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Veterbi Algorithm (revisted)

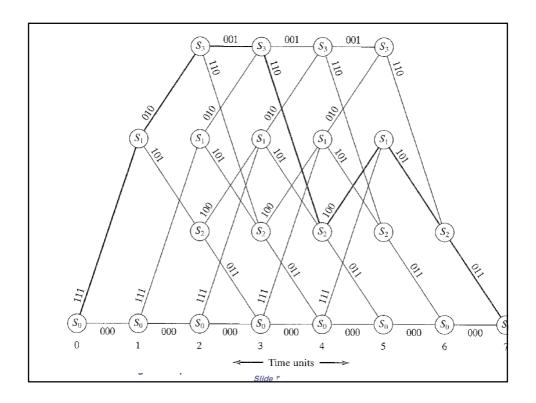
Consider covolutional encoder with

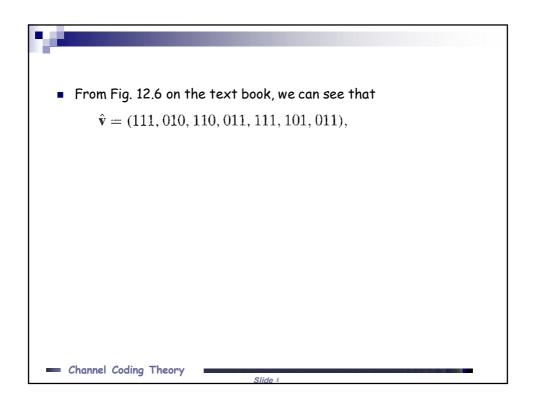
$$G(D) = [1 + D \quad 1 + D^2 \quad 1 + D + D^2]$$

- And information sequences of length h = 5
- The trellis diagram has h + m + 1 timeslots which equals 8 in our case
- Consider received sequence as

$$\mathbf{r} = (110, 110, 110, 111, 010, 101, 101).$$

Channel Coding Theory







SOVA

The Soft-Output Viterbi Algorithm (SOVA) was first introduced in 1989.

We describe SOVA for convolutional code with R = 1/n on binary input, AWGN channel.

We assume that priori probabilities are not equally likely $p(u_L)$ and L = 0,...., h-1.



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Log-likelihood metric

■ Let us define the log-likelihood ratio or the L-value of a received symbol r at the output of channel with binary inputs $v = \pm 1$

$$L(r) = \ln \left[\frac{p(r|v=+1)}{p(r|v=-1)} \right].$$

• Similarly the L-value of an information bit u is defined as

$$L(u) = \ln \left[\frac{p(u=+1)}{p(u=-1)} \right].$$

Using Bay's rule if v is equally likely

$$L(r) = \ln \left[\frac{p(r|v=+1)}{p(r|v=-1)} \right] = \ln \left[\frac{p(v=+1|r)}{p(v=-1|r)} \right].$$

Channel Coding Theory



Log-likelihood metric

- A large positive value of L(r) indicates a high reliability that v = +1.
- A large negative value of L(r) indicates a high reliability that v = -1.
- A close to zero value of L(r) indicates a decision a bout the value of v based only on r is unreliable.
- The same a large positive value of L(u) indicates a high reliability that u = +1



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Log-likelihood metric

■ It can be shown that the L value is equal to (left as exercise for the students)

$$L(r) = (4E_s/N_0)r,$$

• Where $L_c \equiv 4E_s/N_0$ is defined as channel reliability factor

Channel Coding Theory

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BCJR algorithm

- Veterbi Algorithm minimizes the WER $P_{\scriptscriptstyle W}(E)$ that is $P(\hat{v}
 eq v | r)$
- So it is minimizes the error probability between the transmitted and received codeword.
- In BCJR algorithm, we are interested in minimizing the bit error probability.
- This is done by maximizing the posteriori probability $P(\hat{u}_l = u_l | \mathbf{r})$
- That is why BCJR decoder is also called Maximum Posteriori Probability decoder (MAP)



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BCJR algorithm

- We don't assume that the information bits are equally likely.
- The algorithms calculates the a posteriori L-values

$$L(u_l) \equiv \ln \left[\frac{P(u_l = +1|\mathbf{r})}{P(u_l = -1|\mathbf{r})} \right],$$

