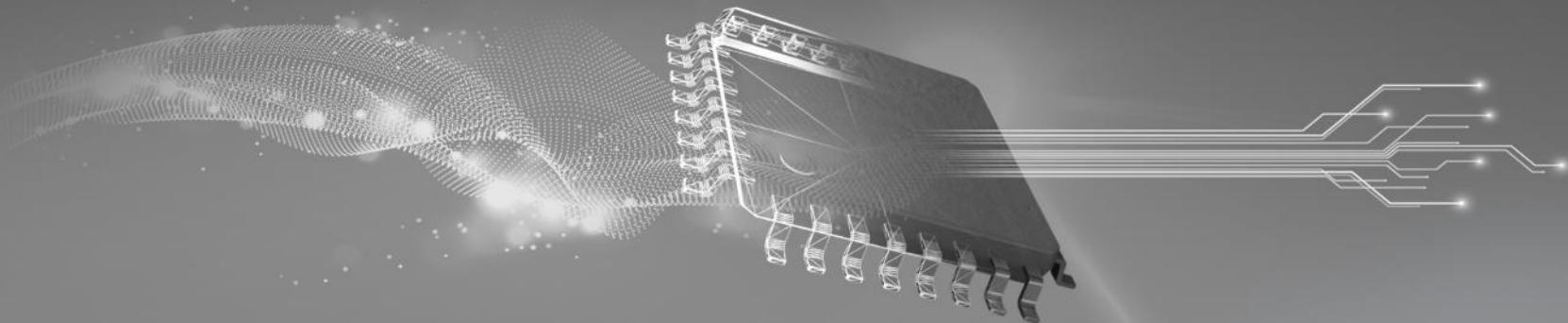


# TI TECH DAYS



## Tips and Tricks for Addressing EMI Issues in Power Supplies

Pradeep Shenoy, Brian King, Bob Sheehan

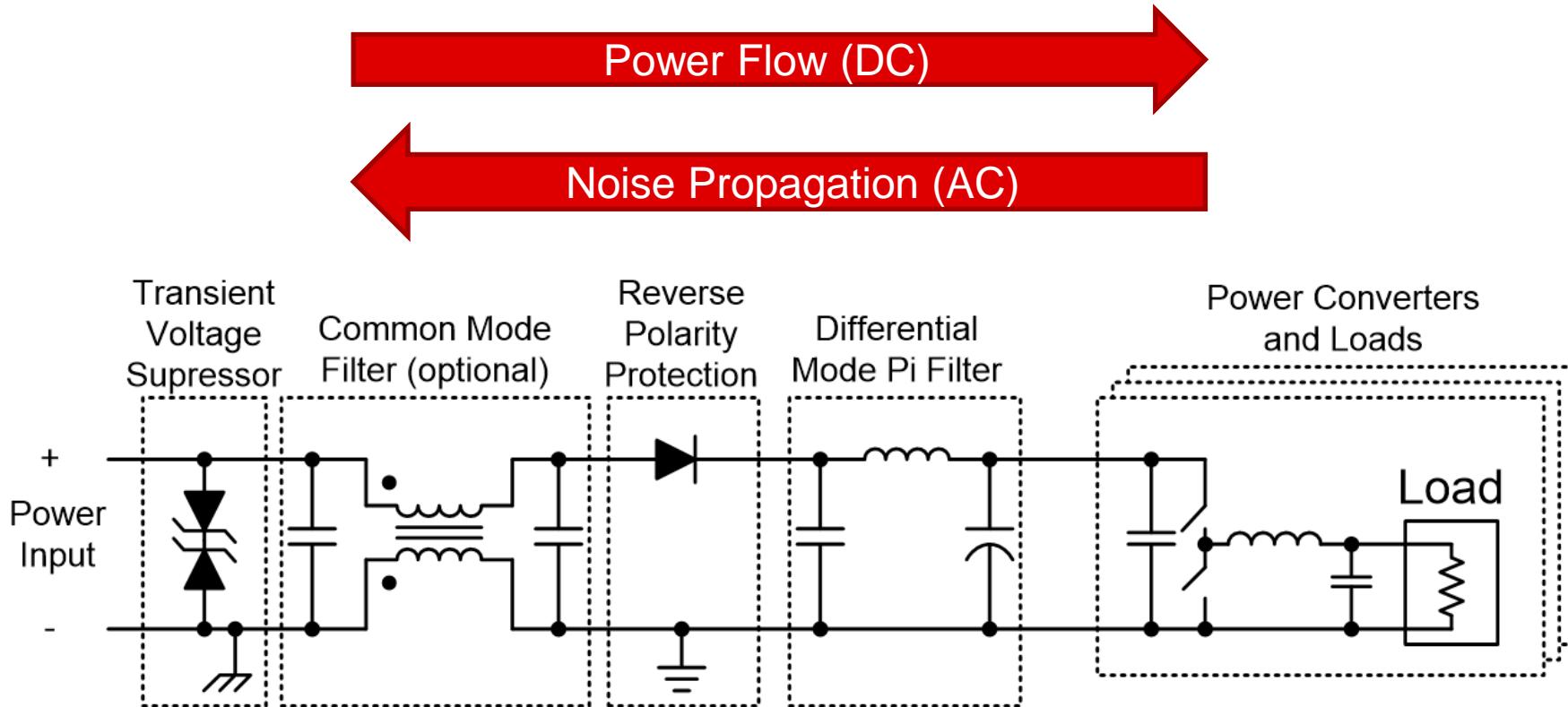
Power Design Services (PDS)

# Agenda

This presentation will cover the unique challenges of designing power converters to pass EMC requirements.

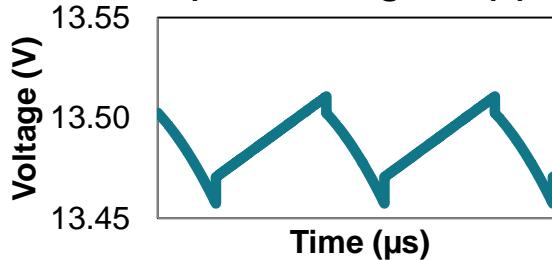
- **Introduction** to EMI: sources, standards, filters, test setup
- Techniques for **debugging** EMI problems
  - Distinguishing between differential mode and common mode noise
- **Design example** showing path to passing CISPR 25

# Overview diagram (automotive)

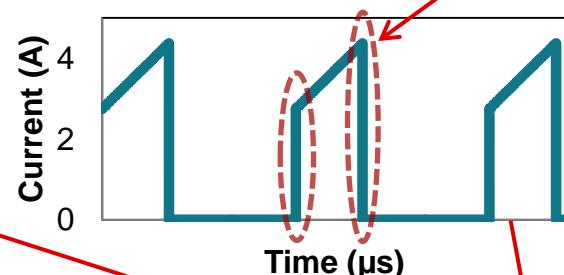


# Sources of emissions from within power supplies

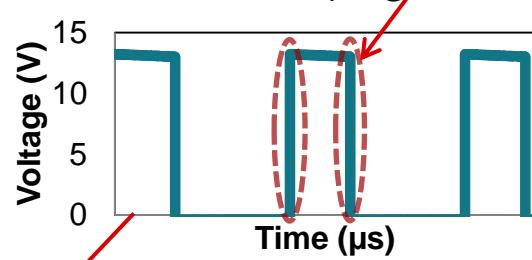
Input Voltage Ripple



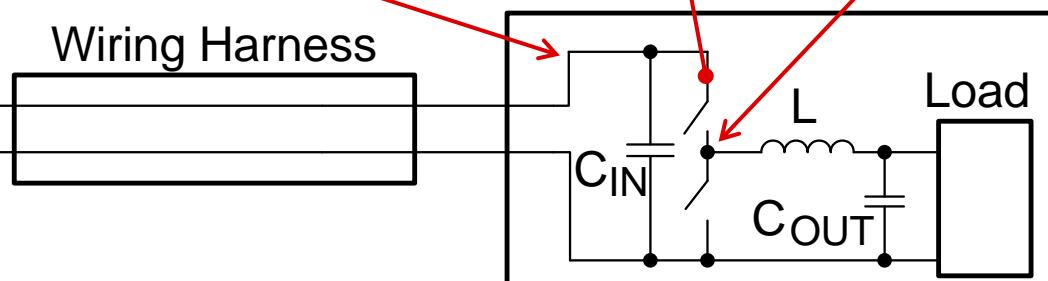
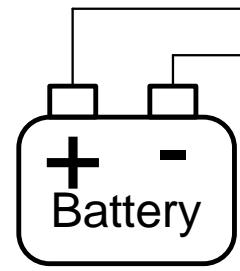
Switch Current (High  $di/dt$ )



Switch Node (High  $dv/dt$ )



Wiring Harness

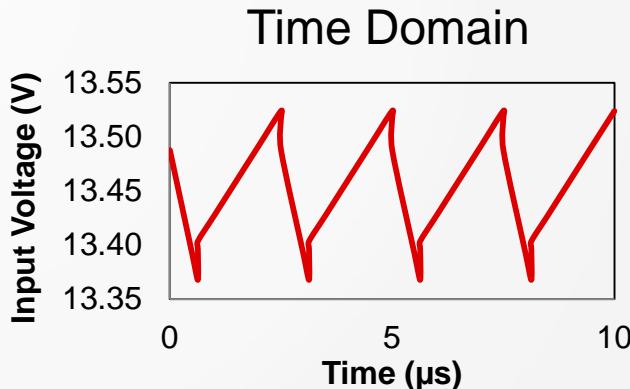


Electronic Control Unit (ECU)

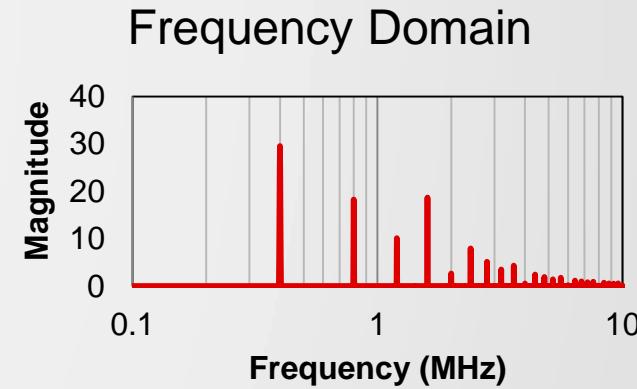
\*EMI filtering and protection circuits not shown

Chassis Ground

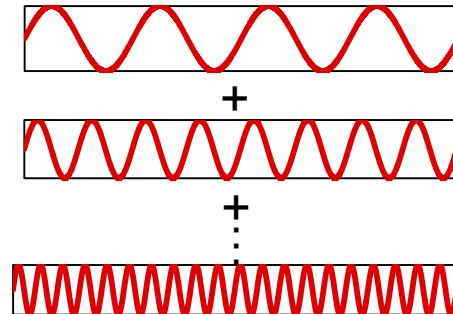
# Frequency spectrum of noise (ripple)



Fourier  
Transform



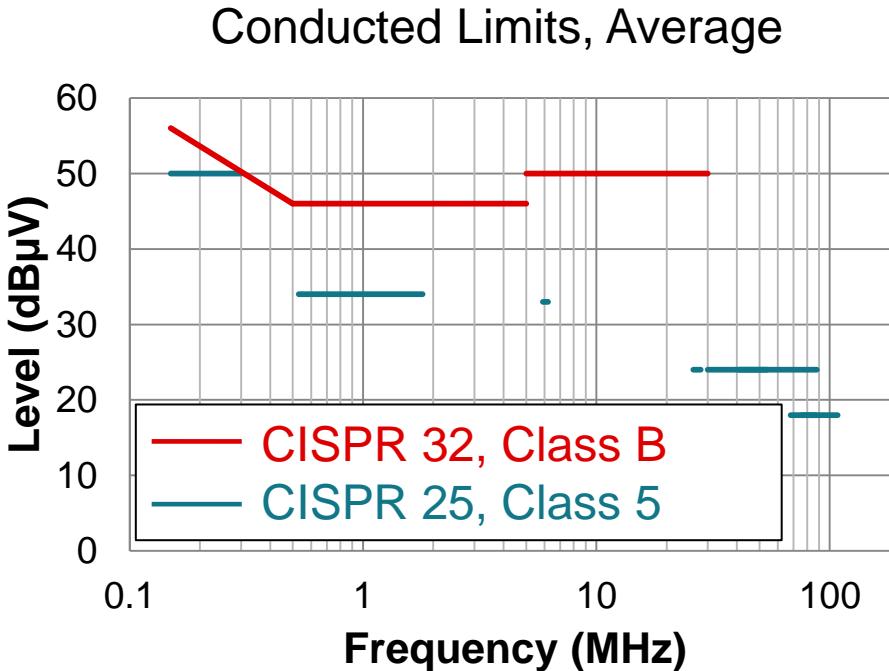
A periodic signal can be expressed as the sum of sinusoidal waveforms



