



## Notes

### Assignments

◆ 5.2, 5.5, 5.15, 5.21(a-f, h)

### Tutorial problems

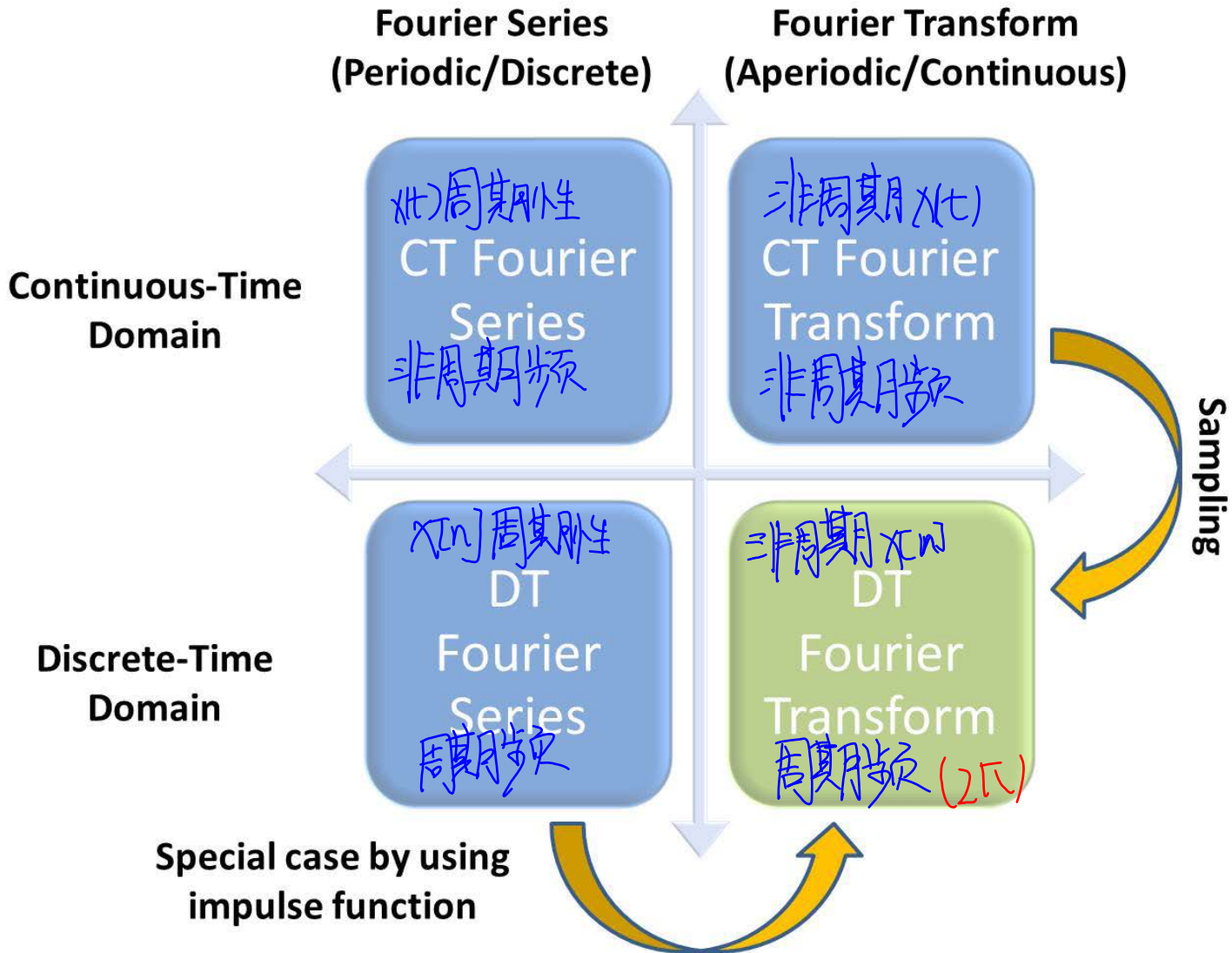
● 5.1, 5.3, 5.4, 5.41

# **Chapter 5**

## **The Discrete-Time Fourier Transform**

# Frequency domain

Time domain



# Thinking...

- How to move from **CT Fourier series** to **CT Fourier transform**?

