

- Homework: 8.4, 8.25, 8.29

- Tutorial: 8.34, 8.35, 8.38

review =  $H_D(e^{j\omega}) \rightarrow H_D(e^{j\omega T}) \rightarrow \tilde{H}_D(e^{j\omega T}) = H_C(j\omega)$

Final Exam  $\rightarrow$

No cheating paper

2023

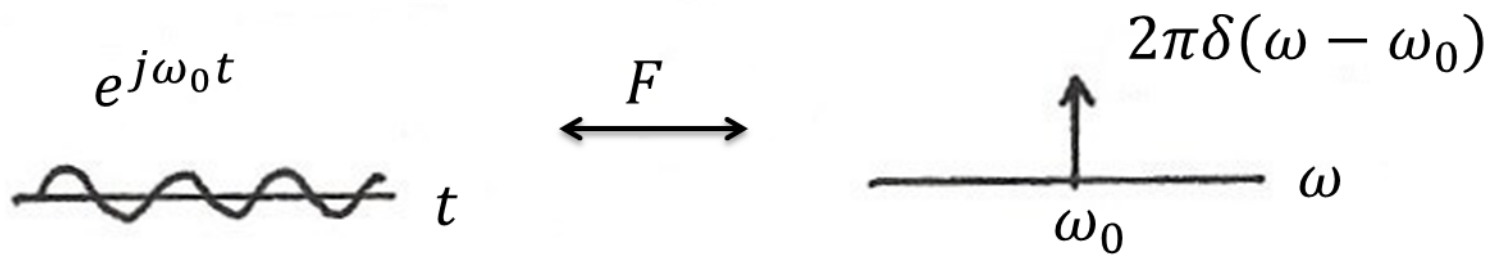


南方科技大学 SUSTC

# Chapter 8

# Communication Systems

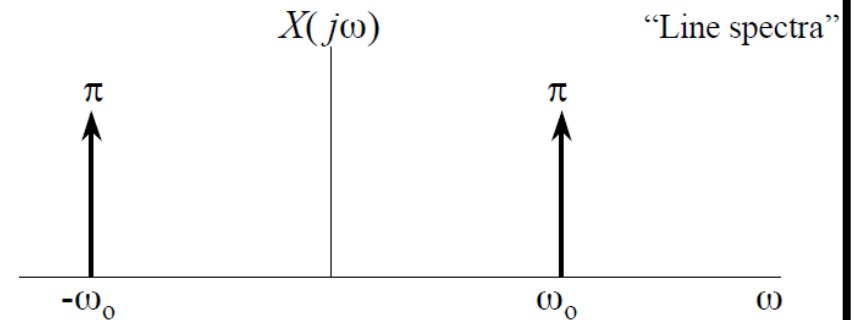
(通信系统)



$$x(t) = \cos \omega_0 t = \frac{1}{2} e^{j\omega_0 t} + \frac{1}{2} e^{-j\omega_0 t}$$

$$\Updownarrow$$

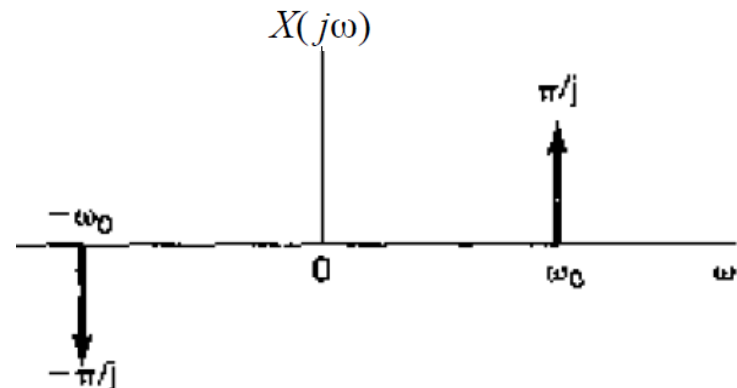
$$X(j\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0)$$



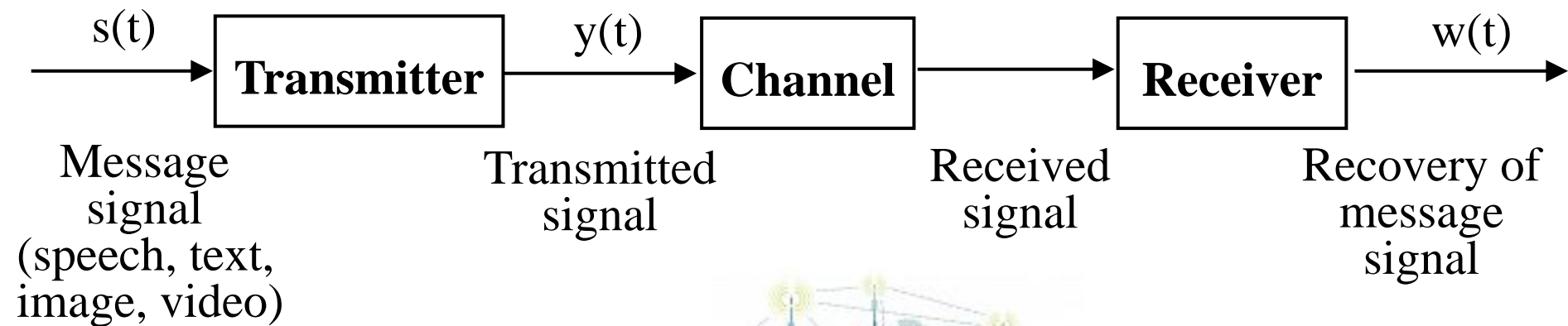
$$x(t) = \sin(\omega_0 t) = \frac{1}{2j} e^{j\omega_0 t} - \frac{1}{2j} e^{-j\omega_0 t}$$

$$\Updownarrow$$

$$X(j\omega) = \frac{\pi}{j} \delta(\omega - \omega_0) - \frac{\pi}{j} \delta(\omega + \omega_0)$$

