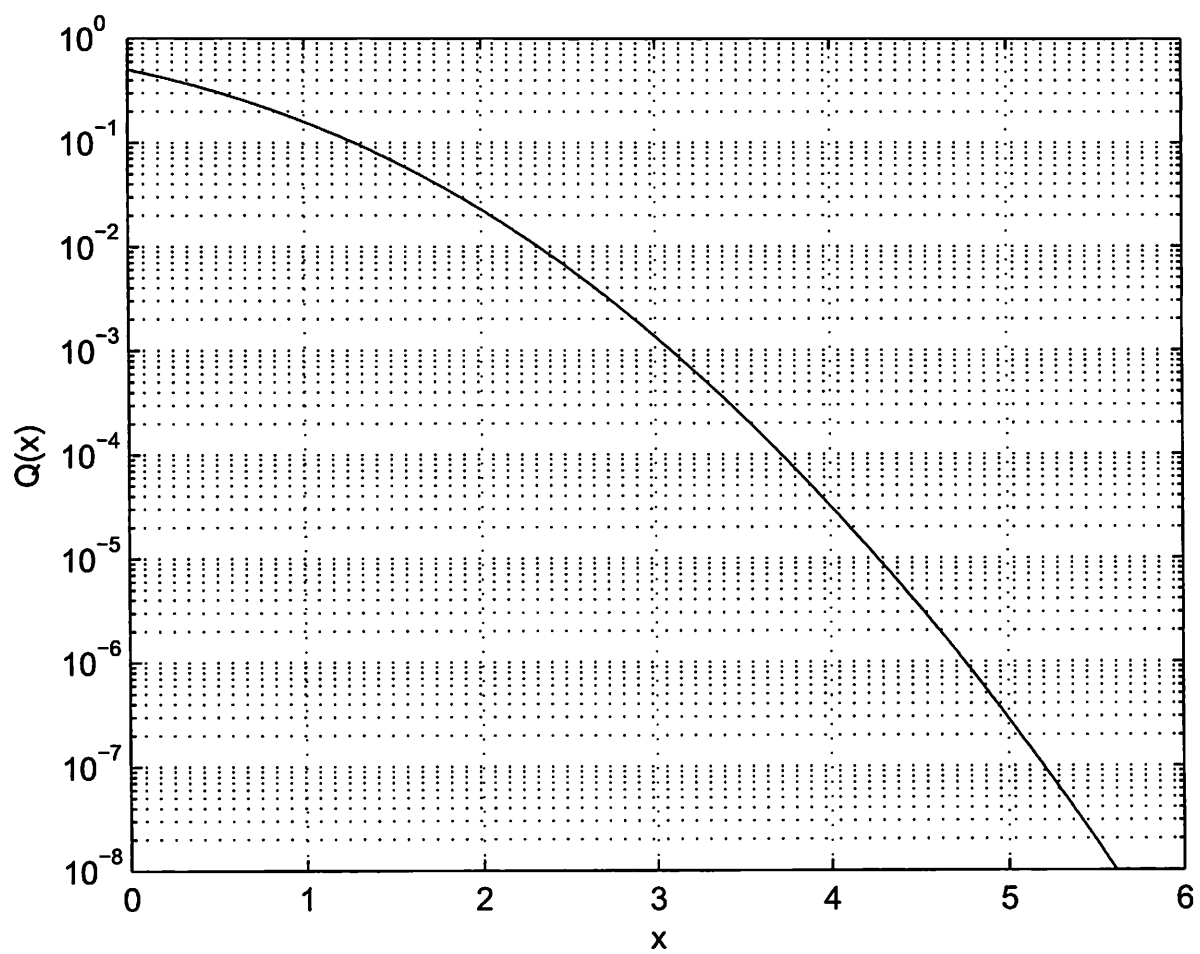


Figure 0.1 The graph of the Q-function.

