

EE206: Communications Principles Tutorial

Assignment 2

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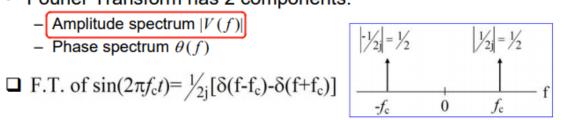
Amplitude Spectrum

In general, the Fourier Transform V(f) is a complex function of frequency *f*, so that it can be rewritten as

$$V(f) = |V(f)| \exp [j \theta(f)]$$

- Fourier Transform has 2 components:

$$\square$$
 F.T. of $\sin(2\pi f_c t) = \frac{1}{2i} [\delta(f-f_c) - \delta(f+f_c)]$



$$\Box$$
 F.T. of $rect\left(\frac{t}{T}\right) = T sinc(f T)$. What is the amplitude spectrum?

Usually, spectrum refers to amplitude spectrum unless otherwise stated.

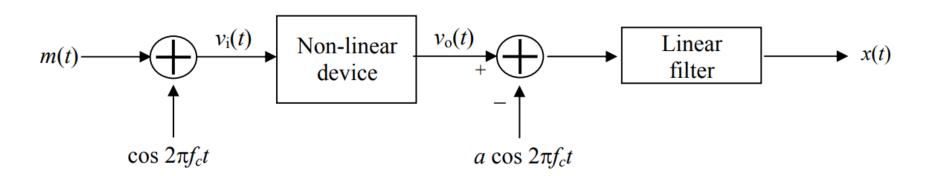
In general, the Fourier transform G(f) is a complex function of frequency f, so that we may express it in the form

$$G(f) = |G(f)| \exp[j\theta(f)]$$
(2.6)

where |G(f)| is called the *continuous amplitude spectrum* of g(t) and $\theta(f)$ is called the continuous phase spectrum of g(t). Here, the spectrum is referred to as a continuous spectrum because both the amplitude and phase of G(f) are defined for all frequencies.

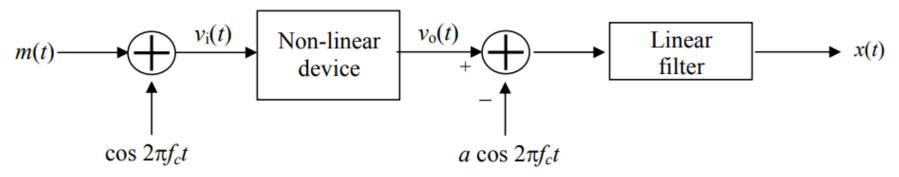


- 1. The message signal $m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$ is input to the system shown below to generate a DSBSC-AM signal x(t). Assume that $v_0(t) = a v_1(t) + b v_1^2(t)$ where a and b are constants, and the carrier frequency $f_c >> 70$ Hz.
 - (a) Sketch the amplitude spectrum of the filter input
 - (b) Determine the center frequency and bandwidth of the filter in this modulator
 - (c) Determine the minimum value of f_c permitted for this modulator





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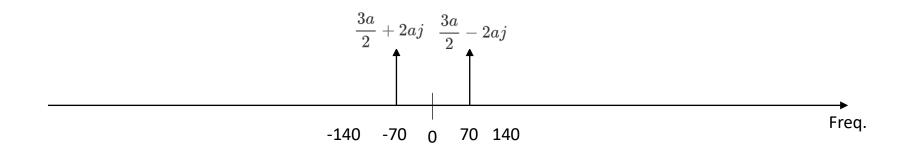


(a) Sketch the amplitude spectrum of the filter input

$$egin{aligned} m(t) &= 3\cos(2\pi 70t) + 4\sin(2\pi 70t) \ v_i(t) &= m(t) + \cos(2\pi f_c t) \ v_o(t) &= av_i(t) + bv_i^2(t) \ &= a[m(t) + \cos(2\pi f_c t)] + b[m(t) + \cos(2\pi f_c t)]^2 \ &= am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + a\cos(2\pi f_c t) + b\cos^2(2\pi f_c t) \ &= am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + a\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t) \ &= am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t) \ &= am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t) \end{aligned}$$



$$egin{aligned} m(t) &= 3\cos(2\pi 70t) + 4\sin(2\pi 70t) \ & ext{filter input} = am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t) \ &am(t) = 3a\cos(2\pi 70t) + 4a\sin(2\pi 70t) \ &= (rac{3a}{2} - 2aj)e^{j2\pi 70t} + (rac{3a}{2} + 2aj)e^{-j2\pi 70t} \end{aligned}$$

