1

Digital Signal Processing

EE3900: Linear Systems and Signal Processing Indian Institute of Technology Hyderabad

Fourier Series

Lokesh Badisa AI21BTECH11005

12 Oct 2022

CONTENTS

1	Periodic Function	1
2	Fourier Series	1
3	Fourier Transform	4
4	Filter	5
5 to	Filter Design Abstract—This manual provides a simple Fourier Series	6 introduction

1. Periodic Function

Let

$$x(t) = A_0 |\sin(2\pi f_0 t)| \tag{1.1}$$

1.1 Plot x(t).

Solution:

wget https://raw.githubusercontent.com/ LokeshBadisa/EE3900-Linear-Systemsand-Signal-Processing/main/charger/ codes/1.1.py python3 1.1.py

1.2 Show that x(t) is periodic and find its period. **Solution:** If a signal x(t) is periodic then

$$x(t+T) = x(t) \tag{1.2}$$

where T is known as fundamental period. Since $|sin\theta|$ function is periodic, x(t) is also periodic.

Fundamental Period =
$$T = \frac{1}{2} \left(\frac{2\pi}{2\pi f_0} \right)$$
 (1.3)

$$=\frac{1}{2f_0}$$
 (1.4)

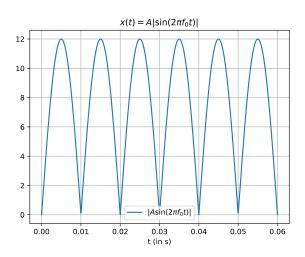


Fig. 1.1.

2. Fourier Series

Consider $A_0 = 12$ and $f_0 = 50$ for all numerical calculations.

2.1 If

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{J^{2\pi k f_0 t}}$$
 (2.1)

show that

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi k f_0 t} dt \qquad (2.2)$$

Solution: From (2.1),

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (2.3)

Mulitply $e^{-j2\pi l f_0 t}$ on both sides

$$x(t)e^{-j2\pi lf_0t} = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi kf_0t} e^{-j2\pi lf_0t}$$
 (2.4)

Integrate on both sides with respect to 't' between -T to T where T is fundamental time period of x(t). Using (1.4),

$$T = \frac{1}{2f_0} \tag{2.5}$$

$$\int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi k f_0 t} dt = \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} \sum_{k=-\infty}^{\infty} c_k e^{j2\pi (k-l)f_0 t} dt$$

$$= \sum_{k=-\infty}^{\infty} c_k \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{j2\pi (k-l)f_0 t} dt$$
(2.6)

The above integral:

$$\int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} e^{j2\pi(k-l)f_0t} dt = \begin{cases} 0 & k \neq l \\ \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} 1 dt & k = l \end{cases}$$
 (2.8)

$$\therefore \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-j2\pi kf_0t} dt = \left(\frac{1}{f_0}\right)c_k \quad (2.9)$$

$$\therefore c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-J2\pi k f_0 t} dt$$
 (2.10)

2.2 Find c_k for (1.1)

Solution: c_k can be calculated even simpler by using

$$c_k = 2f_0 \int_0^{\frac{1}{2f_0}} x(t)e^{-j2\pi k f_0 t} dt$$
 (2.11)

 $x(t) = A_0 \sin(2\pi f_0 t)$ in 0 to $\frac{1}{2f_0}$ region. Also,

$$\sin \theta = \frac{e^{\mathrm{j}\theta} - e^{-\mathrm{j}\theta}}{2\mathrm{1}} \tag{2.12}$$

Using (2.12),

$$c_{k} = 2f_{0} \int_{0}^{\frac{1}{2f_{0}}} A_{0} \left(\frac{e^{j2\pi f_{0}t} - e^{-j2\pi f_{0}t}}{2j}\right) e^{-j2\pi k f_{0}t} dt$$

$$= A_{0}f_{0} \int_{0}^{\frac{1}{2f_{0}}} \left(\frac{e^{j2\pi(1-k)f_{0}t} - e^{j2\pi(-1-k)f_{0}t}}{j}\right) dt$$

$$= A_{0}f_{0} \left[\frac{e^{j2\pi(1-k)f_{0}t}}{-2\pi(1-k)f_{0}}\Big|_{0}^{\frac{1}{2f_{0}}} - \frac{e^{j2\pi(-1-k)f_{0}t}}{-2\pi(-1-k)f_{0}}\Big|_{0}^{\frac{1}{2f_{0}}}\right]$$

$$= A_{0} \left[\frac{e^{j\pi(1-k)} - 1}{2\pi(k-1)} - \frac{e^{-j\pi(1+k)} - 1}{2\pi(k+1)}\right]$$

$$= \begin{cases} \frac{2A_{0}}{\pi(1-k^{2})} & k = even \\ 0 & k = odd \end{cases}$$

$$(2.17)$$

2.3 Verify (1.1) using python.

Solution:

wget https://raw.githubusercontent.com/ LokeshBadisa/EE3900-Linear-Systemsand-Signal-Processing/main/charger/ codes/2.3.py python3 2.3.py