**TABLE 4-2**  
Fourier Transform Pairs

Pair Number	$x(t)$	$X(f)$	Comments on Derivation
1.	$\Pi\left(\frac{t}{\tau}\right)$	$\tau \operatorname{sinc} \pi f$	Direct evaluation
2.	$2W \operatorname{sinc} 2Wt$	$\Pi\left(\frac{f}{2W}\right)$	Duality with pair 1, Example 4-7
3.	$\Lambda\left(\frac{t}{\tau}\right)$	$\tau \operatorname{sinc}^2 \pi f$	Convolution using pair 1
4.	$\exp(-\alpha t)u(t), \alpha > 0$	$\frac{1}{\alpha + j2\pi f}$	Direct evaluation
5.	$t \exp(-\alpha t)u(t), \alpha > 0$	$\frac{1}{(\alpha + j2\pi f)^2}$	Differentiation of pair 4 with respect to $\alpha$
6.	$\exp(-\alpha t ), \alpha > 0$	$\frac{2\alpha}{\alpha^2 + (2\pi f)^2}$	Direct evaluation
7.	$e^{-\pi(t/\tau)^2}$	$\tau e^{-\pi(f\tau)^2}$	Direct evaluation
8.	$\delta(t)$	1	Example 4-9
9.	1	$\delta(f)$	Duality with pair 7
10.	$\delta(t - t_0)$	$\exp(-j2\pi f t_0)$	Shift and pair 7
11.	$\exp(j2\pi f_0 t)$	$\delta(f - f_0)$	Duality with pair 9
12.	$\cos 2\pi f_0 t$	$\frac{1}{2}\delta(f - f_0) + \frac{1}{2}\delta(f + f_0)$	Exponential representation of cos and sin and pair 10
13.	$\sin 2\pi f_0 t$	$\frac{1}{2j}\delta(f - f_0) - \frac{1}{2j}\delta(f + f_0)$	
14.	$u(t)$	$(j2\pi f)^{-1} + \frac{1}{2}\delta(f)$	Integration and pair 7
15.	$\operatorname{sgn} t$	$(j\pi f)^{-1}$	Pair 8 and pair 13 with superposition
16.	$\frac{1}{\pi t}$	$-j \operatorname{sgn}(f)$	Duality with pair 14
17.	$\hat{x}(t) = \frac{1}{\pi} \int_{-\infty}^{\infty} \frac{x(\lambda)}{t - \lambda} d\lambda$	$-j \operatorname{sgn}(f)X(f)$	Convolution and pair 15
18.	$\sum_{m=-\infty}^{\infty} \delta(t - mT_s)$	$f_s \sum_{m=-\infty}^{\infty} \delta(f - mf_s),$ $f_s = T_s^{-1}$	Example 4-10

1.
  - a)
    - i) Consider two random variables  $X$  and  $Y$ . Explain the notions “uncorrelated” and “independent”. Discuss the relation between “uncorrelated” and “independent”. [ 4 ]
    - ii) Draw a diagram showing how to obtain in-phase component  $n_c(t)$  and quadrature component  $n_s(t)$  from the bandpass noise  $n(t)$ . [ 4 ]
  - b)
    - i) With the help of a phasor diagram, write down an equation for the envelope of the output of the AM envelope detector in the presence of noise  $n(t)$ , and give an approximation when the noise is small. [ 6 ]
    - ii) Name three advantages of digital communications, when compared to analog communications. [ 3 ]
  - c)
    - i) Explain and compare source coding and channel coding. Give examples. [ 4 ]
    - ii) What are Hamming codes? What is its minimum Hamming distance? [ 3 ]
    - iii) Given the parameters  $(n, k)$  where  $n$  is the codeword length and  $k$  is the number of information bits, what is the relation between  $n$  and  $k$  for Hamming codes?. [ 2 ]
    - iv) Give the pairs  $(n, k)$  for the first three Hamming codes. [ 2 ]
  - d) Consider a modulated communication system. At the receiver side, the pre-detection filter bandwidth is  $B$ , while the postdetection filter bandwidth is  $W$ . Suppose the predetection signal-to-noise ratio (SNR) is  $SNR_{in}$ , while the post-detection SNR is  $SNR_{out}$ .
    - i) Write down the formula for the maximum rate at which information may arrive at the receiver. [ 3 ]
    - ii) Write down the formula for the maximum rate at which information may leave the receiver. [ 3 ]
    - iii) Derive  $SNR_{out}$  of an ideal communication system as a function of  $SNR_{baseband}$ . Discuss the limit as the ratio  $B/W \rightarrow \infty$ . [ 6 ]

2. a) i) What is a stationary random process? What is the Wiener-Khinchine relation between the power spectral density and the autocorrelation function? [ 5 ]

ii) Given the power spectral density of the ideal low-pass white Gaussian noise

$$S_N(f) = \begin{cases} \frac{N_0}{2}, & |f| < B; \\ 0, & \text{otherwise.} \end{cases}$$

Write down the autocorrelation function using the Wiener-Khinchine relation.

Calculate the autocorrelation between samples taken at the Nyquist rate.