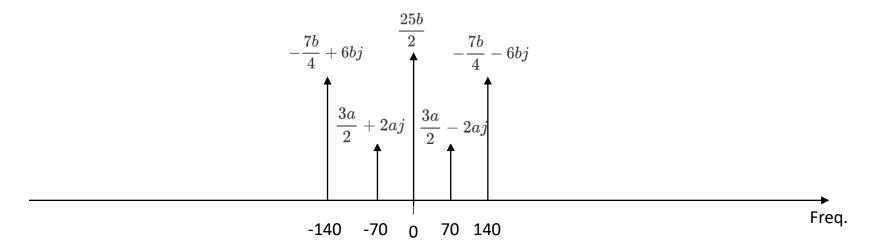




$$m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$$

$$ext{filter input} = am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t)$$

$$egin{aligned} bm^2(t) &= 9b\cos^2(2\pi 70t) + 24b\cos(2\pi 70t)\sin(2\pi 70t) + 16b\sin^2(2\pi 70t) \\ &= rac{9b}{2} + rac{9b}{2}\cos(2\pi 140t) + 12b\sin(2\pi 140t) + 8b - 8b\cos(2\pi 140t) \\ &= rac{25b}{2} - rac{7b}{2}\cos(2\pi 140t) + 12b\sin(2\pi 140t) \\ &= rac{25b}{2} - (rac{7b}{4} + 6bj)e^{j2\pi 140t} - (rac{7b}{4} - 6bj)e^{-j2\pi 140t} \end{aligned}$$

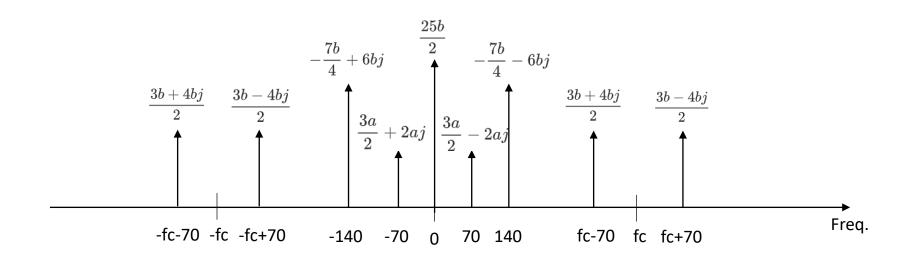




$$m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$$

$$ext{filter input} = am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t)$$

$$\begin{aligned} 2bm(t)\cos(2\pi f_c t) &= 2b(3\cos(2\pi 70t) + 4\sin(2\pi 70t))\cos(2\pi f_c t) \\ &= ((3b - 4bj)e^{j2\pi 70t} + (3b + 4bj)e^{-j2\pi 70t})(\frac{1}{2}e^{j2\pi f_c t} + \frac{1}{2}e^{-j2\pi f_c t}) \\ &= \frac{3b - 4bj}{2}e^{j2\pi(f_c + 70)t} + \frac{3b - 4bj}{2}e^{j2\pi(-f_c + 70)t} + \frac{3b + 4bj}{2}e^{j2\pi(f_c - 70)t} + \frac{3b + 4bj}{2}e^{j2\pi(-f_c - 70)t} \end{aligned}$$