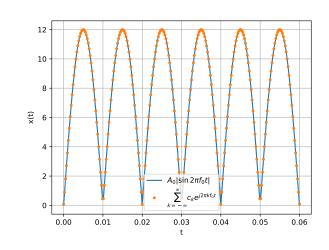


Fig. 2.3.

2.4 Show that

$$x(t) = \sum_{k=0}^{\infty} (a_k \cos 2\pi k f_0 t + b_k \sin 2\pi k f_0 t)$$
(2.18)

and obtain the formulae for a_k and b_k . **Solution:** Using (2.1),

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi k f_0 t}$$
 (2.19)

As,

$$e^{j2\pi k f_0 t} = \cos(2\pi k f_0 t) + j\sin(2\pi k f_0 t)$$
 (2.20)

Substituting leads to

$$x(t) = \sum_{k=-\infty}^{\infty} c_k \left[\cos(2\pi k f_0 t) + j \sin(2\pi k f_0 t) \right]$$
(2.21)

$$= \sum_{k=-\infty}^{\infty} c_k \cos(2\pi k f_0 t) + j c_k \sin(2\pi k f_0 t)$$

$$= \sum_{k=-\infty}^{-1} \left[c_k \cos \left(2\pi k f_0 t \right) + j c_k \sin \left(2\pi k f_0 t \right) \right] + c_0 + \sum_{k=1}^{\infty} \left[c_k \cos \left(2\pi k f_0 t \right) + \sum_{k=1}^{4A_0} \left[c_k \cos \left(2\pi k f_0 t \right) + \sum_{k=1}^{4A_0} \left[c_k \sin \left(2\pi k f_0 t \right) + \sum_{k=1}^{4A_0} \left[c_k \cos \left(2\pi k$$

$$= \sum_{k=1}^{\infty} \left[c_{-k} \cos \left(2\pi k f_0 t \right) - j c_{-k} \sin \left(2\pi k f_0 t \right) \right] + c$$

$$= c_0 + \sum_{k=0}^{\infty} \left[(c_k + c_{-k}) \cos(2\pi k f_0 t) + i(c_k - c_{-k}) \cos(2\pi k f_0 t) \right]$$

Replacing $(c_k + c_{-k}) \rightarrow a_k$ and $j(c_k - c_{-k}) \rightarrow$

$$= c_0 + \sum_{k=1}^{\infty} (a_k \cos 2\pi k f_0 t + b_k \sin 2\pi k f_0 t)$$
(2.26)

$$= \sum_{k=0}^{\infty} (a_k \cos 2\pi k f_0 t + b_k \sin 2\pi k f_0 t) \quad (2.27)$$

$$\therefore a_k = \begin{cases} c_k + c_{-k} & k \neq 0 \\ c_0 & k = 0 \end{cases}$$
 (2.28)

$$b_k = j(c_k - c_{-k}) (2.29)$$

Using (2.2),

$$c_k = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{-J^2\pi k f_0 t} dt$$
 (2.30)

$$c_{-k} = f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t)e^{j2\pi k f_0 t} dt$$
 (2.31) Fig. 2.6.

$$a_{k} = c_{k} + c_{-k} = f_{0} \int_{-\frac{1}{2f_{0}}}^{\frac{1}{2f_{0}}} x(t) \left[e^{-J^{2\pi k f_{0}t}} + e^{J^{2\pi k f_{0}t}} \right] dt$$
(2.32)

$$=2f_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t) \cos(2\pi k f_0 t) dt$$
(2.33)

Parallely,

$$b_k = -2jf_0 \int_{-\frac{1}{2f_0}}^{\frac{1}{2f_0}} x(t) \sin(2\pi k f_0 t) dt \quad (2.34)$$