 Called APP L-values of each information bit, the decoder output is given by

$$\hat{u}_l = \left\{ \begin{array}{l} +1 \text{ if } L(u_l) > 0 \\ -1 \text{ if } L(u_l) < 0 \end{array} \right., \ l = 0, 1, \cdots, h-1.$$

Channel Coding Theory



 We start our development of the BCJR algorithm by rewriting the APP value as

$$P(u_l = +1 | \mathbf{r}) = \frac{p(u_l = +1, \mathbf{r})}{P(\mathbf{r})} = \frac{\sum_{\mathbf{u} \in \mathbf{U}_l^+} p(\mathbf{r} | \mathbf{v}) P(\mathbf{u})}{\sum_{\mathbf{u}} p(\mathbf{r} | \mathbf{v}) P(\mathbf{u})},$$

- Where U_L^+ is the set of all information sequences **u** such as $u_l = 1$, v is the transmitted codeword corresponding to the information sequence u.
- So we can rewrite the expression of the APP L values as

$$L(u_l) = \ln \left[\frac{\sum_{\mathbf{u} \in \mathbb{U}_l^+} p(\mathbf{r}|\mathbf{v}) P(\mathbf{u})}{\sum_{\mathbf{u} \in \mathbb{U}_l^-} p(\mathbf{r}|\mathbf{v}) P(\mathbf{u})} \right],$$

- Where U_L^- is the set of all information sequences **u** such as $u_i = -1$
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- The L values can be calculated using the previous formula but still it suffers from high degree of complexity.
- We can rewrite the a posteriori probability as

$$P(u_l=+1|\mathbb{r})=\frac{p(u_l=+1,\mathbb{r})}{P(\mathbb{r})}=\frac{\sum_{(s',s)\in\Sigma_l^+}p(s_l=s',s_{l+1}=s,\mathbb{r})}{P(\mathbb{r})},$$

- Where $\sum_{i=1}^{+}$ is the set of all state pairs $s_i = s'$ and $s_{i+1} = s$ that corresponds to the input bit $u_i = +1$.
- Reforming $P(u_i = -1/r)$ in the same way and sub in the L value

$$L(u_l) = \ln \left\{ \frac{\sum_{(s',s) \in \Sigma_l^+} p(s_l = s', s_{l+1} = s, \mathbb{I})}{\sum_{(s',s) \in \Sigma_l^-} p(s_l = s', s_{l+1} = s, \mathbb{I})} \right\},\,$$

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- Where $\sum_{i=1}^{n}$ is the set of all state pairs $s_i = s^i$ and $s_{i+1} = s^i$ that corresponds to the input bit $u_i = -1$.
- The joint pdf p(s', s, r) can be found recursively, starting from

$$p(s', s, \mathbb{r}) = p(s', s, \mathbb{r}_{t < l}, \mathbb{r}_{l}, \mathbb{r}_{t > l}),$$

- Where $r_{t < l}$ represents the portion of the received r before the time / and Where $\mathbf{r}_{t > l}$ represents the portion of the received r after the time l.
- Now application of Bay's rule

$$\begin{split} p(s', s, \mathbf{r}) &= p(\mathbf{r}_{t>l}|s', s, \mathbf{r}_{t< l}, \mathbf{r}_{l}) p(s', s, \mathbf{r}_{t< l}, \mathbf{r}_{l}) \\ &= p(\mathbf{r}_{t>l}|s', s, \mathbf{r}_{t< l}, \mathbf{r}_{l}) p(s, \mathbf{r}_{l}|s', \mathbf{r}_{t< l}) p(s', \mathbf{r}_{t< l}) \\ &= p(\mathbf{r}_{t>l}|s) p(s, \mathbf{r}_{l}|s') p(s', \mathbf{r}_{t< l}), \end{split}$$

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Defining

$$\alpha_l(s') \equiv p(s', \mathbf{r}_{t < l})$$

$$\gamma_l(s',s) \equiv p(s, \mathbf{r}_l|s')$$

$$\beta_{l+1}(s) \equiv p(\mathbf{r}_{t>l}|s),$$

• So the joint pdf can be rewritten as

$$p(s', s, \mathbf{r}) = \beta_{l+1}(s)\gamma_l(s', s)\alpha_l(s').$$

• We can write expression for $a_{i+1}(s)$ as

$$\alpha_{l+1}(s) \; = \sum_{s' \in \sigma_l} \gamma_l(s',s) \alpha_l(s'),$$

• Where σ_l is the set of all states at time l.

