
VE320 – Summer 2021

Introduction to Semiconductor Devices

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Chapter 11 Metal-Oxide-Semiconductor Field Effect
Transistors: More Concepts



Outline

Nonideal Effects:

11.1 Subthreshold conduction

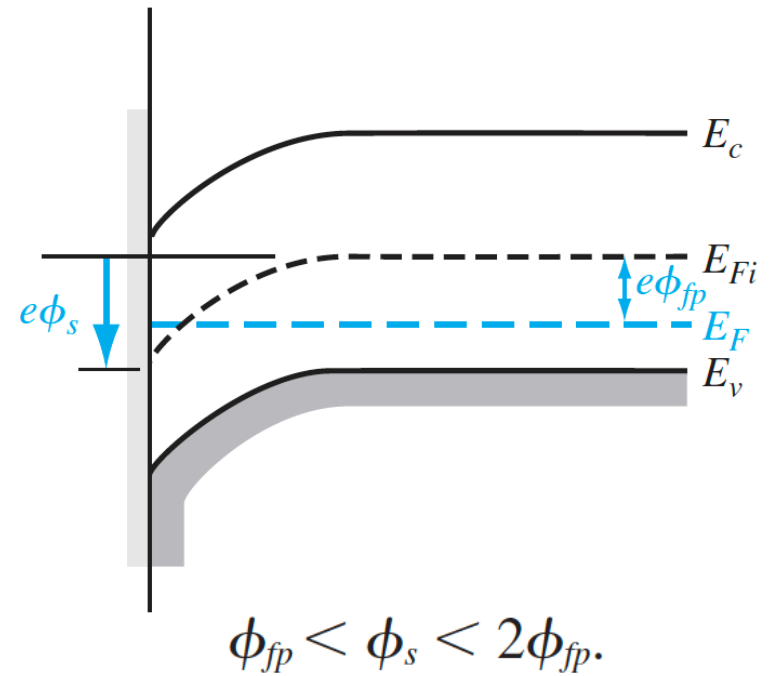
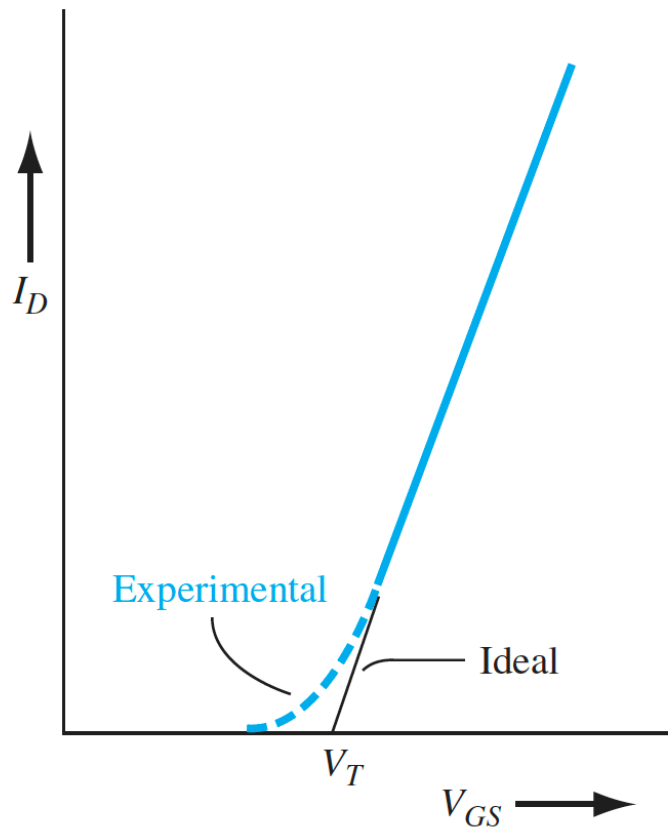
11.2 Channel length modulation

