

SOLUTIONS: DESIGN OF LINEAR MULTIVARIABLE CONTROL SYSTEMS

1. a) i) Since $\begin{bmatrix} A - sI & B \end{bmatrix}$ loses rank for $s = 3$, 3 is an uncontrollable mode, and since $\begin{bmatrix} A' - sI & C' \end{bmatrix}$ loses rank for $s = -4$, -4 is an unobservable mode. Since the uncontrollable mode is unstable, the realisation is not stabilisable and since the unobservable mode is stable, the realisation is detectable.

- ii) By removing the uncontrollable and unobservable parts we get the minimal realisation

$$G(s) \triangleq \left[\begin{array}{c|cc} -1 & 1 & 2 \\ \hline 2 & 0 & 0 \\ 1 & 0 & 0 \end{array} \right].$$

- b) i) Suppose that λ is an eigenvalue of A and let $z \neq 0$ be the corresponding eigenvector. Then $Az = \lambda z$. Pre- and post-multiplying the matrix inequality by z' and z , respectively, we get

$$(\lambda + \bar{\lambda})z'Qz < 0.$$

Since $z \neq 0$ and $Q \succ 0$, this implies that $z'Qz > 0$ so that $\lambda + \bar{\lambda} < 0$ and so A is stable.

- ii) The pair (A, C) is detectable if and only if there exists L such that $A + LC$ is stable. That is, the pair (A, C) is detectable if and only if there exist L and $Q = Q' \succ 0$ such that

$$(A + LC)'Q + Q(A + LC) \prec 0.$$

Comparing this with the inequality in the question, it follows that the pair (A, C) is detectable by identifying Y with QL . It follows that we can define L as $L = Q^{-1}Y$ since $Q \succ 0$ implies Q^{-1} exists.

- iii) Using duality: Suppose there exist $P = P' \succ 0$ and Z such that

$$AP + PA' + BZ + Z'B' \prec 0.$$

Then the pair (A, B) is stabilisable. Furthermore, with $K = ZP^{-1}$, then $A + BK$ is stable.

2. a) i)

Assume first A is stable. Then $\|H\|_\infty < \gamma$ if and only if, with $x(0) = 0$, $J := \int_0^\infty [y'y - \gamma^2 u'u] dt < 0$, for all $u(t)$ such that $\|u\|_2 < \infty$. If $\|u\|_2$ is bounded, then $\lim_{t \rightarrow \infty} x(t) = 0$. Now, $\int_0^\infty \frac{d}{dt} [x'Px] dt = x(\infty)'Px(\infty) - x(0)'Px(0) = 0$. So,

$$0 = \int_0^\infty (\dot{x}'Px + x'P\dot{x}) dt = \int_0^\infty [x'(A'P + PA)x + x'PBu + u'B'Px] dt.$$

Use $y = Cx + Du$ and add the last expression to J

$$\begin{aligned} J &= \int_0^\infty [x'(A'P + PA + C'C)x + 2x'(PB + C'D)u + u'(D'D - \gamma^2 I)u] dt \\ &= \int_0^\infty \begin{bmatrix} x' & u' \end{bmatrix} \overbrace{\begin{bmatrix} A'P + PA + C'C & PB + C'D \\ B'P + D'C & D'D - \gamma^2 I \end{bmatrix}}^M \begin{bmatrix} x \\ u \end{bmatrix} dt. \end{aligned}$$

It follows that $J < 0$, and therefore $\|H\|_\infty < \gamma$, if $M \prec 0$. Finally, the $(1, 1)$ block of the inequality gives the inequality $A'P + PA \prec 0$. Since $P \succ 0$ this proves stability.