Discuss the meaning of your finding. [ 5 ]

iii) Consider the random process

$$X(t) = a \cos(\omega t + \Theta)$$

where  $a$  and  $\omega$  are constants and  $\Theta$  is a binary random variable taking values of 0 or  $\pi$  equiprobably, i.e.,  $P(\Theta = 0) = 1/2$  and  $P(\Theta = \pi) = 1/2$ . Determine whether this is a stationary or nonstationary process, by computing the mean and autocorrelation function. [ 8 ]

b) The output SNR of the FM receiver is given by

$$\text{SNR}_{\text{FM}} = 3\beta^2 \frac{P}{m_p^2} \text{SNR}_{\text{baseband}} \quad (2.1)$$

where  $P$  and  $m_p$  are the power and peak amplitude of the message, respectively. Assume the deviation ratio  $\beta = 5$  and the message  $m(t)$  is a zero-mean Gaussian random process.

Compute  $\text{SNR}_{\text{FM}}$  as a function of  $\text{SNR}_{\text{baseband}}$ , when the overload probability is  $6 \times 10^{-7}$  (i.e., the probability  $P(|m(t)| > m_p) = 6 \times 10^{-7}$ ).

(Hint: use the graph of the Q-function.) [ 12 ]

3. a) Consider a sequence of the English alphabet with their probabilities of occurrence given by

Letter	a	b	i	l	m	o
Probability	0.3	0.1	0.2	0.1	0.1	0.2

- i) Calculate the entropy of this source. [ 2 ]
  - ii) Construct a Huffman code and find the average codeword length. [ 5 ]
- b) Repetition codes represent the simplest type of linear block codes. In particular, a single message bit is encoded into a block of  $n$  identical bits, producing an  $(n, 1)$  code. Let  $n = 5$  in this question.
- i) Do you consider a repetition code to be systematic or non-systematic? Explain your answer. [ 2 ]
  - ii) Write down the generator matrix and parity-check matrix of a  $(5, 1)$  repetition code. [ 4 ]
  - iii) Determine the minimum Hamming distance. [ 2 ]
  - iv) Compute the syndrome table for all possible single-error patterns. [ 5 ]
  - v) Compute the syndrome table for all possible 10 double-error patterns. [ 5 ]
  - vi) Let us apply this  $(5, 1)$  code to a PSK-modulated system with signal-to-noise ratio  $\frac{A}{\sigma} = 4$ . Compute the raw error probability of PSK and the decoding error probability after majority-rule decoding. (Hint: Majority-rule decoding works as follows: if in a block of  $n$  received bits, the number of 0s exceeds the number of 1s, the decoder decides in favor of a 0; otherwise, it decides in favor of a 1.) [ 5 ]

[B] Bookwork

[E] New Example

[A] New Application

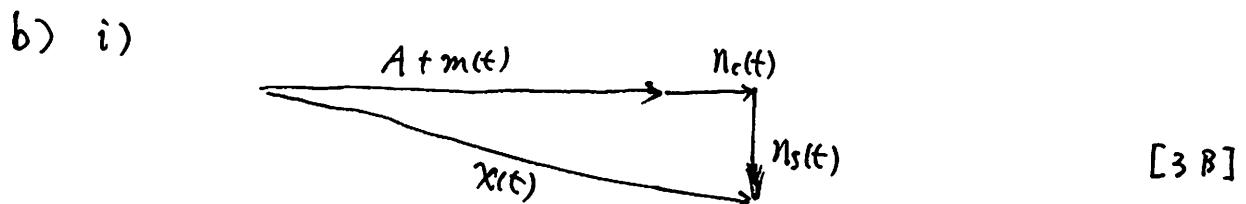
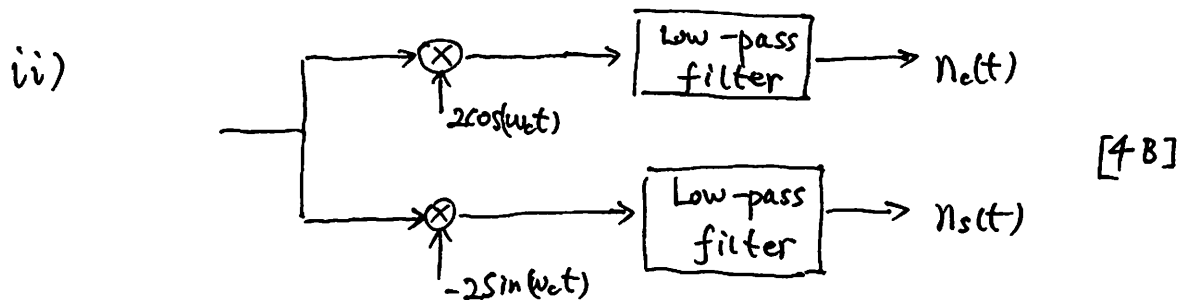
[T] New Theory