Fourier series

$$f(t) = \sum_{n=0}^{\infty} c_n \cos(n\omega t + \theta_n)$$

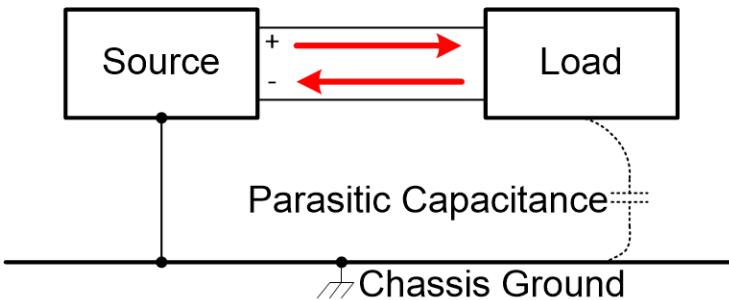
# Conducted EMI limits



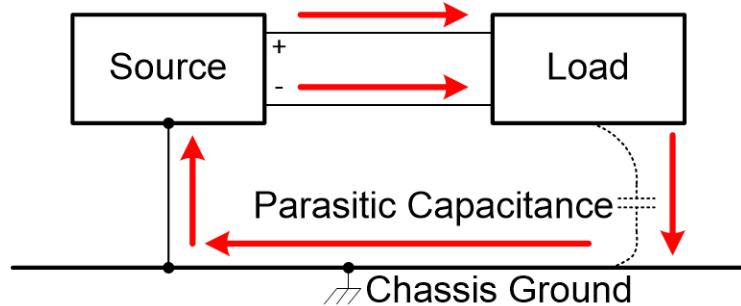
- CISPR 32 (formerly 22) is an EMI standard for **IT/industrial** equipment
  - Continuous line across the frequency spectrum up to 30 MHz
  - Used often for off ac-line applications
- CISPR 25 is a common **automotive** EMI standard
  - Go higher in frequency (108 MHz)
  - Have lower limit levels
  - But have gaps in between frequencies

# Differential mode vs. common mode noise

## Differential Mode



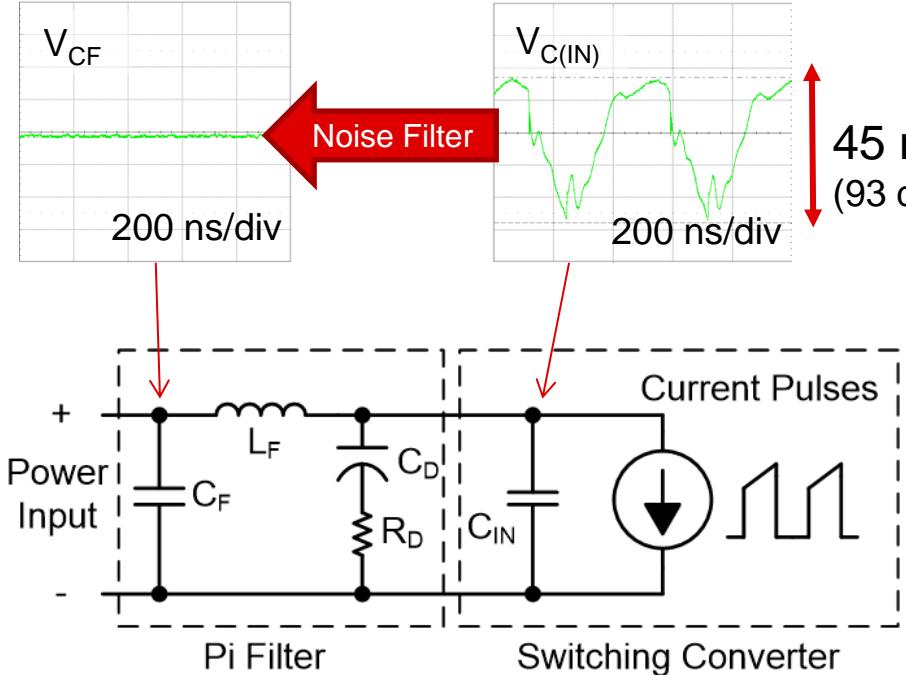
## Common Mode



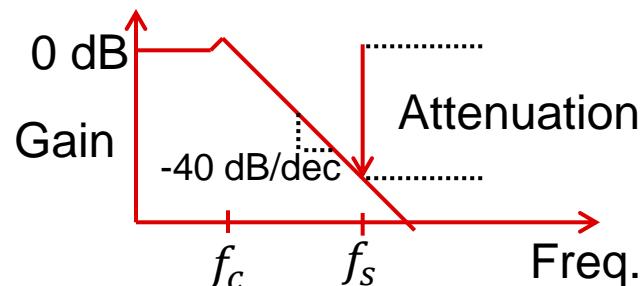
- Signal (noise) flows out on one wire (+) and returns on the other wire (-)
- Increases with load current
- Attenuated with a pi filter

- Noise flows out in the same direction on both wires and returns via stray capacitance and chassis ground
- Mostly independent of load current
- Filtered with a common mode choke

# Differential mode EMI filter

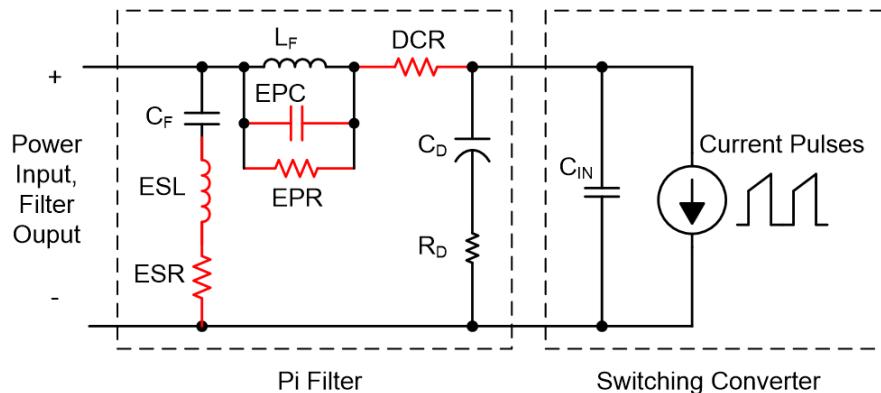


- Buck converters pull pulses of current causing  $C_{IN}$  voltage ripple
- A pi ( $\pi$ ) filter reduces conducted differential mode noise
  - $L_F$  and  $C_F$  form a low pass filter
  - $C_D$  and  $R_D$  included for stability (reduce EMI filter output impedance peak)
- Select  $L_F$  and  $C_F$  to get desired attenuation at  $f_s$

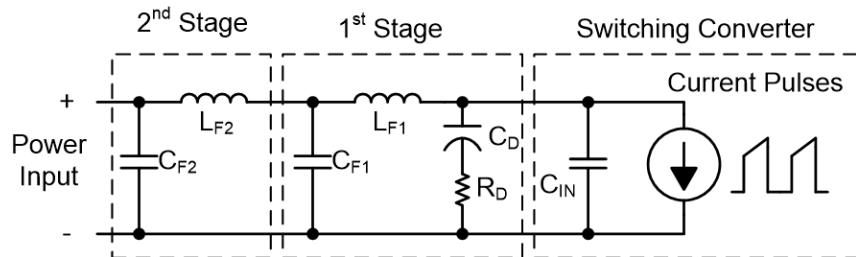


# Impact of parasitics

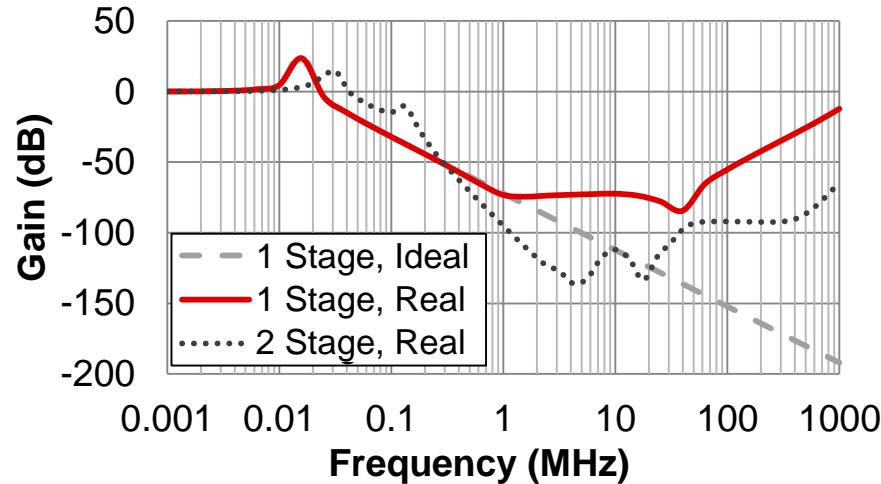
## Parasitic Elements in a Pi Filter



## Two Stage Pi Filter



- High frequency performance of pi filters degrade due to parasitic elements

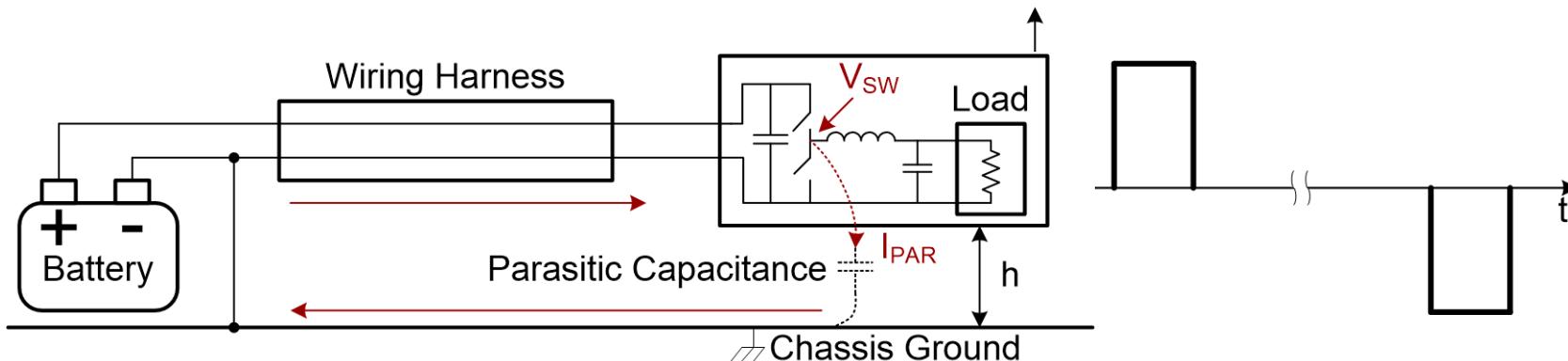
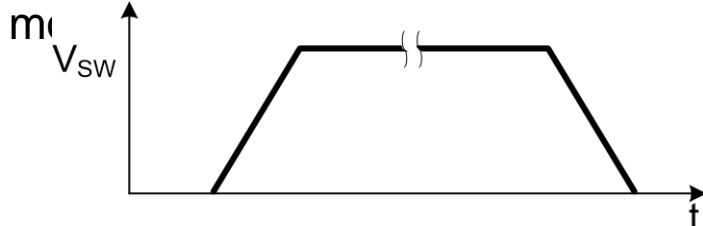


- A two stage pi filter can be used
  - A ferrite bead used in 2<sup>nd</sup> stage ( $L_{F2}$ )

# Source of common mode noise

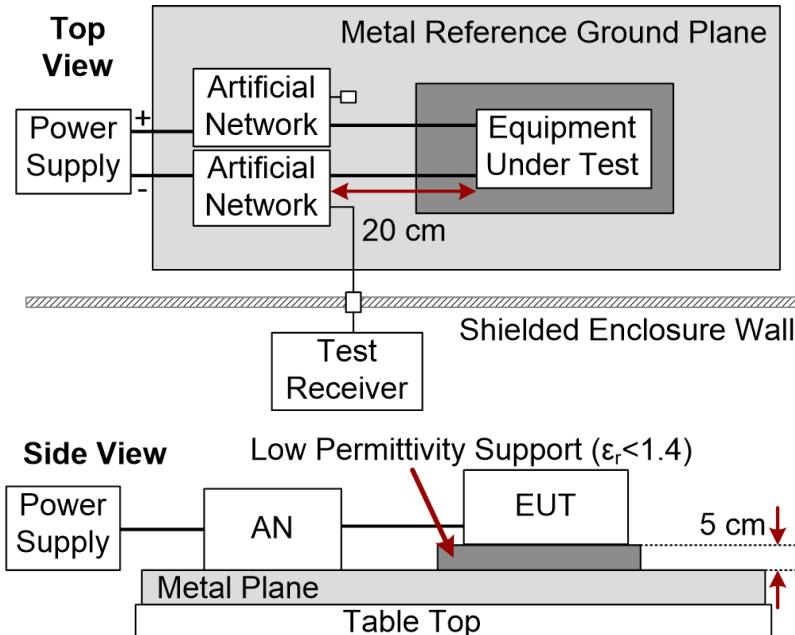
- The primary source of common mode noise is **parasitic capacitive coupling**
  - 5 cm is the standard testing height ( $h$ ) above chassis ground set by CISPR 25
  - Higher/lower would alter parasitic capacitance

- The switch node ( $V_{SW}$ ) pumps common mode noise during the switching cycle