2.5 Find a_k and b_k for (1.1)

Solution: Using (2.28) and (2.29) with (2.17),

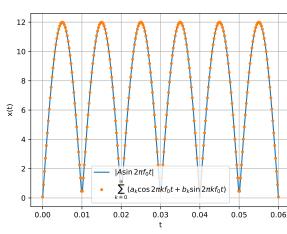
$$+\sum_{k=1}^{\infty} \left[c_{k} \cos \left(2\pi k f_{0} t \right) + \begin{cases} \frac{4A_{0}}{\pi (1-k^{2})} & k = even \\ c_{k} \sin \left(2\pi k f_{0} t \right) \\ \pi & k = 0 \end{cases} \right]$$

$$(2.35)$$

$$= \sum_{k=1}^{\infty} \left[c_{-k} \cos \left(2\pi k f_0 t \right) - j c_{-k} \sin \left(2\pi k f_0 t \right) \right] + c_0 + \sum_{k=1}^{\infty} \left[c_k \cos \left(2\pi k f_0 t \right) + j c_k \sin \left(2\pi k f_0 t \right) \right]^{-1}$$
(2.36)
$$(2.24) \quad 2.6 \text{ Verify (2.18) using python.}$$

python3 2.3.py

 $= c_0 + \sum_{k=0}^{\infty} \left[(c_k + c_{-k}) \cos(2\pi k f_0 t) + j(c_k - c_{-k}) \sin(2\pi k f_0 t) \right]$ wget https://raw.githubusercontent.com/ LokeshBadisa/EE3900-Linear-Systemsand-Signal-Processing/main/charger/ codes/2.6.py



3. Fourier Transform

3.1

$$\delta(t) = 0, \quad t \neq 0 \tag{3.1}$$

$$\int_{-\infty}^{\infty} \delta(t) \, dt = 1 \tag{3.2}$$

3.2 The Fourier Transform of g(t) is

$$G(f) = \int_{-\infty}^{\infty} g(t)e^{-j2\pi ft} dt \qquad (3.3)$$

3.3 Show that

$$g(t-t_0) \stackrel{\mathcal{F}}{\longleftrightarrow} G(f)e^{-j2\pi ft_0}$$
 (3.4)

(3.5)

Solution:

$$\mathcal{F}\{g(t-t_0)\} = \int_{-\infty}^{\infty} g(t-t_0)e^{-j2\pi ft} dt \quad (3.6)$$

$$=e^{-j2\pi ft_0}\int_{-\infty}^{\infty}g(t-t_0)e^{-j2\pi f(t-t_0)}dt \qquad (3.7)$$

$$= G(f)e^{-j2\pi f t_0}$$
 (3.8)

3.4 Show that

$$G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} g(-f)$$
 (3.9)

Solution: From the definition of Inverse Fourier Transform

$$g(t) = \int_{-\infty}^{\infty} G(f)e^{j2\pi ft} df \qquad (3.10)$$

Replace $t \rightarrow f$,

$$g(f) = \int_{-\infty}^{\infty} G(t)e^{j2\pi ft} dt$$
 (3.11)

Replace $f \rightarrow -f$,

$$g(-f) = \int_{-\infty}^{\infty} G(t)e^{-j2\pi ft} dt \qquad (3.12)$$

$$= \mathcal{F}\left\{G(t)\right\} \tag{3.13}$$

$$\therefore G(t) \stackrel{\mathcal{F}}{\longleftrightarrow} g(-f) \tag{3.14}$$

3.5 $\delta(t) \stackrel{\mathcal{F}}{\longleftrightarrow} ?$

Solution:

$$\mathcal{F}\left\{\delta(t)\right\} = \int_{-\infty}^{\infty} \delta(t)e^{-j2\pi ft} dt \qquad (3.15)$$

$$= \int_{-\infty}^{\infty} \delta(t) \, dt \tag{3.16}$$

Since $e^{-j2\pi ft} = 1$ for t=0 and remaining integrand is zero for t \neq 0.

$$= \int_{-\infty}^{\infty} \delta(t) \, dt \tag{3.18}$$

$$= 1$$
 (3.19)

3.6 $e^{-j2\pi f_0 t} \stackrel{\mathcal{F}}{\longleftrightarrow} ?$

Solution:

$$\mathcal{F}\left\{e^{-j2\pi f_0 t}\right\} = \int_{-\infty}^{\infty} 1 \cdot e^{-j2\pi (f + f_0)t} dt \qquad (3.20)$$

$$= \int_{-\infty}^{\infty} \mathcal{F}\left\{\delta(t)\right\} e^{-j2\pi(f+f_0)t} dt \qquad (3.21)$$

Using (3.9),

$$= \delta(-(f + f_0)) = \delta(f + f_0)$$
 (3.22)