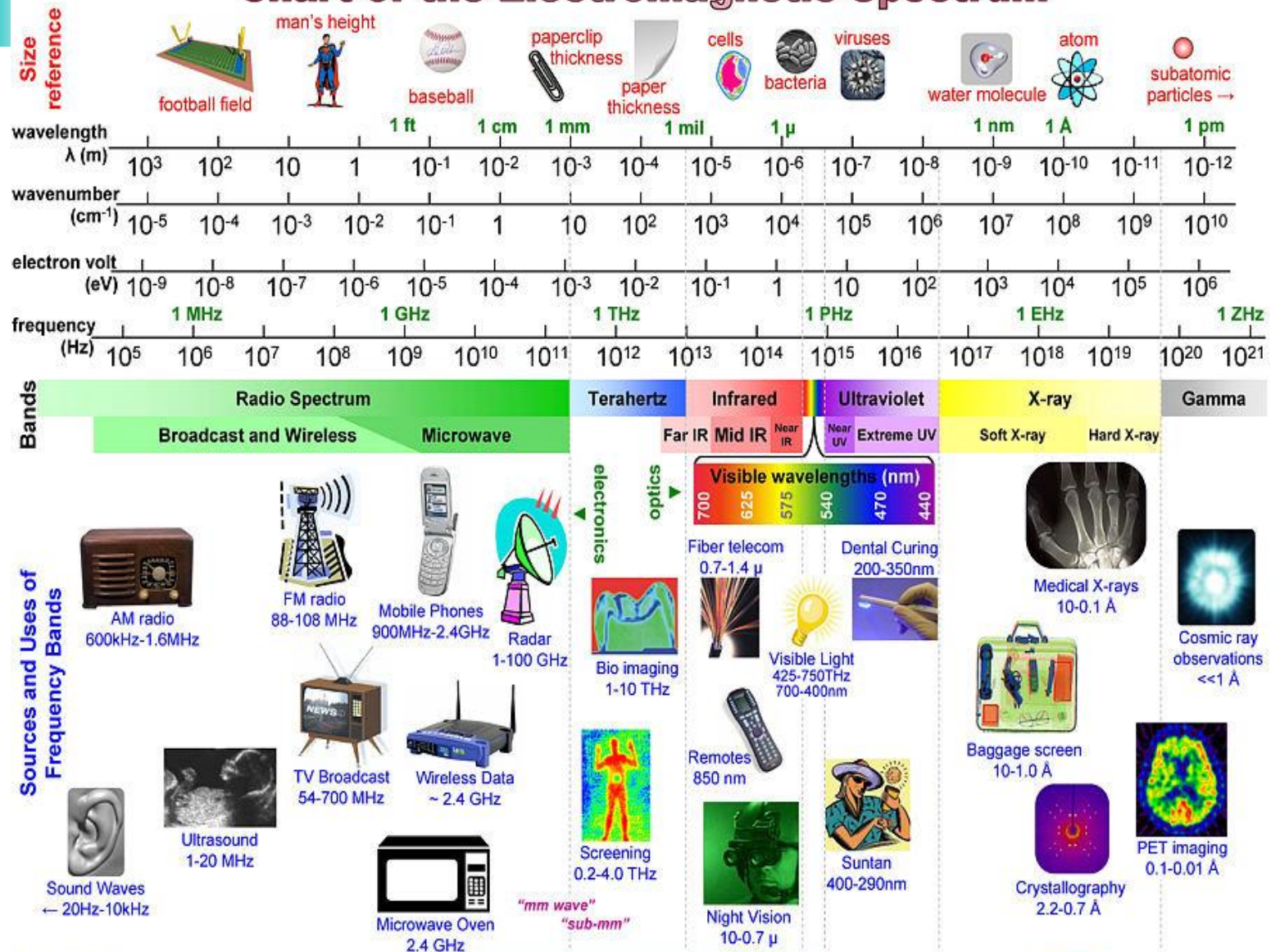


# Communication Systems



# Channel (media) Example

## Chart of the Electromagnetic Spectrum



# Amplitude modulation 调幅

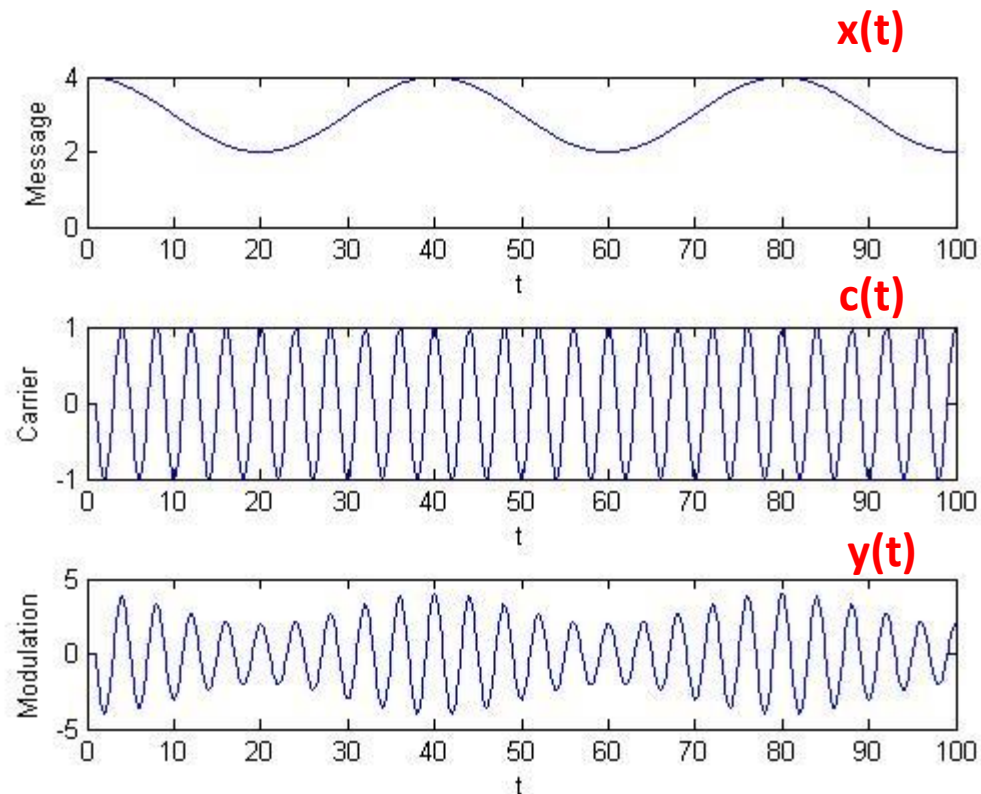
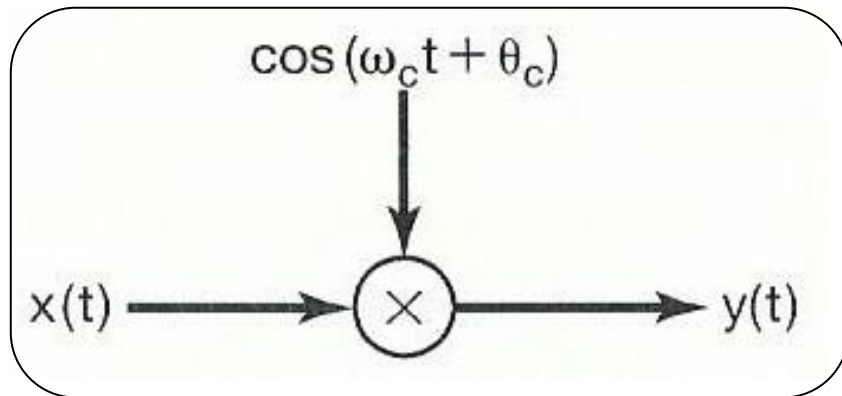
- Modulation: the general process to embed an information-bearing signal into a second signal.

$$y(t) = x(t)c(t)$$

$x(t)$ : **m**odulating (or message) signal

$c(t)$ : **c**arrier signal 载波信号

$y(t)$ : modulated signal



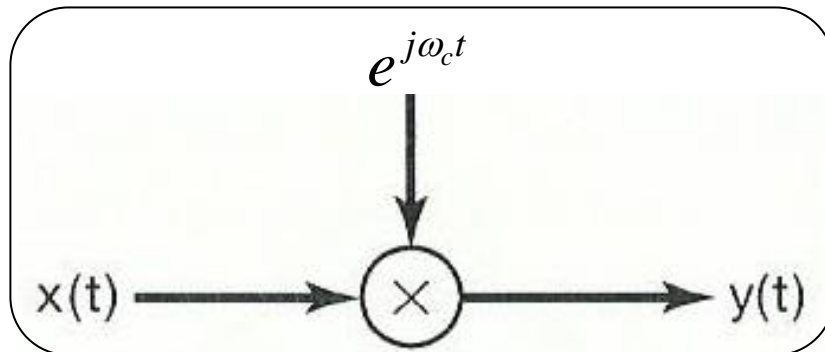
# Amplitude modulation with a complex exponential carrier

$$c(t) = e^{j(\omega_c t + \theta_c)}$$

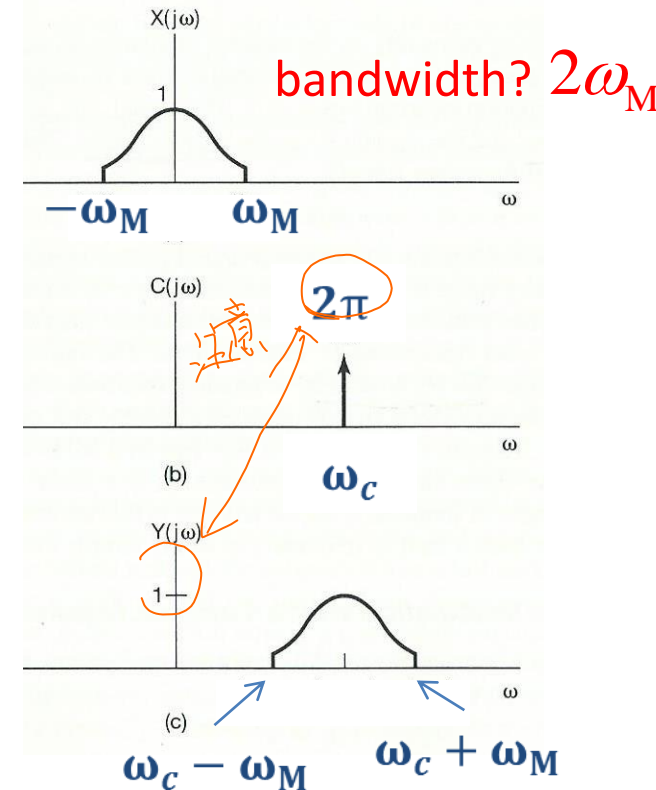
$\omega_c$  : carrier frequency

First, we suppose  $\theta_c = 0$

$$y(t) = x(t)c(t) = x(t)e^{j\omega_c t}$$



$\omega_M$  : the highest frequency in  $x(t)$

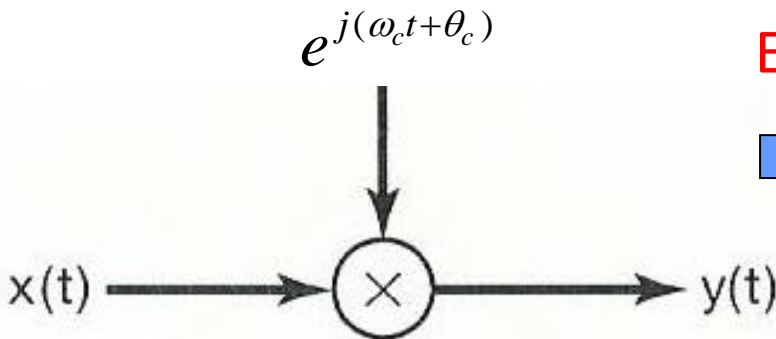


- The spectrum of the modulated output  $y(t)$  is that of modulating input  $x(t)$ , shifted in frequency by amount of carrier frequency  $\omega_c$ .
- How to recover the modulating input  $x(t)$  from modulated signal  $y(t)$ ?

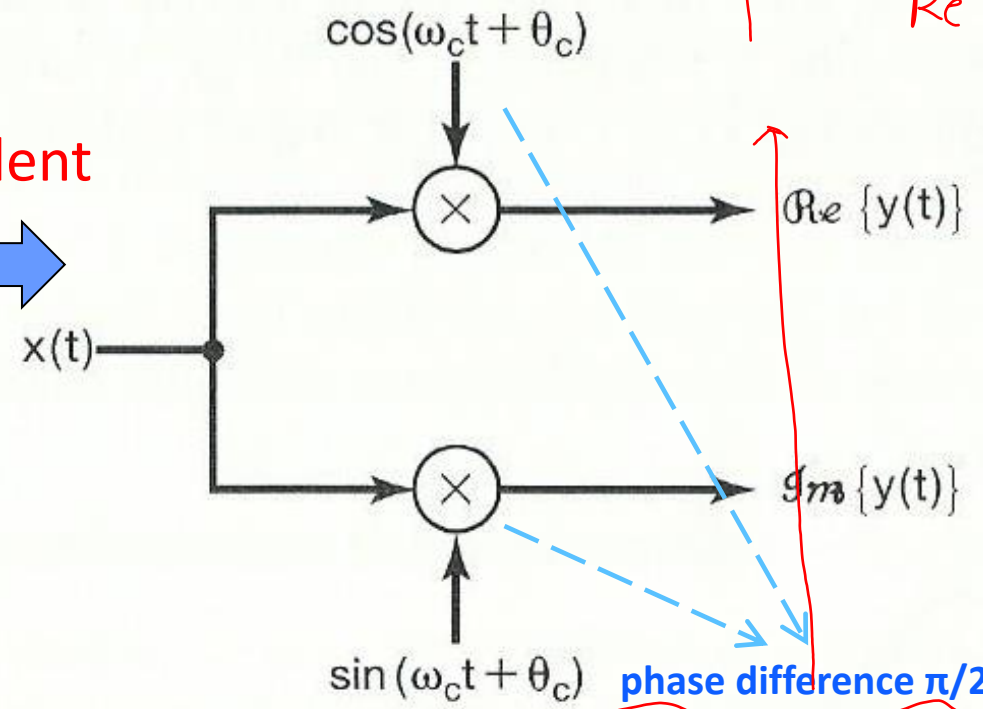
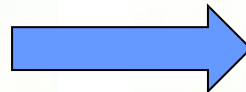