# Discrete-Time Fourier Transform (DTFT)

**DT Fourier Series Pair**  $\left(\omega_o = \frac{2\pi}{N}\right)$

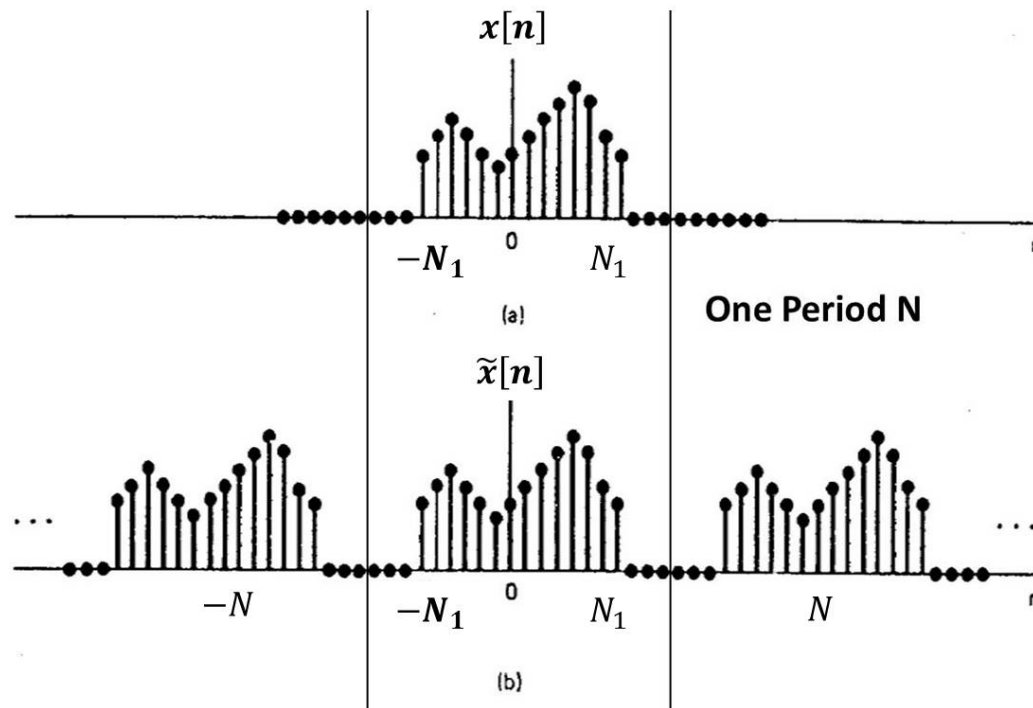
$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_o n} \quad (\text{Synthesis equation})$$

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk\omega_o n} \quad (\text{Analysis equation})$$

Aperiodic signals can be treated as periodic signals with period  $N \rightarrow \infty$

- ▶  $x[n]$  must be like  $\sum_k b_k e^{jk(2\pi/N)n}$  or  $\int_{\omega} b(\omega) e^{j\omega n} d\omega$
- ▶  $b_k$  or  $b(\omega)$  can be calculated from  $x[n]$

# DTFT Derivation (1/3)



Original signal:  $x[n]$

Define new periodic signal with period  $N$ :  $\tilde{x}[n]$ , such that

$$\tilde{x}[n] = x[n], \quad n = -N/2, \dots, N/2 - 1$$

Notice: when  $N \rightarrow \infty$ ,  $\tilde{x}[n]$  becomes  $x[n]$

## DTFT Derivation (2/3)

- Look at the Fourier series of  $\tilde{x}[n]$ :

$$a_k = \frac{1}{N} \sum_{n=-N/2}^{N/2+1} \tilde{x}[n] e^{-jk(2\pi/N)n} = \frac{1}{N} \sum_{n=-\infty}^{\infty} x[n] e^{-jk(2\pi/N)n}$$

# DTFT Derivation (3/3)

- Therefore, we get the discrete-time Fourier transform pair

## Discrete-Time Fourier Transform

$$\text{Synthesis Equation: } x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$\text{Analysis Equation: } X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

- 1) continuous spectrum: Similar to CFT
- 2) Periodic with period  $2\pi$ : Different from CFT
- 3) Low frequency: close to 0 and  $2\pi$ ;  
high frequency: close to  $\pi$



