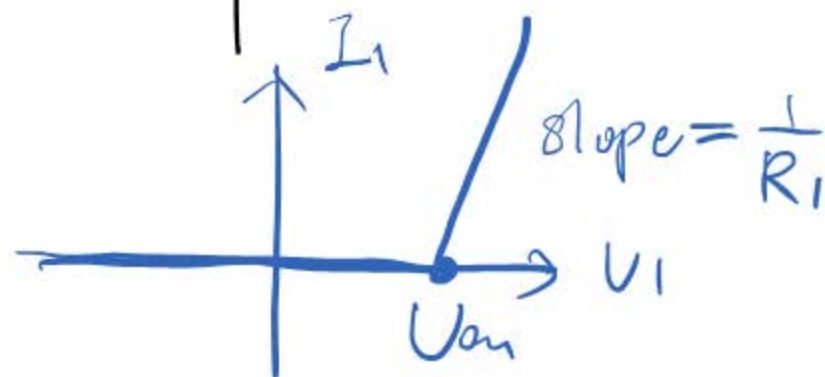
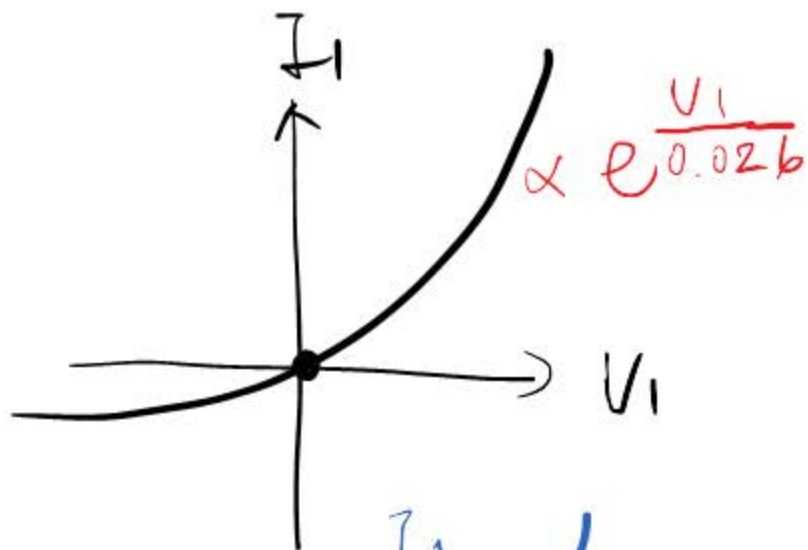
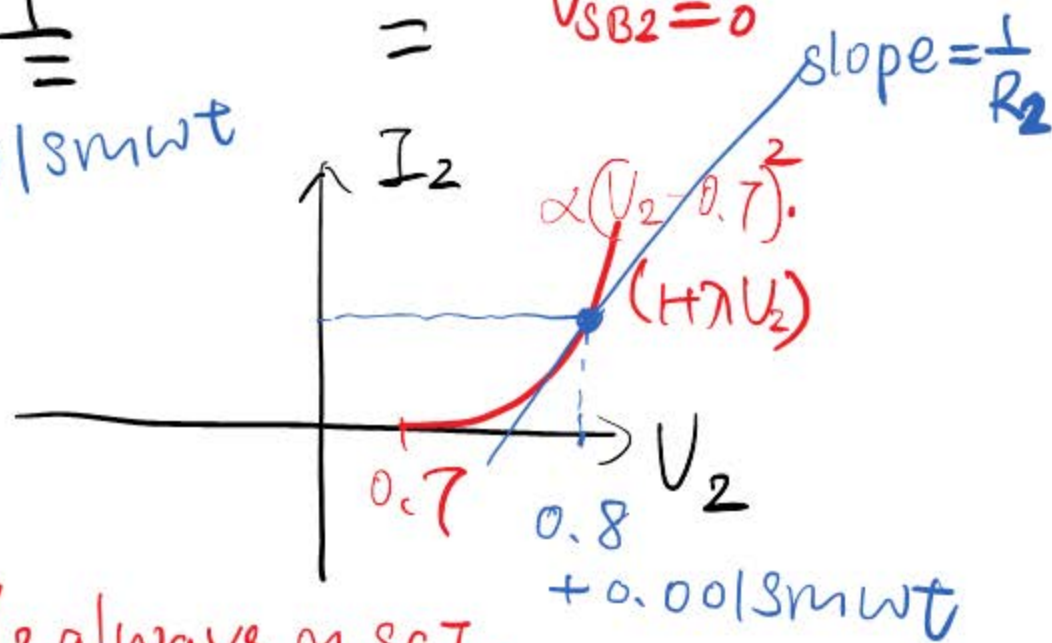
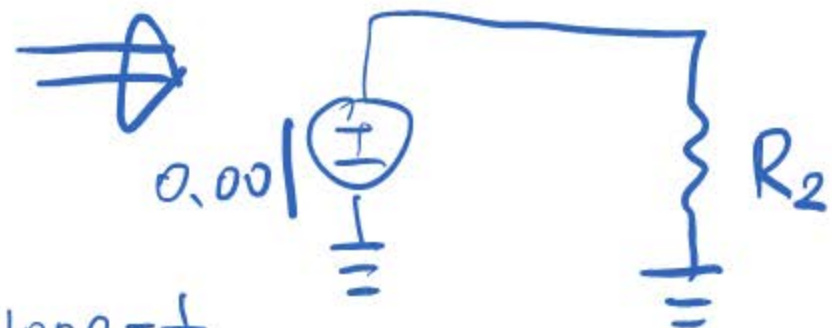
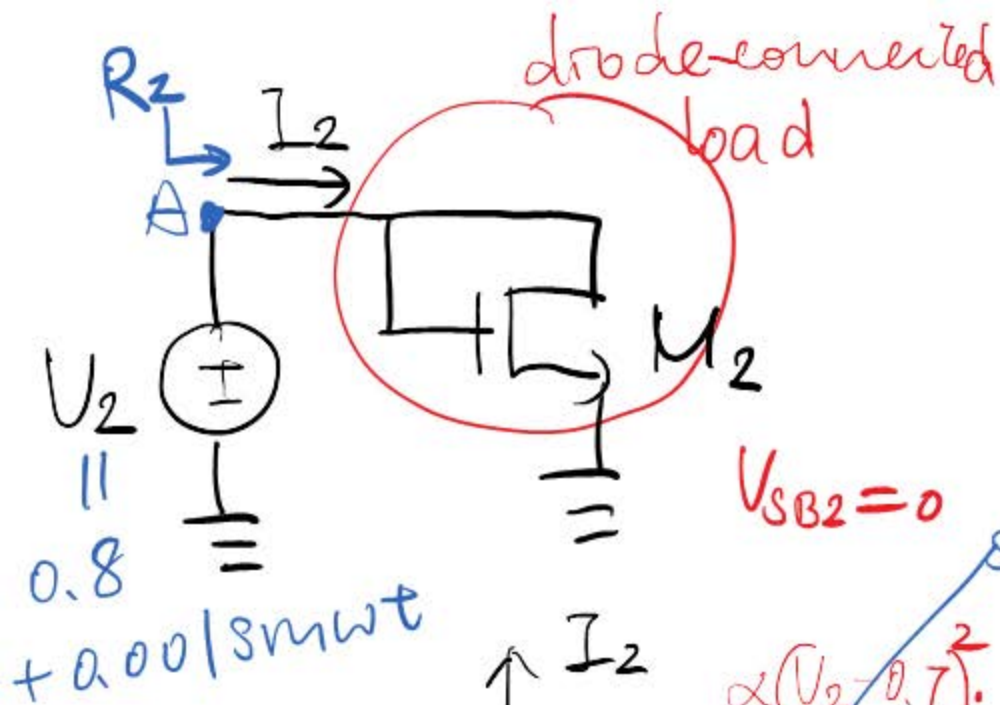


$\frac{1}{R_1}$ (the small signal impedance from node A to small signal ground)

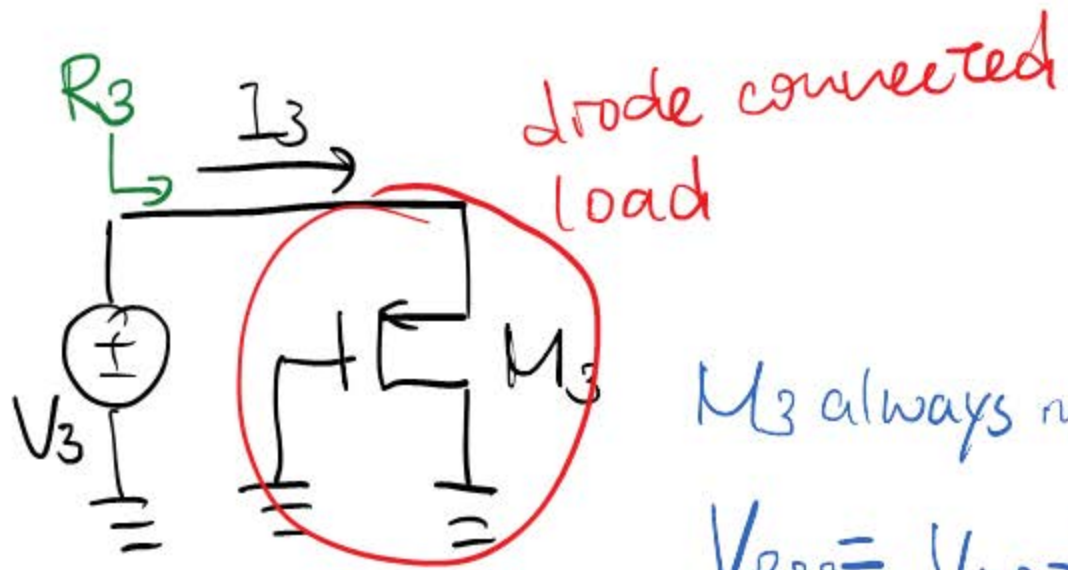




M_2 always in sat.