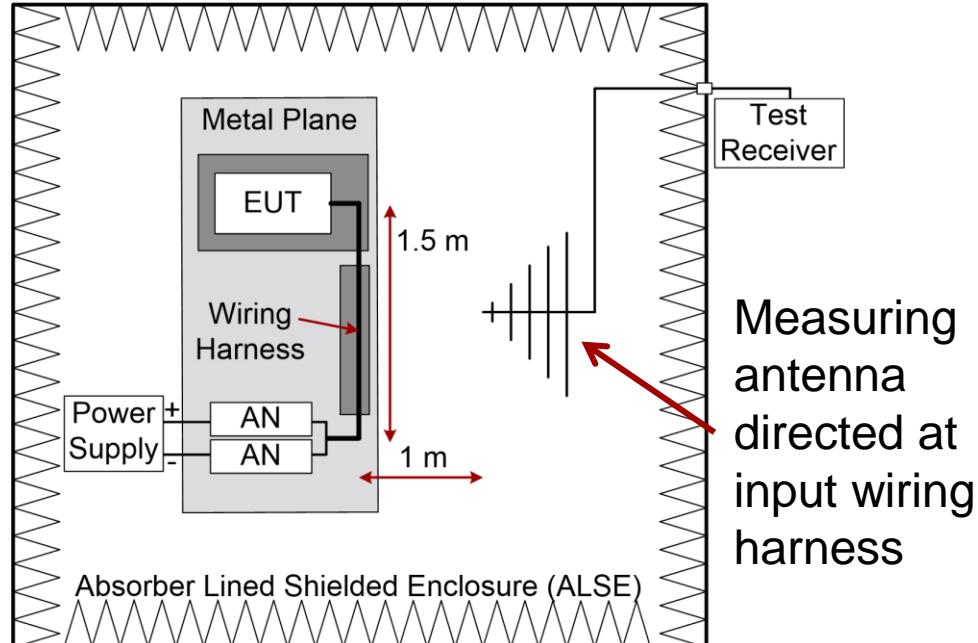


# CISPR 25 emissions test setup

Conducted Emissions Setup



Radiated Emissions Setup

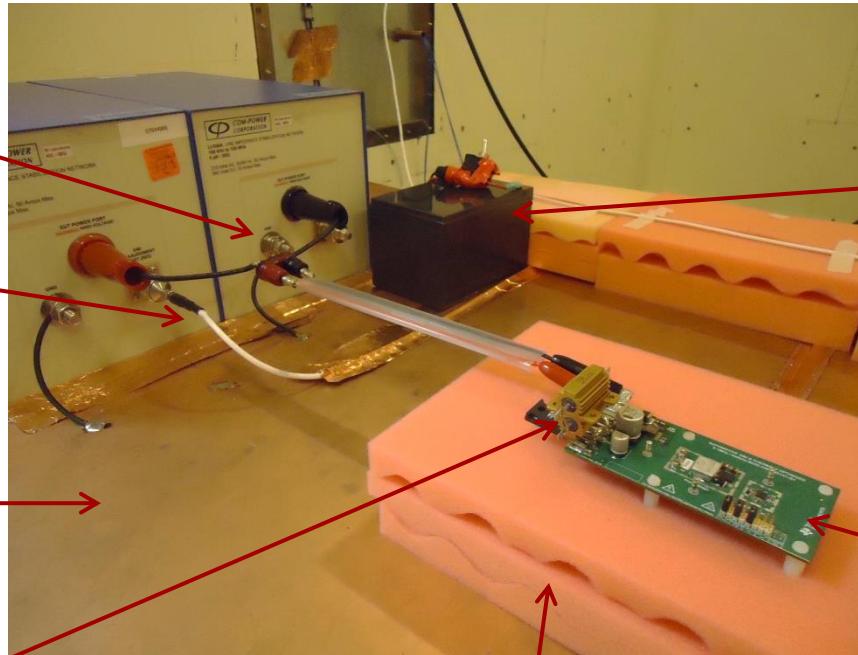


Address conducted emissions before testing radiated emissions

# Conducted EMI test setup (CISPR 25)

2 artificial networks

- One terminated with  $50 \Omega$
- One to receiver



Copper top table tied to back wall (grounded)

Load resistors

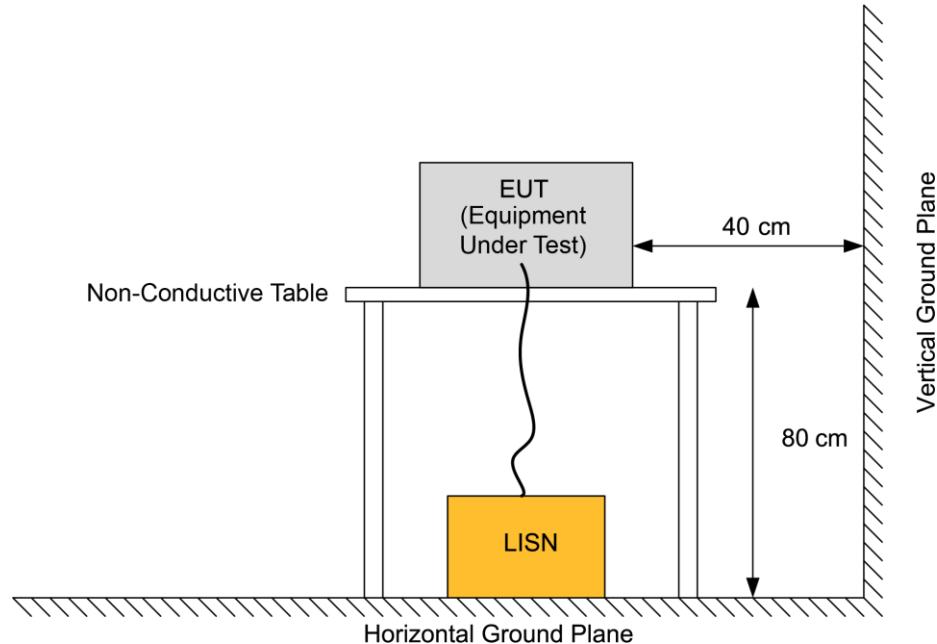
5 cm low permittivity support

Low noise lab source or car battery, connected to AN input

Equipment under test, 20 - 40 cm from ANs

# CISPR 32/22 conducted EMI test setup

- Equipment under test (EUT) placed on non-conductive table
- Horizontal & vertical ground planes
  - Or screened room
- EUT powered through line impedance stabilization network (LISN)
- Measure high-frequency (HF) emissions from LISN

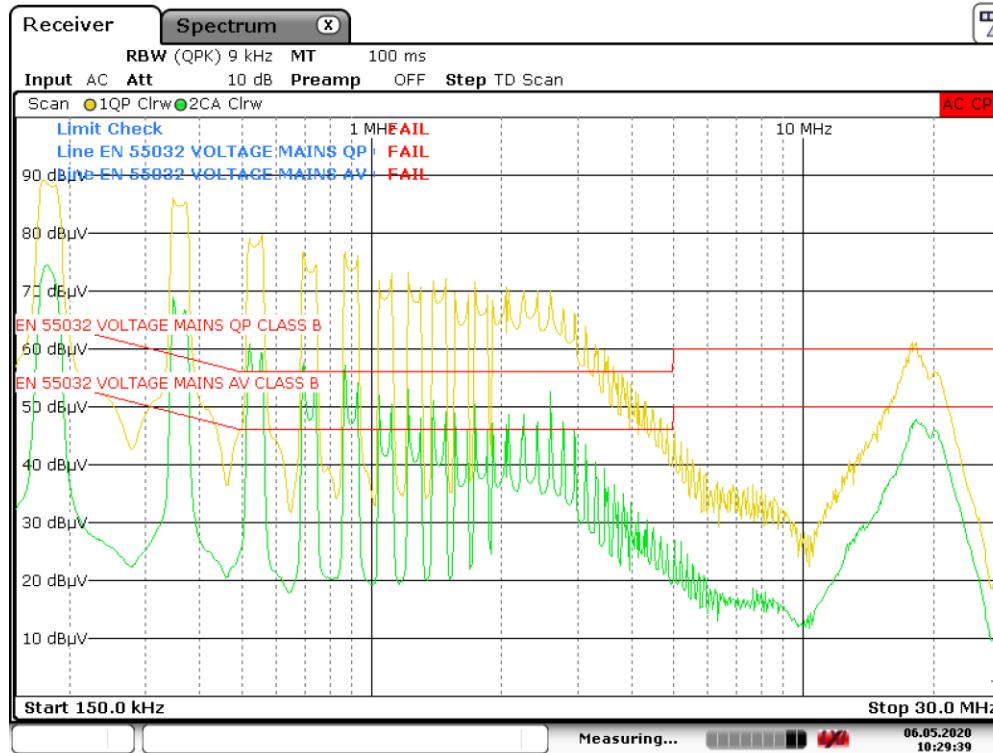


[1] EN55022, 2010, "Information technology equipment – Radio disturbance characteristics – Limits and methods of measurement"

# TECHNIQUES FOR DEBUGGING EMI PROBLEMS

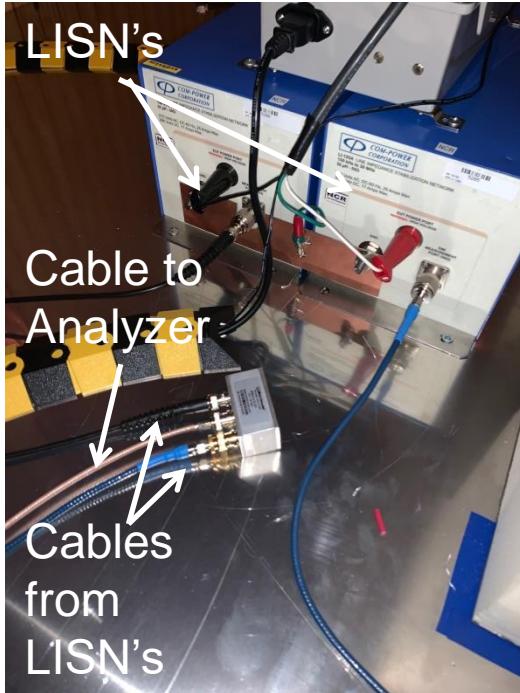
Brian King

# DM or CM, Where to Begin?



Date: 6.MAY.2020 10:29:39

# Powerful EMI Weapons - Splitting CM and DM



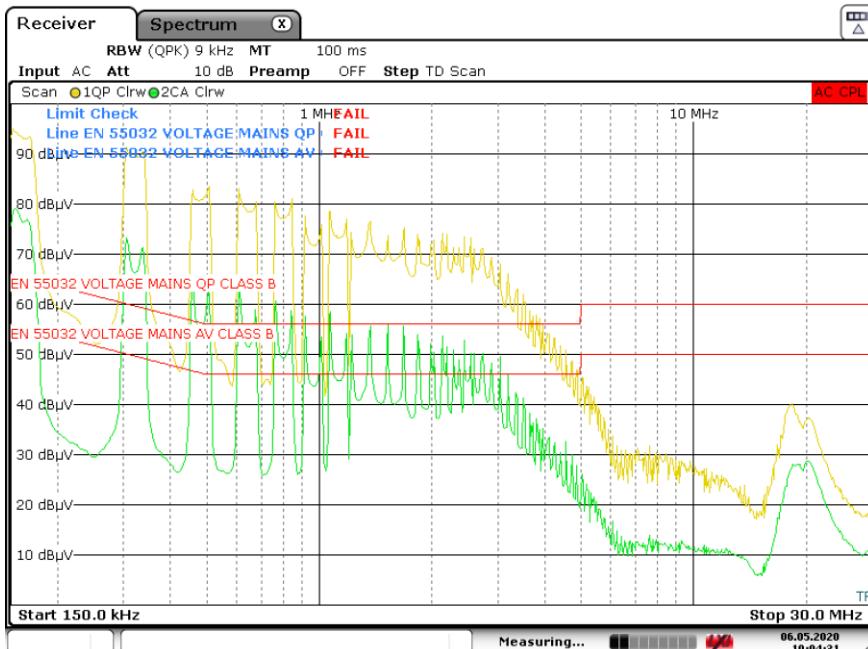
~\$60 on [www.minicircuits.com](http://www.minicircuits.com)  
ZSC-2-2+ for DM mode  
ZSCJ-2-2+ for CM mode

**Bonus Tip:** Can also tie PGND-SGND together to minimize CM signal

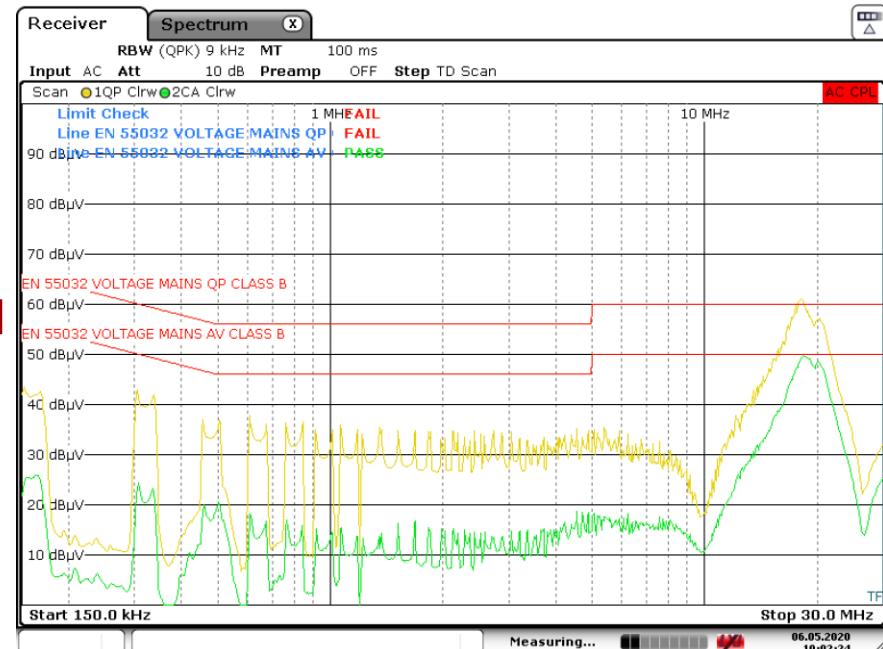
"Separation of common and differential mode conducted emission: Power combiner/splitters"  
Andersen, Michael A. E.; Nielsen, Dennis; Thomsen, Ole Cornelius; Andersen, Michael A. E.  
2012 Proceedings of the International Conference on Renewable Energies and Power Quality