# Periodicity Properties of DT Complex Exponentials

- $x[n]$  - periodic with fundamental period  $N$ , fundamental frequency

$$x[n + N] = x[n] \quad \text{and} \quad \omega_o = \frac{2\pi}{N}$$

$$n = \dots, -1, 0, 1, 2, 3, \dots$$

- For DT complex exponentials, signals are periodic only when  $\omega_o N = k \cdot 2\pi$ ,  $k = 0, \pm 1, \pm 2, \dots$

$$e^{j\omega_o n} = e^{j\omega_o (n+N)} \rightarrow e^{j\omega_o N} = 1 \rightarrow \omega_o N = k \cdot 2\pi$$

- For DT complex exponentials, signals with frequencies  $\omega_o$  and  $\omega_o + k \cdot 2\pi$  are identical. 
$$e^{j(\omega_o + k \cdot 2\pi)n} = e^{j\omega_o n} \cdot e^{jk \cdot 2\pi n} = e^{j\omega_o n}$$
  - We need only consider a frequency interval of length  $2\pi$ , and in most cases, we use the interval:  $0 \leq \omega_o < 2\pi$ , or  $-\pi \leq \omega_o < \pi$

## Cont.

-  $e^{j\omega_0 n}$  does ***not*** have a continually increasing rate of oscillation as  $\omega_0$  is increased in magnitude.

**low-frequency** (slowly varying):  $\omega_0$  near  $0, 2\pi, \dots$ , or  $2k \cdot \pi$

**high-frequency** (rapid variation):  $\omega_0$  near  $\pm \pi, \dots$ , or  $(2k+1) \cdot \pi$

$$e^{j(2k+1)\pi n} = e^{j\pi n} = (e^{j\pi})^n = (-1)^n$$

对比不同的地方,并解释原因

## Discrete-Time Fourier Transform

$$\text{Synthesis Equation: } x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$\text{Analysis Equation: } X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

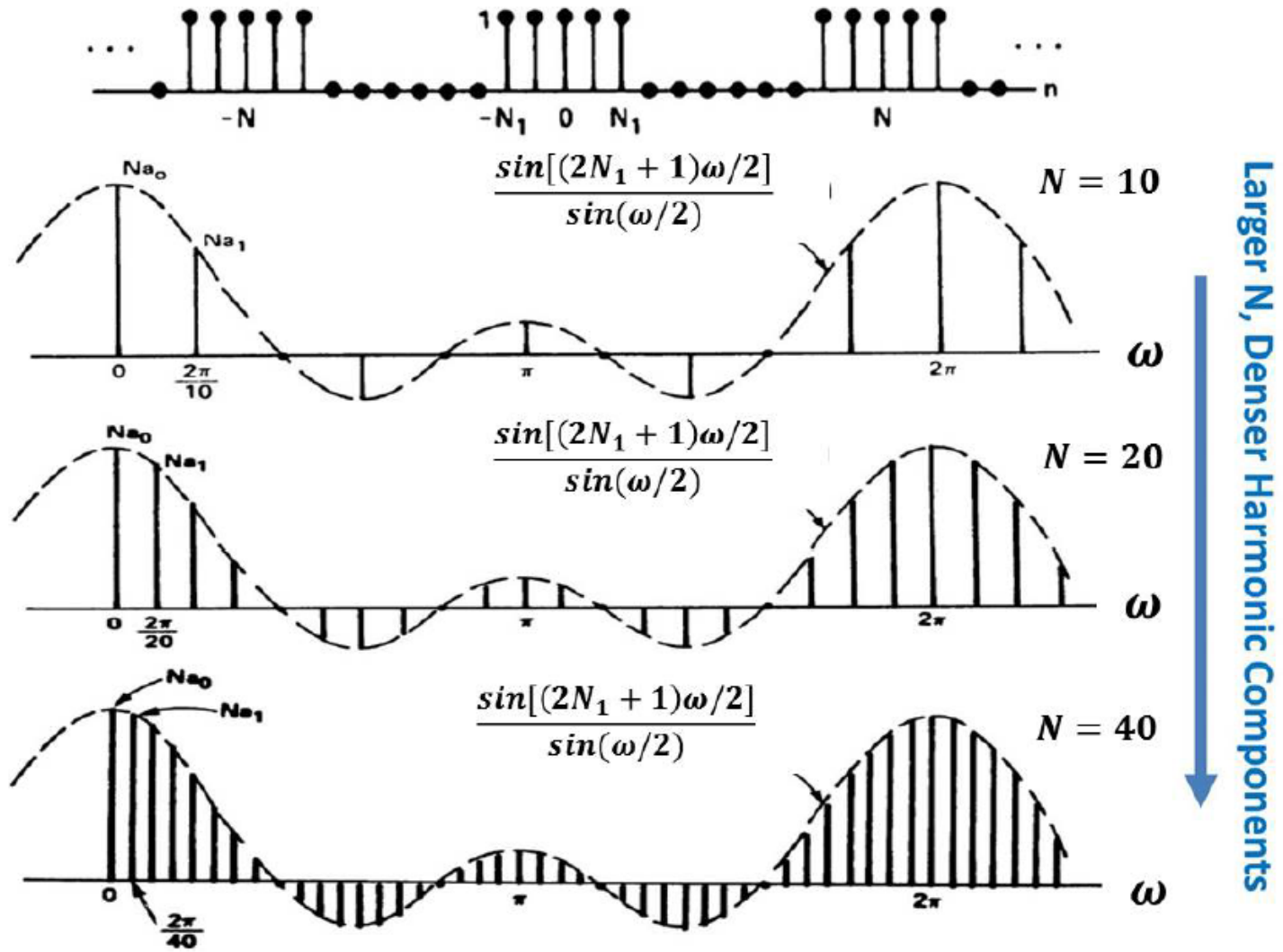
## DT Fourier Series Pair $\left(\omega_o = \frac{2\pi}{N}\right)$