$$R_2 = r_{o2} \parallel \frac{1}{g_{m2}}$$

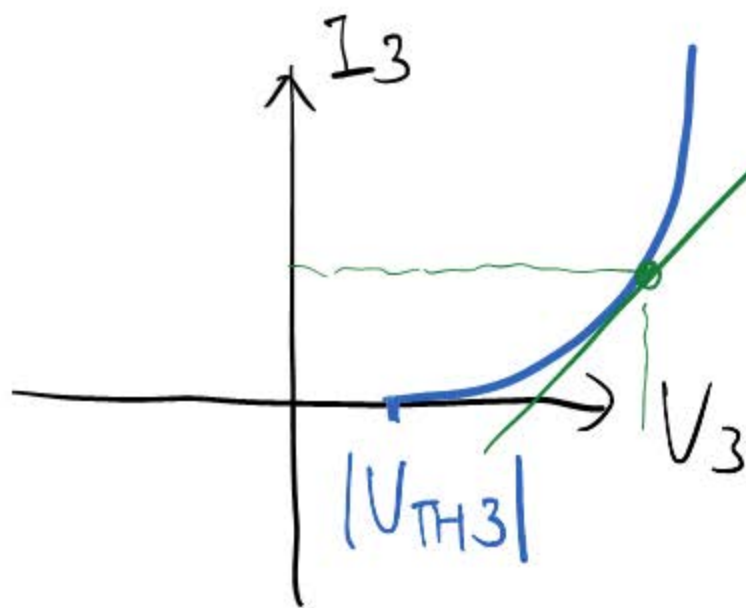
$$I_2 = \frac{1}{2} \mu_n C_{ox} \frac{W}{L_{eff}} (V_2 - 0.7)^2 (1 + \lambda V_2)$$



$$R_3 = r_{o3} \parallel \left(\frac{1}{g_{m3} + g_{mb3}} \right)$$

M_3 always in Sat.

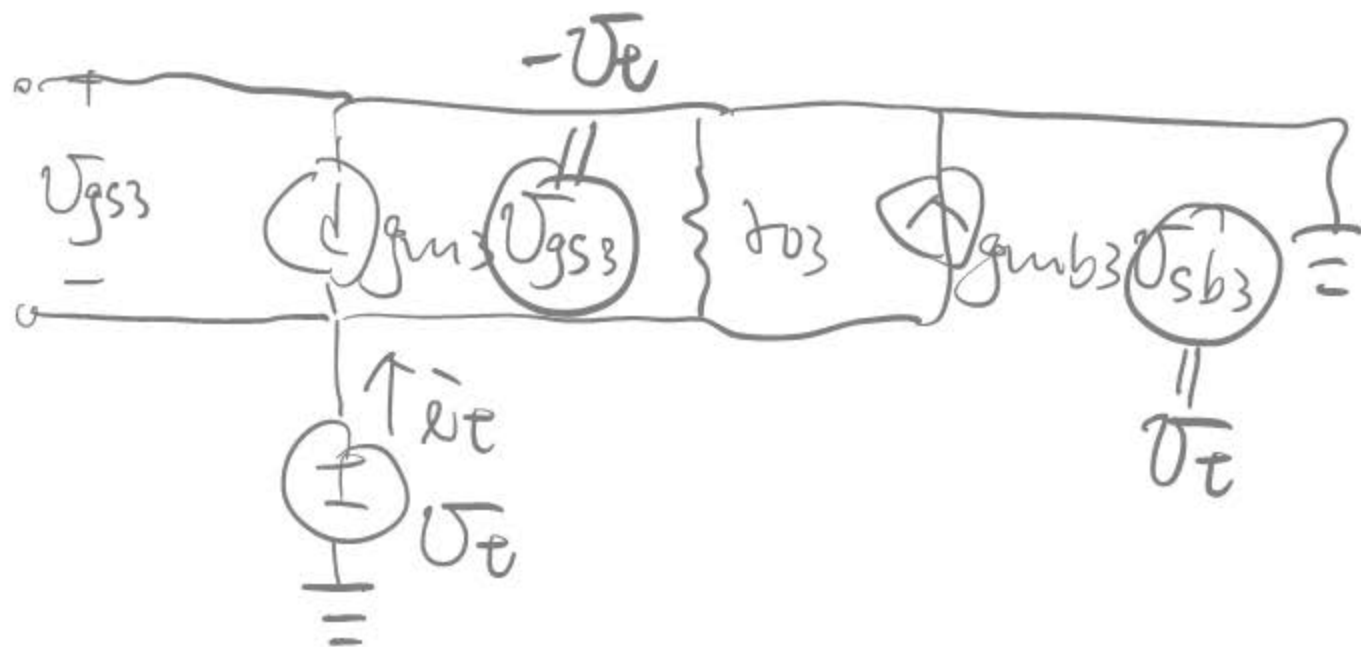
$$V_{BS3} = V_{DD} - V_3 > 0, |V_{TH3}| > 0.8$$



$$\text{slope} = \frac{1}{R_3}$$

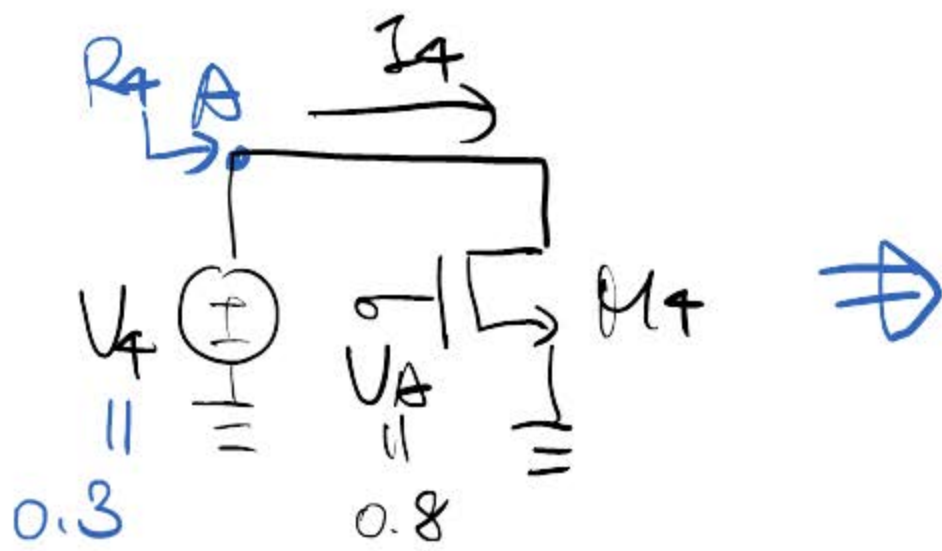
$$I_3 = \frac{1}{2} \mu_p C_{ox} \left(\frac{W}{L_{eff}} \right)_3 (V_3 - |V_{TH3}|)^2 (1 + \lambda V_3)$$