# Divide and Conquer Your EMI Enemies

## Differential Mode

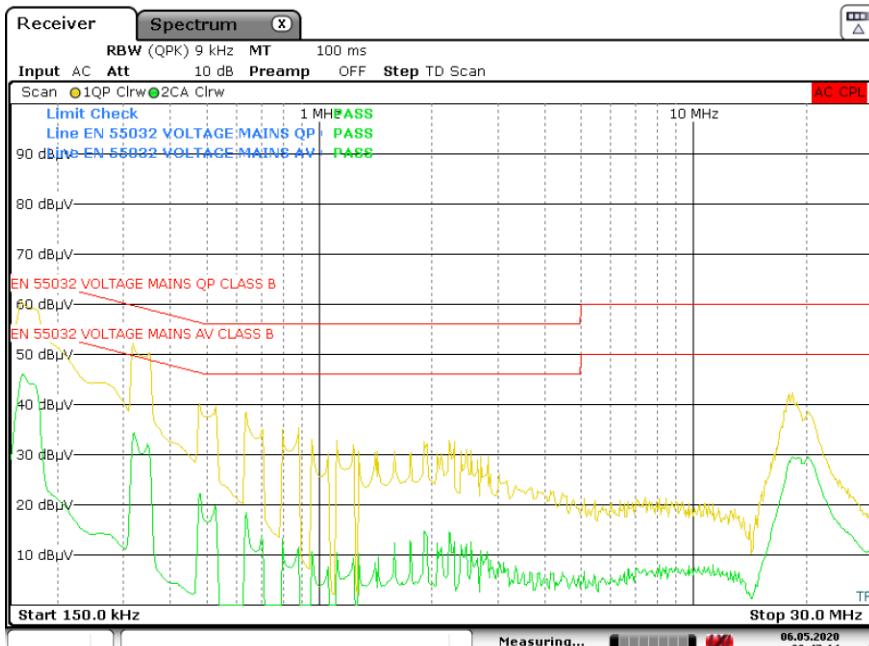


## Common Mode

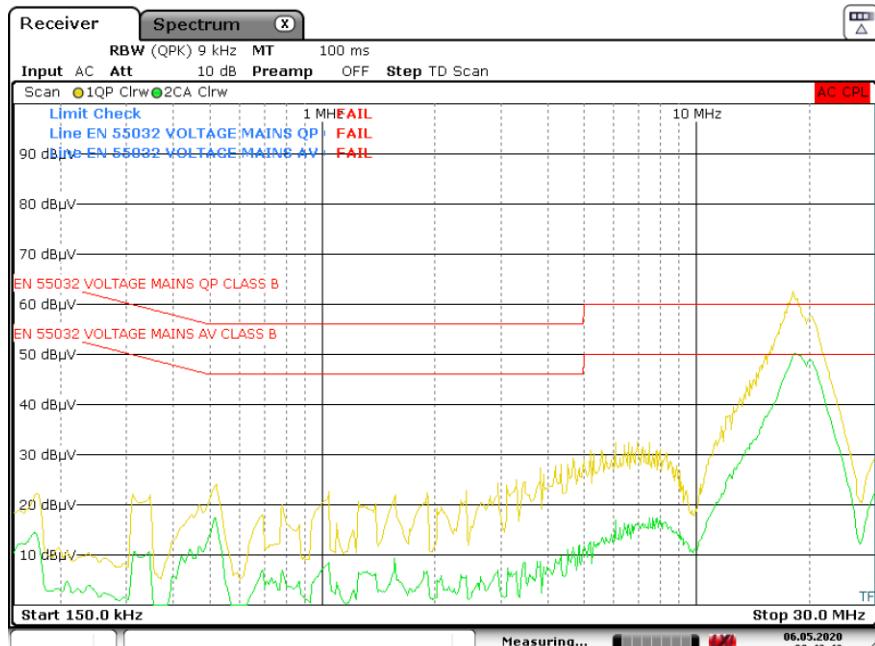


# DM Battle Won!

## Differential Mode

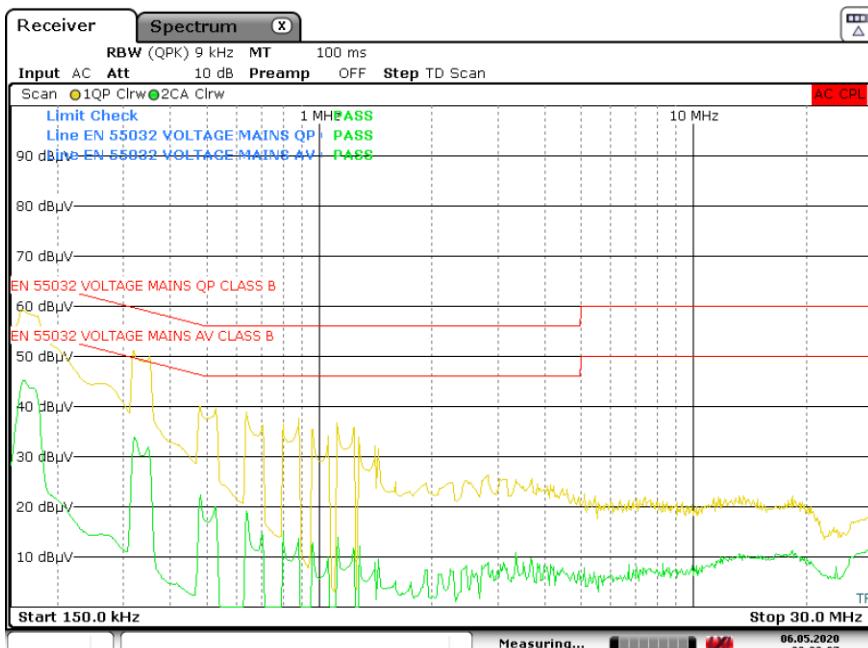


## Common Mode

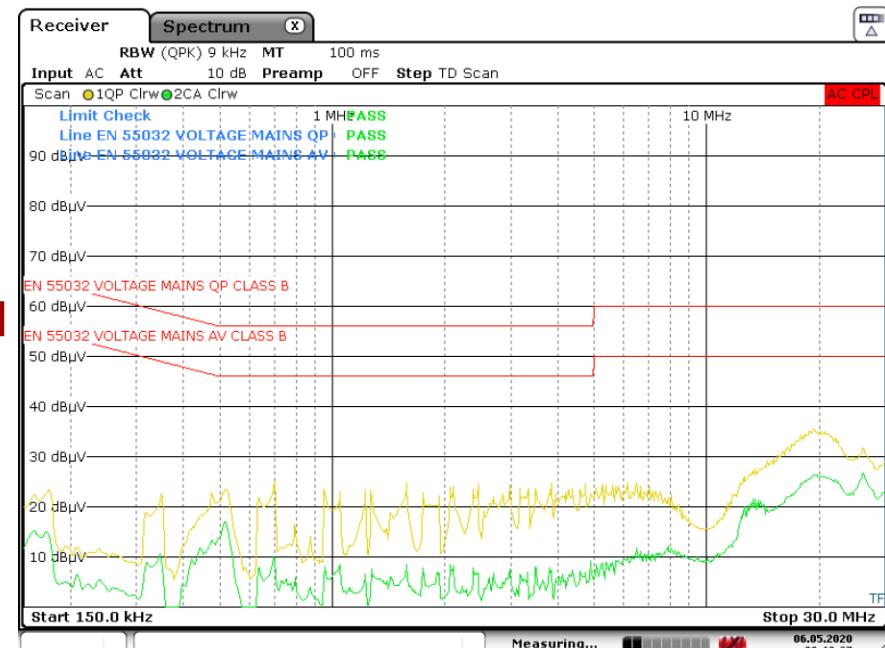


# CM Battle Won!

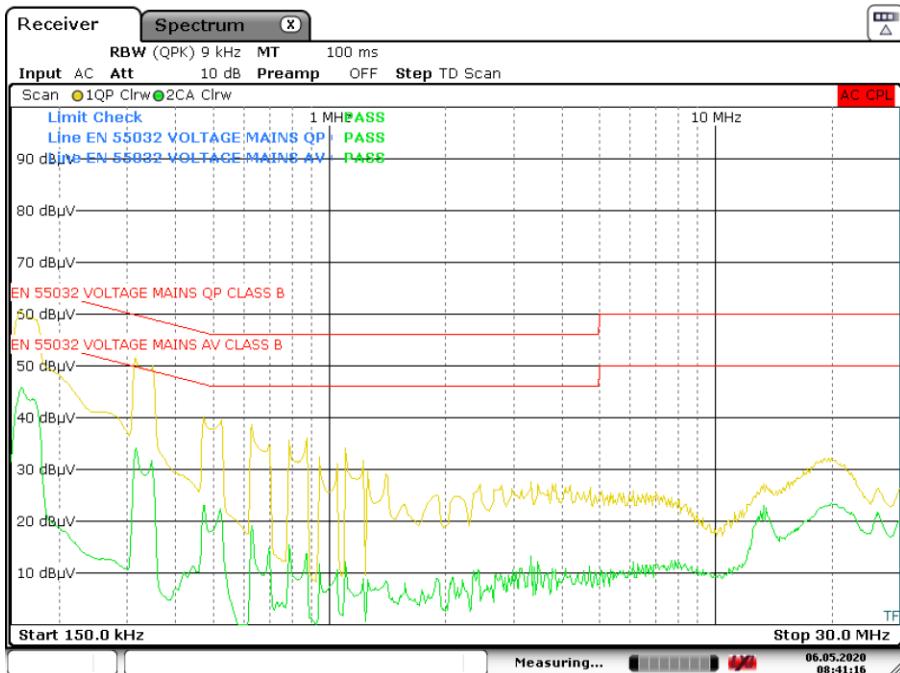
## Differential Mode



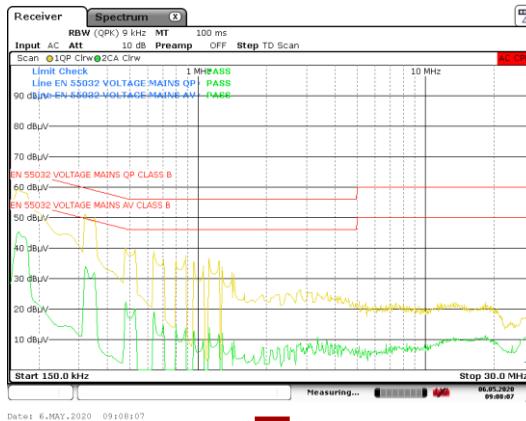
## Common Mode



# Total Victory!



Date: 6.MAY.2020 08:41:16



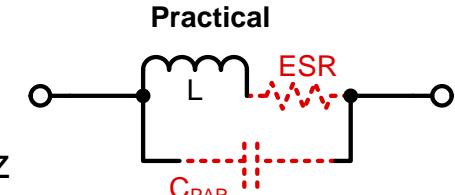
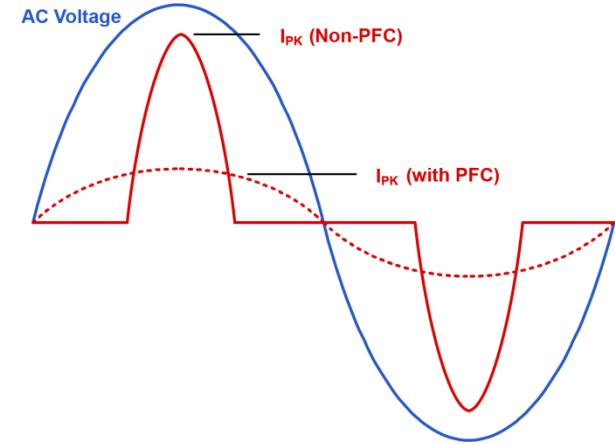
Date: 6.MAY.2020 09:08:07



Date: 6.MAY.2020 09:10:37

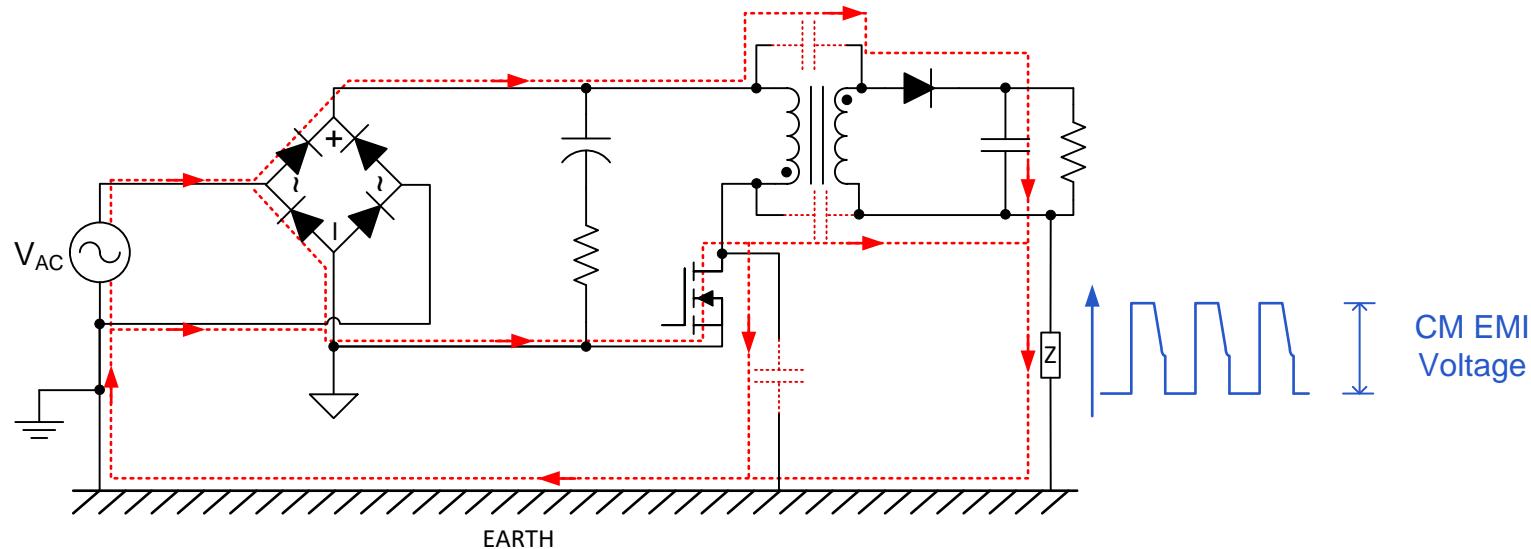
# DM filter choke practical considerations