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_o n} \quad (\text{Synthesis equation})$$

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk\omega_o n} \quad (\text{Analysis equation})$$

\* \* \*  $k = \langle N \rangle$

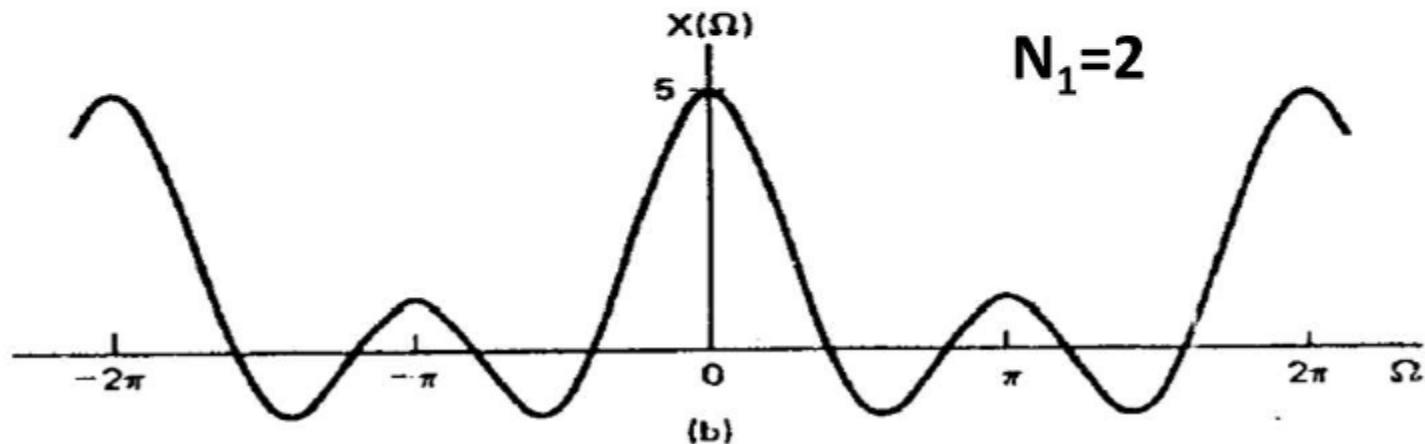
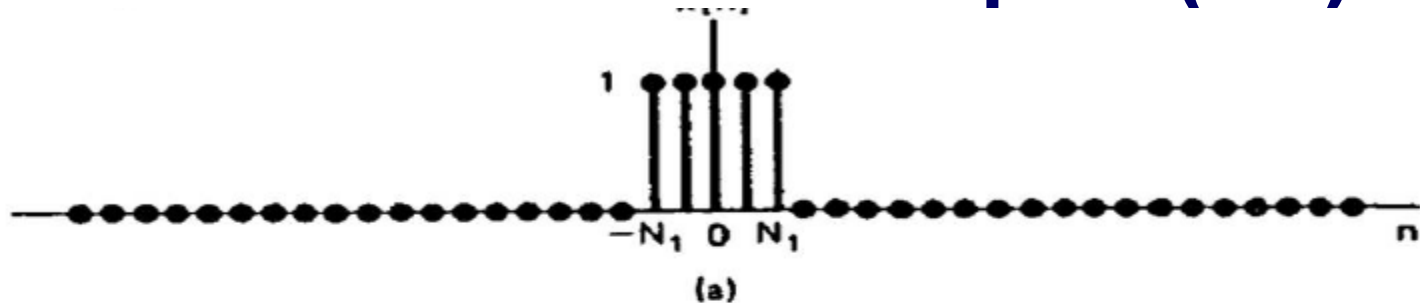
# Example: From Periodic To Aperiodic