11.3 Velocity Saturation

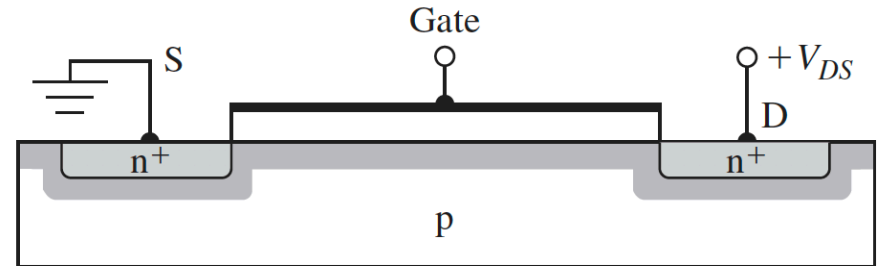
11.4 Short Channel Effect

11.1 Subthreshold conduction

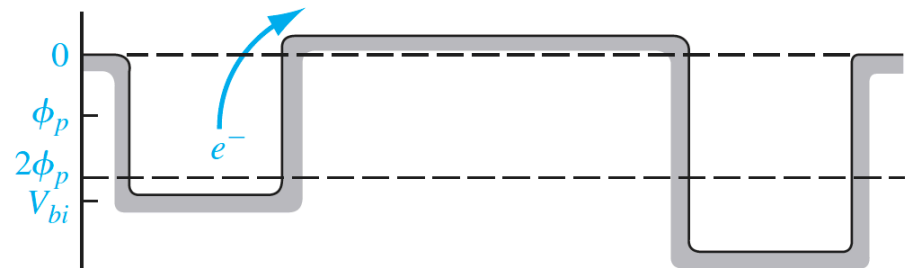
$$V_{GS} < V_T$$



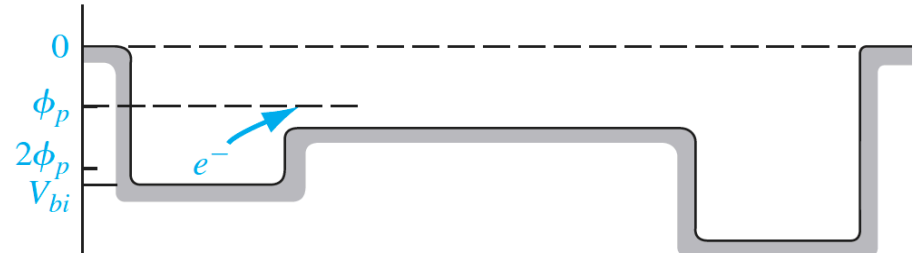
11.1 Subthreshold conduction



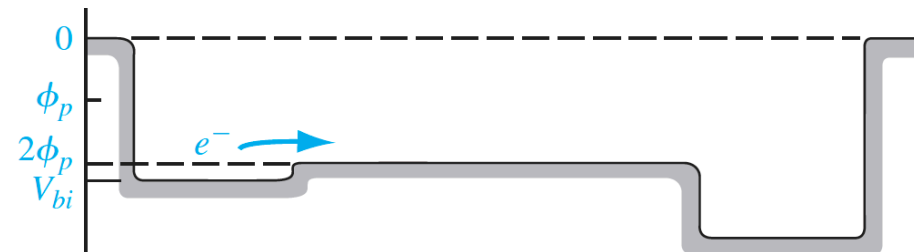
Accumulation



Weak inversion



Inversion



11.1 Subthreshold conduction

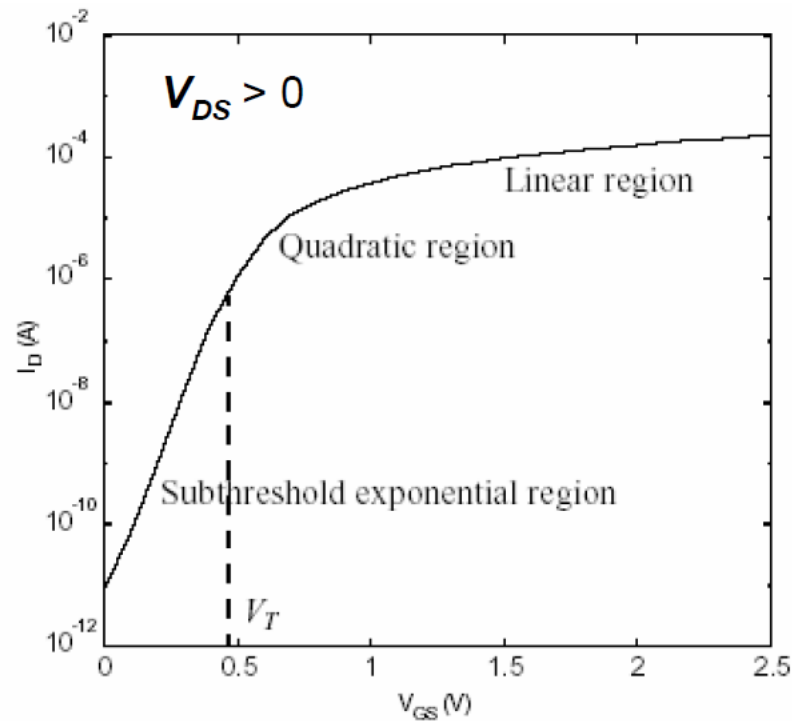
Subthreshold Conduction (Leakage Current)

