EA3036C

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General Description

The EA3036C is a 3CH power management IC for applications powered by one Li-lon battery or a DC 5V adapter. It integrates three synchronous buck converters and can provide high efficiency output at light load and heavy load operation. The internal compensation architecture simplifies the application circuit design. Besides, the independent enable control makes the designer have the greatest flexibility to optimize timing for power sequencing purposes. The EA3036C is available in a 20 pin QFN 3x3 package.

Features

2.7V to 5.5V Input Voltage Range

Three Buck Converters

Output Voltage Range: 0.6V to Vin

Maximum Continuous Loading: 1A (3CH total output power consumption must be less

than 6W)

Fixed 1.5MHz Switching Frequency 100% Duty Cycle Low Dropout

Operation <1uA Shutdown Current

Independent Enable Control

Internal Compensation

Cycle-by-Cycle Current Limit

Short Circuit Protection

Auto Recovery OTP Protection

Available in 20-pin 3mm x 3mm QFN Package

<u>Applications</u>

Smart Phone

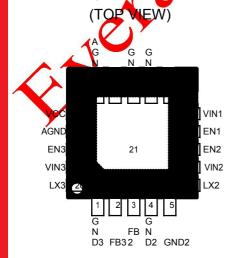
- IP Camera
- Digital Camera







Pin Configuration



QFN 3x3-20



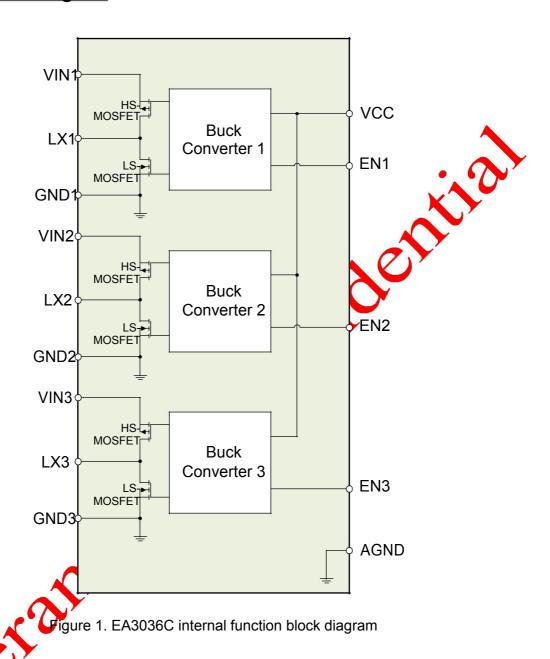
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Pin Description

Function Description	Pin No.
Power ground pin of CH3.	1
Feedback input of CH3. Connect to output voltage with a resistor divider.	2
Feedback input of CH2. Connect to output voltage with a resistor divider.	3
Power ground pin of CH2.	4, 5
Internal MOSFET switching output of CH2. Connect LX2 pin with a low pass filter circuit to obtain a stable DC output voltage.	76
Power input pin of CH2. Recommended to use a 10uF MLCC capacitor between VIN2 pin and GND2 pin.	7
CH2 turns on/turns off control input. Don't leave this pin floating.	8
CH1 turns on/turns off control input. Don't leave this pin floating	9
Power input pin of CH1. Recommended to use a 100F MLCC capacitor between VIN1 pin and GND1 pin.	10
Internal MOSFET switching output of CH1. Connect LX1 pin with a low pass filter circuit to obtain a stable DC output voltage.	11
Power ground pin of CH1.	12, 13
Feedback input of CH1. Connect to output voltage with a resistor divider.	14
Analog ground pin.	15, 17
Input supply pin for internal control circuit.	16
CH3 turns on/turns off control input. Don't leave this pin floating.	18
Power input pin of CH3. Recommended to use a 10uF MLCC capacitor between VIN3 pin and CND3 pin.	19
Internal MOSEET switching output of CH3. Connect SW3 pin with a low pass filter discurbto obtain a stable DC output voltage.	20
The Exposed Pad must be soldered to a large PCB copper plane and connected to GND for appropriate dissipation.	21
	Power ground pin of CH3. Feedback input of CH3. Connect to output voltage with a resistor divider. Feedback input of CH2. Connect to output voltage with a resistor divider. Power ground pin of CH2. Internal MOSFET switching output of CH2. Connect LX2 pin with a low pass filter circuit to obtain a stable DC output voltage. Power input pin of CH2. Recommended to use a 10uF MLCC capacil of between VIN2 pin and GND2 pin. CH2 turns on/turns off control input. Don't leave this pin floating. CH1 turns on/turns off control input. Don't leave this pin floating. CH1 turns on/turns off control input. Don't leave this pin floating. Power input pin of CH1. Recommended to use a 10uF MLCC capacitor between VIN1 pin and GND1 pin. Internal MOSFET switching output of CH1. Connect LX1 pin with a low pass filter circuit to obtain a stable DC output voltage. Power ground pin of CH1. Feedback input of CH1. Connect to output voltage with a resistor divider. Analog ground pin. Input supply pin for internal control circuit. CH3 turns on/turns off control input. Don't leave this pin floating. Power input pin of CH3. Recommended to use a 10uF MLCC capacitor between VIN3 pin and GND3 pin. Internal MOSFET switching output of CH3. Connect SW3 pin with a low pass filter circuit to obtain a stable DC output voltage. The Exposed Pad must be soldered to a large PCB copper plane and