# Amplitude modulation with a complex exponential carrier (cont.)

$$y(t) = x(t)e^{j\omega_c t} = x(t)\cos(\omega_c t) + jx(t)\sin(\omega_c t)$$



Equivalent

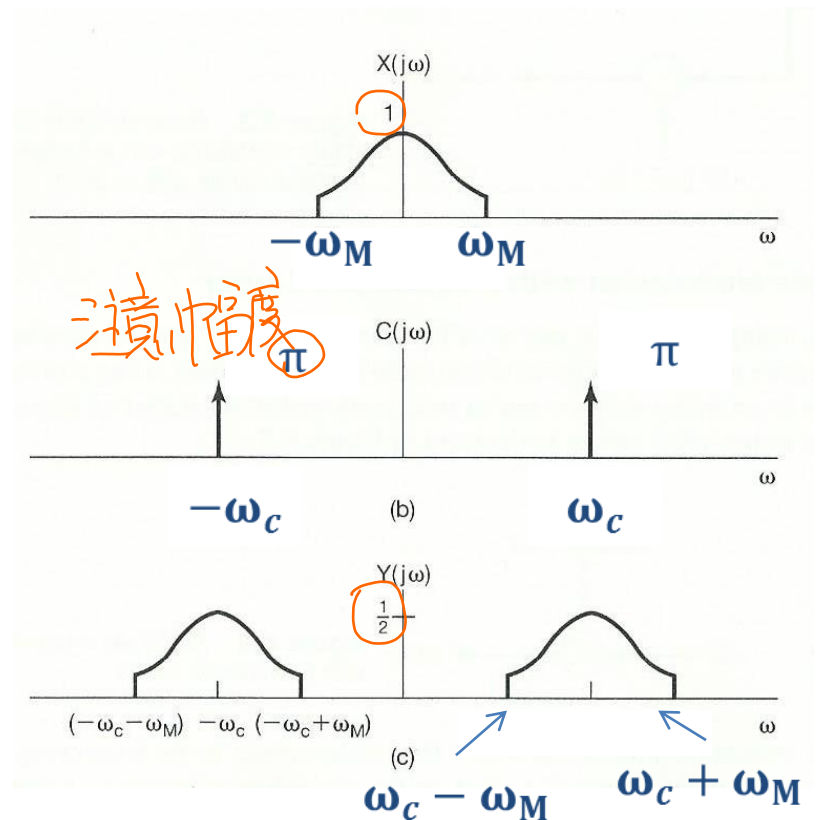
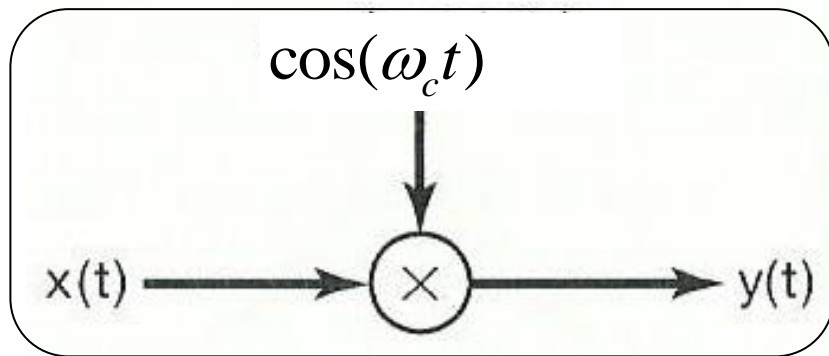




# Amplitude modulation with a sinusoidal carrier

First, we suppose  $\theta_c = 0$

$$y(t) = x(t)c(t) = x(t)\cos(\omega_c t)$$



What is required if  $x(t)$  is recoverable from  $y(t)$ ?

**Condition:  $\omega_c > \omega_M$**

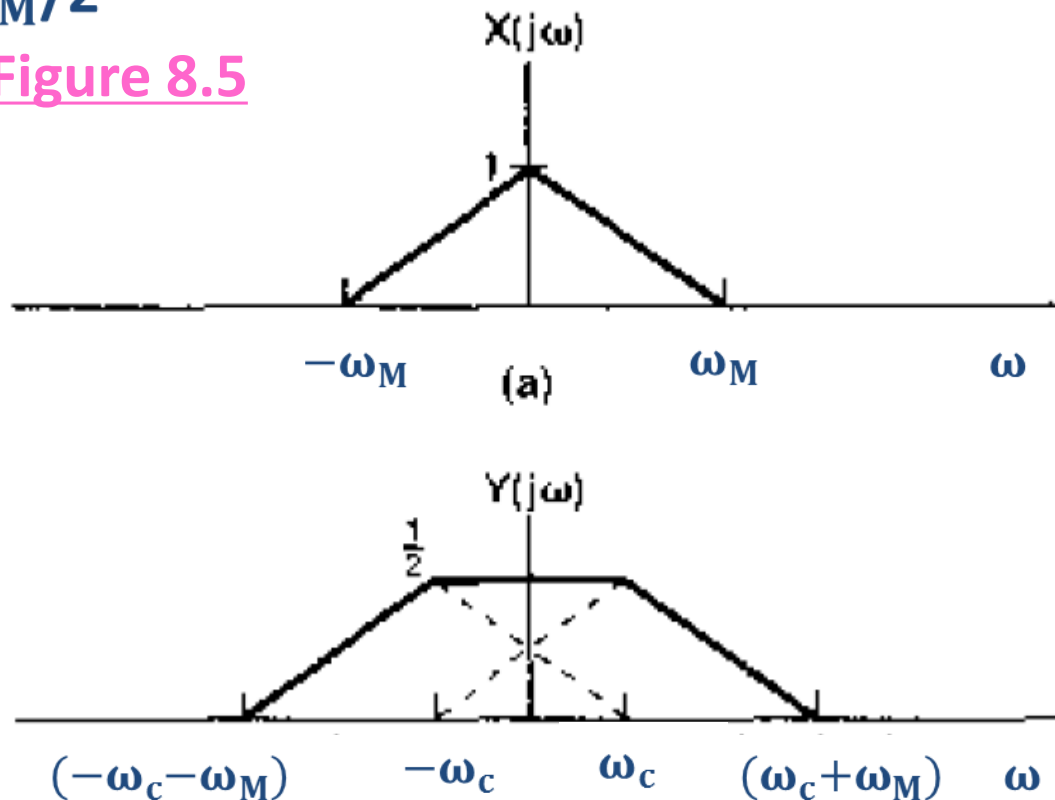
# Amplitude modulation with a sinusoidal carrier (cont.)

What happens if  $\omega_c < \omega_M$ ?

Example of an overlapping between the two replications of  $X(j\omega)$

$$\omega_c = \omega_M/2$$

Figure 8.5

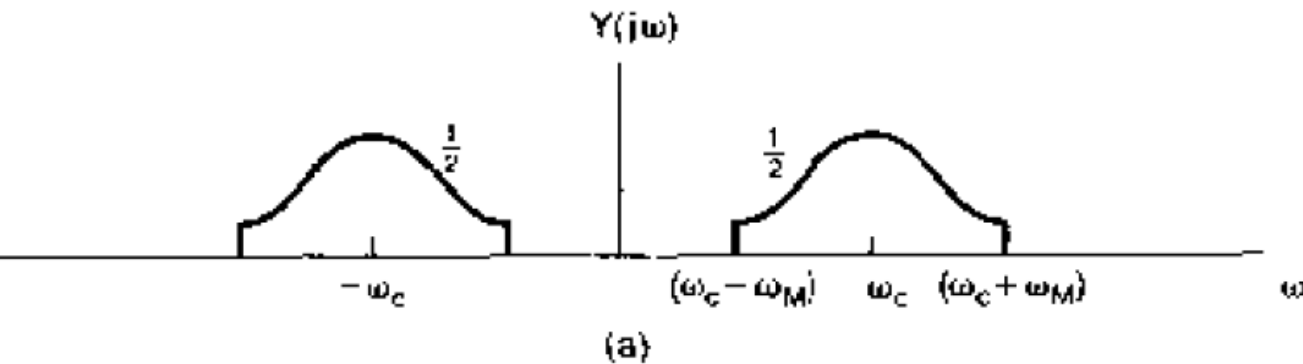


# Demodulation for sinusoidal AM

- Purpose: to **recover** the information-bearing signal  $x(t)$  from the modulated signal  $y(t)$ .
- Two types of AM demodulation:
  - ◆ Synchronous demodulation, in which transmitter and receiver are synchronized in phase and frequency
  - ◆ Asynchronous demodulation

