

ECE212 Homework 9

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Q1.

In the provided spaces, construct the straight-line approximations of the Bode plots for the following transfer functions. Clearly label all critical (corner) frequencies and the dB gains of “flat” regions.

$$a. \quad T(s) = \frac{70}{\left(1 + \frac{s}{300}\right)\left(1 + \frac{s}{70000}\right)}$$

Write the transfer function in standard form:

$$\begin{aligned} T(j\omega) &= \frac{70}{\left(1 + \frac{j\omega}{300}\right)\left(1 + \frac{j\omega}{70000}\right)} \\ &= \frac{70}{\sqrt{1 + \left(\frac{\omega}{300}\right)^2} \cdot e^{j \cdot \arctan\left(\frac{\omega}{300}\right)} \cdot \sqrt{1 + \left(\frac{\omega}{70000}\right)^2} \cdot e^{j \cdot \arctan\left(\frac{\omega}{70000}\right)}} \\ &= \frac{70}{\sqrt{1 + \left(\frac{\omega}{300}\right)^2} \cdot \sqrt{1 + \left(\frac{\omega}{70000}\right)^2}} \cdot e^{j\left(0^\circ - \arctan\left(\frac{\omega}{300}\right) - \arctan\left(\frac{\omega}{70000}\right)\right)} \end{aligned}$$

Two **critical frequencies** can be found at $\omega_1 = 300$, $\omega_2 = 70000$. The full derivations for gain and phase responses are:

$$\begin{aligned} |T(j\omega)|_{dB} &= 20 \log_{10} \left(\frac{70}{\sqrt{1 + \left(\frac{\omega}{300}\right)^2} \cdot \sqrt{1 + \left(\frac{\omega}{70000}\right)^2}} \right) \\ &= 20 \left(\log_{10}(70) - \log_{10} \left(\sqrt{1 + \left(\frac{\omega}{300}\right)^2} \right) - \log_{10} \left(\sqrt{1 + \left(\frac{\omega}{70000}\right)^2} \right) \right) \\ \theta(\omega) &= 0^\circ - \arctan\left(\frac{\omega}{300}\right) - \arctan\left(\frac{\omega}{70000}\right) \end{aligned}$$

Approximated gain calculation based on above derivation:

$$\omega = 0: |T(0)| = 20 \cdot \log_{10}(70) = 36.902 \text{ dB} \Rightarrow f = 0 \text{ Hz}$$

$$\omega = 300: f = \frac{300}{2\pi} = 47.75 \text{ Hz}$$

Approximated phase calculation based on above derivation:

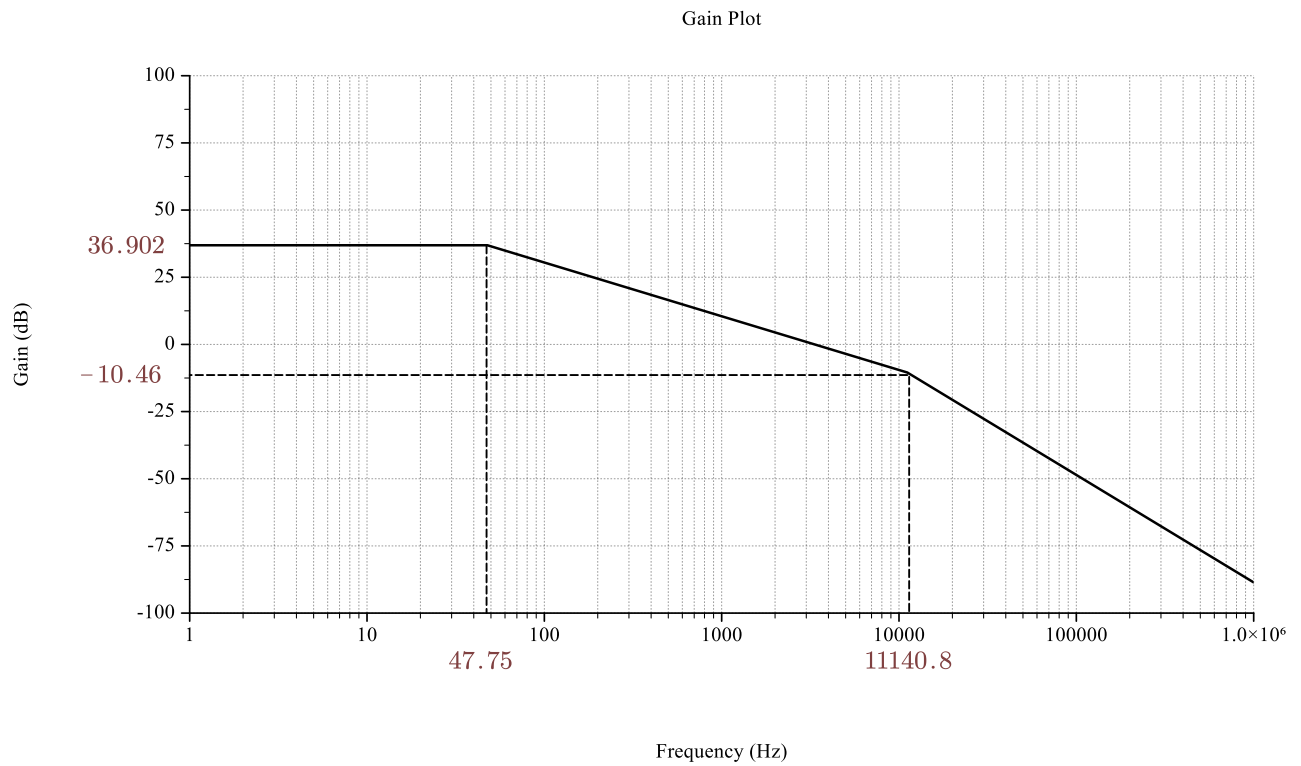
$$\text{when } \omega \ll 300 \Rightarrow f \ll 47.75 \Rightarrow f > \frac{1}{10} \cdot 47.75 = 4.77465 \text{ Hz}$$

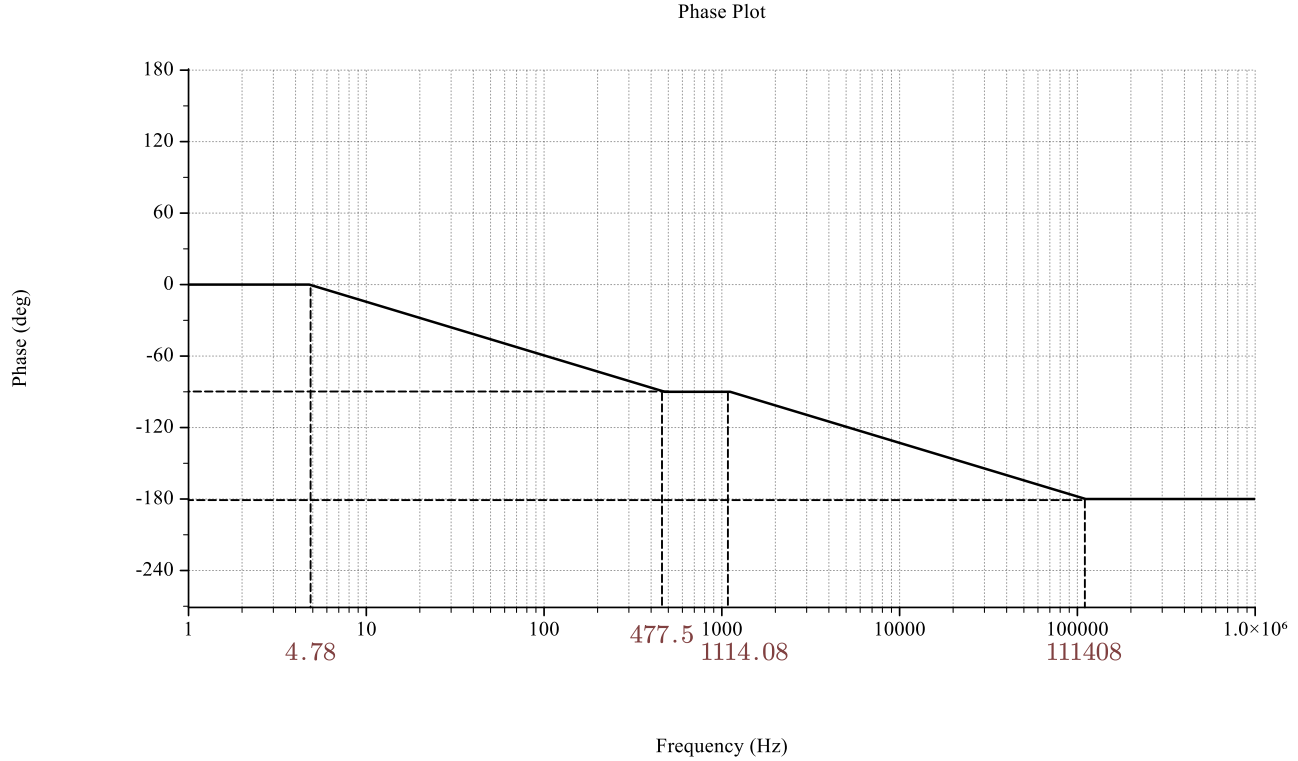
$$300 \gg \omega \Rightarrow f \gg 47.75 \Rightarrow f < 10 \cdot 47.75 = 477.465 \text{ Hz}$$