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• $\beta(s')$ can be written as

$$\beta_l(s') = \sum_{s \in \sigma_{l+1}} \gamma_l(s', s) \beta_{l+1}(s),$$

- Where σ_{l+1} is the set of all states at time l+1.
- The forward recursion starts from

$$\alpha_0(s) = \left\{ \begin{array}{l} 1, \ s = \mathbf{0} \\ 0, \ s \neq \mathbf{0} \end{array} \right.,$$

- And the backward recursion starts from

$$\beta_K(s) = \left\{ \begin{array}{l} 1, \ s = \mathbf{0} \\ 0, \ s \neq \mathbf{0} \end{array} \right.,$$

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• We can write the branch metric as

$$\gamma_{l}(s', s) = p(s, \mathbb{r}_{l}|s') = \frac{p(s', s, \mathbb{r}_{l})}{P(s')}$$

$$= \left[\frac{P(s', s)}{P(s')}\right] \left[\frac{p(s', s, \mathbb{r}_{l})}{P(s', s)}\right]$$

$$= P(s|s')p(\mathbb{r}_{l}|s', s) = P(u_{l})p(\mathbb{r}_{l}|\mathbb{v}_{l}),$$

Which yields

$$\gamma_l(s',s) = P(u_l) p(\mathbf{r}_l | \mathbf{v}_l) = P(u_l) \left(\sqrt{\frac{E_s}{\pi N_0}} \right)^n e^{-\frac{E_s}{N_0} ||\mathbf{r}_l - \mathbf{v}_l||^2},$$

• We can drop the constant to achieve

$$\gamma_I(s',s) = P(u_I)e^{-E_s/N_0||\mathbf{r}_I - \mathbf{v}_I||^2}.$$

Channel Coding Theory

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• The priori probability can be written as

$$\begin{split} P(u_l = \pm 1) &= \frac{[P(u_l = +1)/P(u_l = -1)]^{\pm 1}}{\{1 + [P(u_l = +1)/P(u_l = -1)]^{\pm 1}\}} \\ &= \frac{e^{\pm L_{\alpha}(u_l)}}{\{1 + e^{\pm L_{\alpha}(u_l)}\}} \\ &= \frac{e^{-L_{\alpha}(u_l)/2}}{\{1 + e^{-L_{\alpha}(u_l)}\}} e^{u_l L_{\alpha}(u_l)/2} \\ &= A_l e^{u_l L_{\alpha}(u_l)/2}, \end{split}$$

Channel Coding Theory

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■ The L value depend of the value of u, thus

$$\begin{split} \gamma_{l}(s',s) &= A_{l}e^{u_{l}L_{a}(u_{l})/2}e^{-(E_{s}/N_{0})||\mathbf{r}_{l}-\mathbf{v}_{l}||^{2}}, \\ &= A_{l}e^{u_{l}L_{a}(u_{l})/2}e^{(2E_{s}/N_{0})(\mathbf{r}_{l}\cdot\mathbf{v}_{l})-||\mathbf{r}_{l}||^{2}-||\mathbf{v}_{l}||^{2}} \\ &= A_{l}e^{-(||\mathbf{r}_{l}||^{2}+n)}e^{u_{l}L_{a}(u_{l})/2}e^{(L_{c}/2)(\mathbf{r}_{l}\cdot\mathbf{v}_{l})} \\ &= A_{l}B_{l}e^{u_{l}L_{a}(u_{l})/2}e^{(L_{c}/2)(\mathbf{v}_{l}\cdot\mathbf{v}_{l})}, \quad l = 0, 1, \cdots, h-1, \\ \gamma_{l}(s',s) &= P(u_{l})e^{-(E_{s}/N_{0})||\mathbf{r}_{l}-\mathbf{v}_{l}||^{2}} \\ &= e^{-(E_{s}/N_{0})||\mathbf{r}_{l}-\mathbf{v}_{l}||^{2}}, \\ &= B_{l}e^{(L_{c}/2)(\mathbf{r}_{l}\cdot\mathbf{v}_{l})}, \quad l = h, h+1, \cdots, K-1, \end{split}$$