- Choke requires high attenuation over wide bandwidth:
  - Load current amplitude typically several amps
  - At 50 dB $\mu$ V, current in LISN 50 Ω resistor only ~6.3 μA
- Beware inductance roll-off with DC-bias
  - Must not saturate to be effective – needs high current rating
  - Consider the peak line current for non-PFC – high crest factor
- Switching power stage has fast changing magnetic fields
  - Beware filter bypassing & noise coupling
- Parasitic capacitance across DM inductor very important
  - Reduces effectiveness, especially at high frequency
- Example: To filter 300 kHz component, typically set LC freq. ~30 kHz
  - Expect ~40 dB attenuation at 300 kHz (double-pole  $\Rightarrow$  40 dB/decade)
  - With parasitic cap  $\Rightarrow$  more like 30 dB attenuation only – even worse at higher frequency



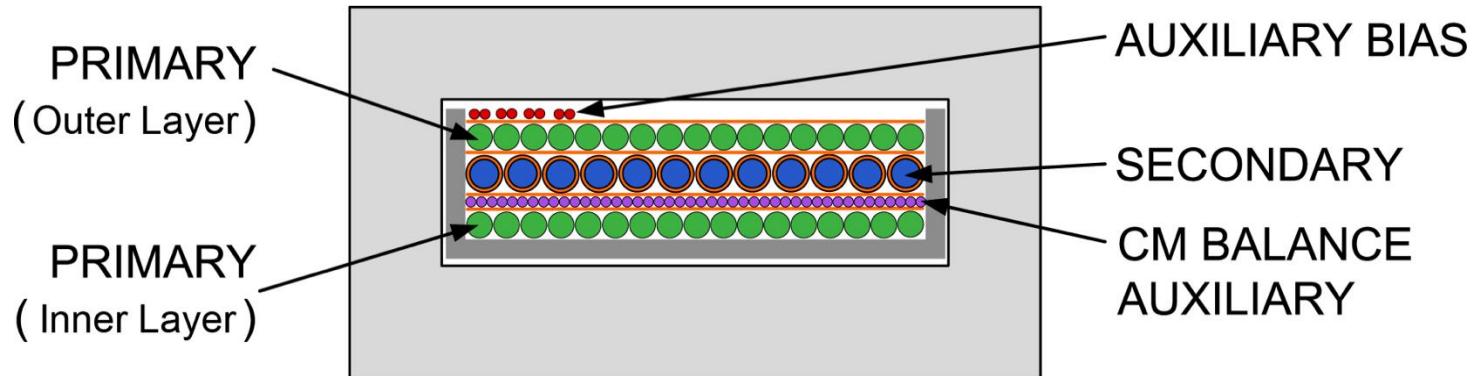
# How is CM EMI generated?

- Switching voltage across parasitic capacitance causes CM current flow to EARTH
- CM noise also radiated to other circuit nodes



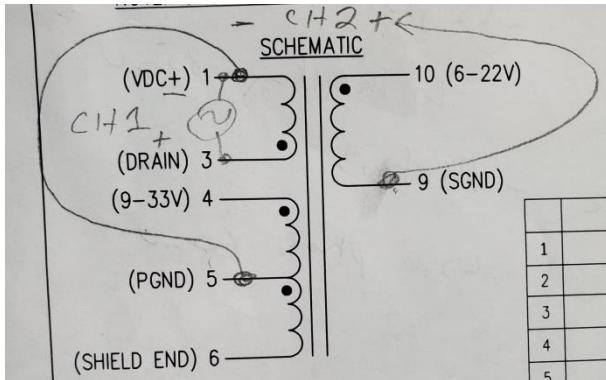
# Transformer CM balance – PMP21479 ACF

- Add CM balance auxiliary layer (purple) in-between inner PRI (noisier) to SEC interface:
  - *Fill layer completely – acts as shield between PRI & SEC*
  - *Add turns to create CM balance, inject current to balance other PRI-SEC interface*

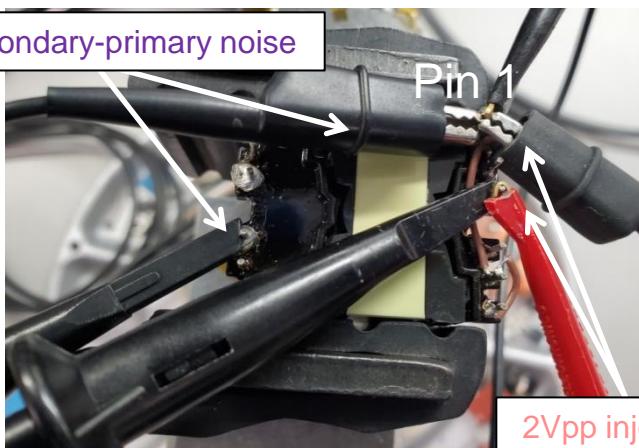


- NOTE: this example shows one way to add CM balance
- But there are many different ways to achieve the same CM result

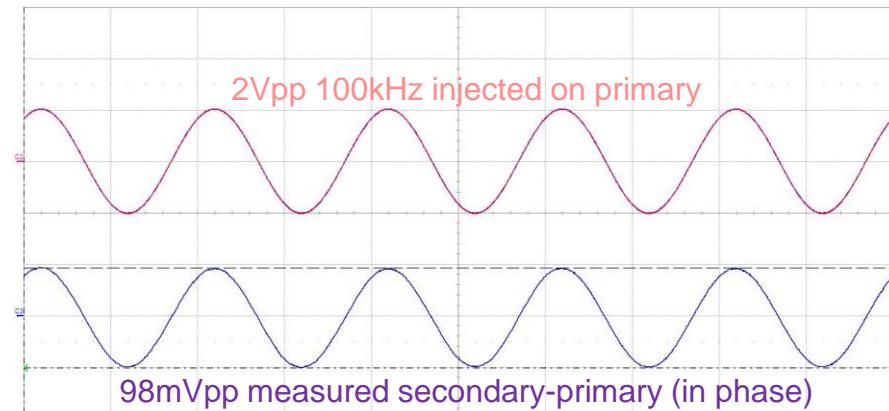
# Checking Transformer CM Balance



secondary-primary noise



1. Tie primary AC quiet pins together.
2. Inject sinewave across main primary winding
3. Measure from secondary AC quiet node to primary AC quiet node

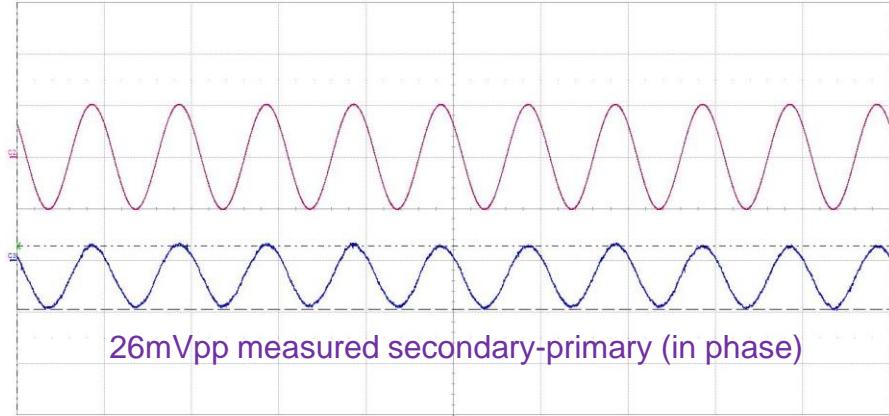


Measure	value	status
BrdL AC1m	FLT DC1m	
	1.00 V/div	
	1.00 mV/div	
	-1.00 V/div	
	-1.00 mV/div	
	-4.02 V	<1.0 mV
	-4.02 mV	
	-2.07 V	46.5 mV
	-2.07 mV	

Timebase 0.0  $\mu$ s  
5.000000 Stop 399 V  
25.000000 End Date Edge Positive

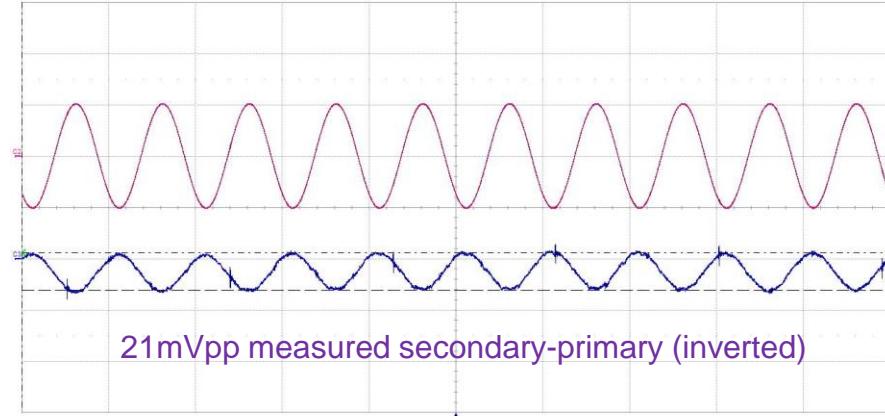
# Checking Transformer CM Balance

+2 Cancellation Turns



Measure value status  
P1: AC11 P1: min(C2) -1.022 V  
P2: AC11 P2: min(C3) -19.3 mV  
P3: freq(C3) 100.15424 kHz  
P4: pkpk(C3) 26.144 mV  
P5: ---  
P6: ---  
P7: ---  
P8: ---

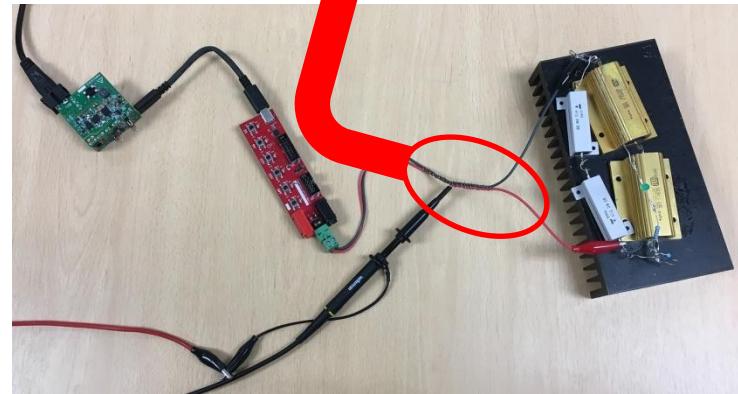
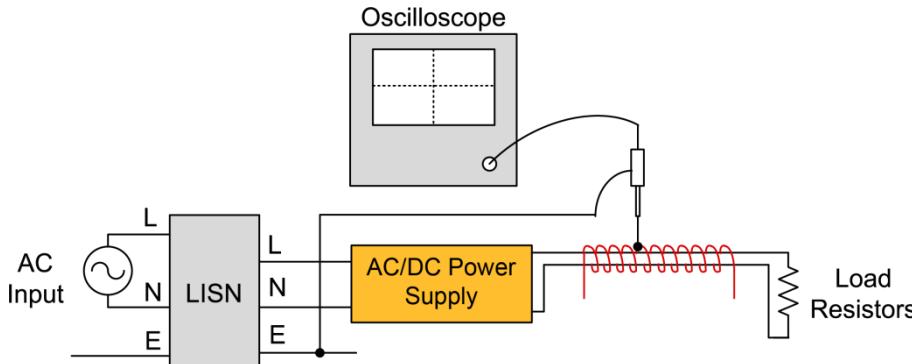
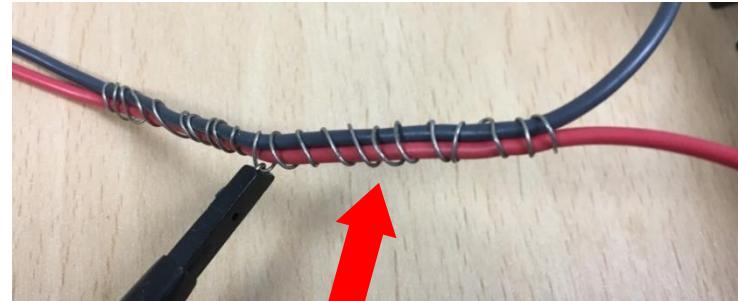
+3 Cancellation Turns



Measure value status  
P1: AC11 P1: freq(C3) 100.43287 kHz  
P2: AC11 P2: pkpk(C3) 21.206 mV  
P3: ---  
P4: ---  
P5: ---  
P6: ---  
P7: ---  
P8: ---