- The transition from the ON state to the OFF state is gradual. This can be seen more clearly when I_D is plotted on a logarithmic scale:

- In the subthreshold ($V_{GS} < V_T$) region,

$$I_D \propto \exp\left(\frac{qV_{GS}}{nkT}\right)$$

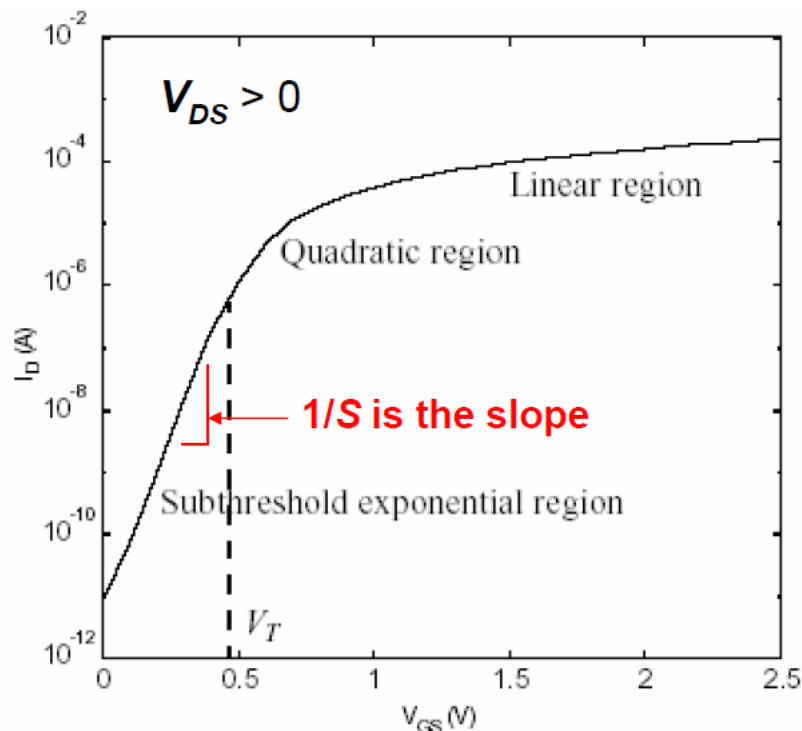
This is essentially the channel-source pn junction current.
(Some electrons diffuse from the source into the channel, if this pn junction is forward biased.)



11.1 Subthreshold conduction

Slope Factor (or Subthreshold Swing) S

- S is defined to be the inverse slope of the log (I_D) vs. V_{GS} characteristic in the subthreshold region:



$$S \equiv n \left(\frac{kT}{q} \right) \ln(10)$$

Units: Volts per decade

Note that $S \geq 60$ mV/dec at room temperature:

$$\left(\frac{kT}{q} \right) \ln(10) = 60 \text{ mV}$$

11.1 Subthreshold conduction

V_T Design Trade-Off

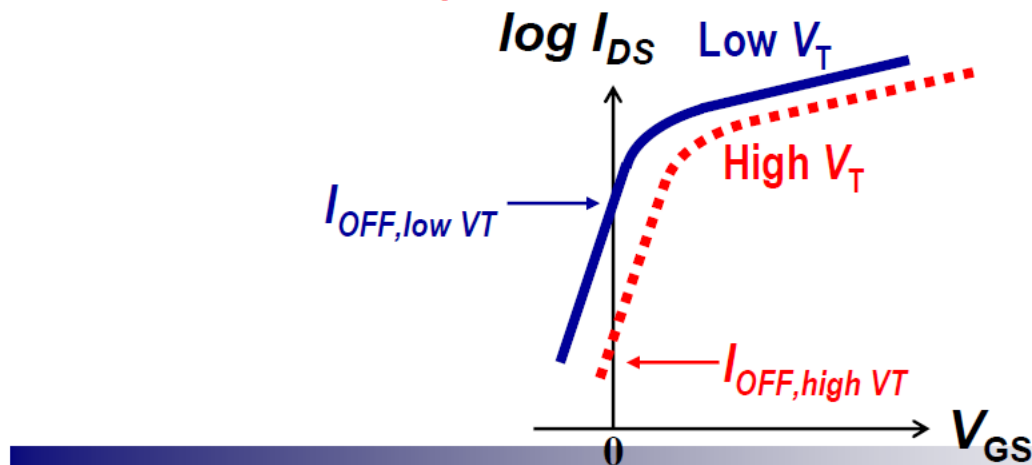
(Important consideration for digital-circuit applications)