Again if we drop the constants

$$\gamma_{l}(s',s) = e^{u_{l}L_{n}(u_{l})/2}e^{(L_{c}/2)(\mathbf{r}_{l}\cdot\mathbf{v}_{l})}, \quad l = 0, 1, \dots, h-1,
\gamma_{l}(s',s) = e^{(L_{c}/2)(\mathbf{r}_{l}\cdot\mathbf{v}_{l})}, \quad l = h, h+1, \dots, K-1,$$

Channel Coding Theory



Using the log-domain enable using

$$\max^*(x, y) \equiv \ln(e^x + e^y) = \max(x, y) + \ln(1 + e^{-|x-y|})$$

And the log-domain metrics are

$$\begin{split} \gamma_l^*(s',s) &\equiv \ln \, \gamma_l(s',s) = \left\{ \begin{array}{l} \frac{u_l L_a(u_l)}{2} + \frac{L_c}{2} \mathbb{I}_l \cdot \mathbb{V}_l, & l = 0, 1, \cdots, h-1, \\ \frac{L_c}{2} \mathbb{I}_l \cdot \mathbb{V}_l, & l = h, h+1, \cdots, K-1, \\ \alpha_{l+1}^*(s) &= \max_{s' \in \sigma_l}^* [\gamma_l^*(s',s) + \alpha_l^*(s')], & l = 0, 1, \cdots, K-1, \\ \alpha_0^*(s) &\equiv \ln \, \alpha_0(s) = \left\{ \begin{array}{l} 0, & s = \emptyset \\ -\infty, & s \neq \emptyset, \\ \end{array} \right. \\ \beta_l^*(s') &= \max_{s \in \sigma_{l+1}}^* [\gamma_l^*(s',s) + \beta_{l+1}^*(s)], & l = K-1, K-2, \cdots, 0, \\ \beta_K^*(s) &\equiv \ln \, \beta_K(s) = \left\{ \begin{array}{l} 0, & s = \emptyset \\ -\infty, & s \neq \emptyset. \end{array} \right. \end{split}$$

Channel Coding Theory

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- Writing the expression for the pdf p(s', s, r) and the APP L-value $L(u_i)$ as:
- $p(s', s, r) = e^{\beta_{l+1}^*(s) + \gamma_l^*(s', s) + \alpha_l^*(s')}$

$$L(u_l) = \ln \left\{ \sum_{(s',s) \in \Sigma_l^+} e^{\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')} \right\}$$
$$- \ln \left\{ \sum_{(s',s) \in \Sigma_l^-} e^{\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')} \right\}$$

Channel Coding Theory



Using the following math expression

$$\max^*(x, y, z) \equiv \ln(e^x + e^y + e^z) = \max^*[\max^*(x, y), z],$$

■ We can formula the L value as

$$L(u_l) = \max_{(s',s) \in \Sigma_l^+} [\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')]$$
$$- \max_{(s',s) \in \Sigma_l^-} [\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')].$$

Channel Coding Theory

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Steps of Log-Domain BCJR algorithm

• Step1: calculate the forward and backward metrics using

$$\alpha_0^*(s) \equiv \ln \, \alpha_0(s) = \left\{ \begin{array}{ll} 0, & s = \emptyset \\ -\infty, & s \neq \emptyset, \end{array} \right.$$

$$\beta_K^*(s) \equiv \ln \beta_K(s) = \begin{cases} 0, & s = 0 \\ -\infty, & s \neq 0. \end{cases}$$

• Step 2 Compute the branch metric using

$$\gamma_l^*(s',s) \equiv \ln \gamma_l(s',s) = \begin{cases} \frac{u_l L_a(u_l)}{2} + \frac{L_c}{2} \mathbb{I}_l \cdot \mathbb{V}_l, & l = 0, 1, \dots, h-1, \\ \frac{L_c}{2} \mathbb{I}_l \cdot \mathbb{V}_l, & l = h, h+1, \dots, K-1, \end{cases}$$

Channel Coding Theory

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Steps of Log-Domain BCJR algorithm

