

2020

2020 年秋季学期第16周

Review:

o CSIR + CSIT + power adaptation
water-filling

$$C = \int_{\gamma_0}^{\infty} B \log\left(\frac{\gamma}{\gamma_0}\right) f_{\gamma}(\gamma) d\gamma$$

28 十五

Monday / 星期一

$$\gamma = \frac{g\bar{P}}{N_0B}$$

o CSIR + CSIT + Channel Inversion

$$\bar{P} = \frac{\sigma}{\gamma} \cdot \bar{P} = \frac{\sigma \cdot N_0 B}{g}$$

29 十六

Tuesday / 星期二

$$C = B \cdot \log_2\left(1 + \frac{1}{E[\gamma/K]}\right)$$

30 十七

Wednesday / 星期三

Δ CSIR & CSIT

Outage Capacity, Truncated Channel Inversion

$$\frac{p(\gamma)}{\bar{p}} = \begin{cases} \frac{\sigma}{\gamma}, & \text{if } \gamma \geq \gamma_0 \\ 0, & \text{if } \gamma < \gamma_0 \end{cases}$$

Outage Capacity:

$$C = B \cdot \log_2(1 + \sigma) \cdot \Pr(\gamma \geq \gamma_0)$$

$$E[p(\gamma)] = \int_{\gamma_0}^{+\infty} p(\gamma) \cdot f(\gamma) d\gamma = \bar{p}$$

$$\Rightarrow \sigma = \frac{1}{\int_{\gamma_0}^{+\infty} \frac{1}{\gamma} \cdot f(\gamma) d\gamma}$$

$$= \frac{1}{E_{\gamma} [1/\gamma]}$$

$$\Rightarrow C = B \log_2 \left(1 + \frac{1}{E_{\gamma} [1/\gamma]} \right) \cdot \Pr(\gamma \geq \gamma_0)$$

maximum outage capacity:

$$C = \max_{\gamma_0} B \cdot \log_2 \left(1 + \frac{1}{E_{\gamma} [1/\gamma]} \right) \cdot \Pr(\gamma \geq \gamma_0)$$

Digital. Modulation and Detection.

△ Digital Modulation.

Mapping from info. bits to analog signals.

message

$$M = \{m_1, m_2, \dots, m_M\}$$

$$K = \log_2 M$$

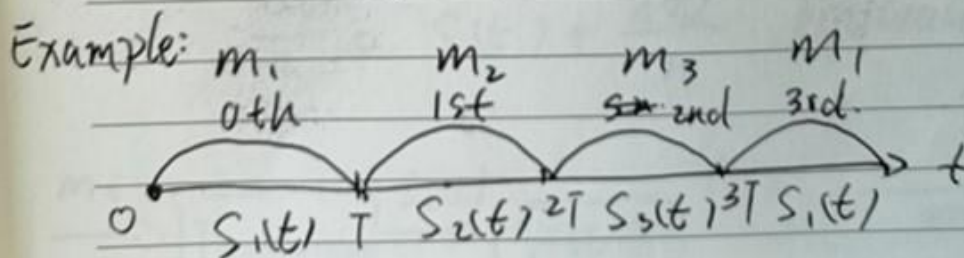


$$m_i = \{b_1, b_2, \dots, b_K\}$$

$$s_i(t) \in \{s_1(t), s_2(t), \dots, s_M(t)\}$$

→ Transmitter →

$$m_i \rightarrow s_i(t) \quad i = 1, \dots, M$$



Let T be the signal duration.

$$s(t) = s_1(t) + s_2(t-T) + s_3(t-2T) + s_4(t-3T)$$

Suppose message $m(k) \in \{1, \dots, M\}$ is transmitted in the k -th period. $\Rightarrow s_{m(k)}(t-kT)$

$$s(t) = \sum_{k=0}^{+\infty} s_{m(k)}(t-kT)$$

Energy of $s(t)$: $E_s = \int_0^T s^2(t) dt$

Δ Geometric Representation of Signals.