# Synchronous demodulation

With sinusoidal carrier

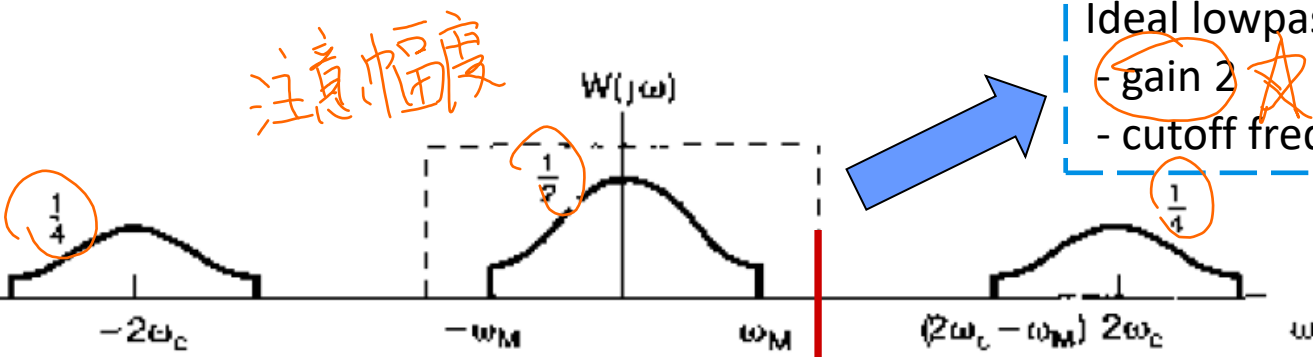


$$y(t) = x(t) \cos(\omega_c t)$$



$$w(t) = y(t) \cos(\omega_c t)$$

$w(t)$  : demodulated signal



Ideal lowpass filtering

- gain 2

- cutoff frequency  $> \omega_M$  and  $< 2\omega_c - \omega_M$



# Synchronous demodulation (cont.)

- Mathematically,

AM modulation  $\rightarrow y(t) = x(t) \cos(\omega_c t)$

AM demodulation  $\rightarrow w(t) = y(t) \cos(\omega_c t) = x(t) \cos^2(\omega_c t)$

$$= x(t) \left[ \frac{1}{2} + \frac{1}{2} \cos(2\omega_c t) \right]$$

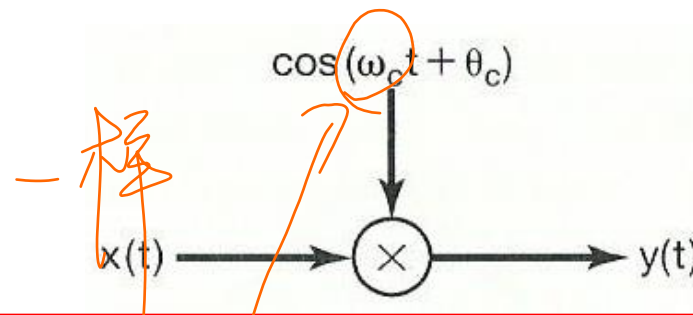
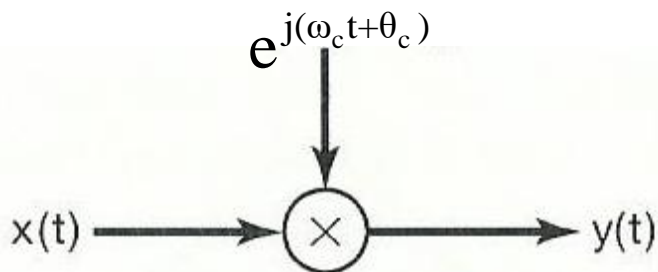
$$= \boxed{\frac{1}{2} x(t)} + \frac{1}{2} x(t) \cos(2\omega_c t)$$

# Synchronous demodulation (cont.)

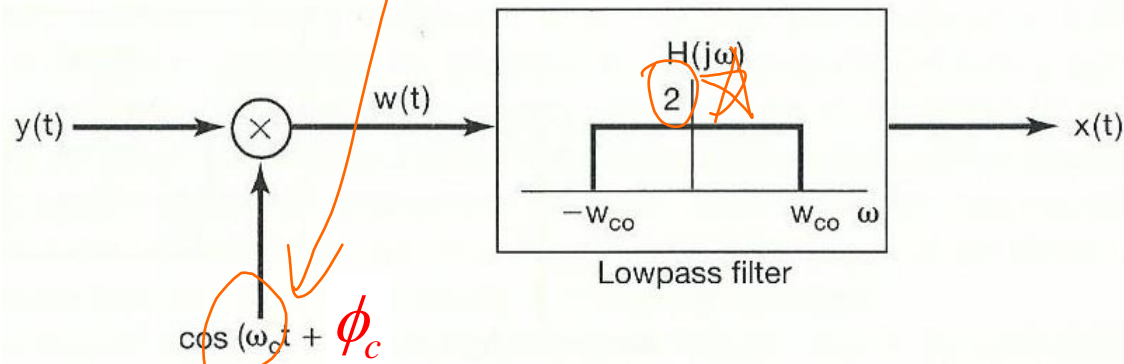
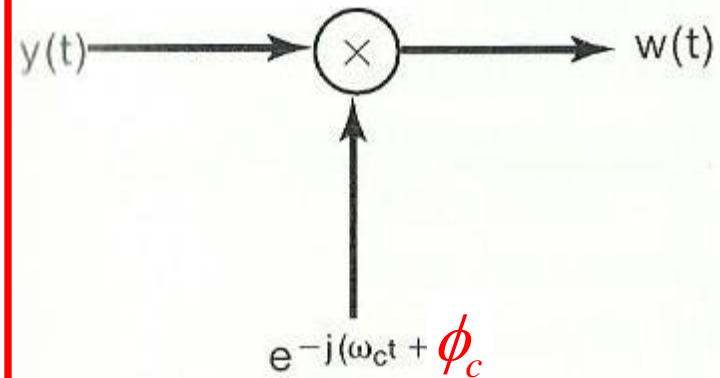
Complex exponential carrier:

Sinusoidal carrier:

AM modulation



De-modulation



$$y(t) = e^{j(\omega_c t + \theta_c)} x(t)$$

$$\begin{aligned} w(t) &= e^{-j(\omega_c t + \phi_c)} y(t) \\ &= e^{j(\theta_c - \phi_c)} x(t) \end{aligned}$$

$$w(t) = x(t) \cos(\omega_c t + \theta_c) \cos(\omega_c t + \phi_c)$$

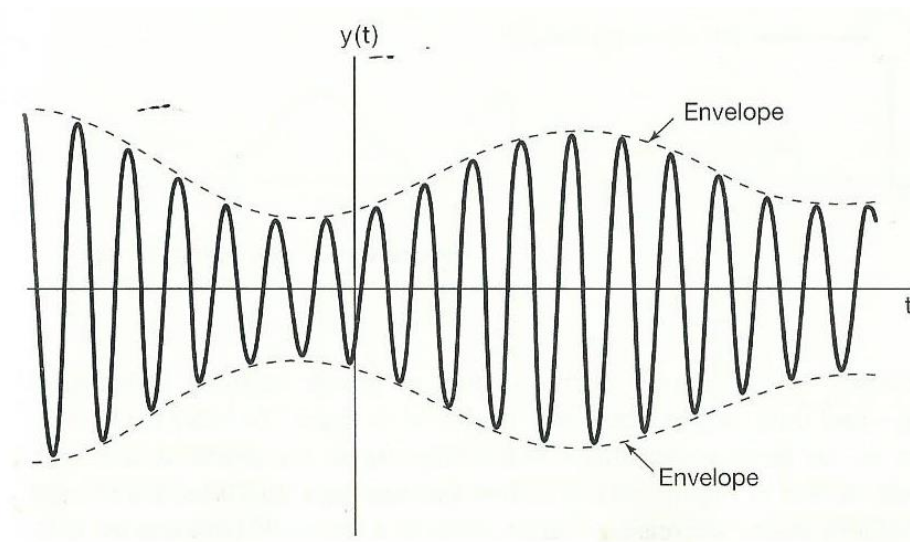
$$\left[ \frac{1}{2} \cos(\theta_c - \phi_c) x(t) \right] + \frac{1}{2} x(t) \cos(2\omega_c t + \theta_c + \phi_c)$$

What is the cost?

# Asynchronous demodulation

- Avoid the need for synchronization between modulator and demodulator.

An example of modulated signal  $y(t)$ , when  $x(t)$  is positive, and  $\omega_c > \omega_M$ .



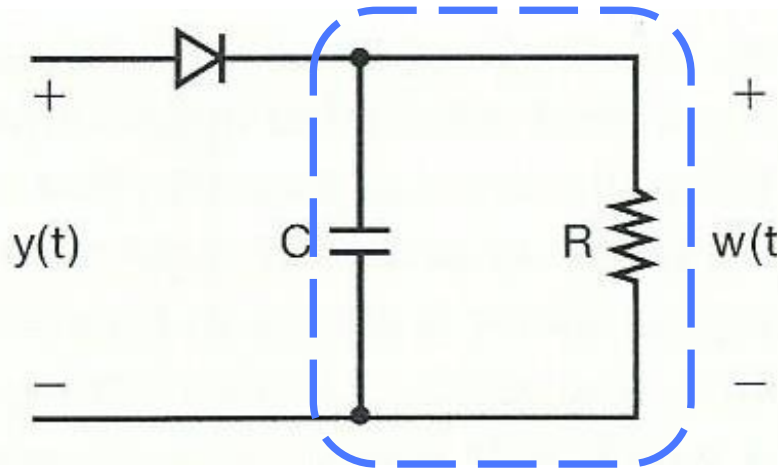
**Envelope signal**, a smooth curve connecting the peaks, approaches to the modulating signal  $x(t)$ .