## ANSWERS

1. a) i) Uncorrelated:  $E[XY] = E[X]E[Y]$

independent:  $f_{XY}(xy) = f_X(x)f_Y(y)$  [4B]

"independent" implies uncorrelated, but the converse is not necessarily true.



$$y(t) = \text{envelope of } x(t)$$

$$= \sqrt{[A + m(t) + n_c(t)]^2 + n_s^2(t)}$$

When noise is small [3B]

$$y(t) \approx A + m(t) + n_c(t)$$

ii) Digital communication is [3B]

more immune to channel noise;

digital signals can be represented in a uniform format;  
easier to process;

flexible and allow for sophisticated functions;

able to provide digital services such as Internet.

c) i) Source coding is to compress the data by reducing the number of source bits, e.g., Huffman code. [4B]

Channel coding is to introduce redundant bits to enable to detection and correction of errors caused by the channel, e.g., Hamming code.

ii) Hamming codes are a class of linear block codes that can correct a single error. [3B]

For Hamming codes,  $d_{\min} = 3$ .

$$\text{iii) } r = n - k = \log_2(n+1) \Rightarrow n = 2^r - 1, \\ k = 2^r - 1 - r. \quad [2B]$$

First few Hamming codes:

$$(n, k) = (7, 4), (15, 11), (31, 26), \dots \quad [2B]$$

$$\text{d) i) } C_{in} = B \log_2(1 + SNR_{in}) \quad [3B]$$

$$\text{ii) } C_{out} = W \log_2(1 + SNR_{out}) \quad [3B]$$

$$\text{iii) For an ideal system, } C_{in} = C_{out}. \quad [2B]$$

$$W \log_2(1 + SNR_{out}) = B \log_2(1 + SNR_{in})$$

Since

$$SNR_{in} = \frac{P}{N_0 B} = \frac{W}{B} \frac{P}{N_0 W} = \frac{W}{B} SNR_{\text{baseband}}, \quad [2B]$$

We have

$$W \log_2(1 + SNR_{out}) = B \log_2\left(1 + \frac{SNR_{\text{baseband}}}{B/W}\right)$$

$$SNR_{out} = \left(1 + \frac{SNR_{\text{baseband}}}{B/W}\right)^{B/W} - 1. \quad [2B]$$

$$\rightarrow e^{SNR_{\text{baseband}}} \text{ as } B/W \rightarrow \infty.$$



2. i) Stationary process: [5B]

$$\begin{aligned} \mu_x(t) &= \mu_x && \text{doesn't depend on } t; \\ R_x(t, t+\tau) &= R_x(\tau) && \text{is a function of } \tau \text{ only.} \end{aligned}$$

Wiener-Khinchine relation: Power spectral density is the Fourier transform of the autocorrelation function.

ii) From the table,

$$R_x(\tau) = N_0 B \operatorname{sinc}(2B\tau). \quad [2B]$$

If taken at Nyquist rate,  $\tau = \frac{k}{2B}$ ,  $k = 0, \pm 1, \pm 2, \dots$

Then,

$$R_x(\tau) = N_0 B \operatorname{sinc}(k) = 0, \quad k = \pm 1, \pm 2, \dots \quad [3B]$$

This means the samples are uncorrelated, hence being independent since they are Gaussian.

iii)

$$\text{Mean: } E[X(t)] = E[a \cos(\omega t + \theta)]$$

$$= \frac{1}{2} [a \cos(\omega t) + a \cos(\omega t + \pi)] \quad [2E]$$

$$= 0$$

Autocorrelation

$$\begin{aligned} R_x(t, t+\tau) &= E[a \cos(\omega t + \theta) a \cos(\omega(t+\tau) + \theta)] \\ &= \frac{1}{2} [a^2 \cos(\omega t) \cos(\omega(t+\tau)) + \\ &\quad a^2 \cos(\omega t + \pi) \cos(\omega(t+\tau) + \pi)] \quad [4E] \end{aligned}$$

$$= a^2 \cos(\omega t) \cos(\omega(t+\tau))$$

$$= \frac{a^2}{2} [\cos(\omega(2t+\tau)) + \cos(\omega\tau)]$$

can't get rid of  $t$ !

Therefore, it is NOT stationary.