ii) Writing $M = M_1 + U'U$ where $U = \begin{bmatrix} C & D \end{bmatrix}$ and using a Schur complement, then

$$M \prec 0 \Leftrightarrow \begin{bmatrix} A'P + PA & PB & C' \\ B'P & -\gamma^2 I & D' \\ C & D & -I \end{bmatrix} \prec 0$$

b) i) Substituting $u = Lw_2 + Cx$, $e = w_2 + Cx$ into the state equation gives

$$\dot{x} = \underbrace{(A + LC)}_{A_c} x + \underbrace{\begin{bmatrix} B & L \end{bmatrix}}_{B_c} w, \quad e = \underbrace{C}_{C_c} x + \underbrace{\begin{bmatrix} 0 & I \end{bmatrix}}_{D_c} w.$$

It follows that $T_{ew}(s) = D_c + C_c(sI - A_c)^{-1}B_c$.

ii) Using the results of part (a), by replacing A, B, C and D by A_c, B_c, C_c and D_c , we have that there exists a feasible L if there exists $P = P' \succ 0$ such that

$$\begin{bmatrix} (A + LC)'P + P(A + LC) + C'C & PB & PL + C' \\ B'P & -\gamma^2 I & 0 \\ L'P + C & 0 & (1 - \gamma^2)I \end{bmatrix} \prec 0.$$

Noting that the only nonlinearity is due to the product PL , we define $Z = PL$ and so there exists a feasible L if there exists $P = P' \succ 0$ and Z such that

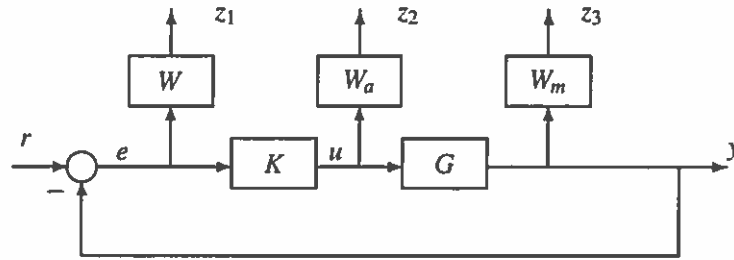
$$\begin{bmatrix} A'P + PA + ZC + C'Z' + C'C & PB & Z + C' \\ B'P & -\gamma^2 I & 0 \\ Z' + C & 0 & (1 - \gamma^2)I \end{bmatrix} \prec 0.$$

3. a) It is clear that we require $K(s)$ to be internally stabilising.

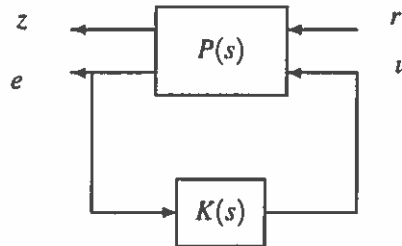
- i) A calculation shows that, when $\Delta_a = 0$ and $\Delta_m = 0$, $e(s) = -S(s)r(s)$ where $S(s) = [I + G(s)K(s)]^{-1}$ is the sensitivity. Thus $\|e(j\omega)\| \leq \|S(j\omega)\| \|r(j\omega)\|$. It follows that a sufficient condition to achieve the first design specification is $\|S(j\omega)\| < |w^{-1}(j\omega)|, \forall \omega$ or equivalently $\|WS\|_\infty < 1$, where $W = wI$.
- ii) Suppose that $\Delta_a = 0$ and let the input to Δ_m be ε_2 while the output from Δ_m be δ_2 . Then a calculation shows that $\varepsilon_2 = -GKS\delta_2$. Using the small gain theorem, to satisfy the first design requirement, it is sufficient that $\|\Delta_m(j\omega)G(j\omega)K(j\omega)S(j\omega)\| < 1, \forall \omega$. This can be satisfied if $\|W_mGKS\|_\infty < 1$, where $W_m = w_mI$.
- iii) An analogous procedure shows that to satisfy the second design requirement, it is sufficient that $\|\Delta_a(j\omega)K(j\omega)S(j\omega)\| < 1, \forall \omega$. This can be satisfied if $\|W_aKS\|_\infty < 1$, where $W_a = w_aI$.