$$X[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} X[n] e^{-j\omega n}$$

## Fourier Transform Examples (1/2)

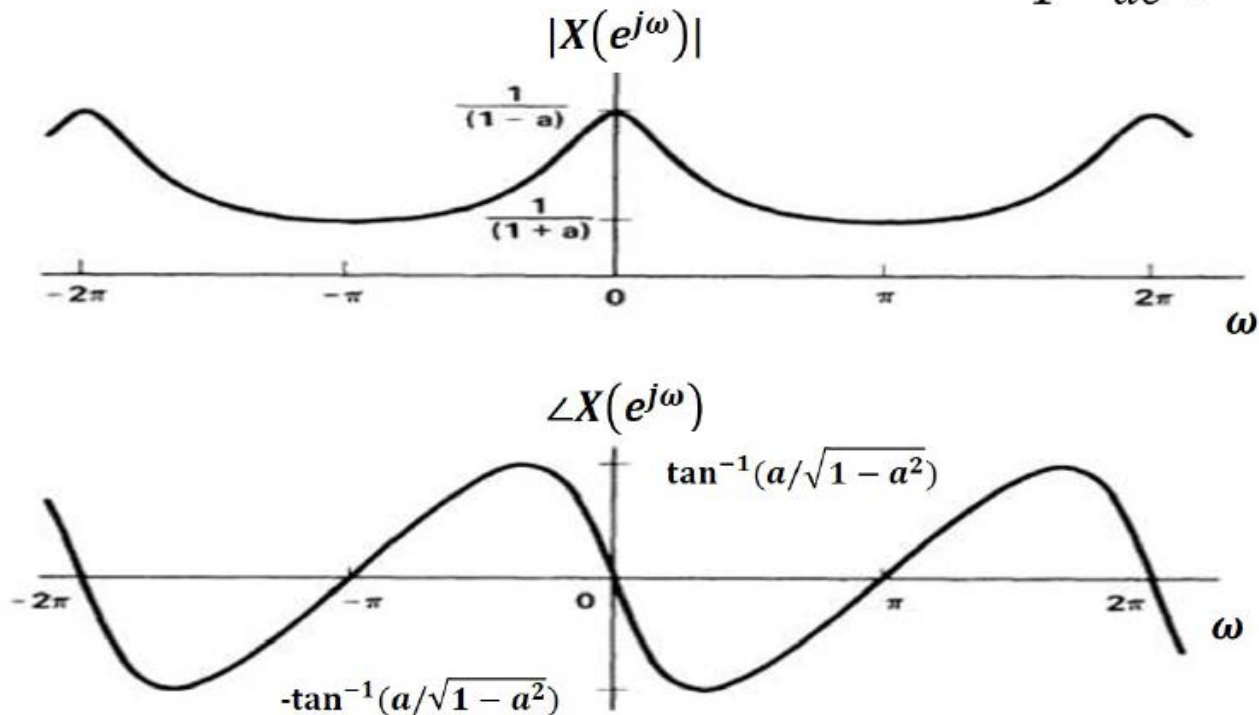


- $x[n] = 1$  ( $n = -N_1, \dots, 0, \dots, N_1$ )
- $X(e^{j\omega}) = \frac{\sin \omega(N_1+1/2)}{\sin(\omega/2)}$
- Width of  $x[n]$ :  $W_t = 2N_1 + 1$ ; width of  $X(e^{j\omega})$ :  $W_f = \frac{4\pi}{2N_1+1}$
- $W_t \times W_f = 4\pi$ , which is a constant