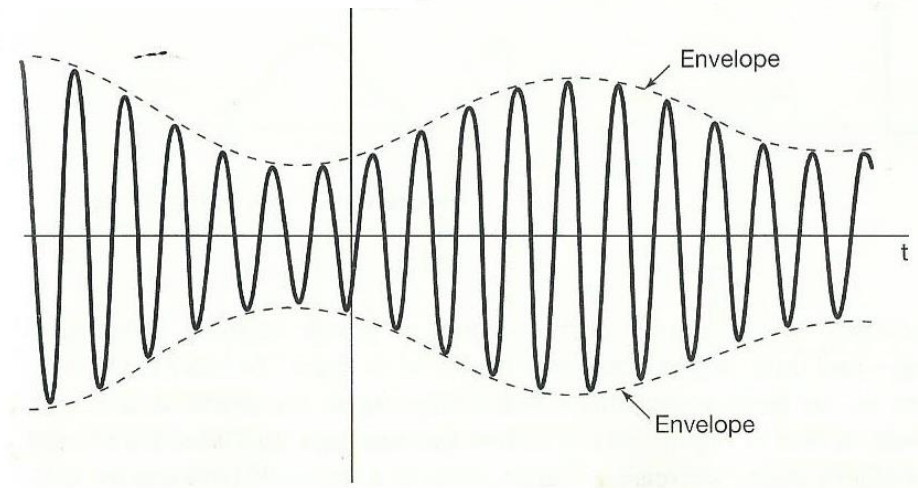
# Envelope detector

Full wave or half-wave envelope detector?

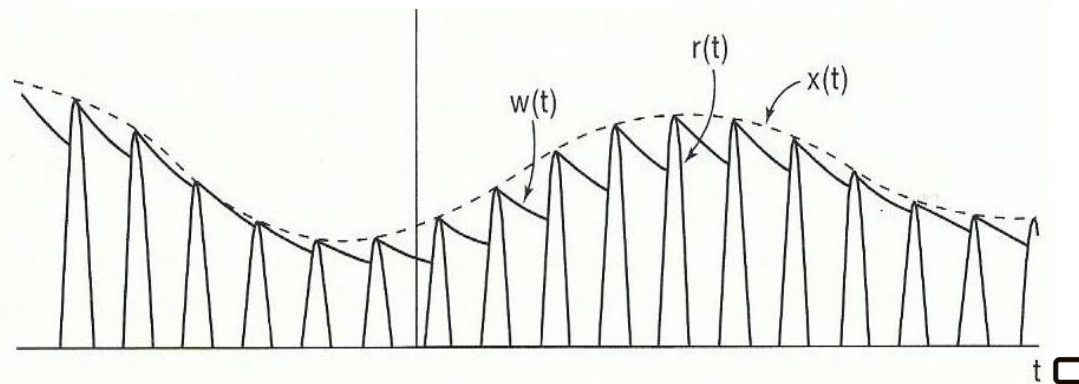
Lowpass filtering



Modulated signal



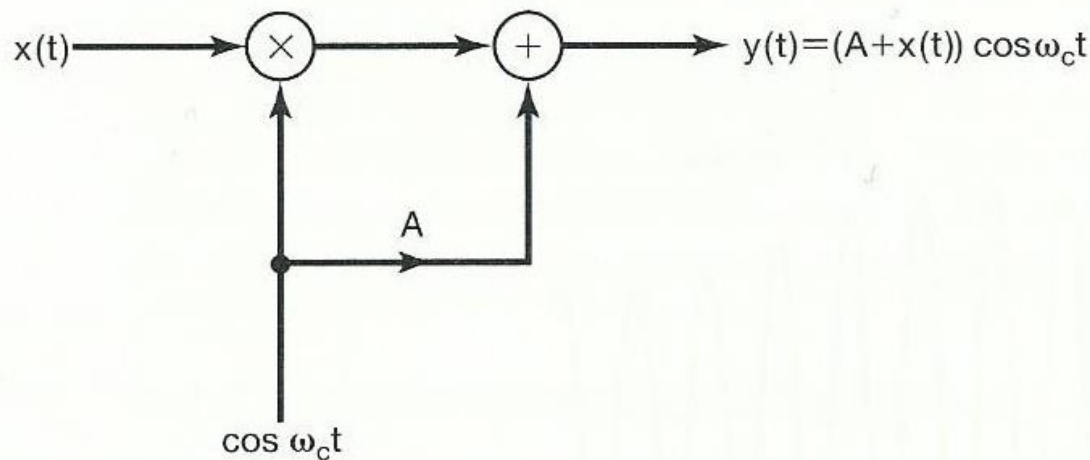
Envelope detector output



# AM in an asynchronous system

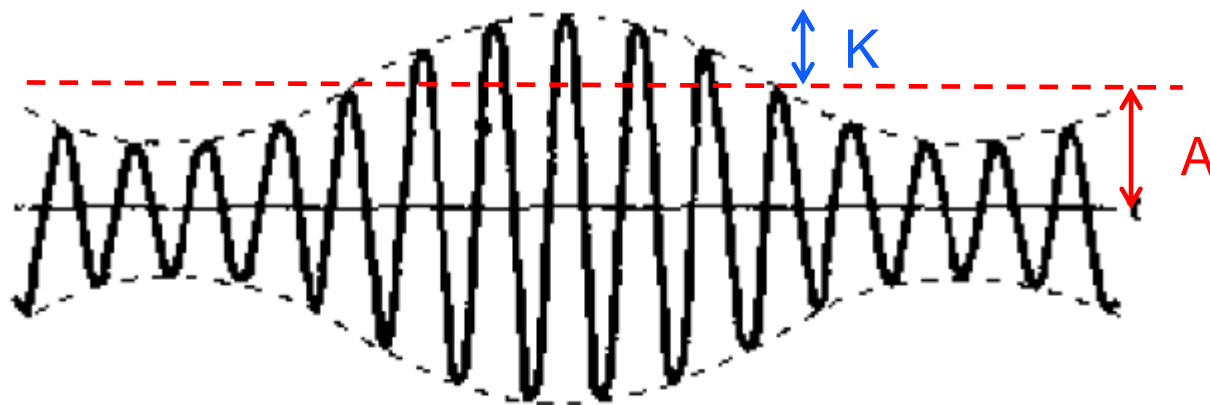
- Two important assumptions:

- ◆  $\omega_c$  is higher than  $\omega_M$
- ◆  $x(t)$  is always positive



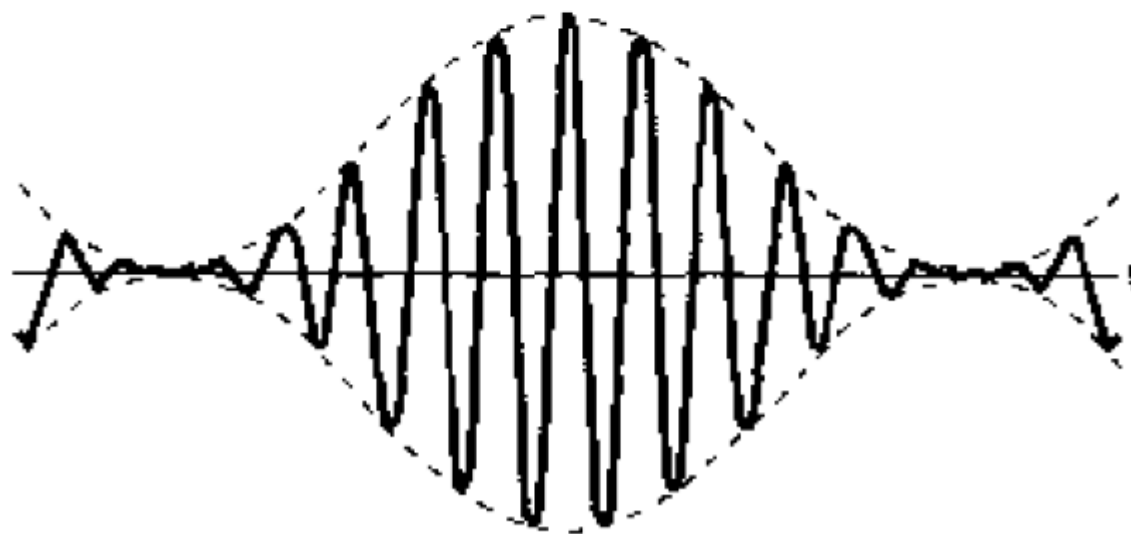
1) Add a constant  $A$  to make sure  $[A + x(t)]$  is positive. Let  $K$  be the maximum amplitude of  $x(t)$ , or  $|x(t)| \leq K$ . We need to have  $A > K$ .

2) Define  $K/A$  as **modulation index  $m$** , which is from 0 to 1.



$m=0.5$

(a)



$m=1$

$\rightarrow x(t)=0$   
What about  $m=0$ ?

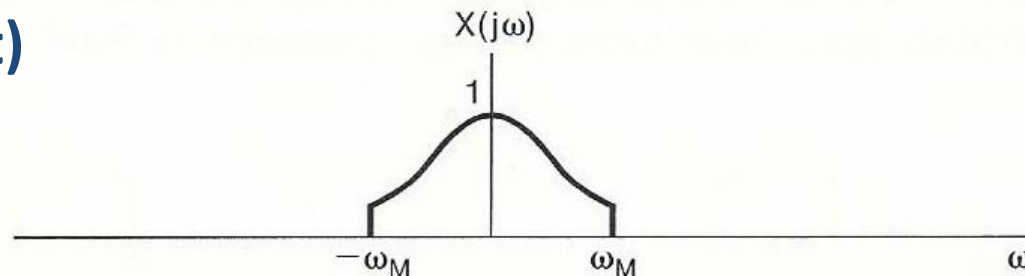
(b)



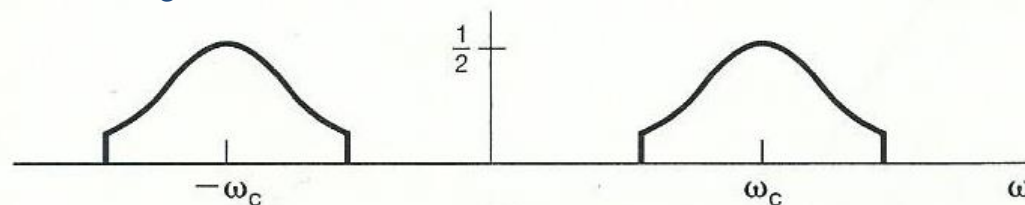
# AM in an asynchronous system (cont.)