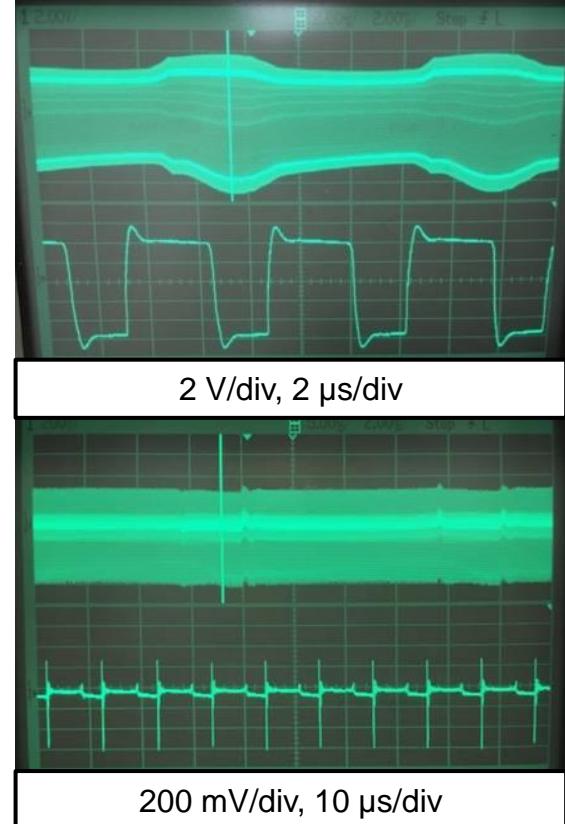
# Observing the time-domain CM signal at the output

- Useful debug technique – ball-park indication of CM performance
  - Remove Y-cap temporarily (maximize signal)
  - Power EUT through LISN, with resistor loads
  - Wind several turns of wire around the load cables to create capacitive sensing coil (pickup coil)
  - Connect scope EARTH lead to LISN EARTH
  - Connect scope tip to sensing coil
  - Scope plot shows how much CM is coupled to output



# Interpretation of time-domain CM signal

- Will see “switch-node” shaped waveform – coupled to output
- Large PK-PK amplitude  $\Rightarrow$  bad CM noise
  - *Will require significant CM filtering to suppress*
  - *Result from ACF example with 100 dB $\mu$ V EMI*
- Small PK-PK amplitude  $\Rightarrow$  good CM noise
  - *Balanced* structure giving low CM
  - *Will require much smaller CM filter*
- Residual HF “spikes”  $\Rightarrow$  should only need small HF CM choke



# **DESIGN EXAMPLE SHOWING PATH TO PASSING CISPR 25**

Bob Sheehan

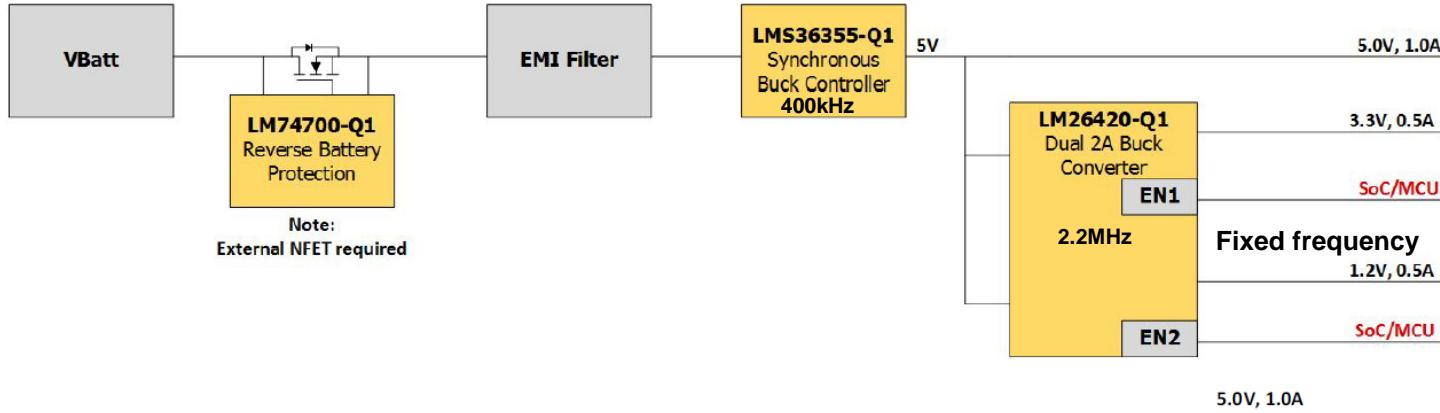
# Design example

PMP21417/PMP21611 showing path to passing CISPR25.

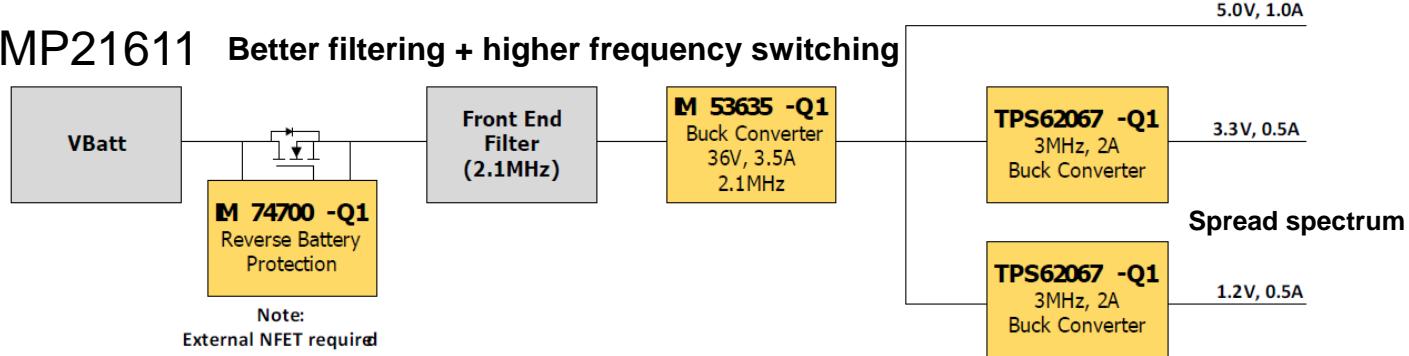
- Automotive front end power solution for a high performance cluster
  - Three rails for point of load power
- PMP21417 designed and sent to customer
  - Our internal conducted emissions test looked good
  - Customer reported **failing radiated emissions**
- PMP21611 redesigned two of the rails
  - Our internal conducted emissions looked better
  - Customer reported **passing radiated emissions**

# Architecture

## PMP21417

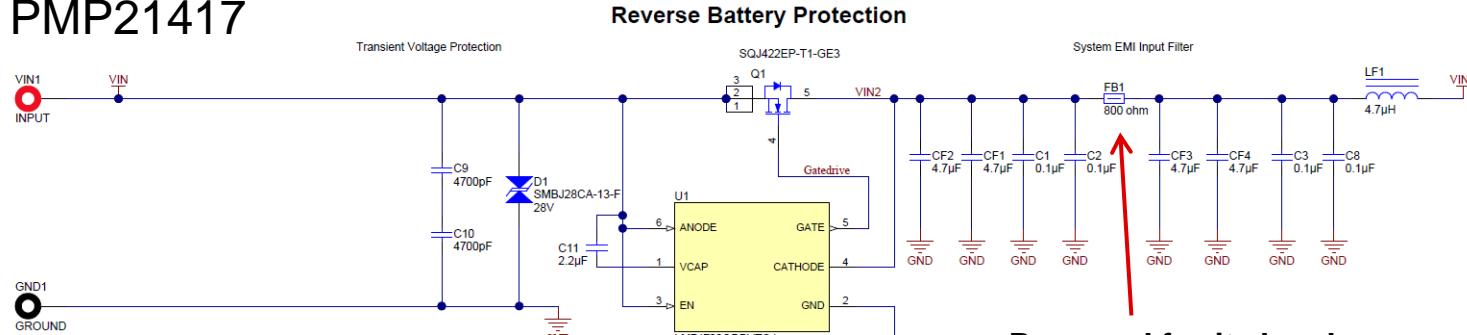


## PMP21611 Better filtering + higher frequency switching



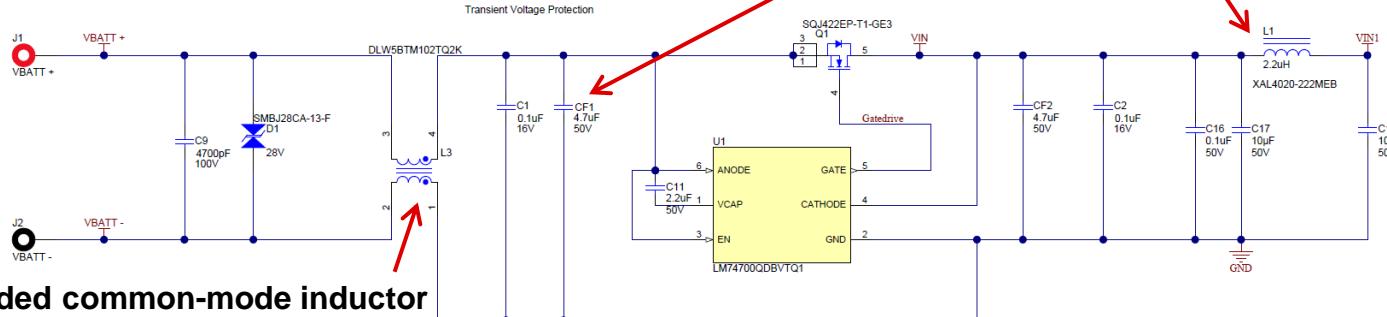
# Reverse protection and filter

PMP21417



Removed ferrite bead  
Re-distributed filter capacitors  
Smaller inductor for higher Fsw