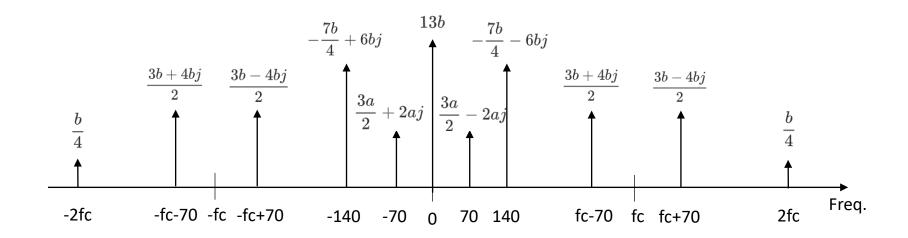




$$m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$$

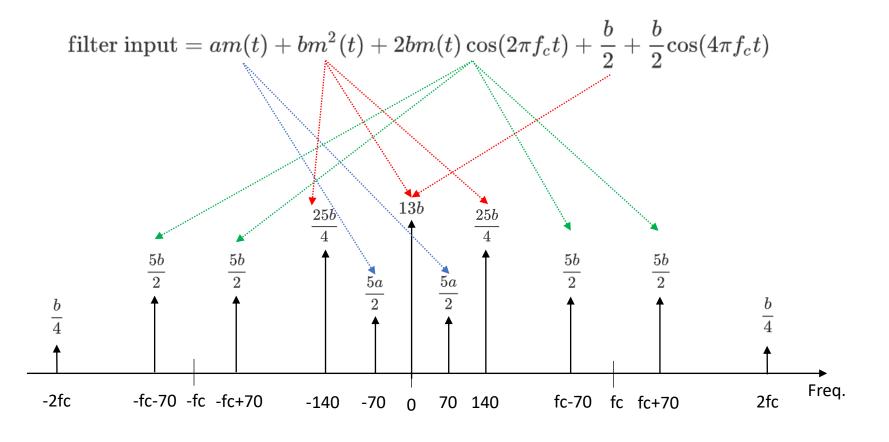
$$ext{filter input} = am(t) + bm^2(t) + 2bm(t)\cos(2\pi f_c t) + rac{b}{2} + rac{b}{2}\cos(4\pi f_c t)$$

$$rac{b}{2} + rac{b}{2} {
m cos}(4\pi f_c t) = rac{b}{2} + rac{b}{4} e^{j2\pi 2 f_c t} + rac{b}{4} e^{-j2\pi 2 f_c t}$$



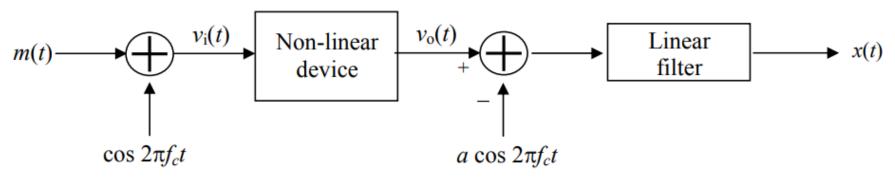


$$m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t) \ = 5\cos(2\pi 70t - \theta) \ heta = an^{-1}rac{4}{3}$$





1. The message signal $m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$ is input to the system shown below to generate a DSBSC-AM signal x(t). Assume that $v_0(t) = a v_1(t) + b v_1^2(t)$ where a and b are constants, and the carrier frequency $f_c >> 70$ Hz.



(b) Determine the center frequency and bandwidth of the filter in this modulator

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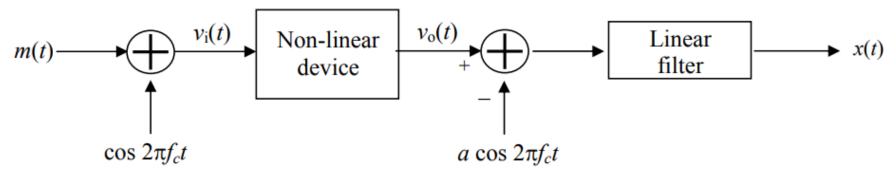
Bandwidth (BW) is defined as range of **positive** frequency occupied by a spectrum

The DSBSC-AM signal should be $m(t)\cos(2\pi f_c t)$.

In order to separate it from the other unwanted components, the center freq $= f_c$, and bandwidth = 140Hz.



1. The message signal $m(t) = 3\cos(2\pi 70t) + 4\sin(2\pi 70t)$ is input to the system shown below to generate a DSBSC-AM signal x(t). Assume that $v_0(t) = a v_1(t) + b v_1^2(t)$ where a and b are constants, and the carrier frequency $f_c >> 70$ Hz.



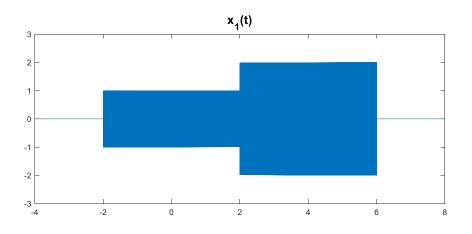
(c) Determine the minimum value of f_c permitted for this modulator

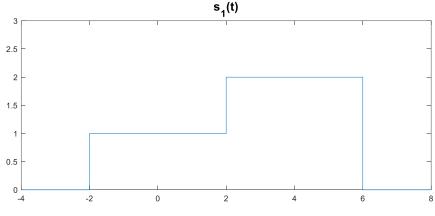
$$f_c-70>140 \
ightarrow f_c>210 Hz$$

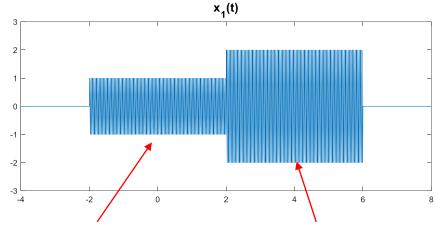


2. A suppressed-carrier AM signal $x_1(t)$ is generated by modulating $s_1(t) = rect\left(\frac{t}{4}\right) + 2rect\left(\frac{t-4}{4}\right)$ with $\sin\left(1000\,\pi t\right)$. Sketch the time waveform of $x_1(t)$.