$$\omega \ll 70000 \Rightarrow f \ll 11140.8 \Rightarrow f > \frac{1}{10} \cdot 11140.8 = 1114.08 \text{ Hz}$$

$$70000 \gg \omega \Rightarrow f \gg 11140.8 \Rightarrow f < 10 \cdot 11140.8 = 111408 \text{ Hz}$$

Therefore, Bode plots for the stated transfer function can be drawn according the above calculations:





$$b. \quad T(s) = \frac{20}{s(s + 0.01)}$$

Write the transfer function in standard form:

$$T(j\omega) = \frac{2000}{j\omega(100j\omega + 1)} = \frac{2000}{\omega \cdot e^{j\frac{\pi}{2}} \cdot \sqrt{1 + \left(\frac{\omega}{0.01}\right)^2} \cdot e^{j \cdot \arctan\left(\frac{\omega}{0.01}\right)}} = \frac{2000}{\omega \cdot \sqrt{1 + \left(\frac{\omega}{0.01}\right)^2}} \cdot e^{j\left(0^\circ - 90^\circ - \arctan\left(\frac{\omega}{0.01}\right)\right)}$$

Two **critical frequencies** can be found at $\omega_1 = 1$, $\omega_2 = 0.01$. The full derivations for gain/ phase responses are:

$$|T(j\omega)|_{dB} = 20 \log_{10} \left(\frac{2000}{\omega \cdot \sqrt{1 + \left(\frac{\omega}{0.01}\right)^2}} \right) = 20 \left(\log_{10}(2000) - \log_{10}(\omega) - \log_{10} \left(\sqrt{1 + \left(\frac{\omega}{0.01}\right)^2} \right) \right)$$

$$\theta(\omega) = -90^\circ - \arctan\left(\frac{\omega}{0.01}\right)$$

Approximated gain calculation based on above derivation:

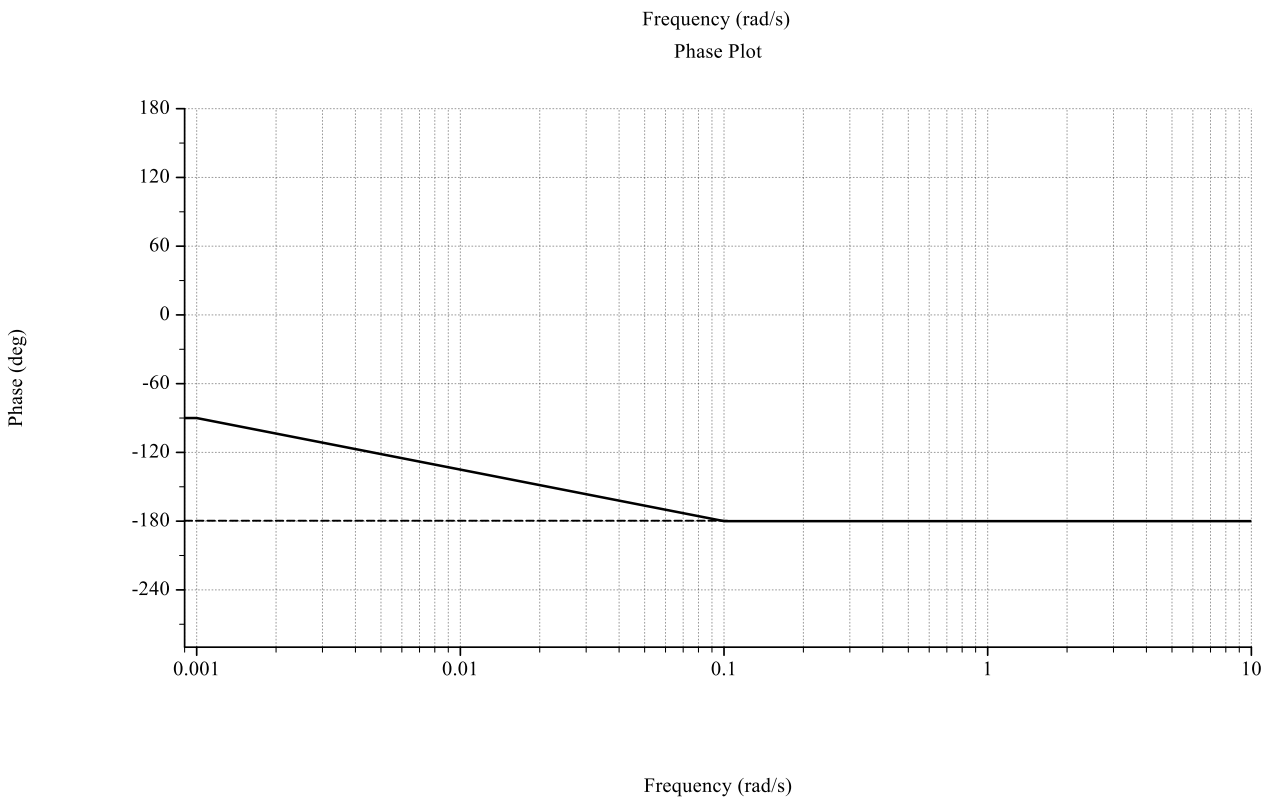
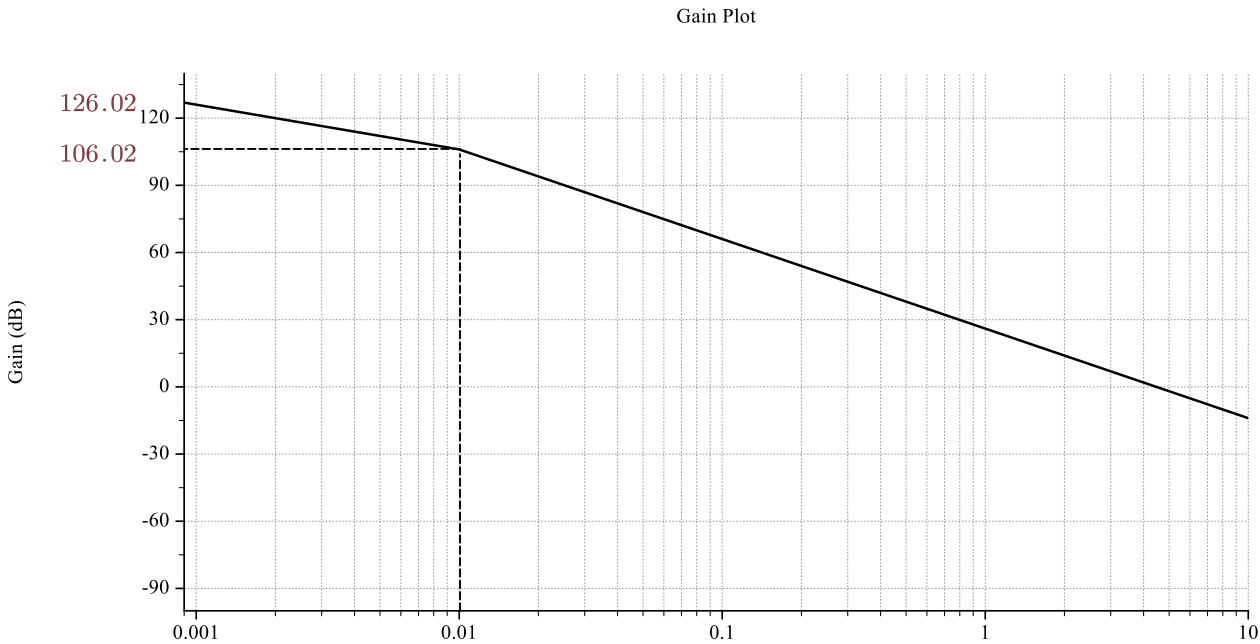
$$\omega = 0.001: |T(0.001)| = 20 \cdot \log_{10}(2000) - 20 \cdot \log_{10}(0.001) = 126.021 \text{ dB}$$

$$\omega = 0.01: |T(0.01)| = 20 \cdot \log_{10}(2000) - 20 \cdot \log_{10}(0.01) = 106.021 \text{ dB}$$