- Low V_T is desirable for high ON current

$$I_{DSAT} \propto (V_{DD} - V_T)^\eta \quad 1 < \eta < 2$$

where V_{DD} is the power-supply voltage

...but high V_T is needed for low OFF current



Outline

Nonideal Effects:

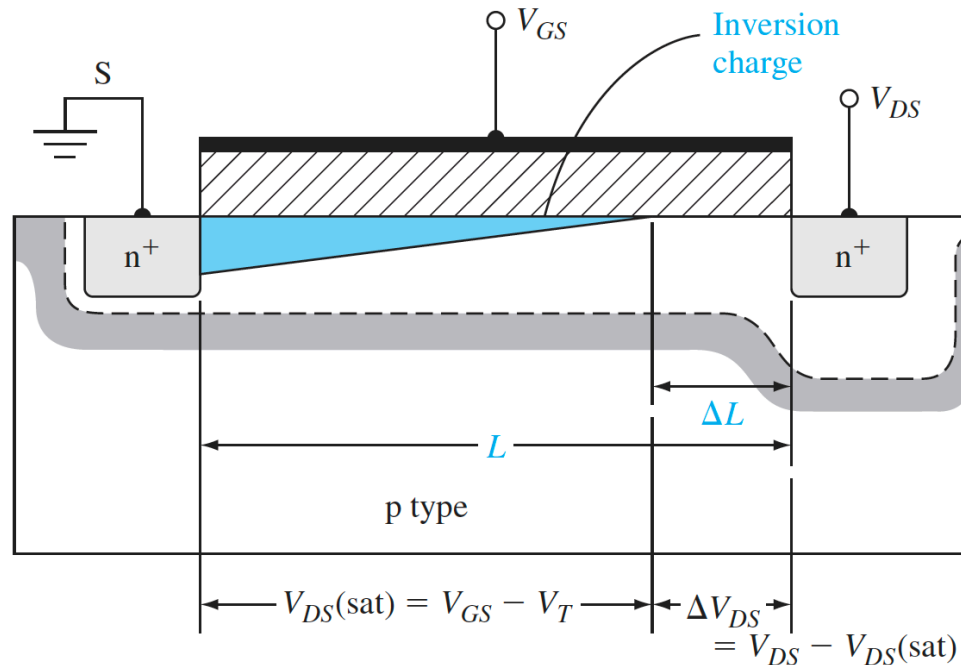
11.1 Subthreshold conduction

11.2 Channel length modulation

11.3 Velocity Saturation

11.4 Short Channel Effect

11.2 Channel length modulation



$$I'_D = \frac{k'_n}{2} \cdot \frac{W}{L} \cdot [(V_{GS} - V_T)^2 (1 + \lambda V_{DS})]$$

11.2 Channel length modulation

$$I_D = \begin{cases} 0 & V_{GS} - V_T < 0 \\ \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_T)^2 & 0 \leq V_{GS} - V_T < V_{DS} \\ \mu_n C_{ox} \frac{W}{L} [(V_{GS} - V_T) V_{DS} - \frac{1}{2} V_{DS}^2] & V_{GS} - V_T \geq V_{DS} \end{cases}$$



$$I_D = \begin{cases} \sim \exp\left(\frac{eV_{GS}}{nkT}\right) [1 - \exp(\frac{-eV_{DS}}{kT})] & V_{GS} - V_T < 0 \\ \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_T)^2 (1 + \lambda V_{DS}) & 0 \leq V_{GS} - V_T < V_{DS} \\ \mu_n C_{ox} \frac{W}{L} [(V_{GS} - V_T) V_{DS} - \frac{1}{2} V_{DS}^2] & V_{GS} - V_T \geq V_{DS} \end{cases}$$

Outline

Nonideal Effects:

11.1 Subthreshold conduction

11.2 Channel length modulation

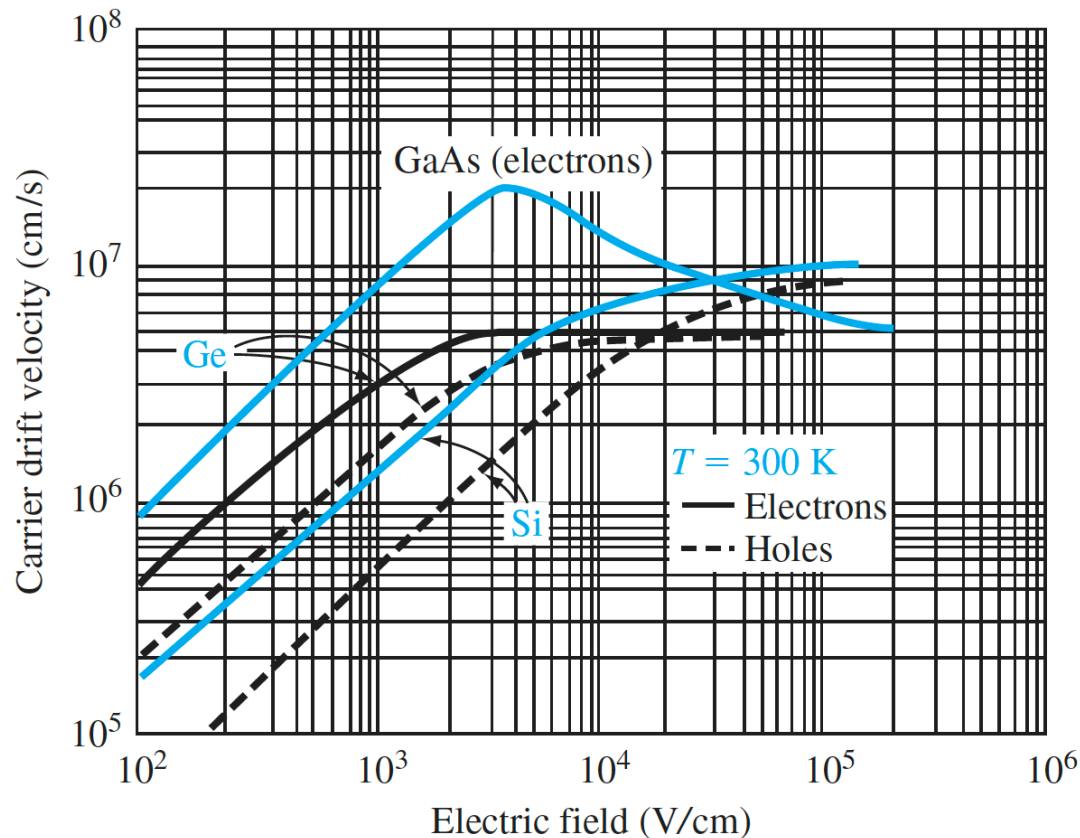
11.3 Velocity Saturation

11.4 Short Channel Effect

11.3 Velocity Saturation

$$v_d \rightarrow v_{th}$$

- Electric field is heating up electrons
- Electrons transfer energy to lattice to reach thermal equilibrium



$$v_n = \frac{v_s}{\left[1 + \left(\frac{E_{on}}{E}\right)^2\right]^{1/2}}$$

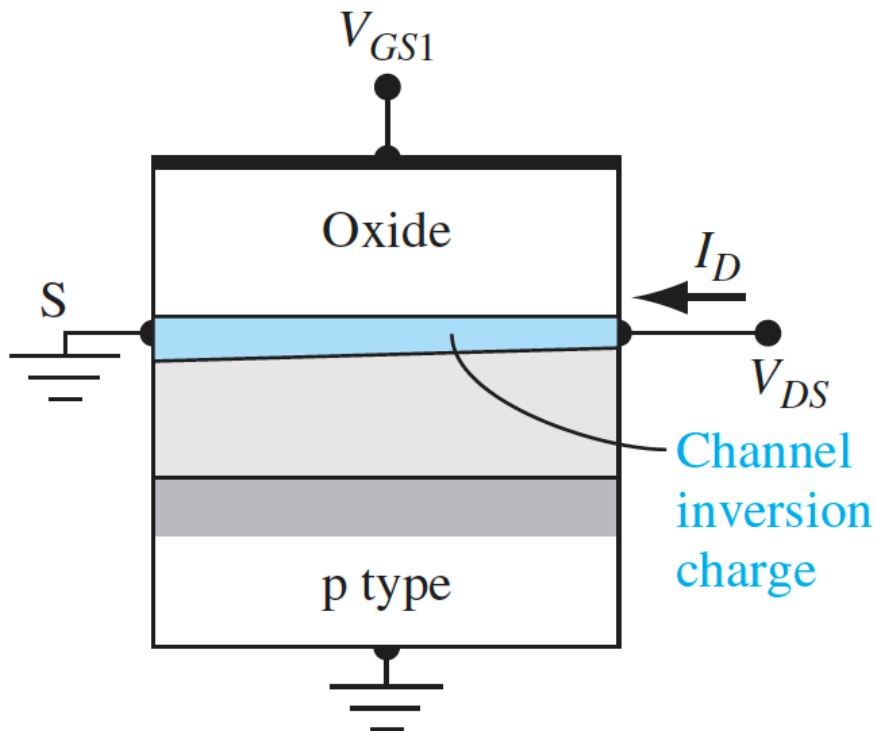
$$v_p = \frac{v_s}{\left[1 + \left(\frac{E_{op}}{E}\right)^2\right]^{1/2}}$$

Probably a typo in textbook

11.3 Velocity Saturation

$$v_d \rightarrow v_{th}$$

- Electric field is heating up electrons
- Electrons transfer energy to lattice to reach thermal equilibrium



$$E_{DS} = \frac{V_{DS}}{L}$$

As the transistor size scales down, the electric field intensity E increases.

$$E_{on} = \frac{V_{DSAT}}{L}$$

11.3 Velocity Saturation

$$v_d \rightarrow v_{th}$$

As the transistor size scales down, the electric field intensity E increases.