■ Step3: calculate the forward metrics using

$$\alpha_{l+1}^*(s) = \max_{s' \in \sigma_l}^* [\gamma_l^*(s', s) + \alpha_l^*(s')], \quad l = 0, 1, \dots, K-1,$$

Step 4 Compute the backward metric using

$$\beta_l^*(s') = \ln \sum_{s \in \sigma_{l+1}} e^{[\gamma_l^*(s',s) + \beta_{l+1}^*(s)]}$$

Step 5 compute the APP-L values using

$$\begin{split} L(u_l) &= \max_{(s',s) \in \Sigma_l^+}^* [\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')] \\ &- \max_{(s',s) \in \Sigma_l^-}^* [\beta_{l+1}^*(s) + \gamma_l^*(s',s) + \alpha_l^*(s')]. \end{split}$$



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Steps of Log-Domain BCJR algorithm

• Step6: (Optional) compute the hard decisions using

$$\hat{u}_l = \left\{ \begin{array}{l} +1 \text{ if } L(u_l) > 0 \\ -1 \text{ if } L(u_l) < 0 \end{array} \right., \ l = 0, 1, \cdots, h-1.$$

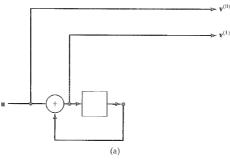
Channel Coding Theory



Example

 We will consider the BCJR decoding of a (2, 1, 1) systematic Recursive Convolutional code on AWGN with generator matrix

$$\mathbb{G}(D) = \begin{bmatrix} 1 & 1/(1+D) \end{bmatrix}$$

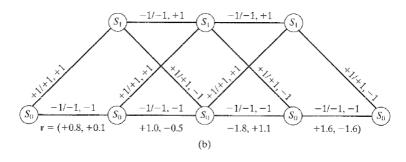


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- Let u = (u0, u1, u2, u3) denote the input vector of length 4 and v = (v0, v1, v2, v3) denotes the codeword of length 8.
- We assume Es/ N0 = $\frac{1}{4}$ (-6.02) dB
- The received vector r = (+0.8, +0.1; +1.0, -0.5; -1.8, +1.1; 1.6, -1.6)



Channel Coding Theory



- The rate of the terminated code is R = h/N = 3/8
- Eb/N0 = Es/RN0 = 2/3
- Assuming that the information bits are equaly likely La(ul) = 0
- Lc = 4 Eb/N0 = 1

$$\gamma_0^*(S_0, S_0) = \frac{-1}{2} L_a(u_0) + \frac{1}{2} \mathbb{r}_0 \cdot \mathbb{v}_0$$

$$= \frac{1}{2} (-0.8 - 0.1) = -0.45$$

$$\gamma_0^*(S_0, S_1) = \frac{+1}{2} L_a(u_0) + \frac{1}{2} \mathbb{r}_0 \cdot \mathbb{v}_0$$

$$= \frac{1}{2} (0.8 + 0.1) = 0.45$$

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$$\gamma_1^*(S_0, S_0) = \frac{-1}{2} L_a(u_1) + \frac{1}{2} \mathbb{I}_1 \cdot \mathbb{V}_1$$

$$= \frac{1}{2} (-1.0 + 0.5) = -0.25$$

$$\gamma_1^*(S_0, S_1) = \frac{+1}{2} L_a(u_1) + \frac{1}{2} \mathbb{I}_1 \cdot \mathbb{V}_1$$

$$= \frac{1}{2} (1.0 - 0.5) = 0.25$$

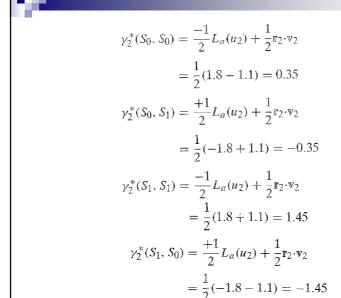
$$\gamma_1^*(S_1, S_1) = \frac{-1}{2} L_a(u_1) + \frac{1}{2} \mathbb{I}_1 \cdot \mathbb{V}_1$$