$$(V_3 - |V_{TH3}|)^2 (1 + \lambda V_3)$$

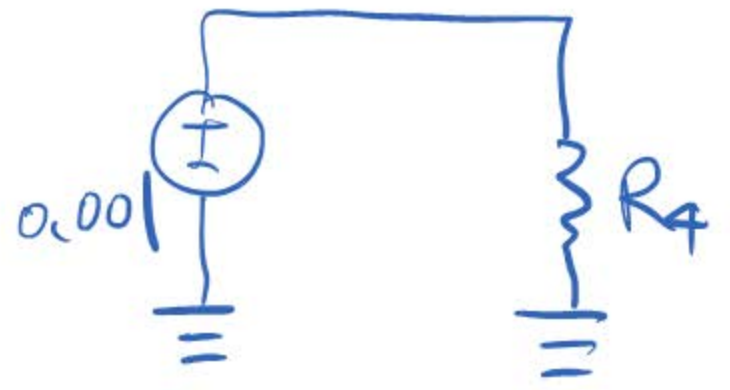


$$\hat{v}_e = (g_{m3} + g_{mb3})V_e + \frac{V_e}{r_{o3}}$$

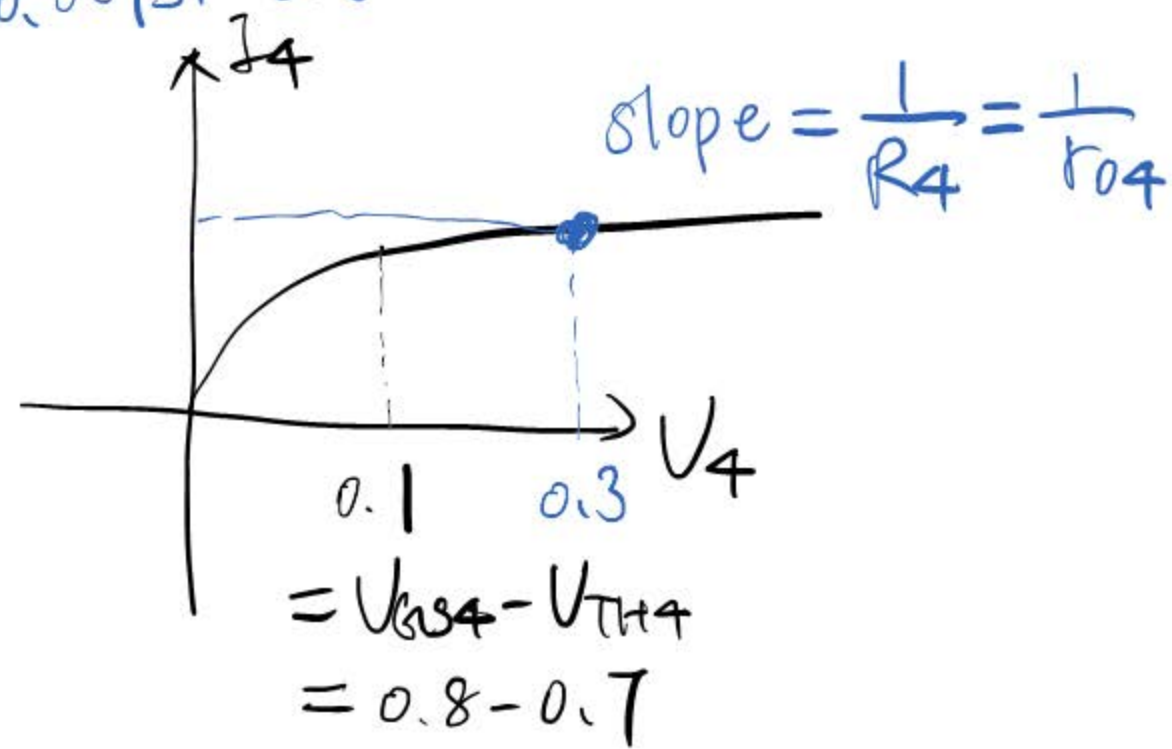
$$R_3 = V_e / \hat{v}_e = r_{o3} \parallel \left(\frac{1}{g_{m3} + g_{mb3}} \right)$$