Approximated phase calculation based on above derivation:

$$\text{when } \omega \ll 0.01 \Rightarrow \omega > \frac{1}{10} \cdot 0.01 = 0.001, \text{ when } 0.01 \gg \omega \Rightarrow \omega < 10 \cdot 0.01 = 0.1$$

Therefore, Bode plots for the stated transfer function can be drawn according the above calculations:



$$c. \quad T(s) = \frac{200}{1 + \frac{s}{8000} + \frac{s^2}{16000}}$$

Write the transfer function in standard form:

$$\begin{aligned} T(j\omega) &= \frac{200}{1 + \frac{j\omega}{8000} - \frac{\omega^2}{16000}} \\ &= \frac{200}{1 - \left(\frac{\omega}{40\sqrt{10}}\right)^2 + j \cdot 2 \cdot \frac{1}{40\sqrt{10}} \frac{\omega}{40\sqrt{10}}} \end{aligned}$$

Two **critical frequencies** can be found at $\omega = 40\sqrt{10} = 126.491$. As the question is asking for straight-line approximation, only approximated calculations (damping factor is neglected) are provided as follows:

$$\omega = 0: |T(0)| = 20 \cdot \log_{10}(200) = 46.0206 \text{ dB}$$

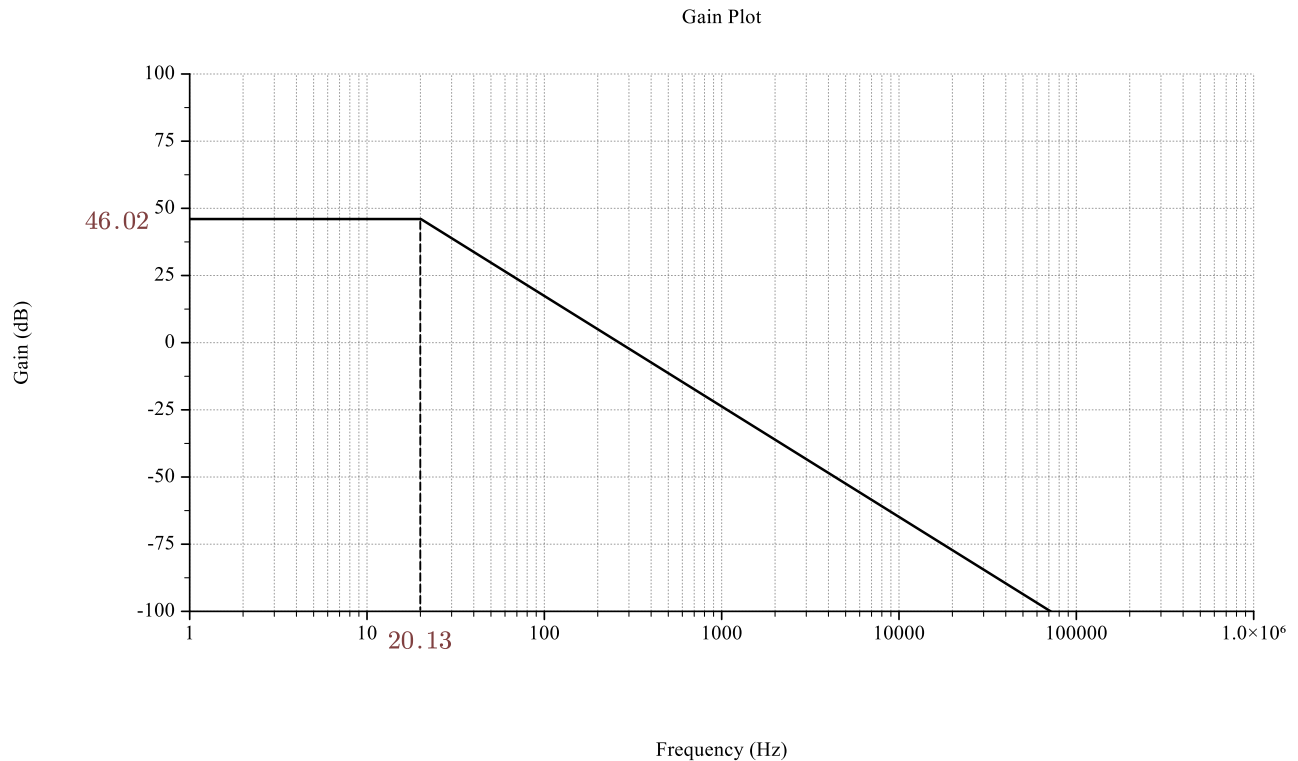
$$\omega = 40\sqrt{10}: f(40\sqrt{10}) = \frac{40\sqrt{10}}{2\pi} = 20.1317 \text{ Hz}$$