 $3.7 \cos(2\pi f_0 t) \stackrel{\mathcal{F}}{\longleftrightarrow} ?$

Solution:

$$\mathcal{F}\left\{\cos(2\pi f_0 t)\right\} = \mathcal{F}\left\{\frac{e^{j2\pi f_0 t} + e^{-j2\pi f_0 t}}{2}\right\} (3.23)$$

Using (3.22),

$$= \frac{\delta(f+f_0) + \delta(f-f_0)}{2}$$
 (3.24)

3.8 Find the Fourier Transform of x(t) and plot it. Verify using python. Using (2.1),

$$\mathcal{F}\left\{x(t)\right\} = \mathcal{F}\left\{\sum_{k=-\infty}^{\infty} c_k e^{\mathrm{J}^2\pi k f_0 t}\right\}$$
(3.25)

$$X(f) = \sum_{k=-\infty}^{\infty} c_k \delta(f - kf_0)$$
 (3.26)

The following verifies the above equations:

wget https://github.com/LokeshBadisa/ EE3900-Linear-Systems-and-Signal-Processing/blob/main/charger/codes/3.8.py python3 3.8.py

3.9 Show that

$$rect(t) \stackrel{\mathcal{F}}{\longleftrightarrow} sinc(t)$$
 (3.27)

Verify using python.

Solution:

$$\mathcal{F}\left\{\operatorname{rect}(t)\right\} = \int_{-\infty}^{\infty} \operatorname{rect}(t)e^{-j2\pi ft} dt \qquad (3.28)$$

$$= \int_{-\frac{1}{2}}^{\frac{1}{2}} 1 \cdot e^{-j2\pi ft} dt \qquad (3.29)$$

$$= \frac{e^{-j2\pi ft}}{-j2\pi f} \Big|_{-\frac{1}{2}}^{\frac{1}{2}}$$
 (3.30)

$$= sinc(t)$$
 (3.31)

python3 3.10.py

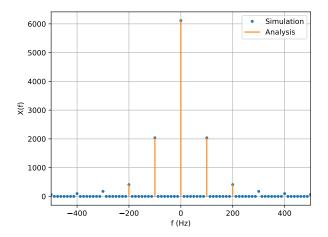


Fig. 3.8.

wget https://github.com/LokeshBadisa/ EE3900-Linear-Systems-and-Signal-Processing/blob/main/charger/codes/3.9.py python3 3.9.py

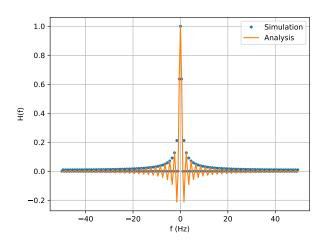


Fig. 3.9.

3.10 $\operatorname{sinc}(t) \stackrel{\mathcal{F}}{\longleftrightarrow}$?. Verify using python. **Solution:** Using (3.31), (3.9) and even property

of rect function,

$$\operatorname{sinc}(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \operatorname{rect}(t)$$
 (3.32)

wget https://github.com/LokeshBadisa/ EE3900-Linear-Systems-and-Signal-Processing/blob/main/charger/codes/3.10. py

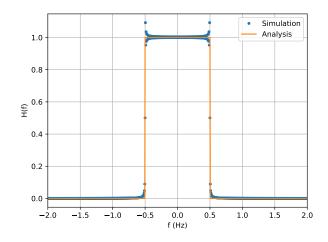
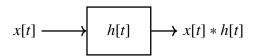


Fig. 3.10.

4. Filter

4.1 Find H(f) which transforms x(t) to DC 5V. **Solution:**



$$X(f) \longrightarrow H(f) \longrightarrow X(f)H(f)$$

$$X(f)H(f) = V_0\delta(f) \tag{4.1}$$

Above equation indicates that H(f) will pass X(f) for f=0.