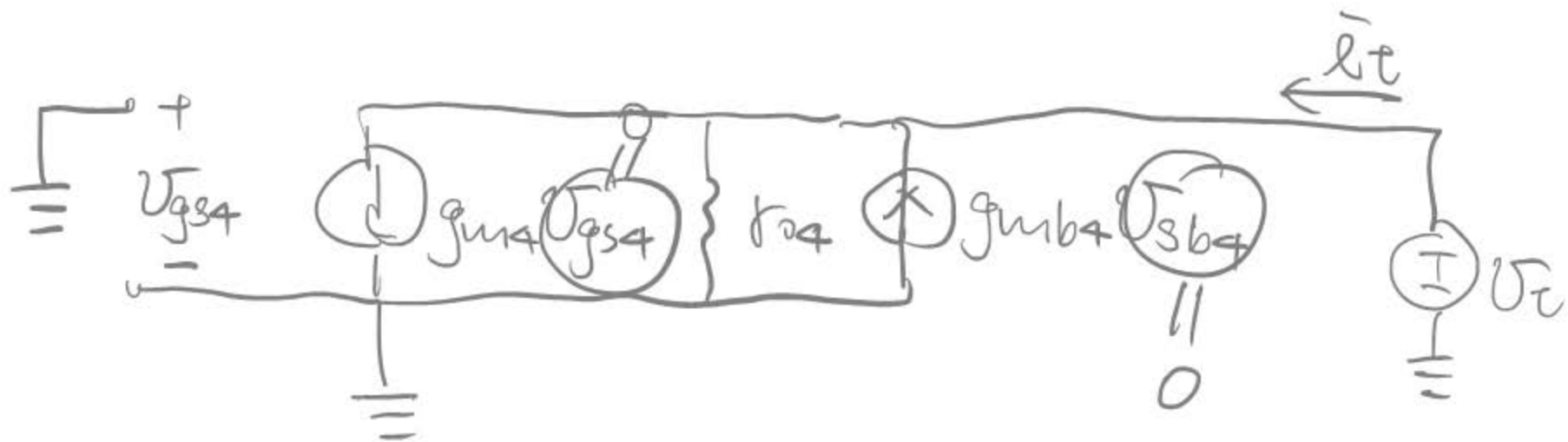


small signal

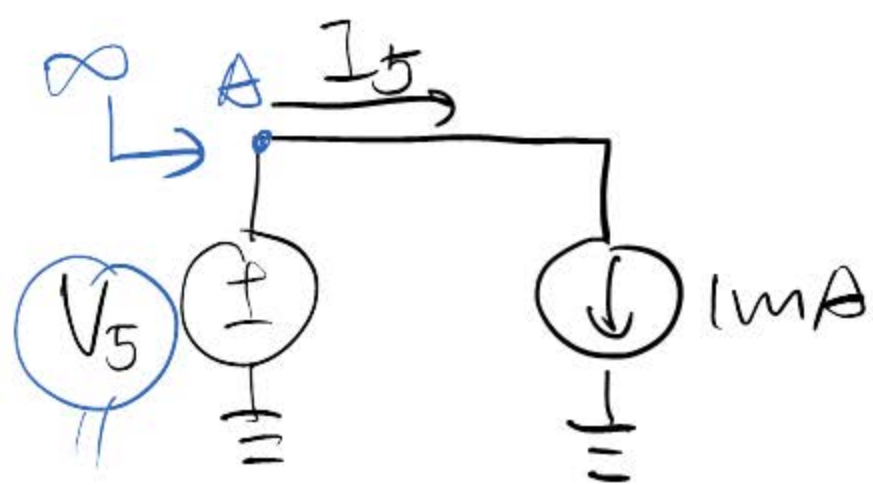


+0.001smwt

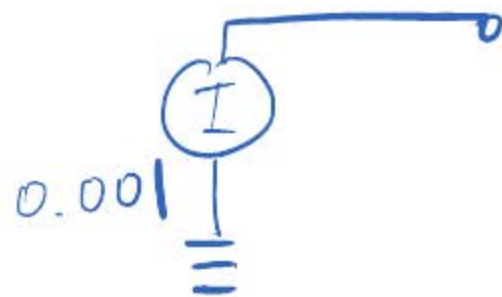




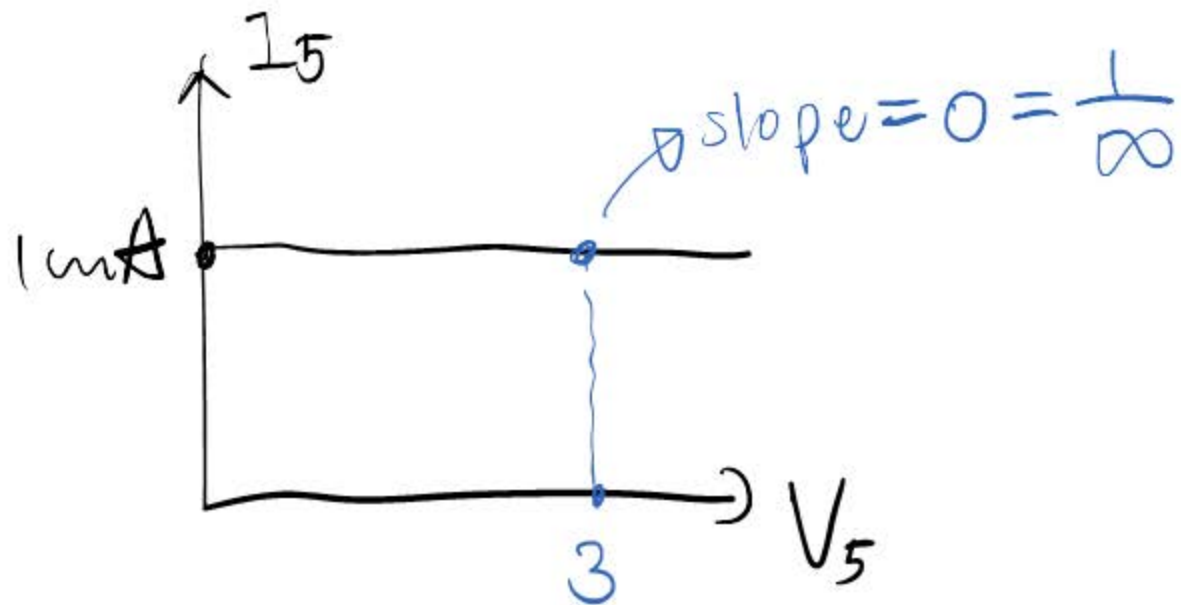
$$R_4 = \frac{V_e}{i_e} = r_{o4}$$

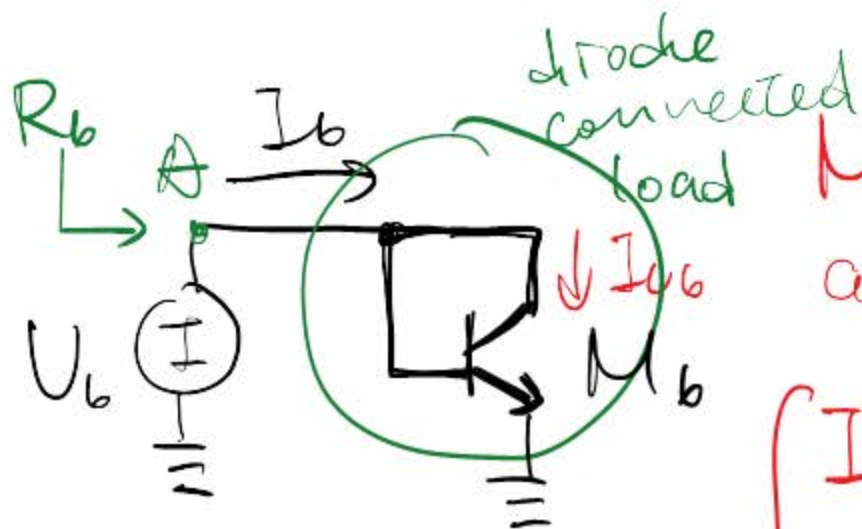


small signal



$3 + 0.001 \text{ smwt}$

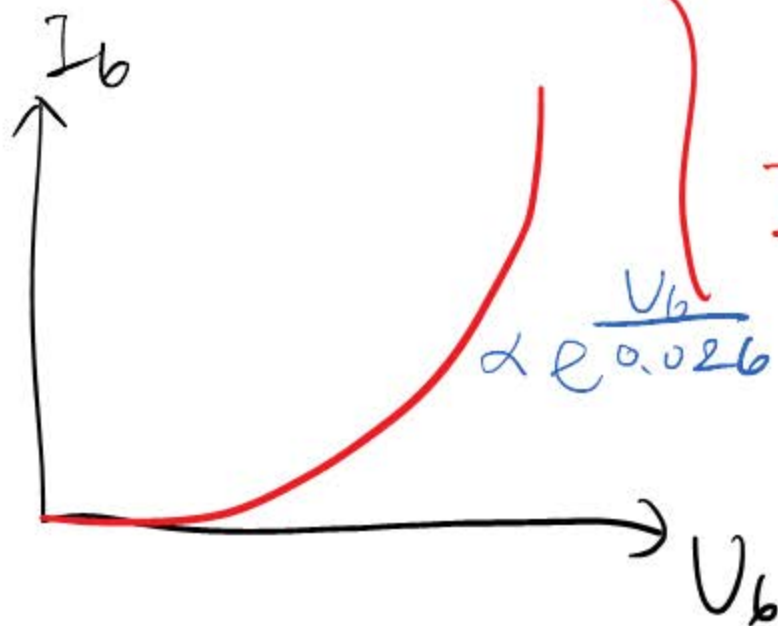




M_b always in forward-active region

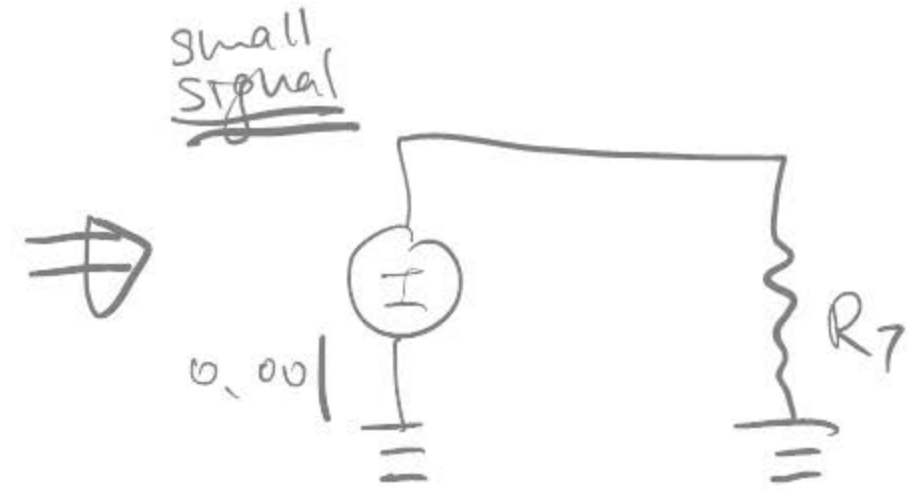
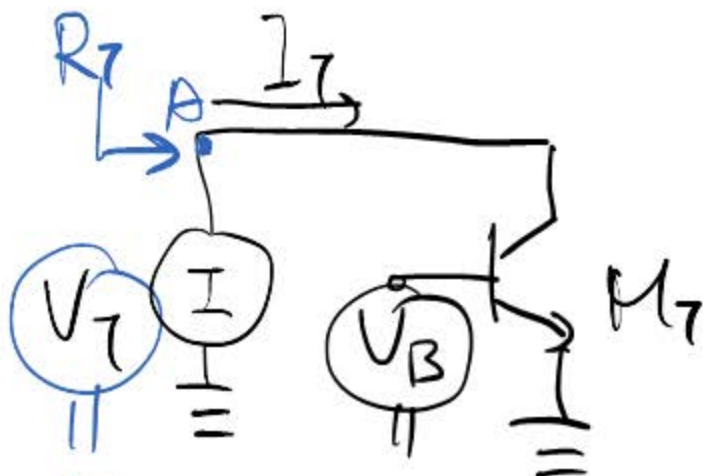
$$I_{c6} = I_s \left(e^{\frac{V_b}{0.026}} - 1 \right)$$

$$I_b = I_{c6} + \frac{I_{c6}}{\beta} \left(1 + \frac{V_b}{V_A} \right)$$

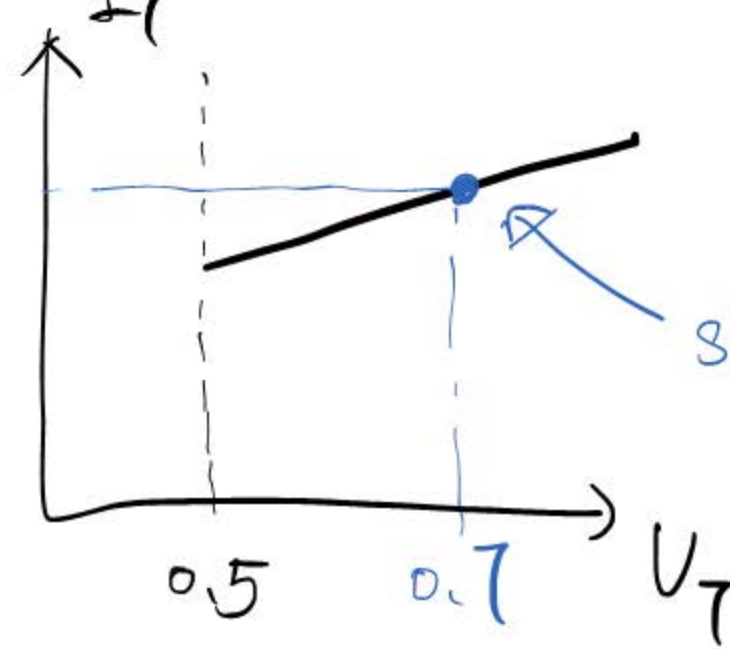




$$R_b = \frac{U_e}{i_e} = r_{\pi b} // r_{o b} // \frac{1}{g_{m b}}$$



0.7
+ 0.001 sin wt



$$I_7 = I_s \left(e^{\frac{0.5}{0.026}} - 1 \right) \left(1 + \frac{V_7}{V_A} \right)$$