Orthonormal basis functions

$$\phi = \{\phi_1(t), \phi_2(t), \dots, \phi_N(t)\} \quad t \in [0, T]$$

$$\langle \phi_i(t), \phi_j(t) \rangle = \int_0^T \phi_i(t) \cdot \phi_j(t) dt = \begin{cases} 1, & i=j \\ 0, & i \neq j \end{cases}$$

e.g. :
$$\begin{cases} \phi_1(t) = \sqrt{\frac{2}{T}} \cos(2\pi f_c t) \\ \phi_2(t) = \sqrt{\frac{2}{T}} \sin(2\pi f_c t) \end{cases}$$

$$s_i(t) = \sum_{j=1}^N s_{ij} \cdot \phi_j(t), \quad 0 \leq t < T$$

$$\int_0^T s_i(t) \cdot \phi_j(t) dt = s_{ij}$$

$$= \langle s_i(t); \phi_j(t) \rangle$$

Given ϕ , signals $\{s_i(t)\}$ can be represented by (s_{i1}, \dots, s_{iN}) by $(s_{11}, \dots, s_{M,N})$

$$S_i(t) \rightarrow (S_{i1}, S_{i2}, \dots, S_{iN}) = \vec{S}_i, i=1, \dots, M$$

signal constellation point: a point in
N-dimensional space.

signal constellation: $\{\vec{S}_1, \vec{S}_2, \dots, \vec{S}_M\}$

signal space

QPSK / QAM

$$s_i(t) = s_{i1} \phi_1(t) + s_{i2} \phi_2(t)$$

$$= s_{i1} \sqrt{\frac{2}{T}} \cos 2\pi f_c t + s_{i2} \sqrt{\frac{2}{T}} \sin 2\pi f_c t.$$

QAM signals

~~$$\text{QPSK: } (s_{i1}, s_{i2}) = \{(1, 1), (1, -1), (-1, 1), (-1, -1)\}$$~~

$$\text{QPSK: } (s_{i1}, s_{i2}) \quad (s_{21}, s_{22}) \quad (s_{31}, s_{32}) \quad (s_{41}, s_{42})$$

$$(1, -1) \quad (1, 1) \quad (-1, 1) \quad (-1, -1)$$

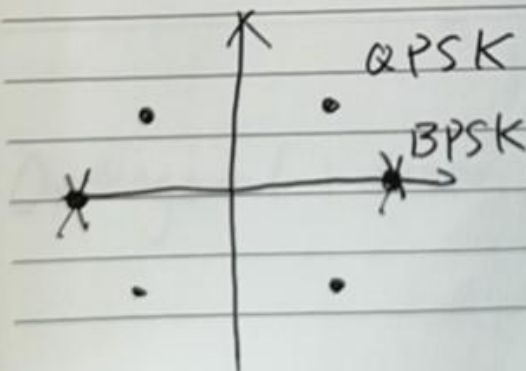
$$\text{BPSK: } s_1(t) = d \cos(2\pi f_c t) \quad \text{"1"}$$

$$s_2(t) = -d \cos(2\pi f_c t) \quad \text{"0"}$$

$$\phi = \left\{ \sqrt{\frac{2}{T}} \cos 2\pi f_c t \right\} \quad N=1$$

$$s_{11} = \int_0^T s_1(t) \cdot \sqrt{\frac{2}{T}} \cos 2\pi f_c t dt = d \sqrt{\frac{T}{2}}$$

$$s_{21} = \int_0^T s_2(t) \cdot \sqrt{\frac{2}{T}} \cos 2\pi f_c t dt = -d \sqrt{\frac{T}{2}}$$



$$\vec{S}_i = (S_{i1}, S_{i2}, \dots, S_{iN})$$

length: $\|\vec{S}_i\| = \sqrt{\sum_{j=1}^N S_{ij}^2}$

distance between \vec{S}_i and \vec{S}_k

$$\|\vec{S}_i - \vec{S}_k\| = \sqrt{\sum_{j=1}^N (S_{ij} - S_{kj})^2}$$

$$= \sqrt{\int_0^T [S_i(t) - S_k(t)]^2 dt}$$

inner ~~po~~ product

$$\langle S_i(t), S_k(t) \rangle \triangleq \int_0^T S_i(t) \cdot S_k(t) dt$$

$$= \langle \vec{S}_i, \vec{S}_k \rangle$$

$$= \sum_{j=1}^N S_{ij} S_{kj}$$

Orthogonal: inner product = 0

Δ AWGN Channel. (move to Rx part)
$$Y(t) = S(t) + n(t)$$

* Flat Fading Channel.

$$Y(t) = h \cdot S(t) + n(t)$$

⇓

$$\frac{Y(t)}{h} = S(t) + \frac{n(t)}{h} \quad \text{equivalent AWGN}$$