Spectrum of modulated signal in an asynchronous system

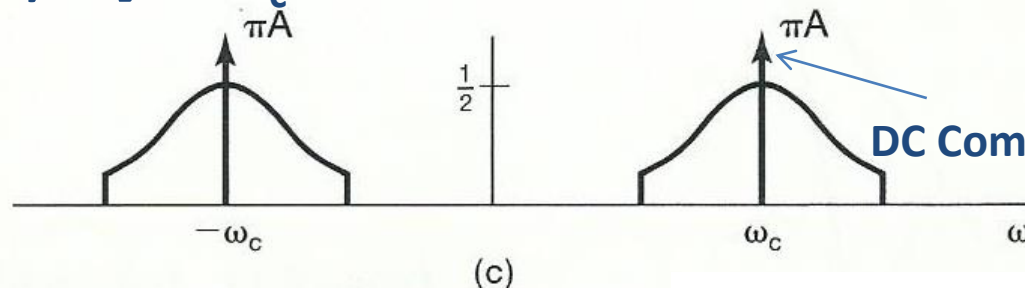
$x(t)$



$x(t)\cos\omega_c t$



$[x(t)+A]\cos\omega_c t$



DC Component/ Carrier

What is the cost ?

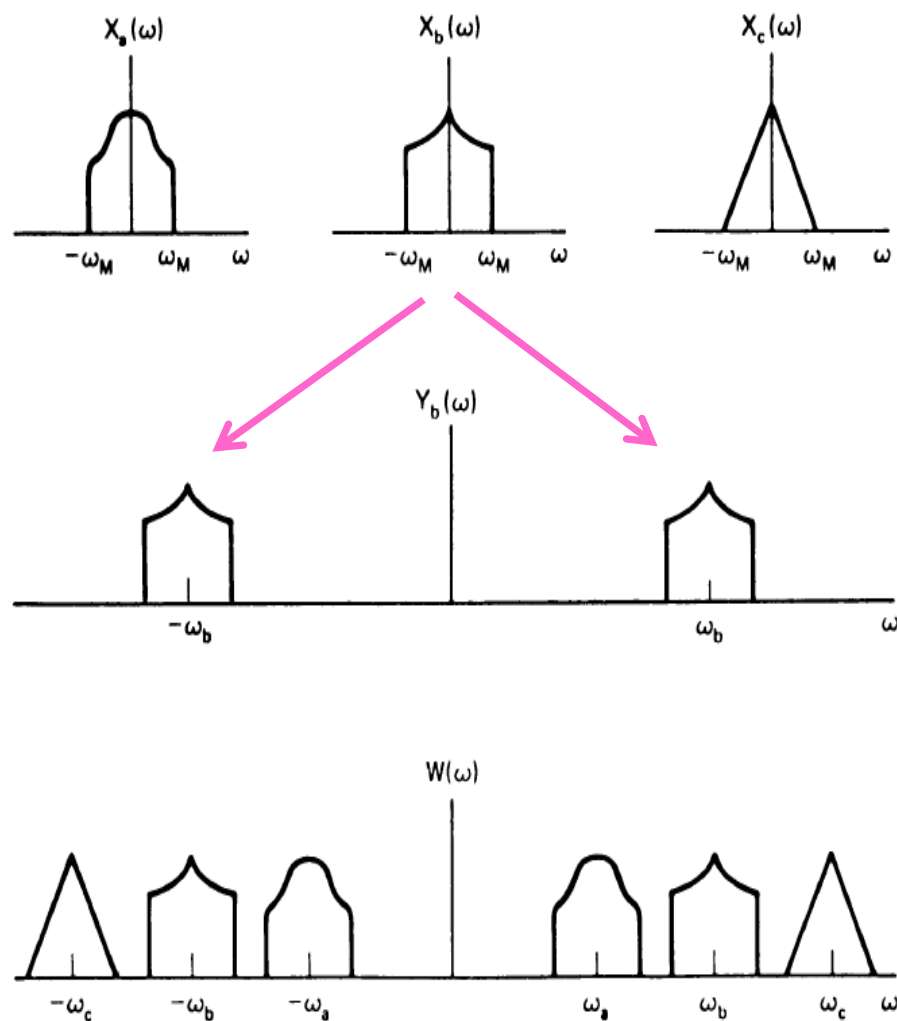
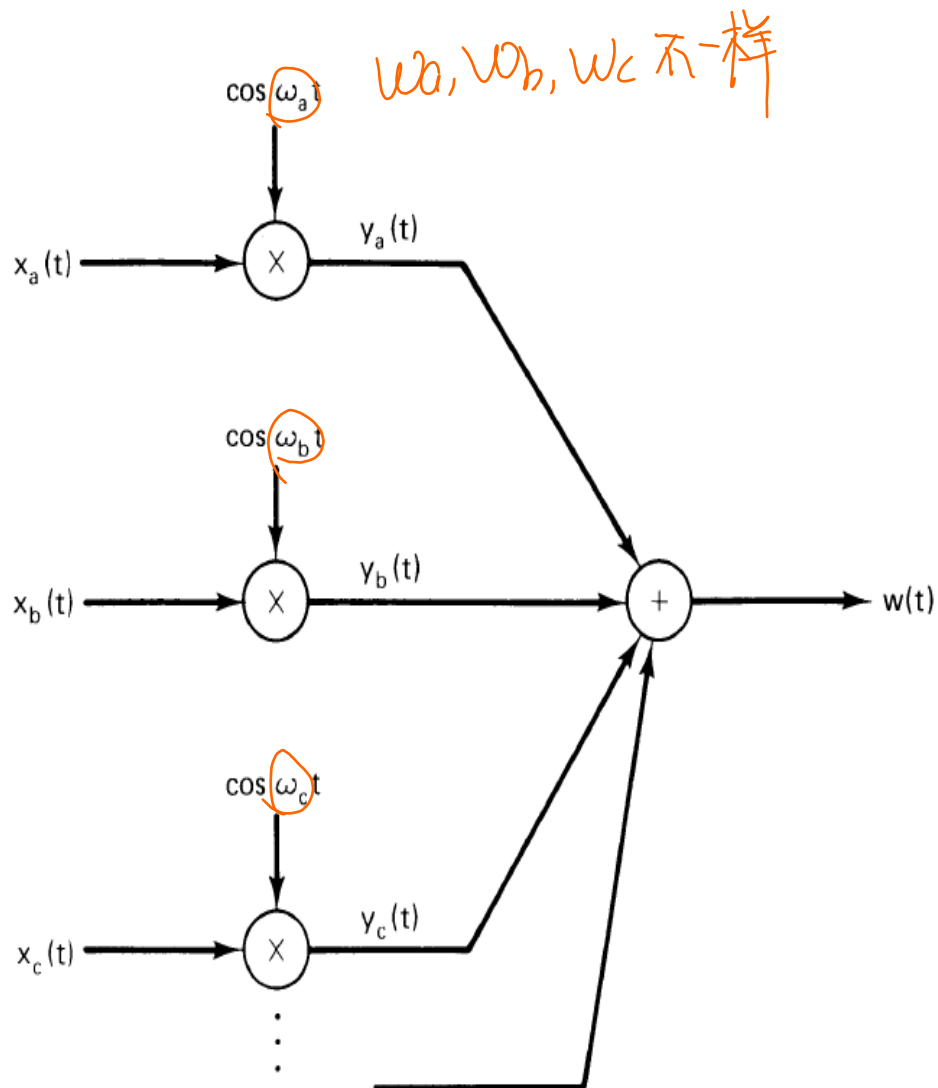
# Frequency-division multiplexing (FDM)

频分复用

- Systems for transmitting signals provide more bandwidth than is required for one signal.
  - ◆ E.g., speech signal → 20 ~ 20 kHz
  - microwave channel → 300 MHz ~ 300 GHz
  - satellite link → a few hundred MHz ~ 40 GHz
  - (more in Fig. 8.18)
- Different modulating signals (e.g., speech), which are overlapping in frequency, can have their spectra **shifted (e.g., by sinusoidal AM) without overlapping**, so they can be transmitted simultaneously over a single wideband channel.