$$\text{slope} = \frac{1}{R_7} = \frac{1}{r_{o7}}$$

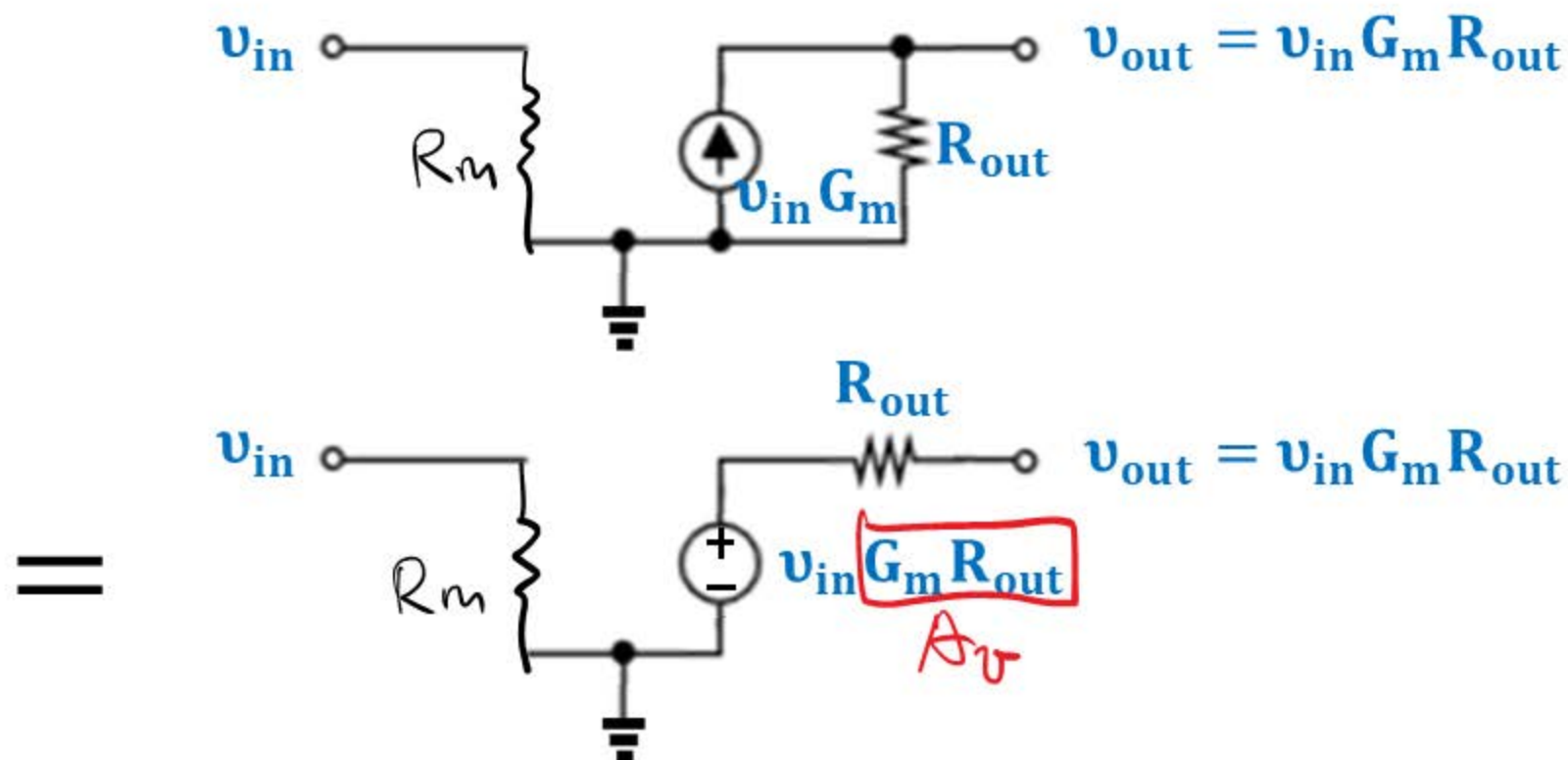


$$R_7 = \frac{U_e}{\hat{i}_e} = r_{o1}$$

Degeneration

Common-Source with Source Degradation

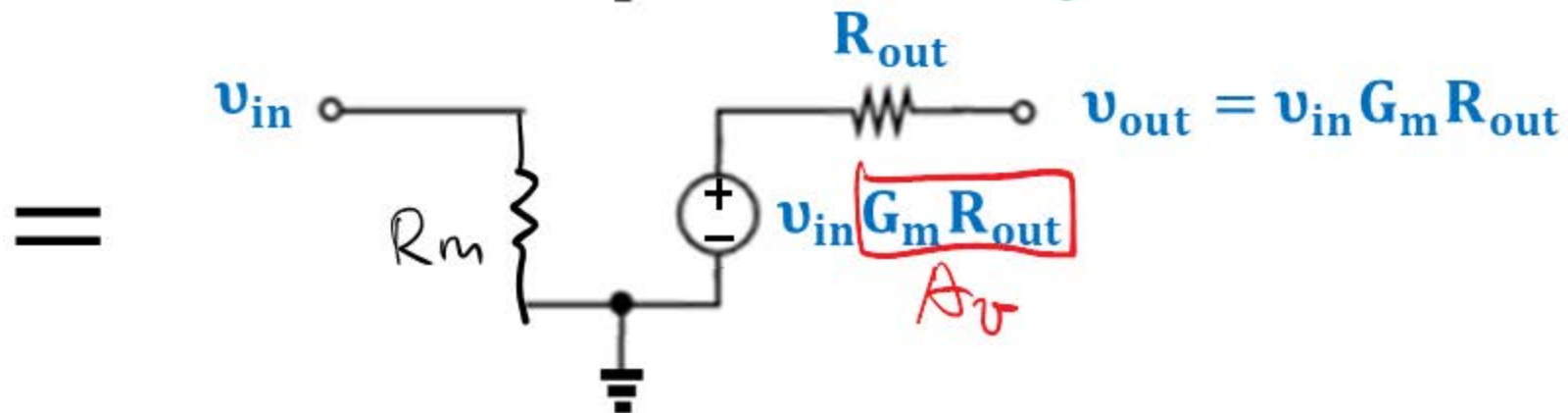
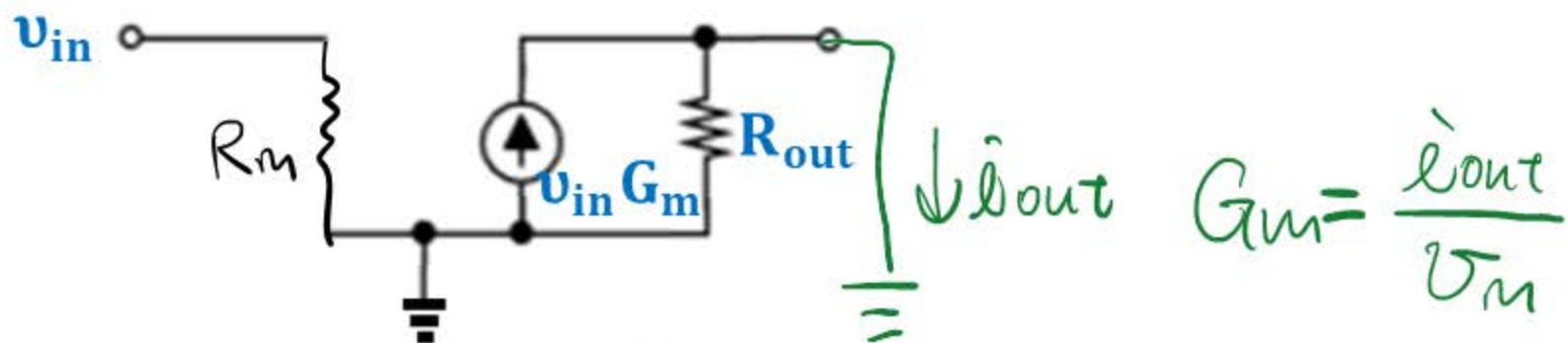
Amplifier Equivalent Circuit