Function Block Diagram



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Absolute Maximum Ratings

Parameter	Value
Input Voltage (V _{VIN1} , V _{VIN2} , V _{VIN3} , V _{VCC})	-0.3V to +6.5V
SW Pin Voltage (V _{LX1} , V _{LX2} , V _{LX3})	-0.3V to V _{VINX} +0.3V
All Other Pins Voltage	-0.3V to +6.5V
Ambient Temperature operating Range (T _A)	-40°C/to +85°C
Maximum Junction Temperature (T _{Jmax})	+150°C
Lead Temperature (Soldering, 10 sec)	2 60°C
Storage Temperature Range (T _S)	-65°C to +150°C

Note (1): Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device.

Exposure to "Absolute Maximum Ratings" conditions for extended periods may affect device reliability and lifetime.

Package Thermal Characteristics

Parameter	Value
QFN 3x3-20 Thermal Resistance (θ _{JC})	7.5°C/W
QFN 3x3-20 Thermal Resistance (θJA)	67°C/W
QFN 3x3-20 Power Dissipation at T _A =25°C (P _D	0max) 1.87W

Note (1): PDmax is calculated according to the formula: PDMAX=(TJMAX-TA)/ θJA.

Recommended Operating Conditions

Parameter	Value	
Input Voltage (V _{VIN1} V _{VIN2} V _{VIN3} , V _{VCC})	+2.7V to +5.5V	
Junction Temperature Range (T _J)	-40°C to +125°C	





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Electrical Characteristics

V_{VINX}=3.6V, V_{VCC}=3.6V, T_A=25°C, unless otherwise noted

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Supply Voltage						
Input Voltage	Vinx		2.7		5.5	V
Control Circuit Input Voltage	Vvcc		2.7		5.5	V
Buck Converter 1, 2, 3					• 6	77
Shutdown Supply Current	Isp	V _{EN} = 0V		0.1	X1	uA
Quiescent Current	IQ	Non-switching, No Load		40	80	uA
UVLO Threshold	Vuvlo	V _{VIN} Rising	1.7	1.9	2.1	V
UVLO Hysteresis	Vuv-HYST		٠, (Ø.1		V
Output Load Current	load				1	Α
Reference Voltage	V_{REF}		0.588	0.6	0.612	V
Switching Frequency	Fsw	I _{LOAD} = 100mA	1	1.5	2	MHz
Short Frequency	Fsw-short	Vour = 0V		300		KHz
PMOS Current Limit	ILIM-P		1.5	2		Α
PMOS On-Resistance	RDS(ON)-P	LOAD = 100mA		120		mΩ
NMOS On-Resistance	Rds(on)-n	100mA		110		mΩ
Enable Pin Input Low Voltage	V _{EN-1}				0.4	V
Enable Pin Input High Voltage	Ven-H		2			V
Maximum Duty Cycle	Dмах		100			%
Thermal Shutdown						
Thermal Shutdown Threshold	Тотр			165		°C
Thermal Shutdown Hysteresis	Тнүзт			30		°C

Note MOSFET on-resistance specifications are guaranteed by correlation to wafer level measurements.

^{(2):} Thermal shutdown specifications are guaranteed by correlation to the design and characteristics analysis.