[2E]

b

Since  $m(t)$  is Gaussian,

$$P(|m(t)| > m_p) = 2 Q\left(\frac{m_p}{\sigma}\right) = 6 \times 10^{-7} \quad [3A]$$

where  $\sigma$  is the standard deviation. Thus,  $P = \sigma^2$ .

$$\text{From the graph, if } Q(x) = 3 \times 10^{-7}, \quad [3A]$$

$$x = 5.$$

$$\text{Therefore, } \frac{m_p}{\sigma} = 5.$$

$$\text{This means } \frac{P}{m_p^2} = \frac{\sigma^2}{m_p^2} = \frac{1}{25}. \quad [3A]$$

Therefore,

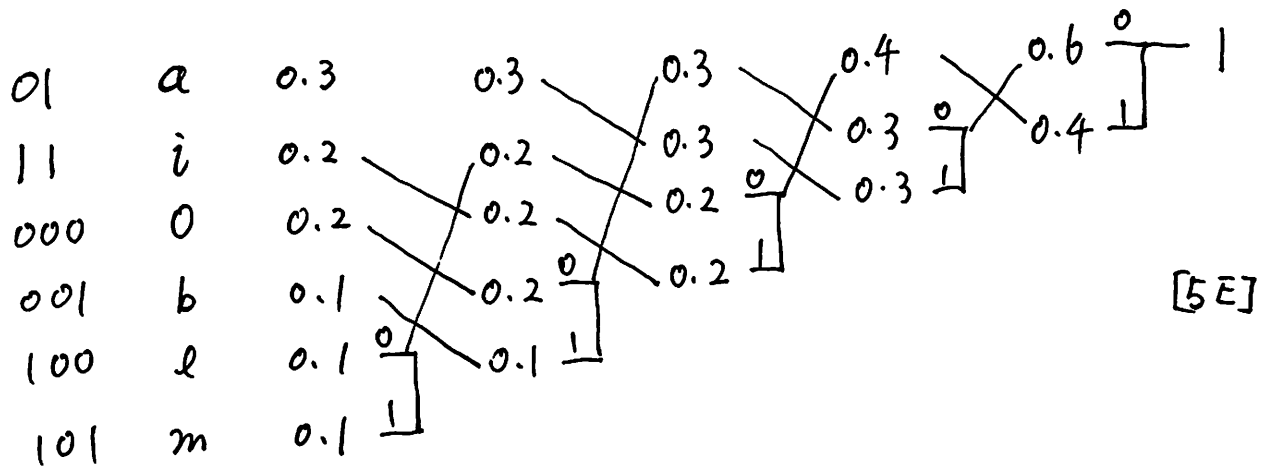
$$\begin{aligned} SNR_{FM} &= 3 \times 5^2 \times \frac{1}{25} SNR_{\text{baseband}} \\ &= 3 SNR_{\text{baseband}}. \end{aligned} \quad [3A]$$

3. a) i)  $H(S) = - \sum p_i \log_2 p_i$  [2E]

$$= - (0.3 \log_2 0.3 + 2 \times 0.2 \times \log_2 0.2 + 3 \times 0.1 \times \log_2 0.1)$$

$$= 2.45$$

ii)



average length

$$L = 2 \times 0.5 + 3 \times 0.5$$

$$= 2.5$$

b) i) Systematic, because the information bit appears as is. [2A]

ii)  $G = [1 \ 1 \ 1 \ 1 \ 1]$  [2A]

$H = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 \\ 1 & 0 & 1 & 1 & 0 \\ 1 & 0 & 1 & 0 & 1 \end{bmatrix}$  [2A]

iii)  $d_{\min} = 5$  [2A]

iv)  $S = eH^T$

s	e
1 1 1 1	1 0 0 0 0
1 0 0 0	0 1 0 0 0
0 1 0 0	0 0 1 0 0
0 0 1 0	0 0 0 1 0
0 0 0 1	0 0 0 0 1

[5A]

v)

s	e
0 1 1 1	1 1 0 0 0
1 0 1 1	1 0 1 0 0
1 1 0 1	1 0 0 1 0
1 1 1 0	1 0 0 0 1
1 1 0 0	0 1 1 0 0
1 0 1 0	0 1 0 1 0
1 0 0 1	0 1 0 0 1
0 1 1 0	0 0 1 1 0
0 1 0 1	0 0 1 0 1
0 0 1 1	0 0 0 1 1

[5A]

vi) For coherent PSK, the raw error probability is

$$P_e = Q\left(\frac{A}{2\sigma}\right) \\ = Q(2) \approx 2.2 \times 10^{-2}$$

With majority-rule decoding, ( $p = 2.2 \times 10^{-2}$ )

$$P_e = \sum_{i=3}^5 \binom{5}{i} p^i (1-p)^{5-i} \quad [5A]$$

$$= \binom{5}{3} p^3 (1-p)^2 + \binom{5}{4} p^4 (1-p) + \binom{5}{5} p^5$$

$$= 10 p^3 (1-p)^2 + 5 p^4 (1-p) + p^5$$

$$= 1 \times 10^{-4}$$