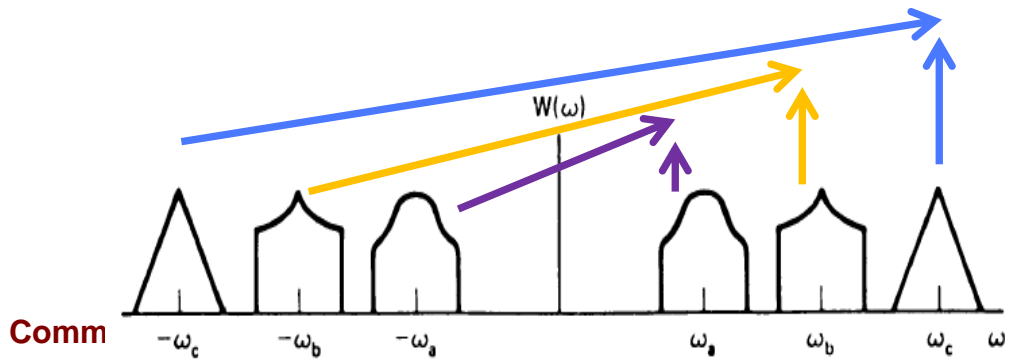
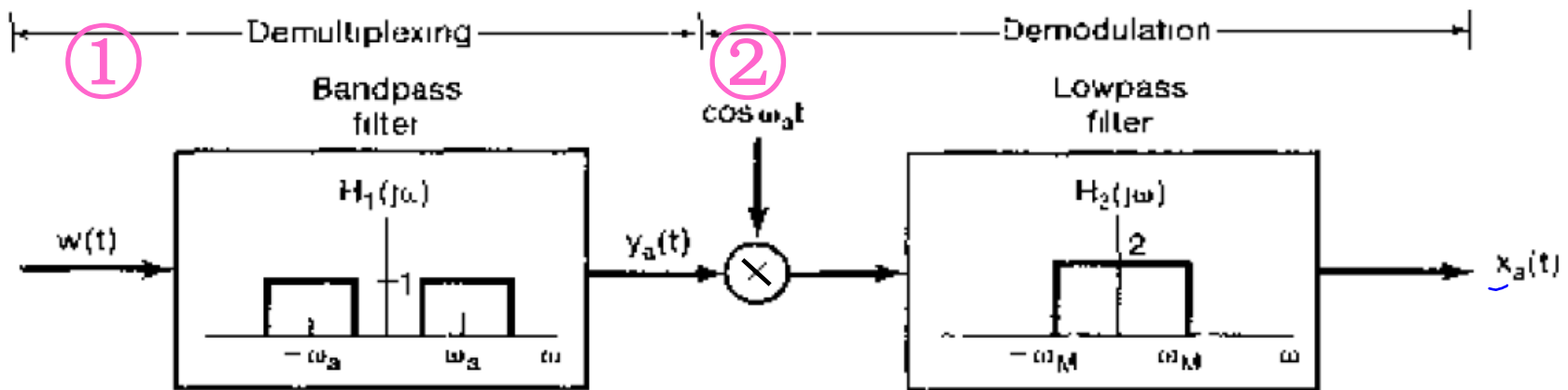
↓  
同时地

# FDM (cont.)



# FDM demultiplexing and demodulation

- ① bandpass filtering to have the modulated signal from one channel
- ② demodulation to recover the modulating signal



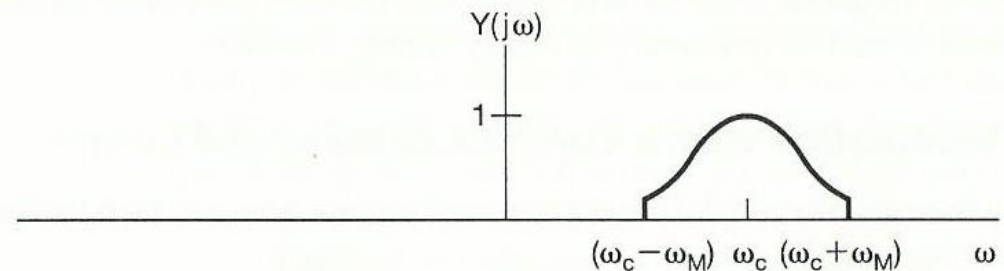
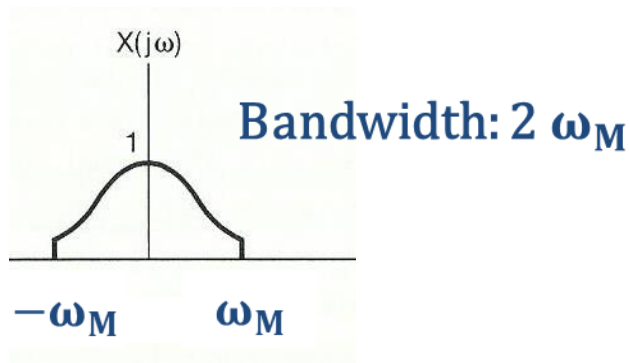
Occupy twice the original bandwidth  
 → insufficient use of bandwidth



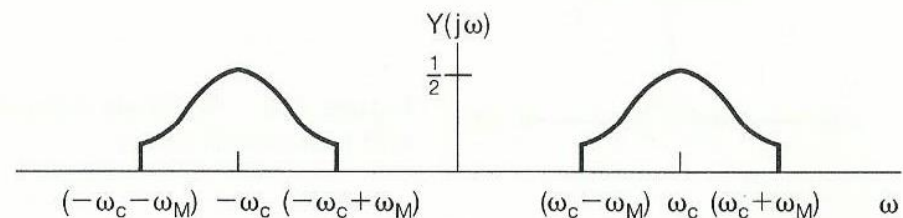
# Single-sideband (SSB) sinusoidal AM

Occupied bandwidth:

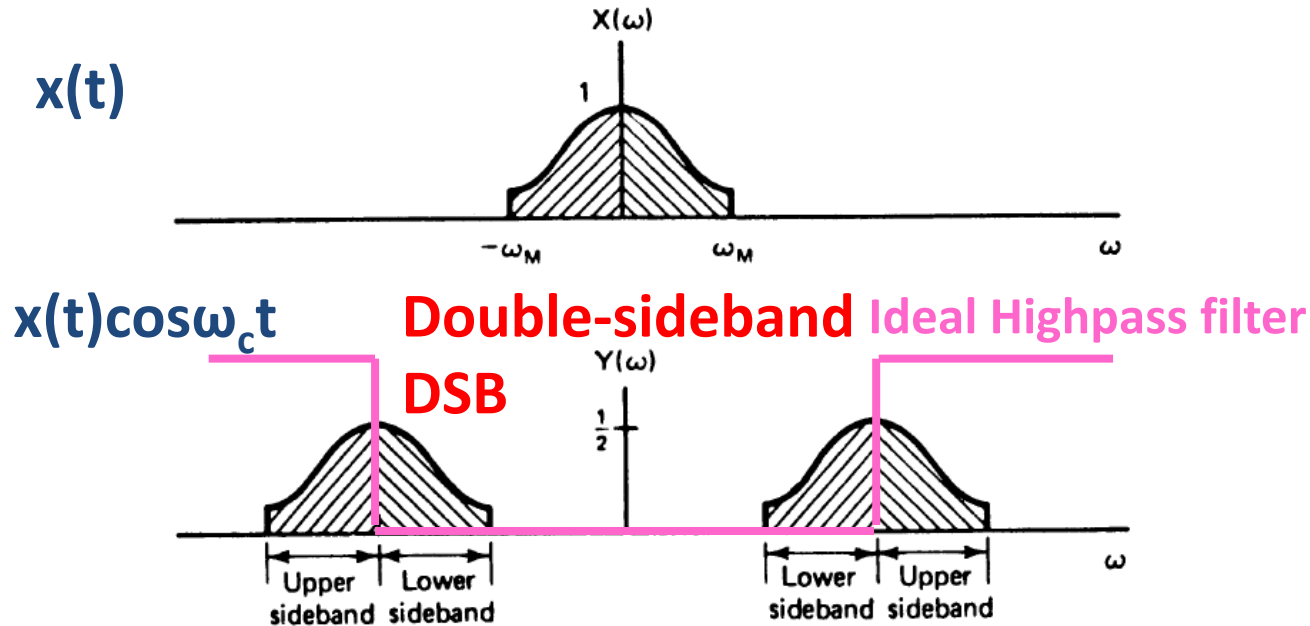
- With exponential carrier, the bandwidth is still  $2\omega_M$



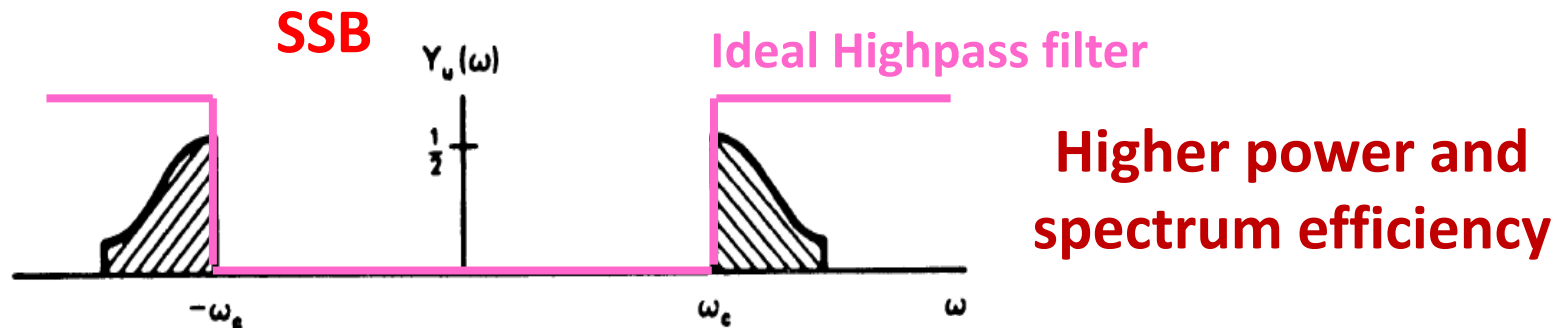
- With sinusoidal carrier, twice bandwidth is required.