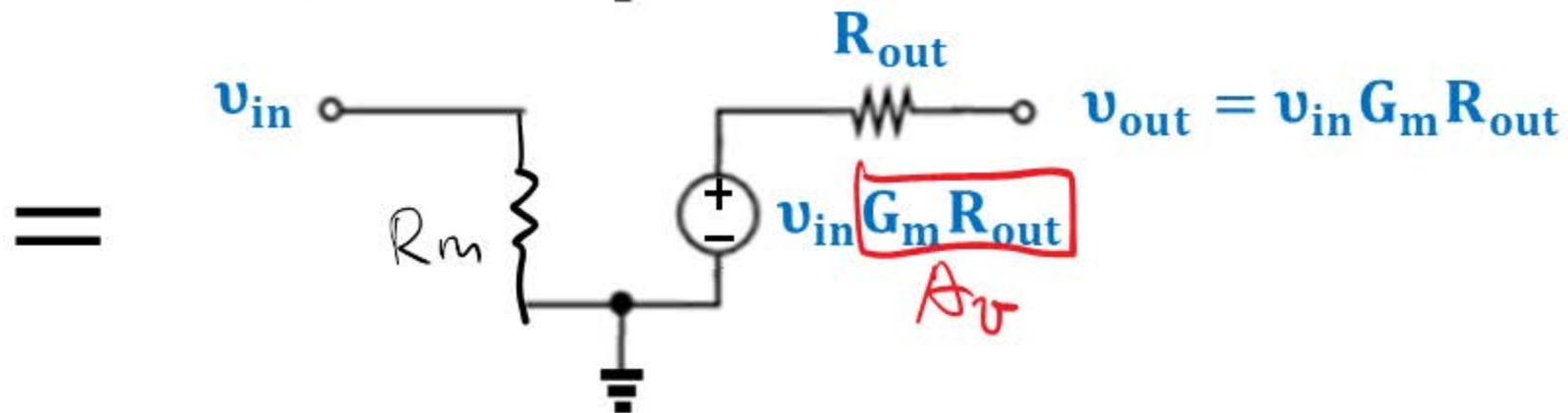
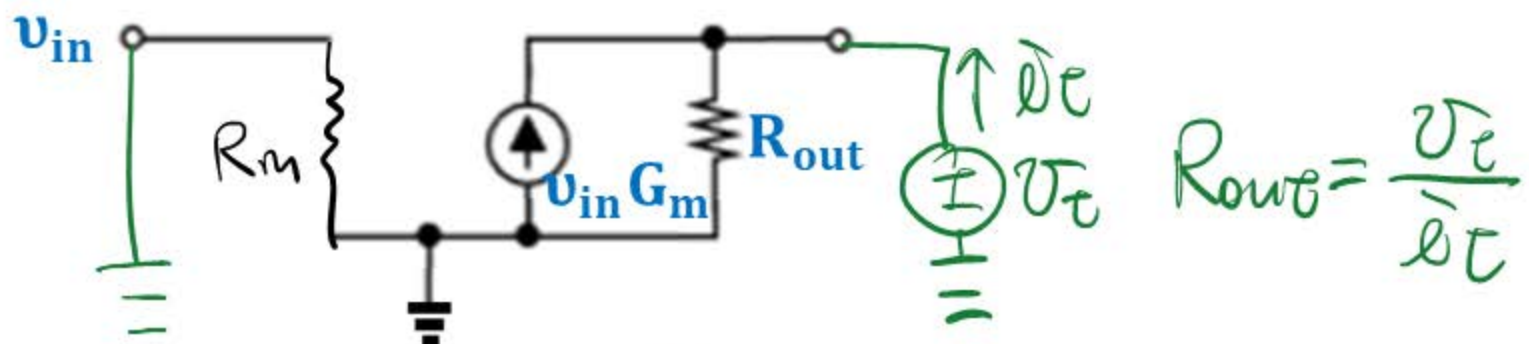
- How to calculate G_m ? v_{out} shorted to ground. $G_m = i_{out}/v_{in}$
- How to calculate R_{out} ? v_{in} shorted to ground and v_{out} connected to v_{test} .

$$R_{out} = v_{test}/i_{test}$$

Amplifier Equivalent Circuit



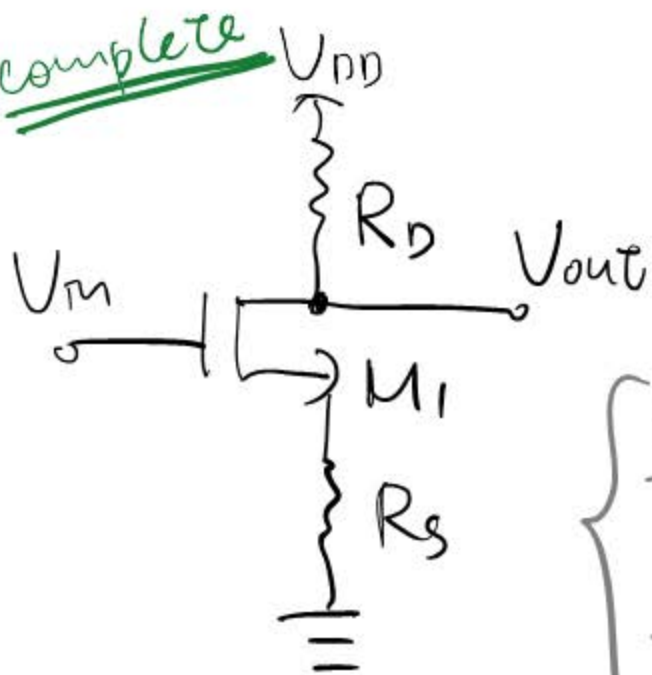
Amplifier Equivalent Circuit



complete

$$\eta \neq 0, r \neq 0$$

$$A_v = ?$$

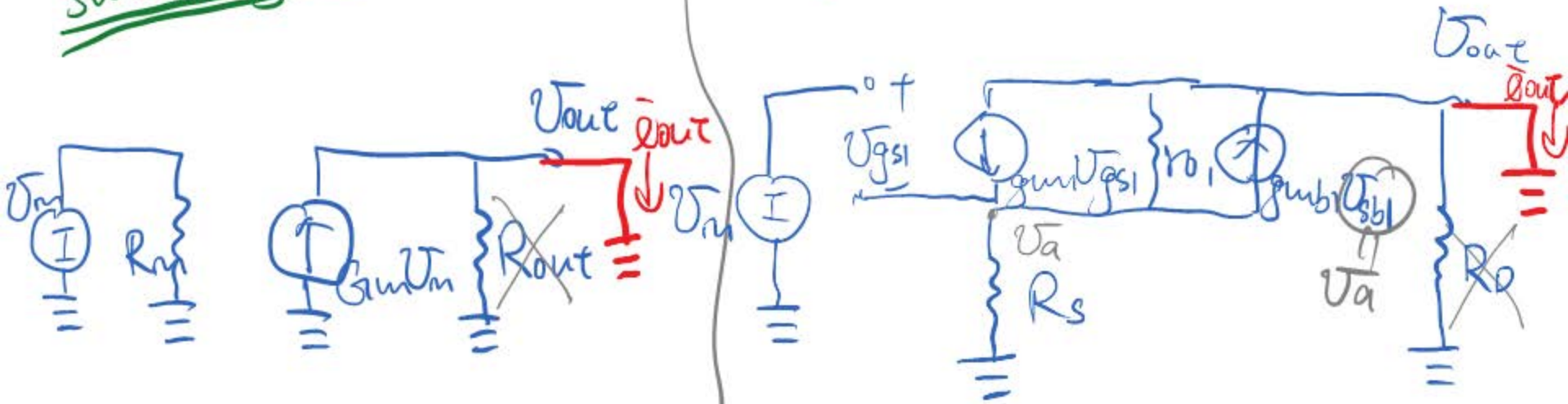


$$G_m = \bar{Q}_{out} / V_m$$

$$\begin{cases} \frac{V_a}{R_s} + (V_a - V_m)g_{m1} + \frac{V_a}{r_{o1}} + g_{m1}V_a = 0 \\ -\frac{V_a}{R_s} = \bar{Q}_{out} \end{cases}$$

Equivalent small signal

Small signal

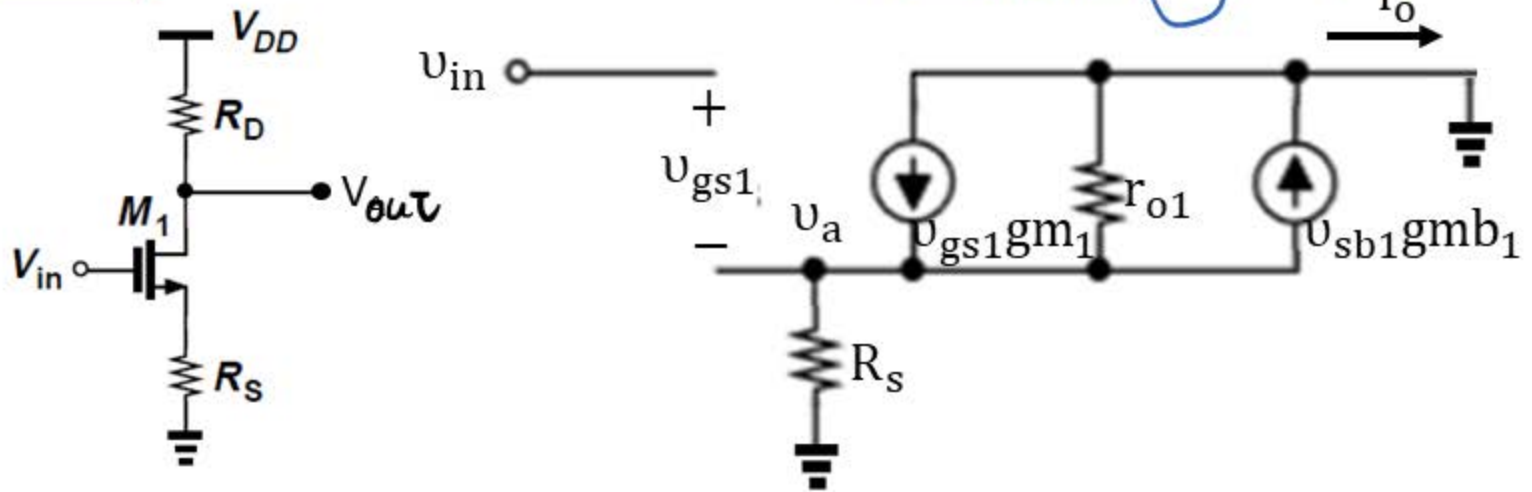


Common-Source with Source Degradation

Small-signal
Analysis

$\lambda \neq 0$ $\gamma \neq 0$

small signal circuit
for calculating G_m



$$\begin{cases} i_o = \frac{-v_a}{R_S} \\ (v_{in} - v_a)g_{m1} + i_o = \frac{v_a}{r_{o1}} + v_a g_{mb1} \end{cases}$$