- Electric field is heating up electrons
- Electrons transfer energy to lattice to reach thermal equilibrium

$$\begin{aligned} I_{DSAT} &= \mu_n C_{ox} \frac{W}{L} \left(V_{GS} - V_T - \frac{1}{2} V_{DS} \right) V_{DS} \\ &= C_{ox} W \left(V_{GS} - V_T - \frac{1}{2} V_{DSAT} \right) \frac{V_{DSAT}}{L} \mu_n \end{aligned}$$

$$\begin{aligned} I_{DSAT} &= W C_{ox} \left[V_{GS} - V_T - \frac{V_{DSAT}}{2} \right] v_{sat} \\ \text{where } V_{DSAT} &= \frac{L}{\mu_n} v_{sat} \end{aligned}$$

11.3 Velocity Saturation

Unified model

$$I_D = 0 \text{ for } V_{GT} \leq 0$$

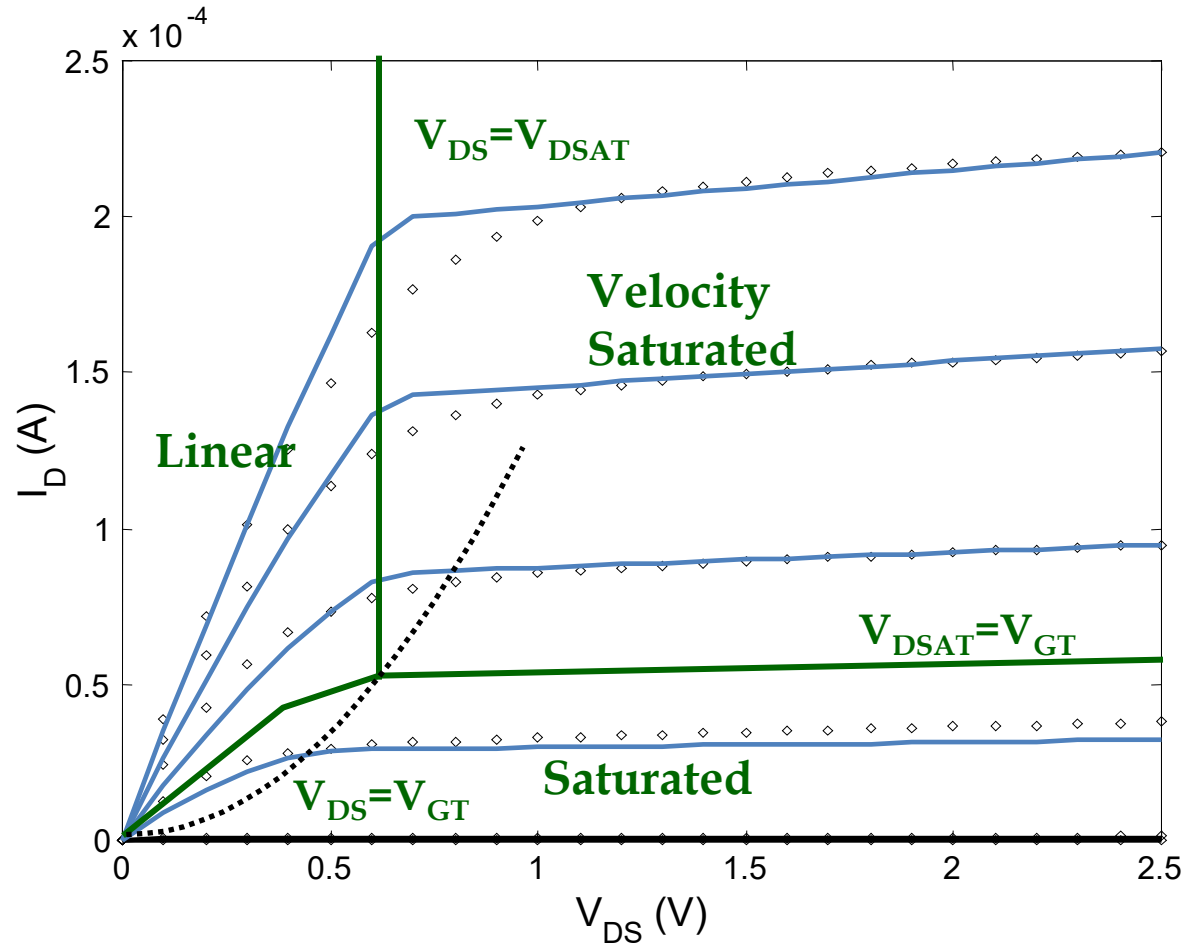
$$I_D = k' \frac{W}{L} \left(V_{GT} V_{min} - \frac{V_{min}^2}{2} \right) \quad \text{for } V_{GT} \geq 0$$