$$= \frac{1}{2} (-1.0 - 0.5) = -0.75$$

$$\gamma_1^*(S_1, S_0) = \frac{+1}{2} L_a(u_1) + \frac{1}{2} \mathbb{I}_1 \cdot \mathbb{V}_1$$

$$= \frac{1}{2} (1.0 + 0.5) = 0.75$$

Channel Coding Theory



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$$\begin{aligned} \gamma_3^*(S_0, S_0) &= \frac{+1}{2} \mathbb{I}_3 \cdot \mathbb{V}_3 \\ &= \frac{1}{2} (-1.6 + 1.6) = 0 \\ \gamma_3^*(S_1, S_0) &= \frac{+1}{2} \mathbb{I}_3 \cdot \mathbb{V}_3 \\ &= \frac{1}{2} (1.6 + 1.6) = 1.60. \end{aligned}$$

Now compute the log-domain forward metrics

$$\alpha_1^*(S_0) = [\gamma_0^*(S_0, S_0) + \alpha_0^*(S_0)] = -0.45 + 0 = -0.45$$

$$\alpha_1^*(S_1) = [\gamma_0^*(S_0, S_1) + \alpha_0^*(S_0)] = 0.45 + 0 = 0.45$$

$$\alpha_2^*(S_0) = \max^* \{ [\gamma_1^*(S_0, S_0) + \alpha_1^*(S_0)], [\gamma_1^*(S_1, S_0) + \alpha_1^*(S_1)] \}$$

$$= \max^* \{ [(-0.25) + (-0.45)], [(0.75) + (0.45)] \}$$

$$= \max^* (-0.70, +1.20) = 1.20 + \ln(1 + e^{-|-1.9|}) = 1.34$$

- Channel Coding Theory



$$\alpha_2^*(S_1) = \max^* \{ [\gamma_1^*(S_0, S_1) + \alpha_1^*(S_0)], [\gamma_1^*(S_1, S_1) + \alpha_1^*(S_1)] \}$$

= \text{max}^*(-0.20, -0.30) = -0.20 + \ln(1 + e^{-|0.1|}) = 0.44.

Similarly compute the log-domain backward metrics

$$\begin{split} \beta_3^*(S_0) &= \left[\gamma_3^*(S_0, S_0) + \beta_4^*(S_0) \right] = 0 + 0 = 0 \\ \beta_3^*(S_1) &= \left[\gamma_3^*(S_1, S_0) + \beta_4^*(S_0) \right] = 1.60 + 0 = 1.60 \\ \beta_2^*(S_0) &= \max^* \{ \left[\gamma_2^*(S_0, S_0) + \beta_3^*(S_0) \right], \left[\gamma_2^*(S_0, S_1) + \beta_3^*(S_1) \right] \} \\ &= \max^* \{ \left[(0.35) + (0) \right], \left[(-0.35) + (1.60) \right] \} \\ &= \max^* (0.35, 1.25) = 1.25 + \ln(1 + e^{-|-0.90|}) = 1.59 \\ \beta_2^*(S_1) &= \max^* \{ \left[\gamma_2^*(S_1, S_0) + \beta_3^*(S_0) \right], \left[\gamma_2^*(S_1, S_1) + \beta_3^*(S_1) \right] \} \\ &= \max^* (-1.45, 3.05) = 3.05 + \ln(1 + e^{-|-4.5|}) = 3.06 \end{split}$$

Channel Coding Theory

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$$\beta_1^*(S_0) = \max^* \{ [\gamma_1^*(S_0, S_0) + \beta_2^*(S_0)], [\gamma_1^*(S_0, S_1) + \beta_2^*(S_1)] \}$$

$$= \max^* (1.34, 3.31) = 3.44$$

$$\beta_1^*(S_1) = \max^* \{ [\gamma_1^*(S_1, S_0) + \beta_2^*(S_0)], [\gamma_1^*(S_1, S_1) + \beta_2^*(S_1)] \}$$

$$= \max^* (2.34, 2.31) = 3.02.$$

Finally we compute the app L -values

$$L(u_0) = [\beta_1^*(S_1) + \gamma_0^*(S_0, S_1) + \alpha_0^*(S_0)] - [\beta_1^*(S_0) + \gamma_0^*(S_0, S_0) + \alpha_0^*(S_0)]$$