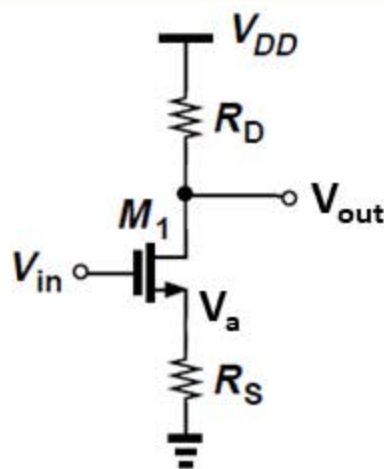
$$G_m = \frac{i_o}{v_{in}} = \frac{-g_{m1}r_{o1}}{R_S + r_{o1} + (g_{m1} + g_{mb1})r_{o1}R_S} \approx -\frac{1}{R_S}$$

intrinsic g_m of M_1

if $g_{mb1} \ll g_{m1}$
if $(g_{m1} + g_{mb1})r_{o1}R_S \gg r_{o1}$ and R_S

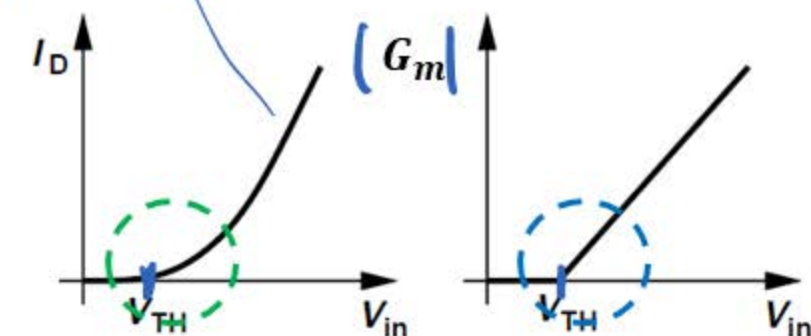
Common-Source with Source Degradation

DC Analysis

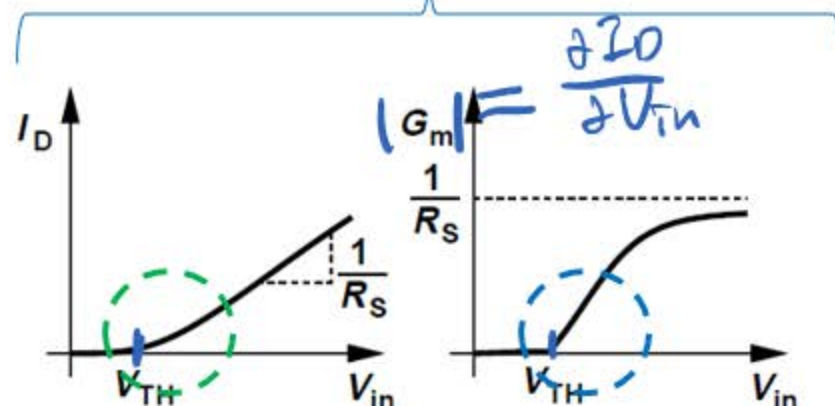


$\propto (V_m - V_{TH})^2$

$R_S = 0$

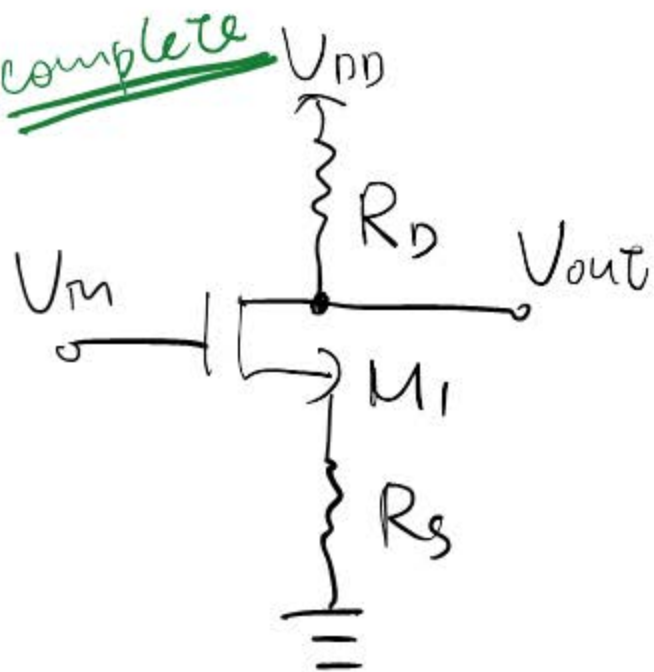


$R_S \neq 0$



- At low V_{in} (gm small), turn-on behavior of $R_S \neq 0$ is similar to that of $R_S = 0$.
- At large V_{in} (gm large), the effect of R_S , i.e. degradation, becomes more significant.
- $V_{in} = 0 \text{ V} \rightarrow M_1$ off, no current flowing $\rightarrow V_a = 0 \text{ V}$ and $V_{out} = V_{DD}$