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Application Circuit Diagram

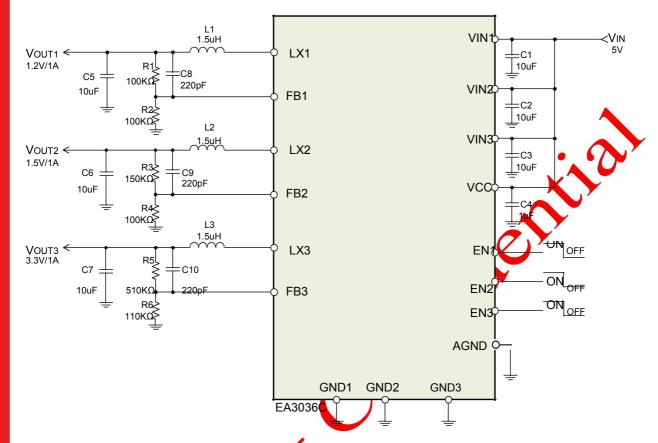


Figure 2. Typical application circuit diagram

Ordering Information

Part Number	Package Type	Packing Information
EA3036CQBR	QFN 3mm x 3mm-20	Tape & Reel / 3000
Note (1): "QB": Package type cod	Y	

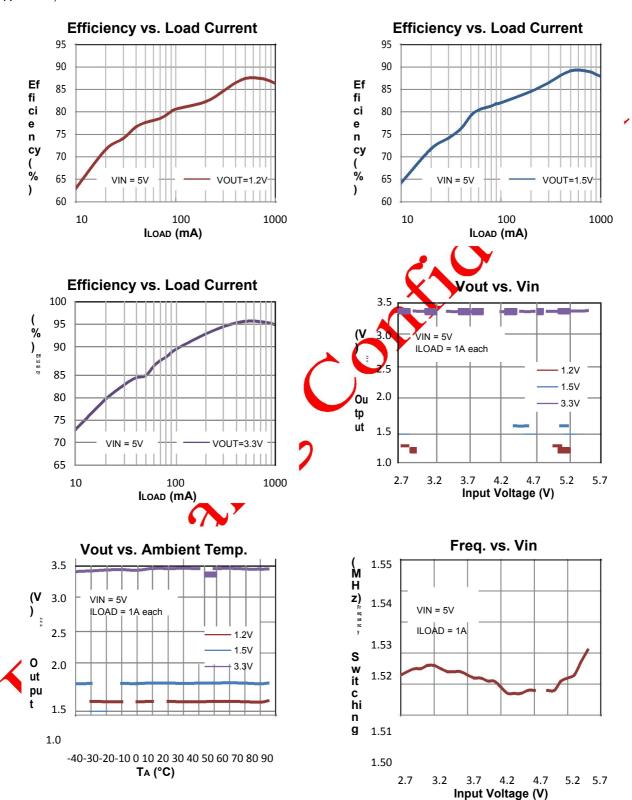




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Typical Operating Characteristics

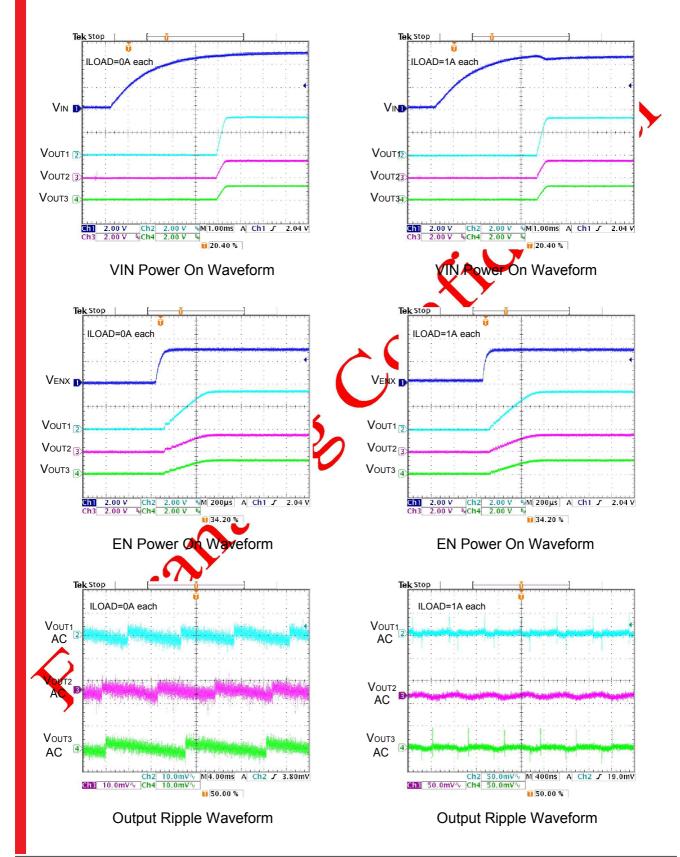
 V_{IN} =5V, V_{VCC} =5V, V_{OUT1} =3.3V, V_{OUT2} =1.5V, V_{OUT3} =1.2V, L1=1.5uH, L2=1.5uH, L3=1.5uH, T_A =25°C, unless otherwise noted