$$= (3.47) - (2.99) = +0.48$$

$$L(u_1) = \max^* \{ [\beta_2^*(S_0) + \gamma_1^*(S_1, S_0) + \alpha_1^*(S_1)], [\beta_2^*(S_1) + \gamma_1^*(S_0, S_1) + \alpha_1^*(S_0)] \}$$

$$- \max^* \{ [\beta_2^*(S_0) + \gamma_1^*(S_0, S_0) + \alpha_1^*(S_0)], [\beta_2^*(S_1) + \gamma_1^*(S_1, S_1) + \alpha_1^*(S_1)] \}$$

$$= \max^* [(2.79), (2.86)] - \max^* [(0.89), (2.76)]$$

$$= (3.52) - (2.90) = +0.62$$

Channel Coding Theory



$$\begin{split} L(u_2) &= \max^* \{ [\beta_3^*(S_0) + \gamma_2^*(S_1, S_0) + \alpha_2^*(S_1)], [\beta_3^*(S_1) + \gamma_2^*(S_0, S_1) + \alpha_2^*(S_0)] \} \\ &- \max^* \{ [\beta_3^*(S_0) + \gamma_2^*(S_0, S_0) + \alpha_2^*(S_0)], [\beta_3^*(S_1) + \gamma_2^*(S_1, S_1) + \alpha_2^*(S_1)] \} \\ &= \max^* [(-1.01), (2.59)] - \max^* [(1.69), (3.49)] \\ &= (2.62) - (3.64) = -1.02. \end{split}$$

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Example

In the Max-log-MAP algorithm, the branch metrics remain as computed in (12.137). The approximation $\max^*(x, y) \approx \max(x, y)$ of (12.133) affects the computation of the forward and backward metrics (see (12.138) and (12.139), respectively), as follows:

$$\alpha_2^*(S_0) = \max(-0.70, +1.20) = 1.20$$

$$\alpha_2^*(S_1) = \max(-0.20, -0.30) = -0.20$$

$$\beta_2^*(S_0) = \max(0.35, 1.25) = 1.25$$

$$\beta_2^*(S_1) = \max(-1.45, 3.05) = 3.05$$

$$\beta_1^*(S_0) = \max(1.34, 3.31) = 3.31$$

$$\beta_1^*(S_1) = \max(2.34, 2.31) = 2.34.$$

Channel Coding Theory



Finding the L-values

Channel Coding Theory

$$L(u_0) = [\beta_1^*(S_1) + \gamma_0^*(S_0, S_1) + \alpha_0^*(S_0)] - [\beta_1^*(S_0) + \gamma_0^*(S_0, S_0) + \alpha_0^*(S_0)]$$

$$= (2.79) - (2.86) = -0.07$$

$$L(u_1) = \max\{[\beta_2^*(S_0) + \gamma_1^*(S_1, S_0) + \alpha_1^*(S_1)], [\beta_2^*(S_1) + \gamma_1^*(S_0, S_1) + \alpha_1^*(S_0)]\}$$

$$- \max\{[\beta_2^*(S_0) + \gamma_1^*(S_0, S_0) + \alpha_1^*(S_0)], [\beta_2^*(S_1) + \gamma_1^*(S_1, S_1) + \alpha_1^*(S_1)]\}$$

$$= \max\{[2.79), (2.86)] - \max\{[0.89), (2.76)\}$$

$$= (2.86) - (2.76) = +0.1000$$

$$L(u_2) = \max\{[\beta_3^*(S_0) + \gamma_2^*(S_1, S_0) + \alpha_2^*(S_1)], [\beta_3^*(S_1) + \gamma_2^*(S_0, S_1) + \alpha_2^*(S_0)]\}$$

$$- \max\{[\beta_3^*(S_0) + \gamma_2^*(S_0, S_0) + \alpha_2^*(S_0)], [\beta_3^*(S_1) + \gamma_2^*(S_1, S_1) + \alpha_2^*(S_1)]\}$$

$$= \max\{[-1.65), (2.45)] - \max[(1.55), (2.85)]$$

$$= (2.45) - (2.85) = -0.4000.$$
(12.143)



- Then he hard decision outputs of the Max-log-MAP decoder is
- u = (-1, +1, -1)

Channel Coding Theory