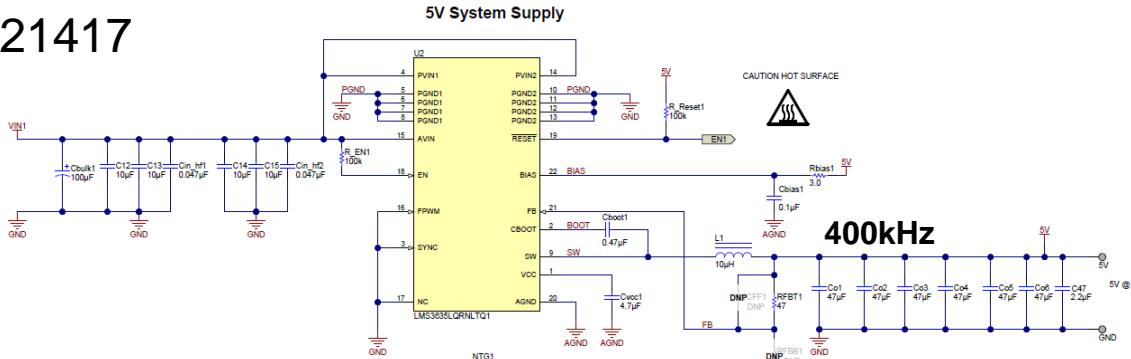
PMP21611



Added common-mode inductor

# 5 V system supply

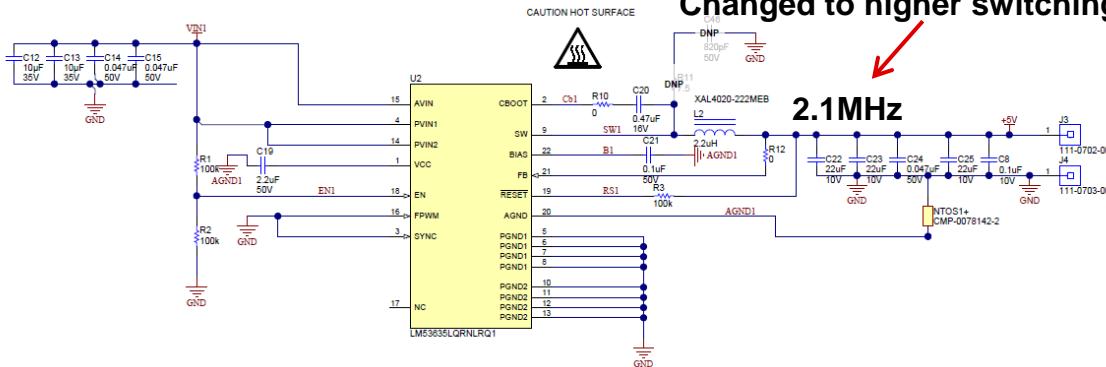
PMP21417



400kHz

Both are spread spectrum

PMP21611



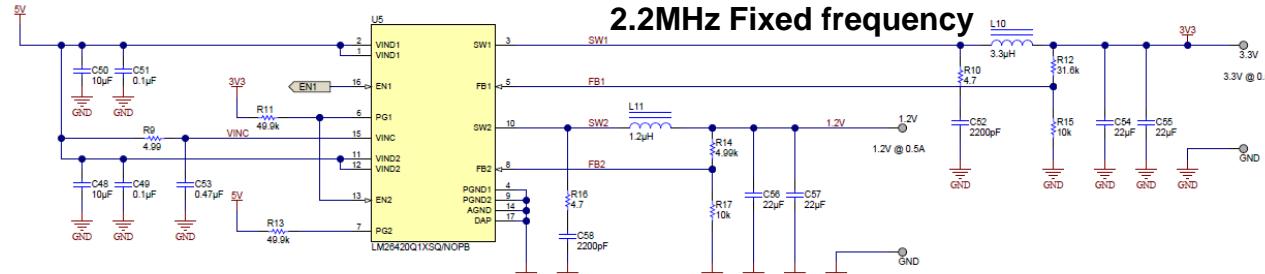
2.1MHz

Changed to higher switching frequency

# 3.3 V and 1.2 V supplies

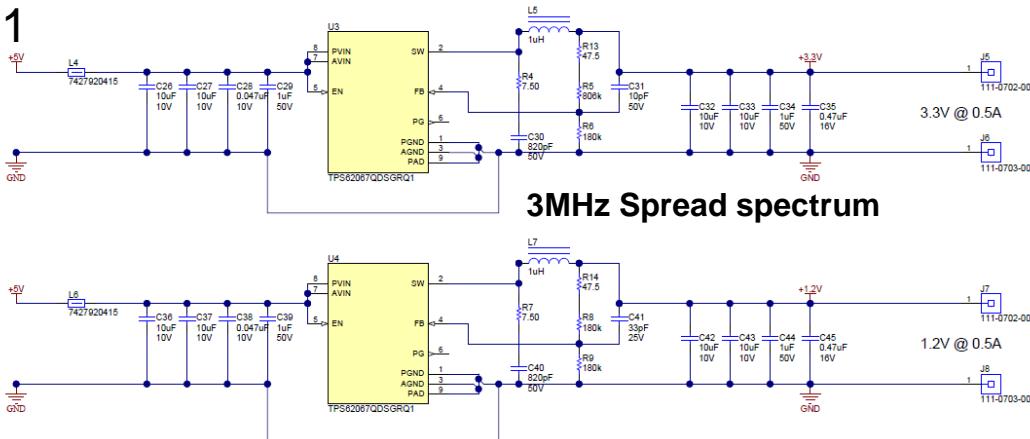
PMP21417

3.3V & 1.2V Supplies



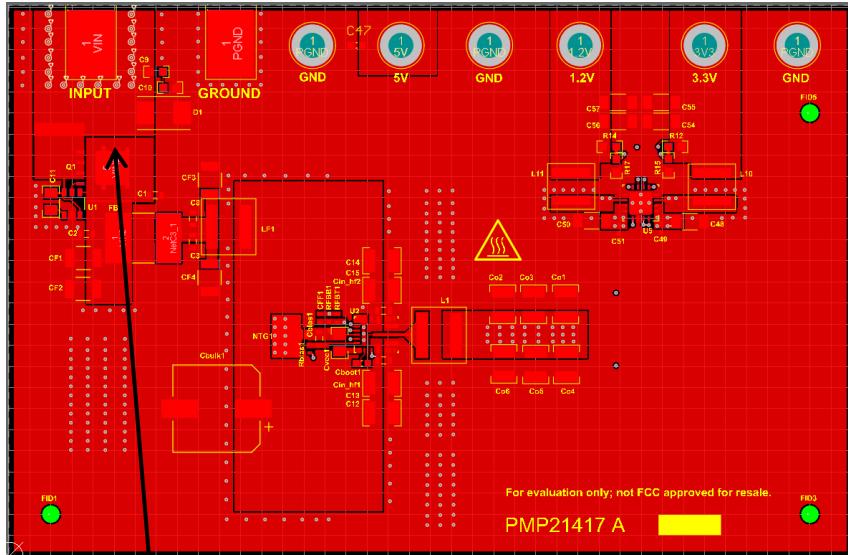
PMP21611

3MHz Spread spectrum



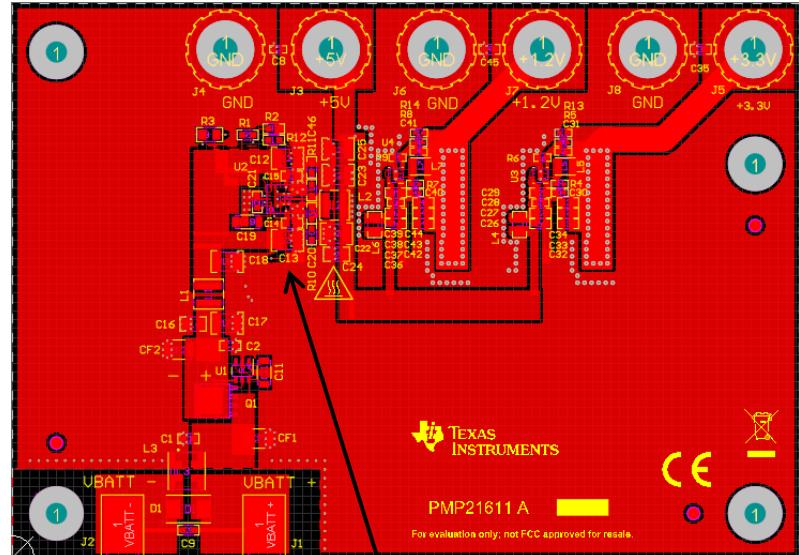
# Top layer

PMP21417



Moved input and filter to opposite edge of the board

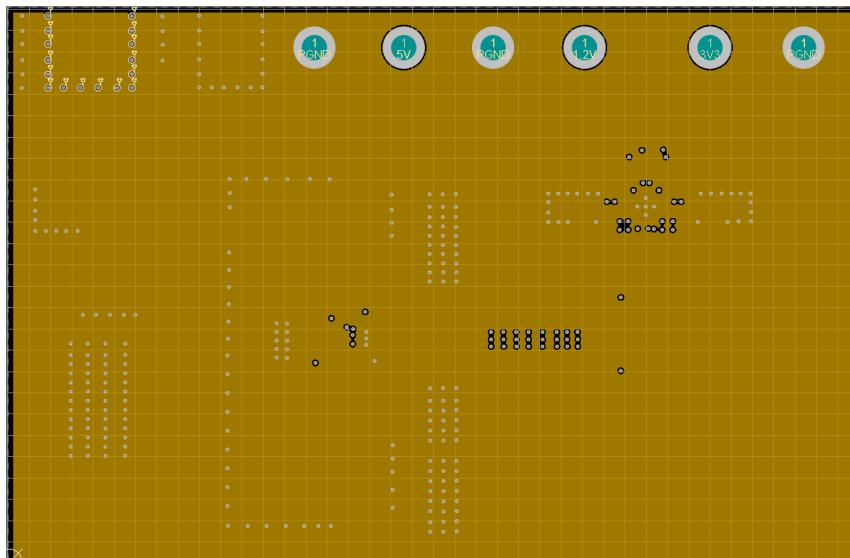
PMP21611



Ceramic caps feed 5V converter from both sides

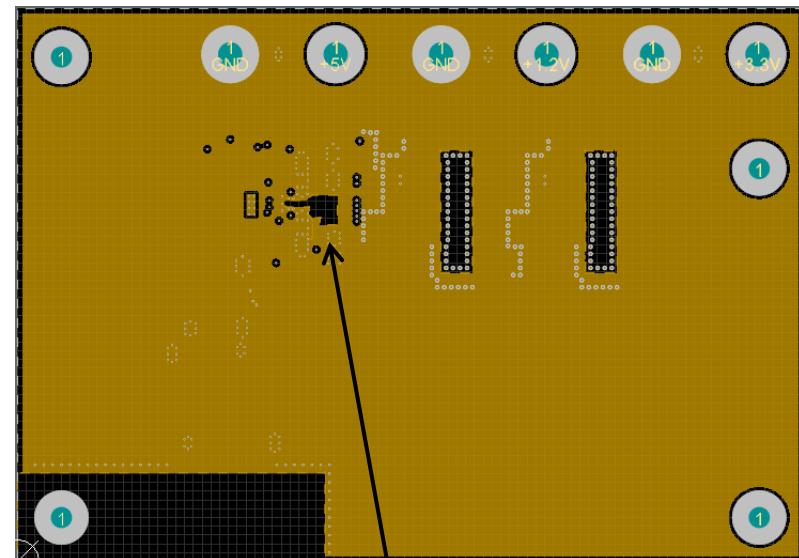
# Mid-layer 1

PMP21417



Input common-mode filter area kept clear

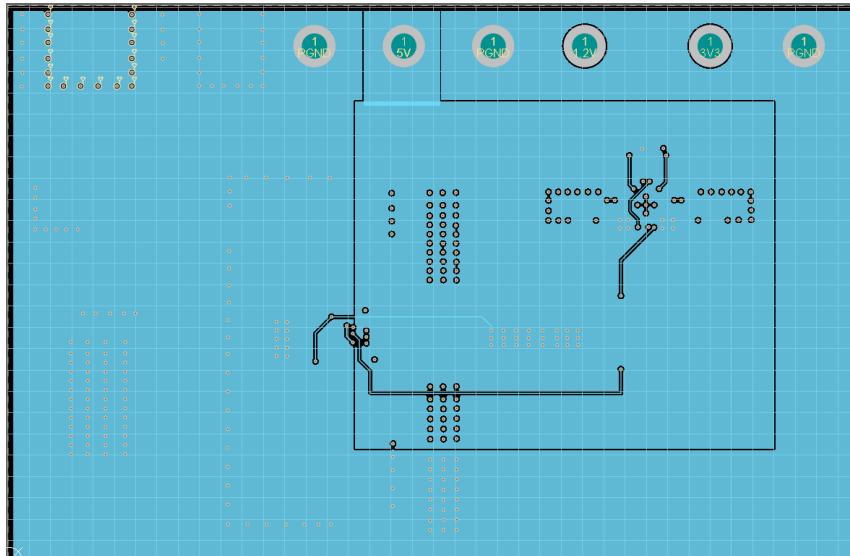
PMP21611



Removed copper under 5V inductor

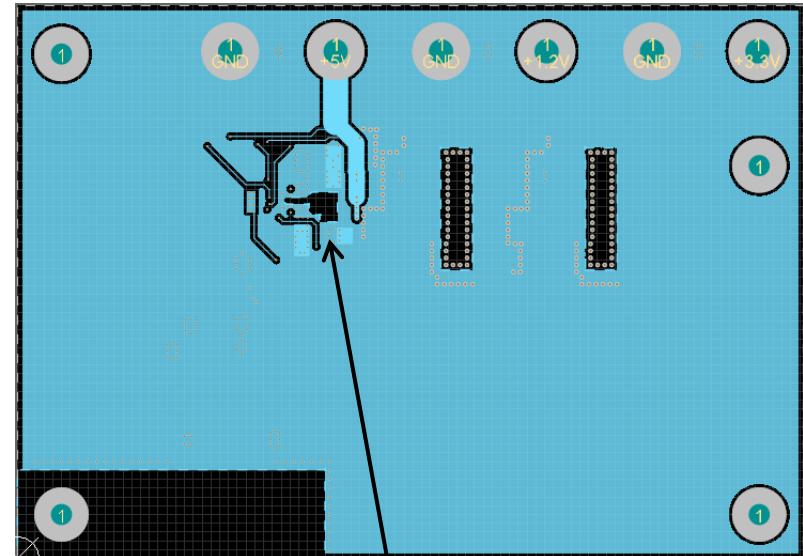
# Mid-layer 2

PMP21417



Input common-mode filter area kept clear

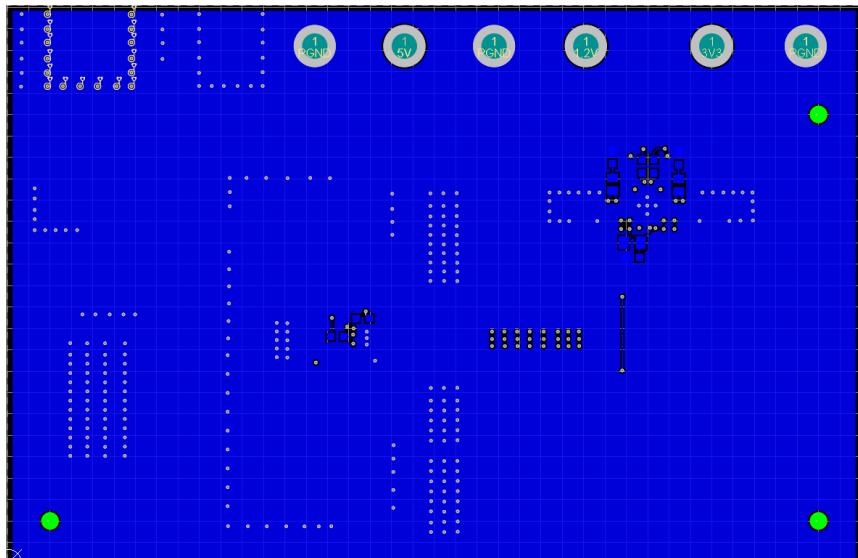
PMP21611



Removed copper under 5V inductor

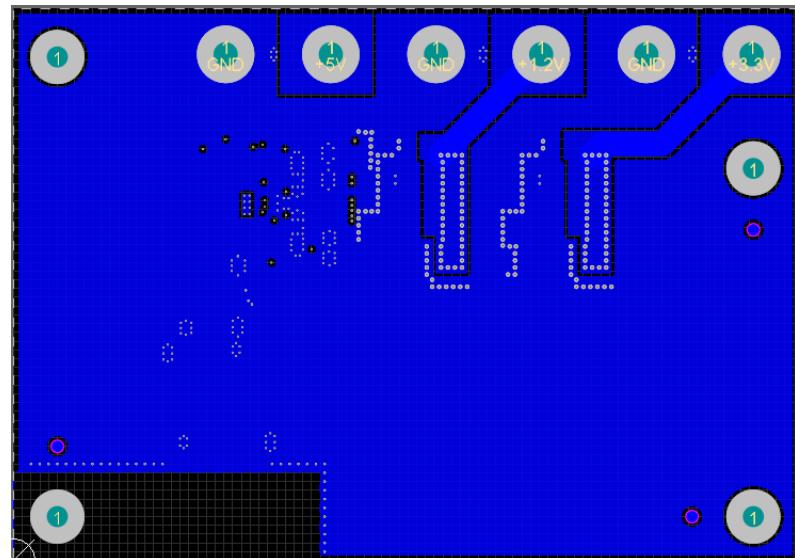
# Bottom layer

PMP21417



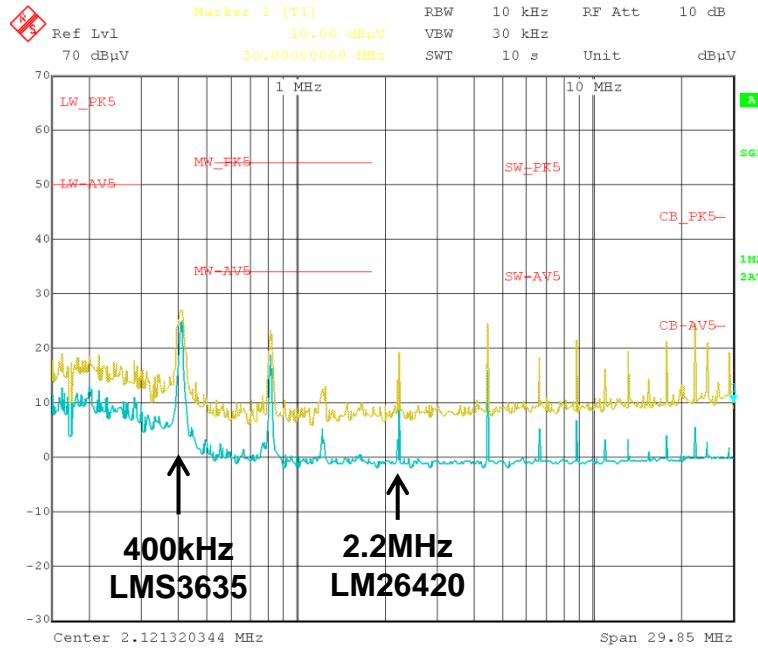
Input common-mode filter area kept clear