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Typical Operating Characteristics

 V_{IN} =5V, V_{VCC} =5V, V_{OUT1} =3.3V, V_{OUT2} =1.5V, V_{OUT3} =1.2V, L1=1.5uH, L2=1.5uH, L3=1.5uH, T_A =25°C, unless otherwise noted



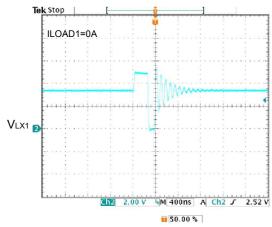
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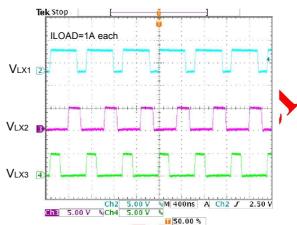
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Typical Operating Characteristics

 V_{IN} =5V, V_{VCC} =5V, V_{OUT1} =3.3V, V_{OUT2} =1.5V, V_{OUT3} =1.2V, L1=1.5uH, L2=1.5uH, L3=1.5uH, T_A =25°C, unless otherwise noted



V_{LX1} Switching Waveform



V_{LX1,2,3} Switching Waveform

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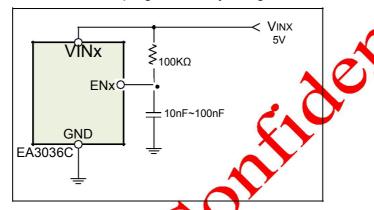
Functional Description

PFM/PWM Operation

Each of the buck regulators can be operated at PFM/PWM mode. If the output current is less than 150mA (typ.), the regulators automatically enters the PFM mode. The output voltages and output ripples at PFM mode are higher than the output voltages and output ripples at PWM mode. But at very light load, the PFM mode operation provides higher efficiency than PWM mode operation.

Enable Control

The EA3036C is a high efficiency Power Management IC which is designed for IPC applications. It incorporates three 1A synchronous buck regulators and can be controlled by individual EN pms. The start-up time for each channel can be programmed by using the circuit shown as below:



180° Phases Shifted Architecture

In order to reduce the input ripple current, the EA3036C applied 180° phases shifted architecture. Buck1 and Buck3 have the same phase and Buck2 is 180° out of phase. This architecture allows the system board has less ripple current, and those can reduce EMI.

Over Current Protection

The EA3036C internal three regulators have their own cycle-by-cycle current limit circuits. When the inductor peak current exceeds the current limit threshold, the output voltage starts to drop until FB pin voltage is below the threshold, typically 30% below the reference. Once the threshold is triggered, the switching frequency is reduced to 300KHz (typ.).

Thermal Shutdown

The EA3036C will automatically disabled if the die temperature is higher than the thermal shutdown threshold point. To avoid unstable operation, the hysteresis of thermal shutdown is about 30°C.



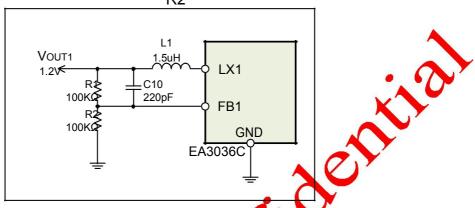
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Application Information

Output Voltage Setting

Each of the regulators output voltage can be set via a resistor divider (ex. R1, R2). The output voltage is calculated by following equation:

$$V_{OUT1} = 0.6 \times \frac{R1}{R2} + 0.6 \text{ V}$$



The following table lists common output voltage and the corresponding R1, R2 resistance value for reference.