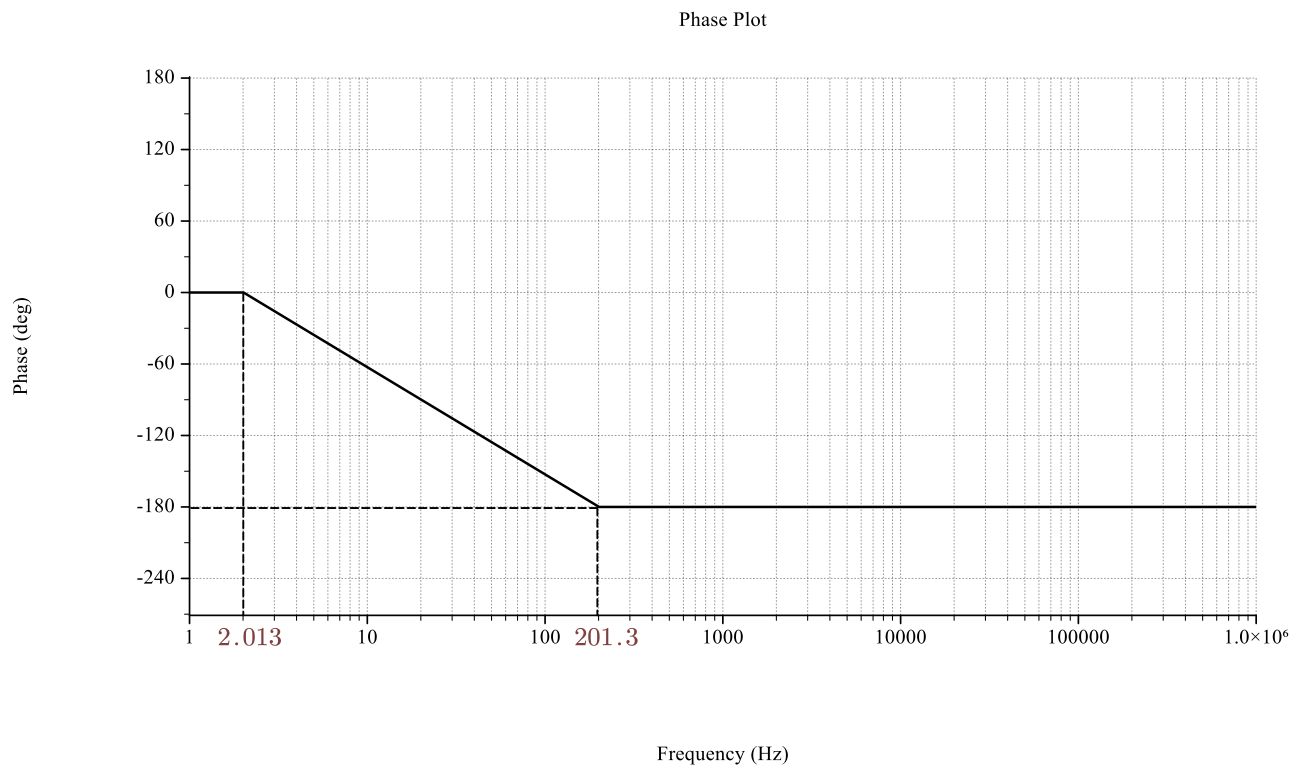
Approximated phase calculation based on above derivation:

$$\text{when } f \ll 20.13 \Rightarrow \omega > \frac{1}{10} \cdot 20.13 = 2.013,$$

$$20.13 \gg \omega \Rightarrow \omega < 10 \cdot 20.13 = 201.3$$

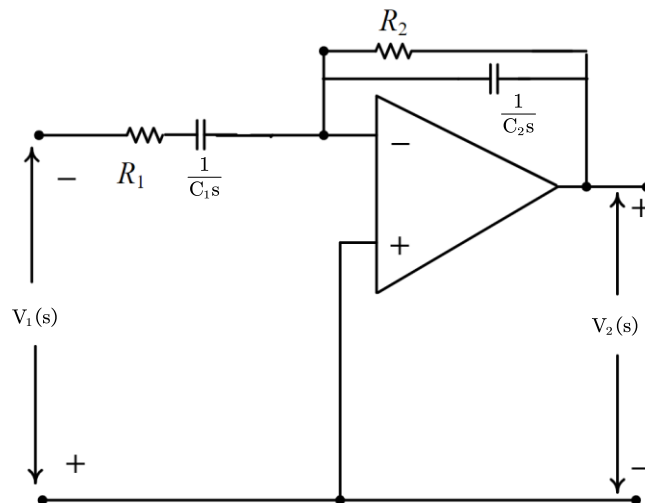
Therefore, Bode plots for the stated transfer function can be drawn according the above calculations:





Q2.

a. Transfer the circuit in s-domain



b. find the feedback and forward path impedance Z2 and Z1

Forward path impedance:

$$Z_1 = Z_{R1} + Z_{C1} = R_1 + \frac{1}{C_1 s}$$

Feedback path impedance:

$$Z_2 = Z_{R2} || Z_{C2} = \frac{R_2 \cdot \frac{1}{C_2 s}}{R_2 + \frac{1}{C_2 s}} = \frac{R_2}{R_2 C_2 s + 1}$$

c. find the Gain $T(s) = v_2(s)/v_1(s) = Z_2(s)/Z_1(s)$

$$T(s) = \frac{v_2(s)}{v_1(s)} = \frac{Z_2(s)}{Z_1(s)} = \frac{\frac{R_2}{R_2 C_2 s + 1}}{R_1 + \frac{1}{C_1 s}} = \frac{R_2 C_1 s}{(R_1 C_1 s + 1)(R_2 C_2 s + 1)}$$

d. Select the values of R_s and C_s such that the critical (corner) frequencies and the gains of the OP.Amp. circuit match those shown with the Bode plots.

Write the transfer function in standard form:

$$T(s) = \frac{R_2 C_1 s}{(R_1 C_1 s + 1)(R_2 C_2 s + 1)} = \frac{R_2 C_1 s}{\left(\frac{s}{\frac{1}{R_1 C_1}} + 1\right)\left(\frac{s}{\frac{1}{R_2 C_2}} + 1\right)}$$

Two critical points at $\omega_1 = 20 \text{ Hz} = 125.664 \text{ rad/s}$ and $\omega_1 = 10000 \text{ Hz} = 62831.9 \text{ rad/s}$ can be examined from the given gain plot. Furthermore, the gain plot suggests that the gain at $\omega_1 = 0 \text{ rad/s}$ is approximately 14. This signifies that $20 \cdot \log(R_2 C_1) \approx 14 \Rightarrow R_2 C_1 = 10^{14/20} = 5.01187$.

Therefore, a system of equations can be constructed as follows:

$$\begin{cases} \frac{1}{R_1 C_1} = 125.664 \\ \frac{1}{R_2 C_2} = 62831.9 \\ R_2 C_1 = 5.01187 \end{cases} \Rightarrow \text{solve by setting } R_2 = 1 \Omega \Rightarrow \begin{cases} R_1 = 1.588 \text{ m}\Omega \\ C_2 = 15.92 \text{ }\mu\text{F} \\ C_1 = 5.01187 \text{ F} \end{cases}$$

Select $R_1 = 1.588 \text{ m}\Omega$, $R_2 = 1 \Omega$, $C_1 = 5.01 \text{ F}$, $C_2 = 15.92 \text{ }\mu\text{F}$.