 \therefore H(f) should be a low pass filter.

$$|H(f)| = \frac{V_0}{\left(\frac{2A_0}{\pi}\right)} = \frac{V_0\pi}{2A_0}$$
 (4.2)

$$H(f) = \frac{V_0 \pi}{2A_0}$$
 in $-2f_0 \le f \le 2f_0$ (4.3)

$$H(f) = \frac{V_0 \pi}{2A_0} \operatorname{rect}\left(\frac{f}{4f_0}\right)$$
 (4.4)

4.2 Find h(t).

Solution: Using (4.4) and (3.32),

$$h(t) = \frac{2V_0 \pi f_0}{A_0} \text{sinc} (4f_0 t)$$
 (4.5)

4.3 Verify your result using through convolution. **Solution:**

wget https://github.com/LokeshBadisa/ EE3900-Linear-Systems-and-Signal-Processing/blob/main/charger/codes/4.3.py python3 4.3.py

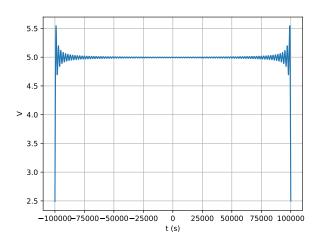


Fig. 4.3.

5. FILTER DESIGN

5.1 Design a Butterworth filter for H(f). Solution:

$$|H(f)| = \frac{1}{\sqrt{1 + \left(\frac{f}{f}\right)^{2n}}}$$
 (5.1)

n = Order of the filter

$$A = 10log_{10}|H(f)|^2 (5.2)$$

$$A_1 = -10log_{10} \left[1 + \left(\frac{f_1}{f_c} \right)^{2n} \right]$$
 (5.3)

$$A_2 = -10log_{10} \left[1 + \left(\frac{f_2}{f_c} \right)^{2n} \right]$$
 (5.4)

Solving for n,

$$n = \frac{\log\left(\frac{10^{-A_1/10} - 1}{10^{-A_2/10} - 1}\right)}{2\log\left(\frac{f_1}{f_2}\right)}$$
(5.5)

Assuming values,

$$A_1 = -1dB \tag{5.6}$$

$$A_2 = -10dB \tag{5.7}$$

$$f_1 = 50Hz \tag{5.8}$$

$$f_2 = 100Hz$$
 (5.9)

We get,

$$n = 2.5596 \approx 3 \tag{5.10}$$

Putting n = 3 in above equations, We get

$$f_1^{'} = 62.628Hz \tag{5.11}$$

$$f_2' = 69.336Hz \tag{5.12}$$

$$f_c = \sqrt{f_1' f_2'} = 65.897 Hz \tag{5.13}$$

5.2 Design a Chebyschev filter for H(f).

Solution:

$$|H_n(f)| = \frac{1}{\sqrt{1 + \epsilon^2 c_n^2 \left(\frac{f}{f_0}\right)}}$$
 (5.14)

where

 c_n =nth order Chebyshev polynomial, ϵ =ripple factor which is related to passband ripple in δ as $\sqrt{10^{\delta/10}-1}$

$$A_1 = -10\log_{10}\left[1 + \epsilon^2 c_n^2 \left(\frac{f}{f_0}\right)\right]$$
 (5.15)

$$\implies c_n \left(\frac{f}{f_0} \right) = \frac{\sqrt{10^{-\frac{A_1}{10}} - 1}}{\epsilon} \tag{5.16}$$

$$\implies n = \frac{\cosh^{-1}\left(\frac{\sqrt{10^{-\frac{A_1}{10}}-1}}{\epsilon}\right)}{\cosh^{-1}\left(\frac{f}{f_0}\right)} \tag{5.17}$$

Considering

$$f_0 = 65Hz (5.18)$$

$$f = 120Hz \tag{5.19}$$

$$A_1 = 10dB \tag{5.20}$$

$$\delta = 0.2dB \tag{5.21}$$

$$\epsilon = 0.217 \tag{5.22}$$

On solving we get n=2.71302, $\implies n=3$.

5.3 Design a circuit for your Butterworth filter. **Solution:** Using Cauer Topology,

$$C_k = 2\sin\left[\frac{(2k-1)}{2n}\pi\right] \tag{5.23}$$

$$L_k = 2\sin\left[\frac{(2k-1)}{2n}\pi\right] \tag{5.24}$$

5.4 Design a circuit for your Chebyschev filter.

Solution: the normalised values obtained are $C_1 = 1.2276F$, $L_2 = 1.1525H$, $C_3 = 1.2276F$. De-normalizing the values we get:

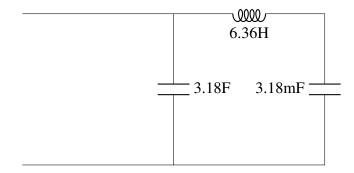


Fig. 5.3.

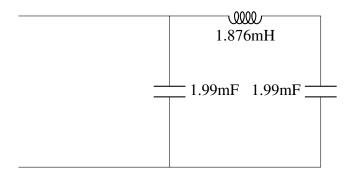


Fig. 5.4.