$$\text{with } V_{min} = \min(V_{GT}, V_{DS}, V_{DSAT}),$$

$$V_{GT} = V_{GS} - V_T,$$

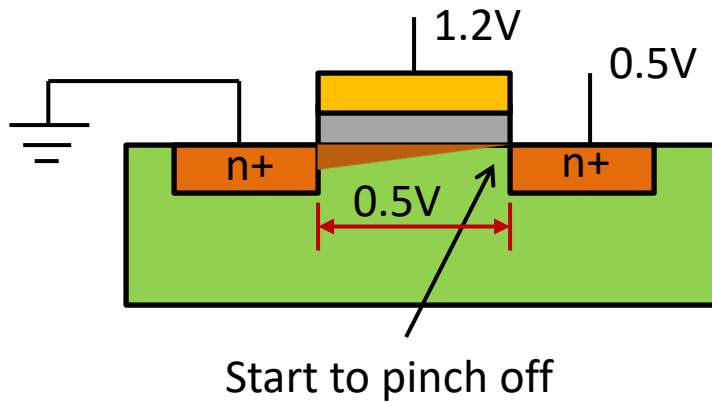
$$\text{and } V_T = V_{T0} + \gamma (\sqrt{|-2\phi_F + V_{SB}|} - \sqrt{|-2\phi_F|})$$

11.3 Velocity Saturation



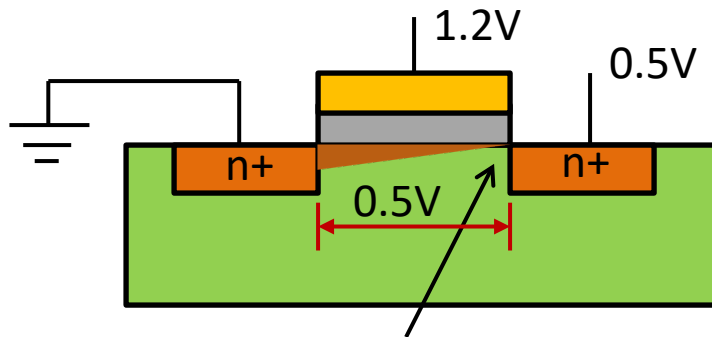
Example

$$V_T = 0.7, \quad V_{gs} = 1.2 \text{ V}, \quad V_{sat} = 1 \text{ V}$$

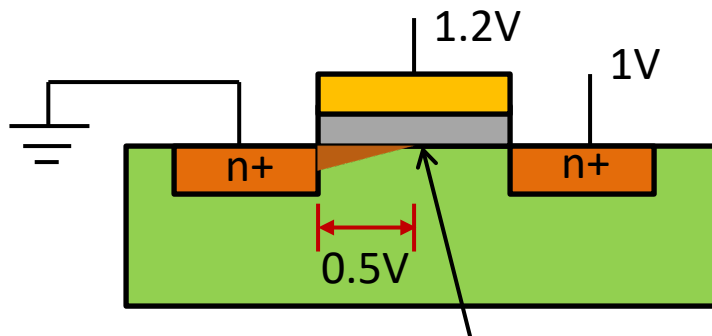


Example

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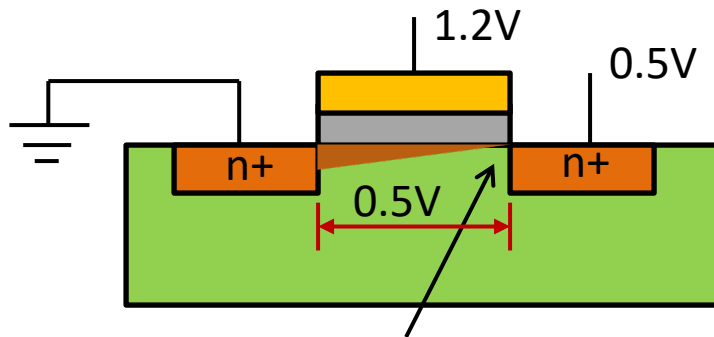
Start to pinch off