## Fourier Transform Examples (2/2)

$$x[n] = a^n u[n] \quad 0 < a < 1 \quad \Leftrightarrow \quad X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}}$$



- See textbook, Example 5.1
- What's the shape of magnitude when  $a \rightarrow 1$  or  $a \rightarrow 0$ ?

# Convergence Issue of Analysis Equation

## Sufficient Condition of Convergence

*The analysis equation  $X(e^{j\omega}) = \sum_{-\infty}^{\infty} x[n]e^{-j\omega n}$  will converge either if  $x[n]$  is absolutely summable or if the sequence has finite energy, thus,*

$$\sum_{-\infty}^{\infty} |x[n]| < \infty \text{ or } \sum_{-\infty}^{\infty} |x[n]|^2 < \infty$$

## Cont.

- Do the following signals have Fourier transform:

- ▶  $a^n u[n]$  ( $0 < a < 1$ )

- ▶  $\delta[n]$   $\leftrightarrow 1$

- ▶  $u[n]$   $\nexists$

- ▶  $e^{j\frac{2}{5}\pi n}$ ,  $\cos(\frac{2}{5}\pi n)$

- ▶  $a^n u[n]$  ( $a > 1$ )



# Can Periodic Signals Have DTFT?

- Definition of DTFT:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

- Justification of divergence:

- ▶ Let  $\omega = 2k\pi$ , we have  $X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]$

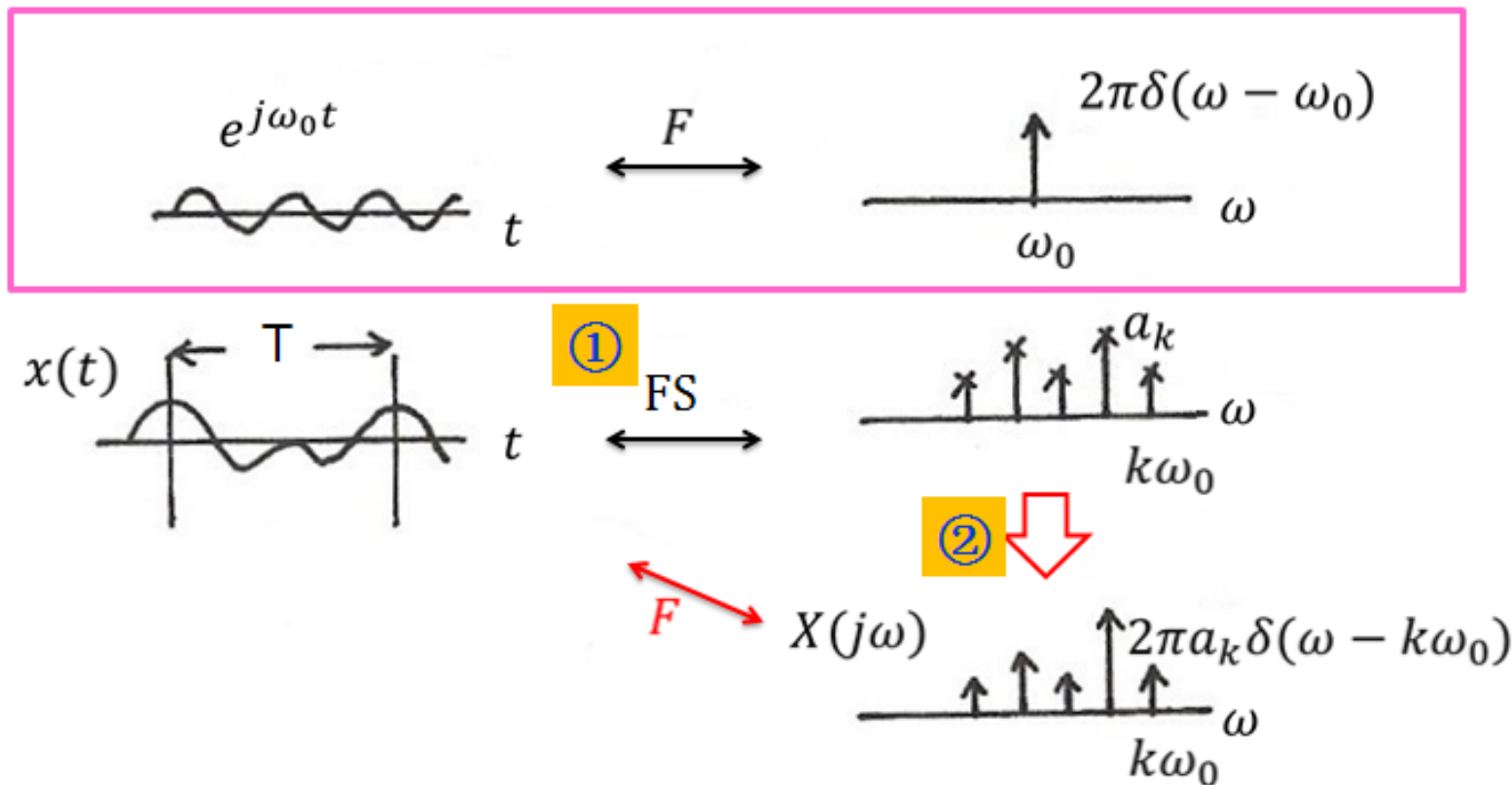
- ▶ Since  $x[n]$  is periodic, the summation  $\sum_{n=-\infty}^{\infty} x[n]$  will never converge unless  $x[n] = 0$