Output Voltage	R1 Resistance	R2 Resistance	Tolerance
3.3V	510ΚΩ	110ΚΩ	1%
1.8V	200ΚΩ	100ΚΩ	1%
1.5V	150ΚΩ	100ΚΩ	1%
1.2V	100ΚΩ	100ΚΩ	1%

Input / Output Capacitors Selection

The input capacitors are used to suppress the noise amplitude of the input voltage and provide a stable and clean DC input to the device. Because the ceramic capacitor has low ESR characteristic, so it is suitable for input capacitor use. It is recommended to use X5R or X7R MLCC capacitors in order to have better temperature performance and smaller capacitance tolerance. In order to suppress the output voltage ripple, the MLCC capacitor is also the best choice. The suggested part numbers of input voltage ripple are as follows:

	Vendor	Part Number	Capacitance	Edc	Parameter	Size
	TDK	C2012X5R1A106M	10uF	10V	X5R	0805
1	TDK	C3216X5R1A106M	10uF	10V	X5R	1206
	TDK	C2012X5R1A226M	22uF	10V	X5R	0805
	TDK	C3216X5R1A226M	22uF	10V	X5R	1206

Output Inductor Selection

The output inductor selection mainly depends on the amount of ripple current through the inductor ΔI_L . Large ΔI_L will cause larger output voltage ripple and loss, but the user can use a smaller inductor to save cost and space. On the contrary, the larger inductance can get smaller ΔI_L and

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thus the smaller output voltage ripple and loss. But it will increase the space and the cost. The inductor value can be calculated as:

$$L = \frac{V_{\text{PWR}} - V_{\text{OUT}}}{I_{\text{L}} \times F_{\text{SW}}} \times \frac{V_{\text{OUT}}}{V_{\text{PWR}}}$$

For most applications, 1.0uH to 2.2uH inductors are suitable for EA3036C.

Power Dissipation

The total output power dissipation of EA3036C should not to exceed the maximum 6W range. The total output power dissipation can be calculated as:

$$P_D$$
 (total) = $V_{OUT1} \times I_{OUT1} + V_{OUT2} \times I_{OUT2} + V_{OUT3} \times I_{OUT3}$

PCB Layout Recommendations

Layout is very critical for PMIC designs. For EA3036C PCB layout considerations, please refer to the following suggestions to get best performance.

It is suggested to use 4-layer PCB layout and place LX plane and output plane on the top layer, place VIN plane in the inner layer.

The top layer SMD input and output capacitors ground plane should be connected to the internal ground layer and bottom ground plane individually by using vias

The AGND should be connected to inner ground layer directly by using via.

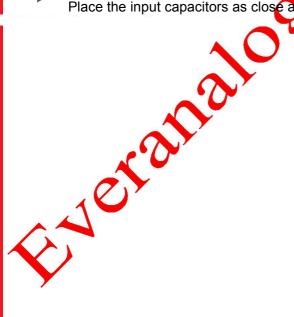
High current path traces need to be widened.

Place the input capacitors as close as possible to the VIXx pin to reduce noise interference.

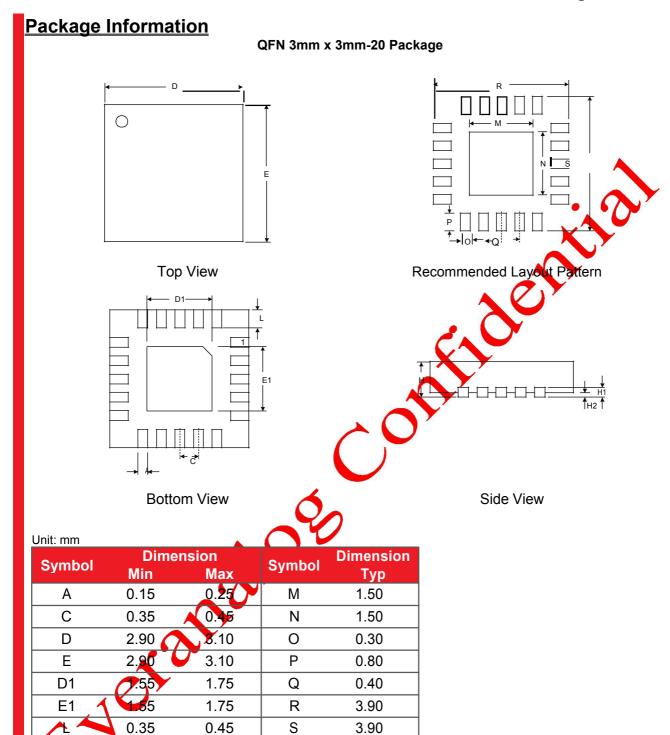
Keep the feedback path (from Voutx to FBx) away from the noise node (ex. LXx). LXx is a high current noise node. Complete the layout by using short and wide traces.

The top layer exposed pad ground plane should be connected to the internal ground layer and bottom ground plane by using a number of vias to improve thermal performance.

Place the input capacitors as close as possible to the VINx pin to reduce noise interference.



3CH Power Management IC



H2

0.70

0.18

0.00

0.80

0.25

0.05