PMP21611



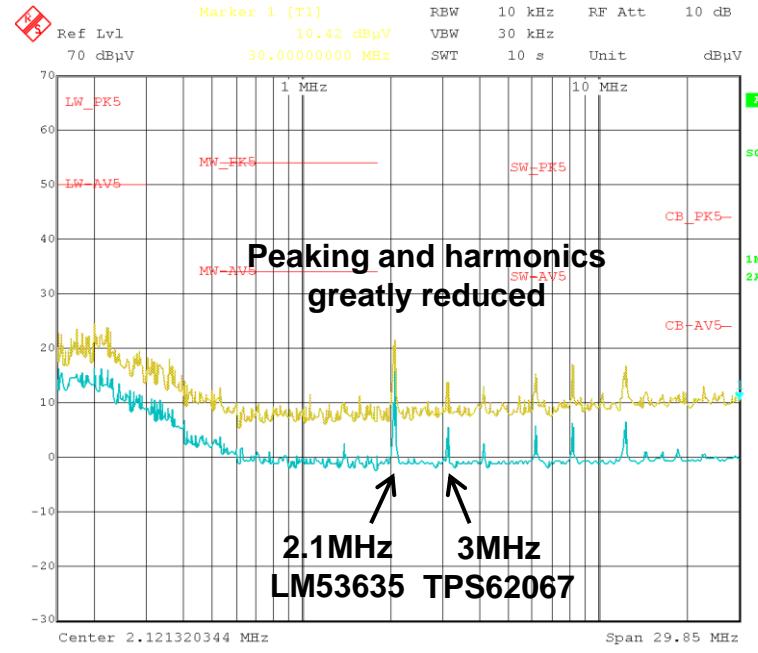
# Conducted emissions – low frequency

PMP21417



Date: 6.MAR.2018 08:58:39

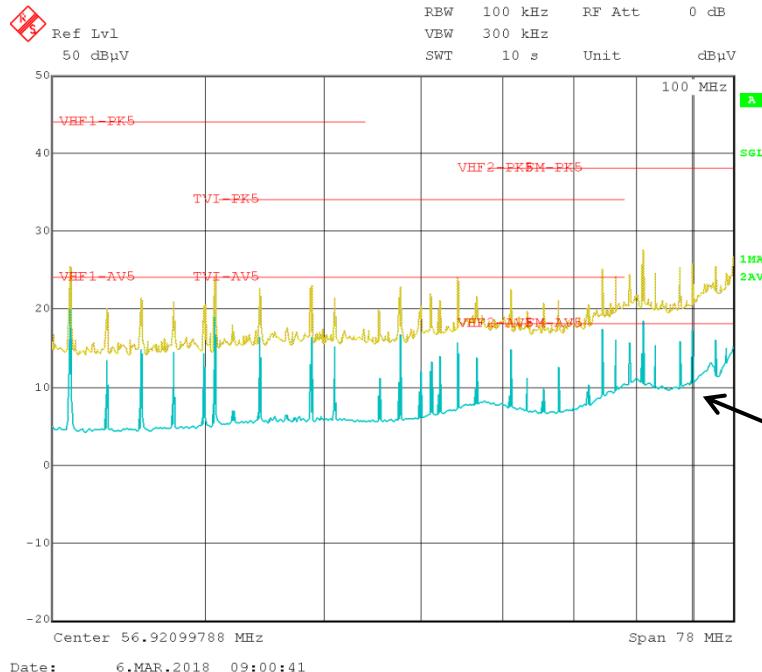
PMP21611



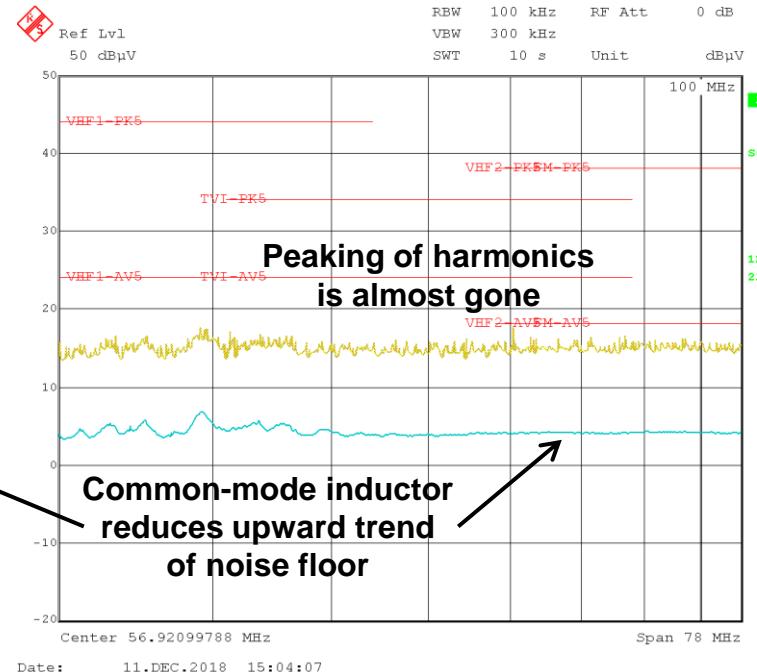
Date: 11.DEC.2018 15:01:54

# Conducted emissions – high frequency

PMP21417

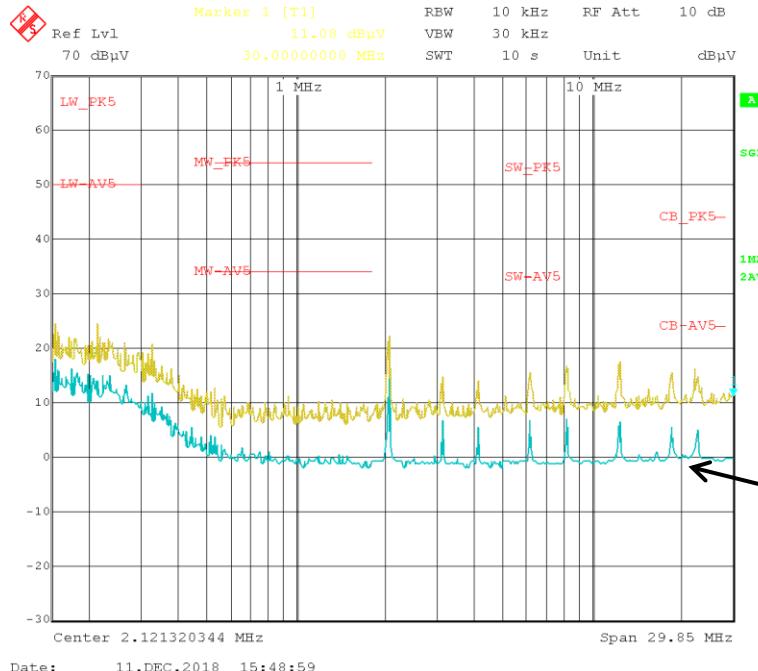


PMP21611

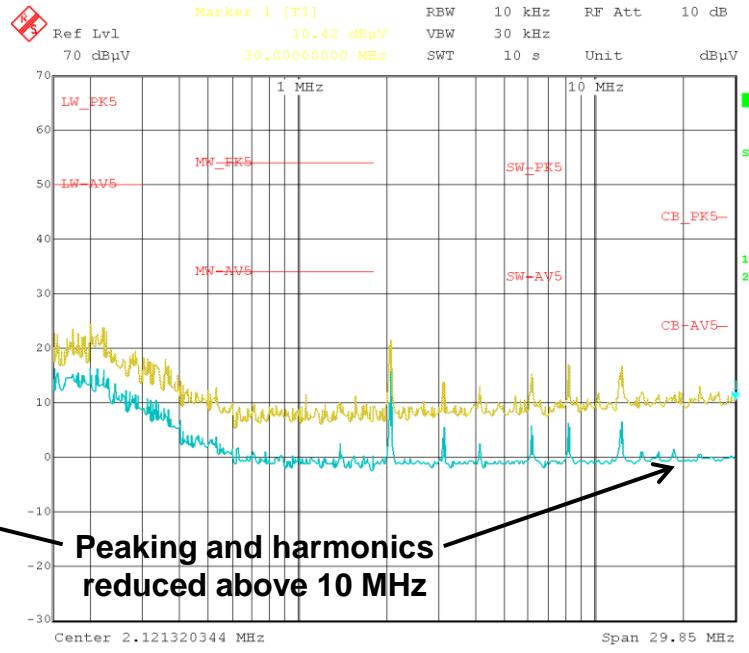


# Common-mode inductor – low frequency

PMP21611 without CM inductor



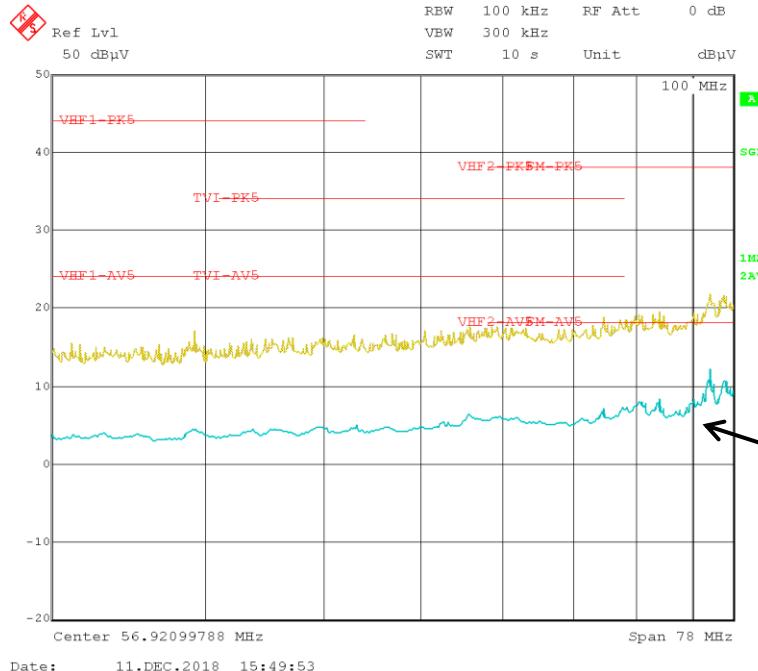
PMP21611 with CM inductor



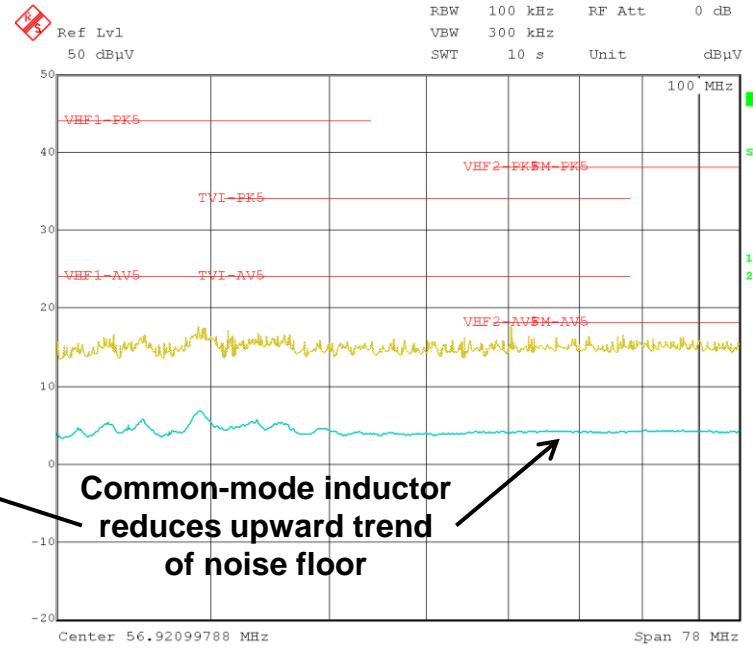
Peaking and harmonics reduced above 10 MHz

# Common-mode inductor – high frequency

PMP21611 without CM inductor



PMP21611 with CM inductor



Common-mode inductor  
reduces upward trend  
of noise floor

# Conclusion

- Designing power conversion systems to meet EMC requirements can be challenging.
- Understanding EMI sources, standards, filters, test setups is important
- Distinguishing between differential mode and common mode noise helps when debugging EMI problems and identifying solutions
- A path to meeting CISPR 25 requirements was demonstrated with a design example



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