- **Conclusion:** Most of periodic signals do NOT have DTFT according to the definition
- However, it's of significant engineering importance to extend Fourier transform to periodic signals

# Can Periodic Signals Have DTFT?

## Review

### Fourier Transform for **Periodic Signals** – Unified Framework



# DTFT with Periodic Signals (1/2)

## Fourier Transform of $e^{j\omega n}$

*The following transform pair is actually NOT rigorously defined:*

$$x[n] = e^{j\omega_0 n} \longleftrightarrow X(e^{j\omega}) = \sum_{l=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 - 2\pi l)$$

- Synthesis:

$$\frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega = \frac{1}{2\pi} \int_{2\pi} \sum_{l=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 - 2\pi l) e^{j\omega n} d\omega = e^{j\omega_0 n}$$

- Analysis:

$$\sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} = \sum_{n=-\infty}^{\infty} e^{j(\omega_0 - \omega)n} \quad \text{converge??}$$

## DTFT with Periodic Signals (2/2)

- According to the Fourier series, for a periodic signal with period  $N$ :

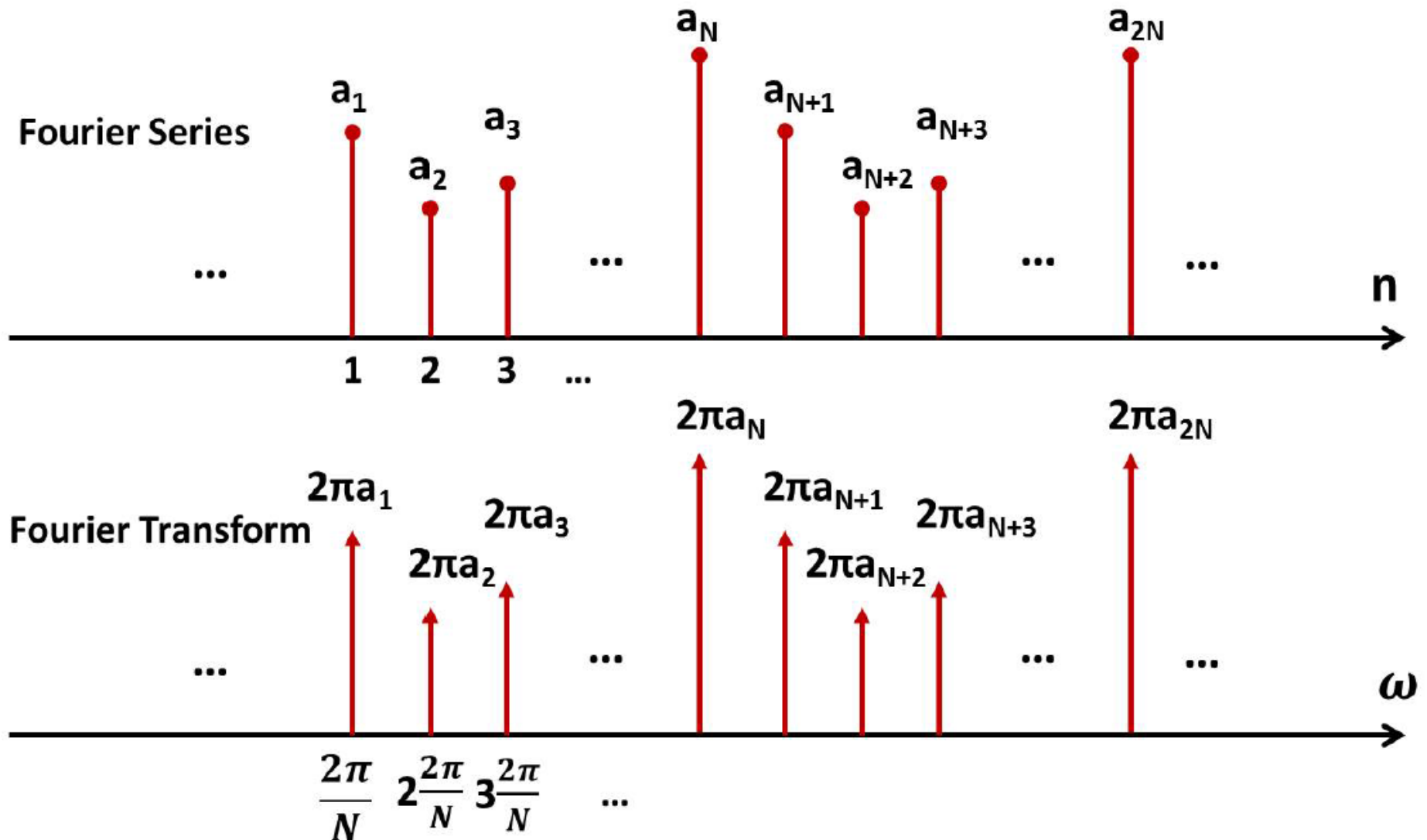
$$x[n] = a_0 + a_1 e^{j(2\pi/N)n} + \dots + a_k e^{jk(2\pi/N)n} + \dots + a_{N-1} e^{j(N-1)(2\pi/N)n}$$