To satisfy all design requirements, it is sufficient that $\left\| \begin{bmatrix} WS \\ W_aKS \\ W_mGKS \end{bmatrix} \right\|_\infty < 1$.

- b) Since $e = -Sr$, the design specifications reduce to the requirement that the transfer matrix from r to $z = [z_1^T \ z_2^T \ z_3^T]^T$ in the following diagram has \mathcal{H}_∞ -norm less than 1.



- c) The corresponding generalized regulator formulation is to find an internally stabilizing K such that $\|\mathcal{F}_l(P, K)\|_\infty < 1$:



$$P = \begin{bmatrix} P_{11} & P_{12} \\ P_{21} & P_{22} \end{bmatrix} = \left[\begin{array}{c|c} W & -WG \\ 0 & W_a \\ 0 & W_mG \\ \hline I & -G \end{array} \right]$$

4. a) The generalized regulator formulation is given by

$$\begin{bmatrix} z(s) \\ y(s) \end{bmatrix} = P(s) \begin{bmatrix} w(s) \\ u(s) \end{bmatrix}, u(s) = Fy(s), P(s) = \begin{bmatrix} P_{11}(s) & P_{12}(s) \\ P_{21}(s) & P_{22}(s) \end{bmatrix} \triangleq \begin{bmatrix} A & B_1 & B_2 \\ C_1 & 0 & 0 \\ 0 & 0 & I \\ I & 0 & 0 \end{bmatrix}.$$

- b) Let $V = x'Xx$ and set $u = Fx$. Provided that $X = X' > 0$ and $\dot{V} < 0$ along the closed-loop trajectory, we can assume $\lim_{t \rightarrow \infty} x(t) = 0$. Then

$$\dot{V} = x'X\dot{x} + x'\dot{X}x = x'(A'X + XA + F'B_2'X + XB_2F)x + x'XB_1w_1 + w_1'B_1'Xx.$$

Integrating from 0 to ∞ and using $x(\infty) = 0$,

$$-x_0'Xx_0 = \int_0^\infty [x'(A'X + XA + F'B_2'X + XB_2F)x + x'XB_1w_1 + w_1'B_1'Xx]dt.$$

Using the definition of J , adding the last equation, and completing two squares,

$$J = x_0'Xx_0 + \int_0^\infty \{x'[A'X + XA + C_1'C_1 - X(B_2B_2' - \gamma^{-2}B_1B_1')X]x + \|Zx\|^2 - \|v\|^2\}dt$$

where $Z = F + B_2'X$ and $v = \gamma w - \gamma^{-1}B_1'Xx$.

- c) Setting $x_0 = 0$, the requirement $\|H\|_\infty \leq \gamma$ is equivalent to $J \leq 0$. Thus sufficient conditions for $J \leq 0$ are $Z = 0$ and the existence of X such that

$$A'X + XA + C_1'C_1 - X(B_2B_2' - \gamma^{-2}B_1B_1')X = 0, \quad X = X' > 0.$$

Setting $Z = 0$, the control policy is $F = -B'X$. Setting $v = 0$, the worst case disturbance is $w^* = \gamma^{-2}B_1'X$. The closed-loop with $u = Fx$ and $w = w^*$ is $\dot{x} = A_c x$ where $A_c = A - (B_2B_2' - \gamma^{-2}B_1B_1')$ and a further condition is $\text{Re } \lambda_i(A_c) < 0, \forall i$.

- d) Using the expression for J in Part b) and the solution in Part c), it follows that $J \leq x_0'Xx_0$, and so we can take $\gamma_2^2 = x_0'Xx_0$ as the tightest upper bound on the regulation cost.
- e) Since $X(\gamma)$ is decreasing in γ and since $x_0'X(\gamma)x_0$ is an upper bound on the regulation cost, then we can use the value of γ as a trade-off in the design between maximum robustness (the minimum value of γ achieving the sufficient conditions in Part c)) and maximum regulation (the maximum value of γ , i.e., $\gamma \rightarrow \infty$).