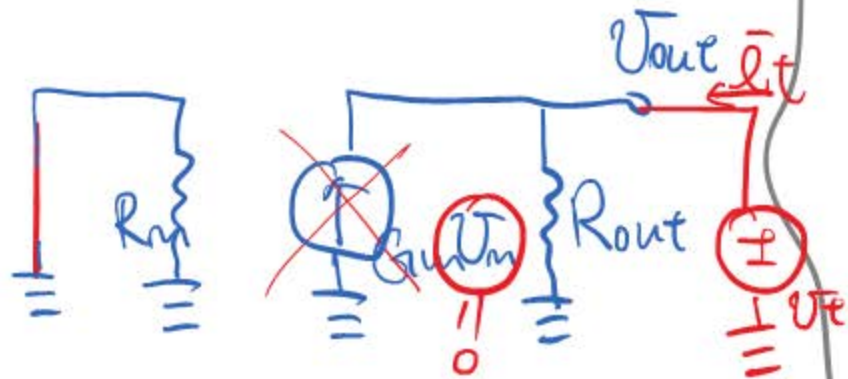
complete



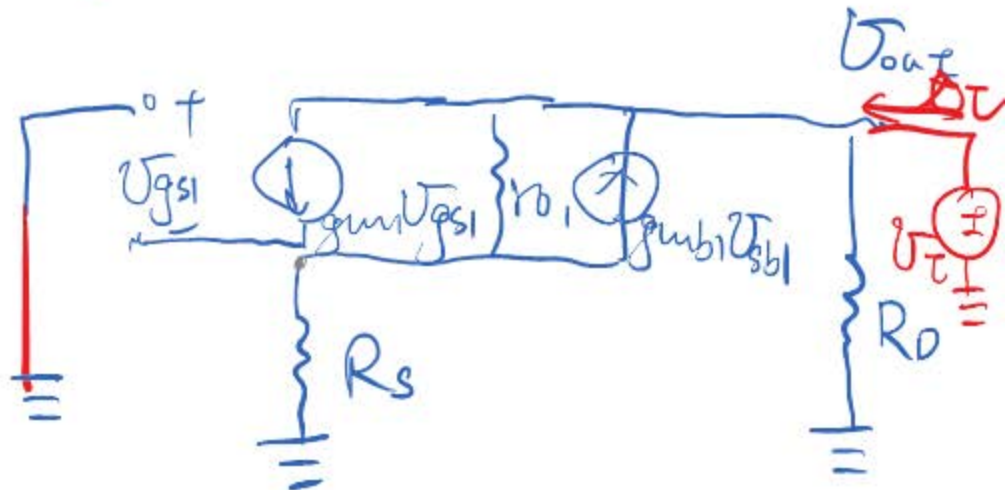
$\lambda \neq 0, r \neq 0$
 $A_v = ?$

$R_{out} = \cancel{v_e} / \bar{i}_e$

Equivalent small signal



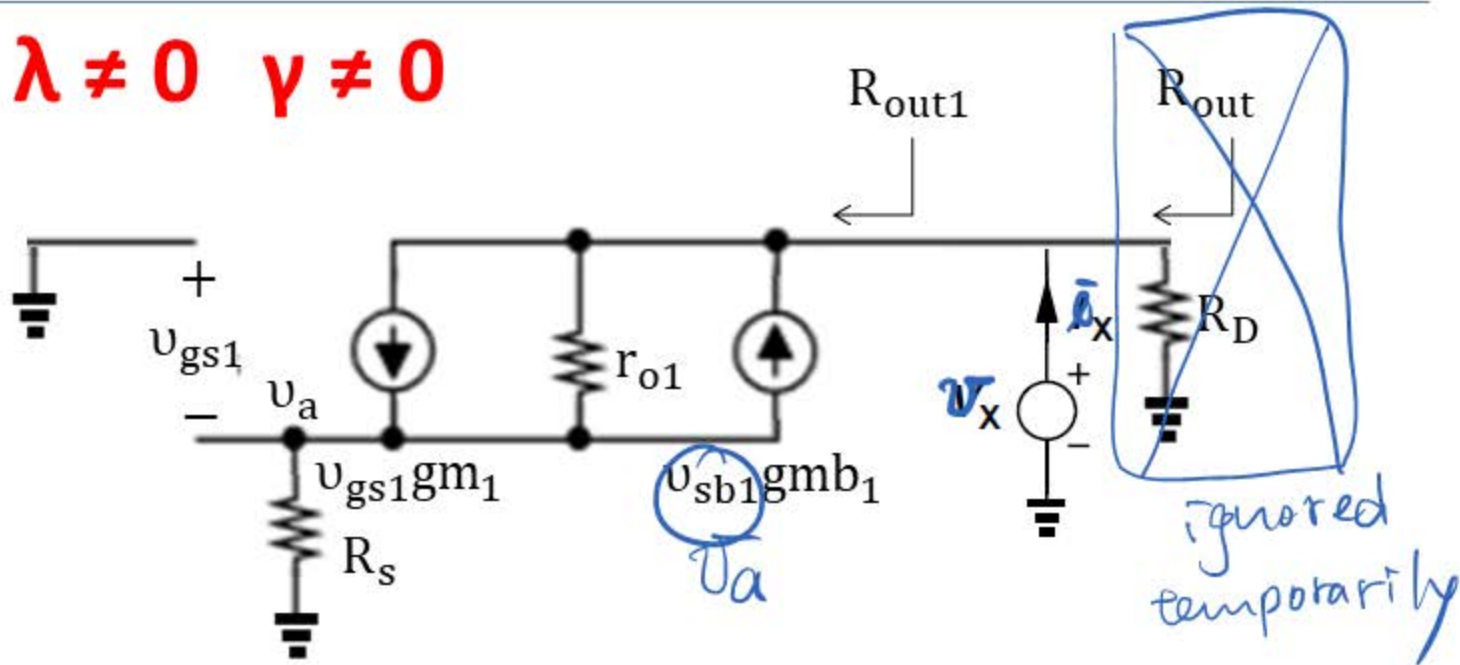
Small signal



Common-Source with Source Degradation

Small-signal
Analysis

$\lambda \neq 0$ $\gamma \neq 0$



$$\begin{cases} i_x = \frac{v_a}{R_S} \\ v_a g_{m1} + v_a g_{mb1} + \frac{v_a - v_x}{r_o} + i_x = 0 \end{cases}$$

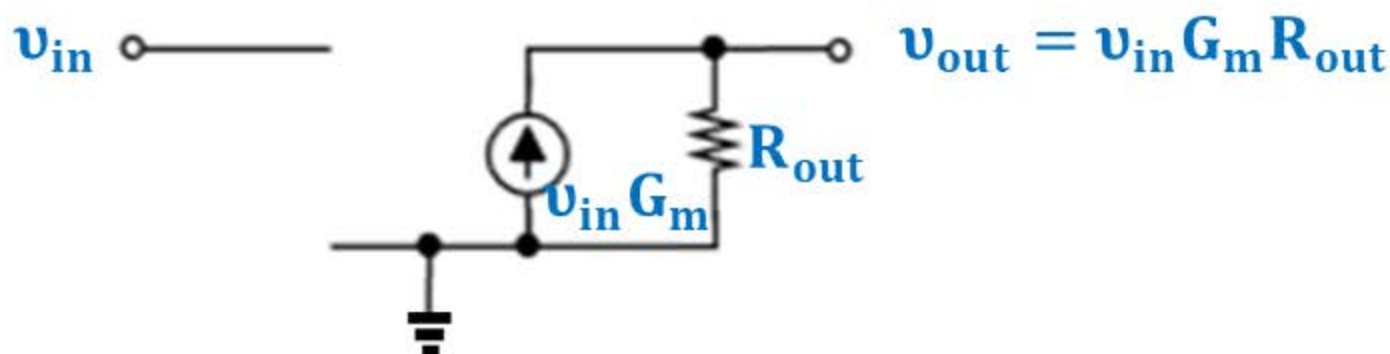
$$R_{out} = R_{out1} \parallel R_D = [R_S + r_{o1} + (g_{m1} + g_{mb1})r_{o1}R_S] \parallel R_D \approx R_D$$

$v_x / i_x = R_{out1}$ if $(g_{m1} + g_{mb1})r_{o1}R_S \gg R_D$

Common-Source with Source Degradation

Small-signal
Analysis

$$\lambda \neq 0 \quad \gamma \neq 0$$



$$A_v = \frac{v_{out}}{v_{in}} = G_m R_{out}$$

$$= \frac{-g_{m1} r_{o1}}{(R_S + r_{o1} + (g_{m1} + g_{mb1}) r_{o1} R_S)} \cdot \frac{(R_S + r_{o1} + (g_{m1} + g_{mb1}) r_{o1} R_S) R_D}{[R_S + r_{o1} + (g_{m1} + g_{mb1}) r_{o1} R_S] + R_D}$$

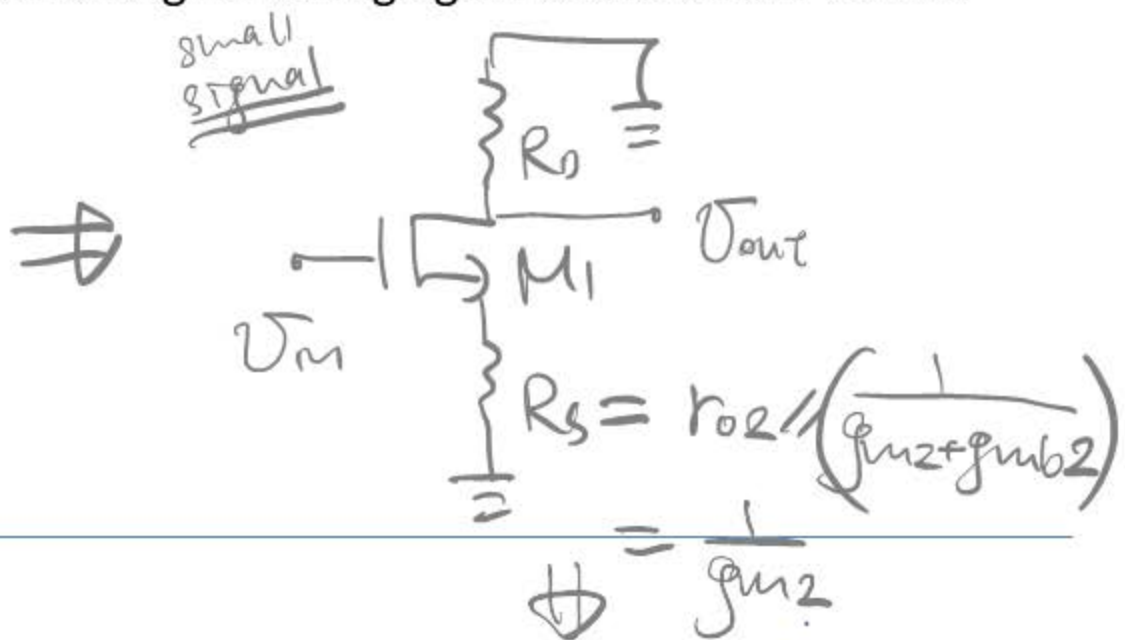
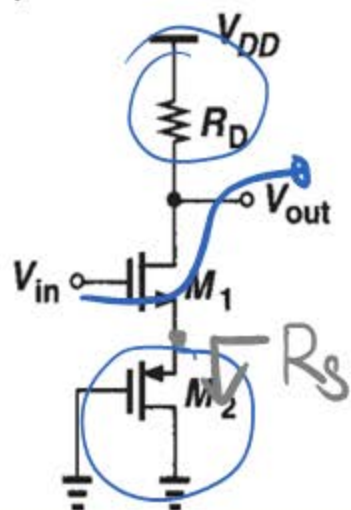
$$\approx -\frac{R_D}{R_S}$$

If $(g_{m1} + g_{mb1}) r_{o1}$, the intrinsic gain, is large.

$$f_{gmb} \ll g_m$$

Example

Assuming $\lambda = \gamma = 0$, calculate the small signal voltage gain of the circuit below.



Solution:

$$G_m = - \frac{1}{\frac{1}{g_{m1}} + \frac{1}{g_{m2}}}$$

$$R_{out} = R_D$$

$$A_v = G_m R_{out}$$