- $e^{jk(2\pi/N)n} \longleftrightarrow \sum_{l=-\infty}^{\infty} 2\pi \delta(\omega - k(2\pi/N) - 2\pi l)$
- Then, due to the linearity of Fourier transform

$$\begin{aligned} \mathcal{F}\{x[n]\} &= \sum_{k=0}^{N-1} a_k \mathcal{F}\{e^{jk(2\pi/N)n}\} = \sum_{k=0}^{N-1} a_k \sum_{l=-\infty}^{\infty} 2\pi \delta(\omega - k(2\pi/N) - 2\pi l) \\ &= \sum_{l=-\infty}^{\infty} \sum_{k=0}^{N-1} a_k 2\pi \delta(\omega - k(2\pi/N) - 2\pi l) \end{aligned}$$

- Fourier transform of a periodic signal is a periodic sequence of impulses
  - ▶ What's the period? How many impulses within one period?

# Fourier Series v.s. Fourier Transform



# Example: Discrete-Time Impulse Chain

- What's the Fourier transform of  $x[n] = \sum_{k=-\infty}^{+\infty} \delta[n - kN]$  ?
- First of all, we calculate the Fourier series:

$$\begin{aligned} a_k &= \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk \frac{2\pi}{N} n} \\ &= \frac{1}{N} \sum_{n=\langle N \rangle} \sum_{k=-\infty}^{+\infty} \delta[n - kN] e^{-jk \frac{2\pi}{N} n} \\ &= \frac{1}{N} \sum_{n=\langle N \rangle} \delta[n] e^{-jk \frac{2\pi}{N} n} = \frac{1}{N} \end{aligned}$$

# Frequency domain

Time domain