# SSB sinusoidal AM (cont.)

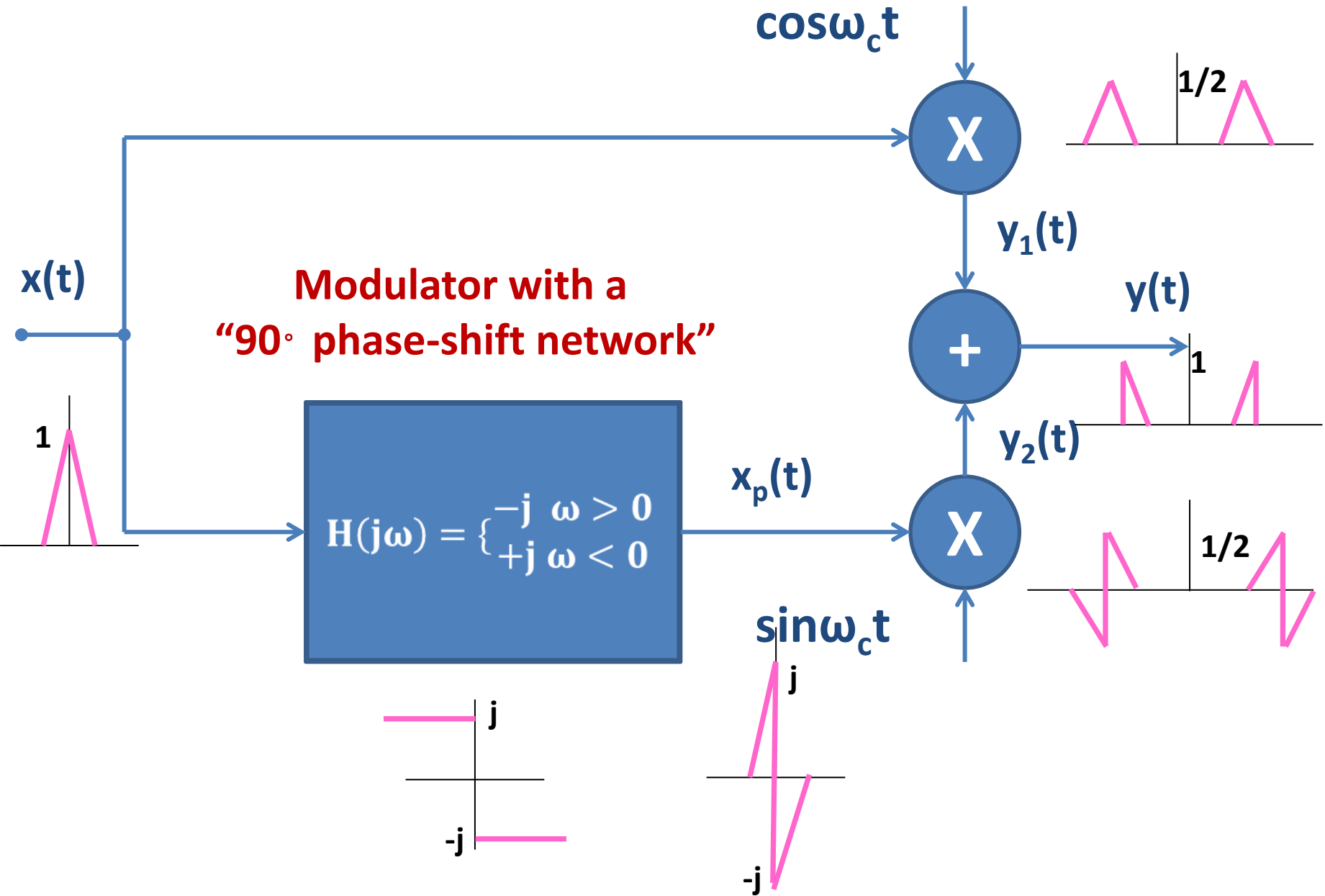


Observation:  $x(t)$  can be recovered if two upper (or lower) sidebands are retained.



What is the cost ?

# SSB sinusoidal AM (cont.)



# Amplitude modulation with a pulse train

- Carrier signal could be a sinusoidal signal, or a pulse train.

先把  $c(t)$  写成 CTFS,  
再 CTFS  $\rightarrow$  CTFT

$X(j\omega)$   $\uparrow$

$$C(j\omega) = 2\pi \sum_{k=-\infty}^{\infty} a_k \delta(\omega - k\omega_c)$$

$$a_k = \frac{\sin(k\omega_c \Delta / 2)}{\pi k}$$

Read Example 3.5

$$Y(j\omega) = \sum_{k=-\infty}^{\infty} a_k X(j(\omega - k\omega_c))$$

Figure 8.23

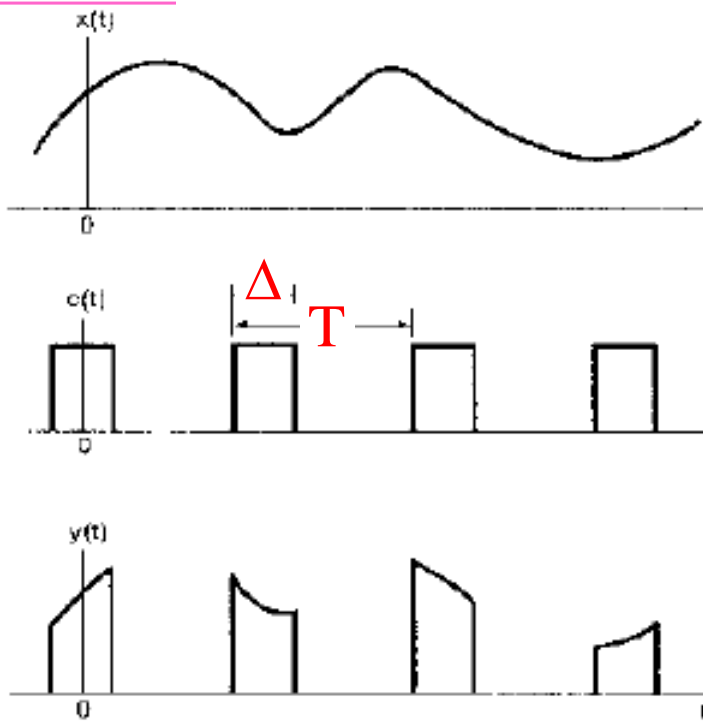
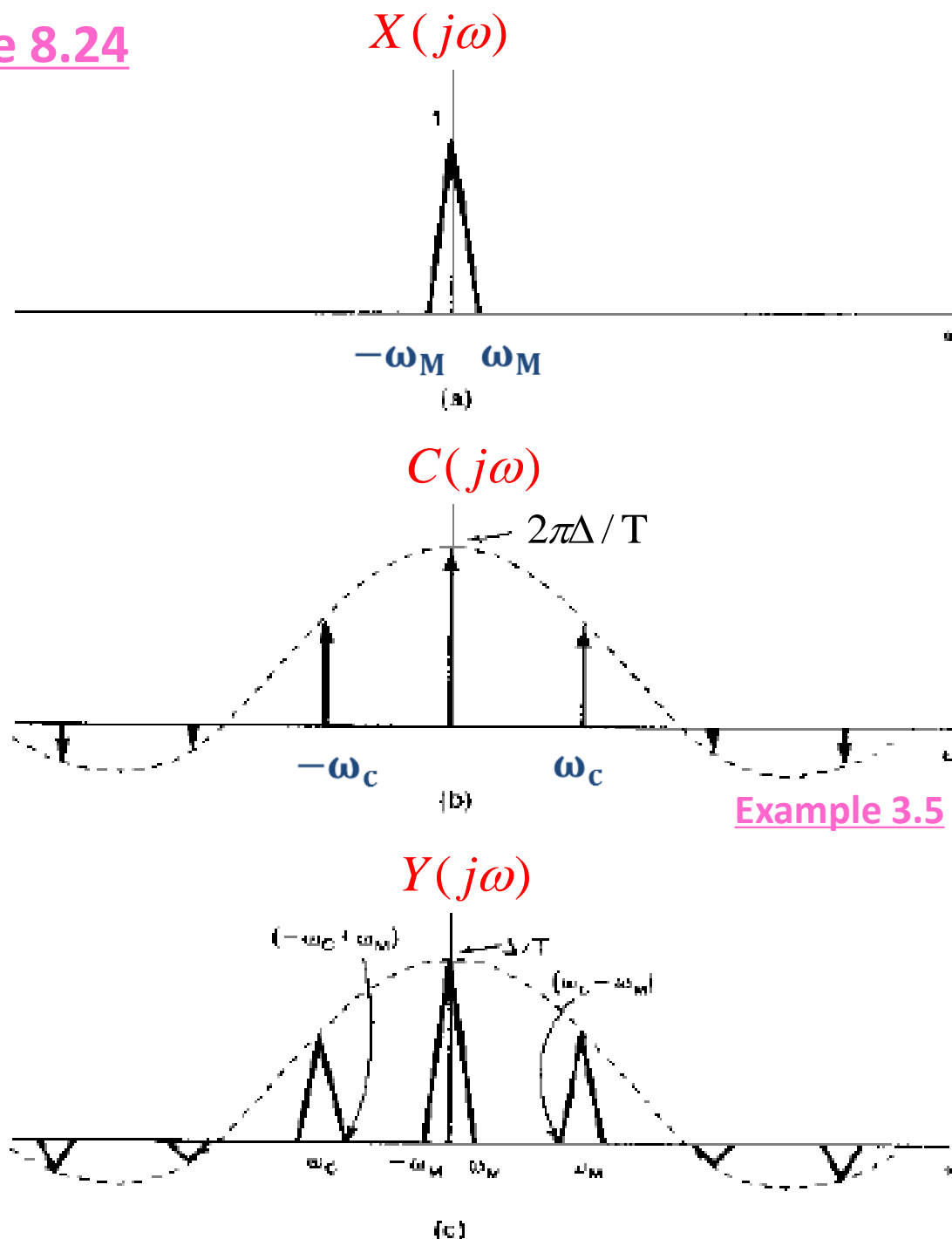


Figure 8.24

When  $\omega_c > 2\omega_M$ ,  $X(j\omega)$  can be recovered by lowpass filtering.

Note: similar to the condition in sampling.

# Summary

- Meaning of amplitude modulation
  - ◆ with a complex exponential carrier
  - ◆ with a sinusoidal carrier
- Demodulation for sinusoidal AM
  - ◆ Synchronous demodulation
  - ◆ Asynchronous demodulation, and its two important assumptions
- Frequency-division multiplexing (FDM)