$$x_1(t) = s_1(t)\sin(1000\pi t)$$





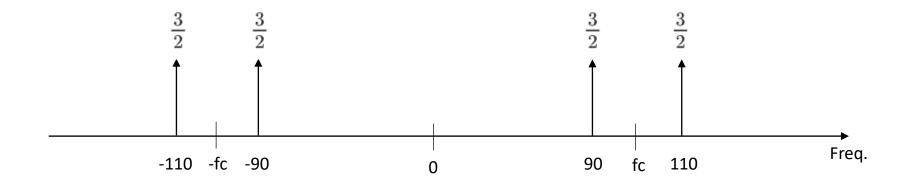


 $500 \times 4 = 2000 \text{ cycles} \qquad 500 \times 4 = 2000 \text{ cycles}$



$$x(t) = 3 \sin 180\pi t + 3 \sin 220\pi t$$

(a) Sketch the amplitude spectrum of x(t) to deduce the carrier frequency in x(t)

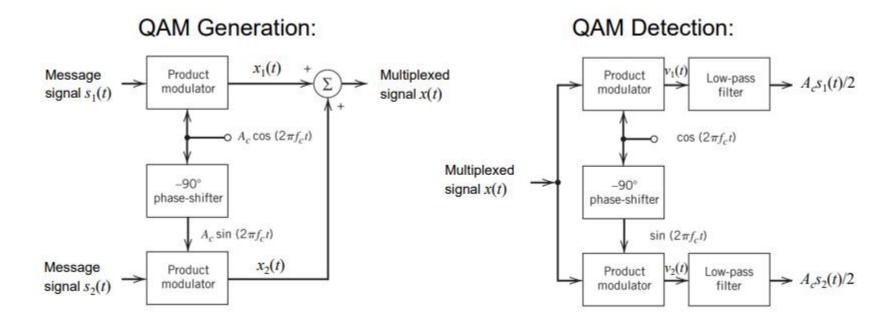


carrier frequency $f_c = 100Hz$.



$$x(t) = 3 \sin 180\pi t + 3 \sin 220\pi t$$

(b) Given that x(t) was generated using a sine carrier signal with phase 0, demodulate x(t).





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(b) Given that x(t) was generated using a sine carrier signal with phase 0, demodulate x(t).

QAM Generation:

$$x_1(t) = A_c s_1(t) \cos 2\pi f_c t = \text{in-phase (I) signal}$$

 $x_2(t) = A_c s_2(t) \sin 2\pi f_c t = \text{quadrature (Q) signal}$
 $\therefore x(t) = x_1(t) + x_2(t) = A_c s_1(t) \cos 2\pi f_c t + A_c s_2(t) \sin 2\pi f_c t$
where $s_1(t)$ and $s_2(t)$ are two different message signals.

QAM Detection:

$$v_1(t) = x(t)\cos 2\pi f_c t = A_c \, s_1(t)\cos^2 2\pi f_c t + A_c \, s_2(t)\sin 2\pi f_c t\cos 2\pi f_c t$$
$$= \frac{1}{2}A_c \, s_1(t) + \frac{1}{2}A_c \, s_1(t)\cos 4\pi f_c t + \frac{1}{2}A_c \, s_2(t)\sin 4\pi f_c t$$

Similarly, we get

$$v_2(t) = x(t) \sin 2\pi f_c t$$

= $\frac{1}{2} A_c s_2(t) - \frac{1}{2} A_c s_2(t) \cos 4\pi f_c t + \frac{1}{2} A_c s_1(t) \sin 4\pi f_c t$

Thus, $\frac{1}{2}A_c s_1(t)$ and $\frac{1}{2}A_c s_2(t)$ can be obtained after the lowpass filters.

Hence, the two message signals $s_1(t)$ and $s_2(t)$ can be recovered.



$$x(t) = 3 \sin 180\pi t + 3 \sin 220\pi t$$

(b) Given that x(t) was generated using a sine carrier signal with phase 0, demodulate x(t).

$$\begin{aligned} (3\sin 180\pi t + 3\sin 220\pi t)\sin 200\pi t &= 3(\sin 180\pi t \sin 200\pi t + \sin 220\pi t \sin 200\pi t) \\ &= \frac{3}{2}(\cos(180-200)\pi t - \cos(180+200)\pi t + \cos(220-200)\pi t - \cos(220+200)\pi t) \\ &= \frac{3}{2}(2\cos 20\pi t - \cos 380\pi t - \cos 420\pi t) \end{aligned}$$

after lowpass filter (10Hz bandwidth)

output =
$$3\cos 20\pi t$$