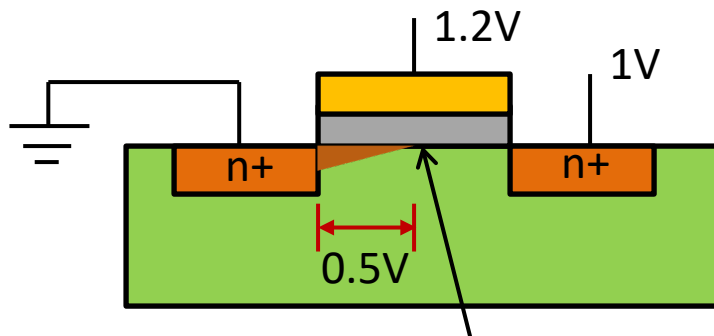
Start to pinch off

Example

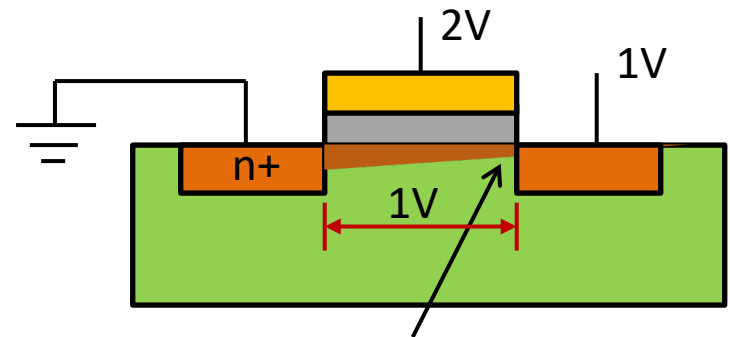
$$V_T = 0.7, \quad V_{gs} = 1.2 \text{ V}, \quad V_{sat} = 1 \text{ V}$$



Start to pinch off



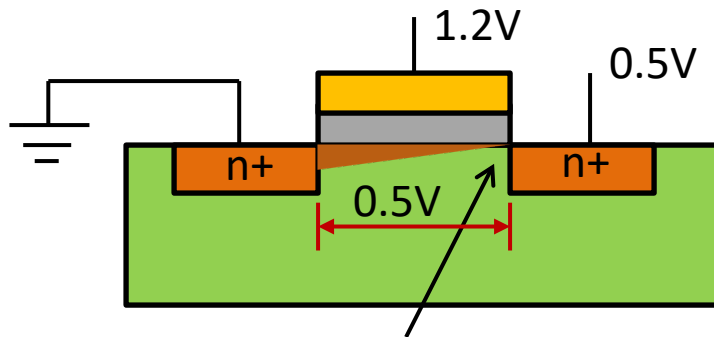
Start to pinch off



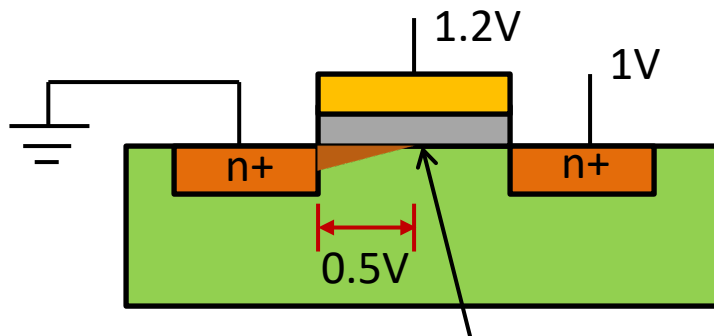
- NO pinch off
- Velocity saturation starts

Example

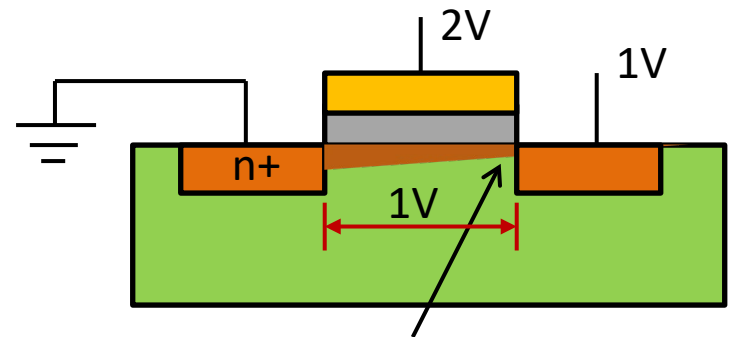
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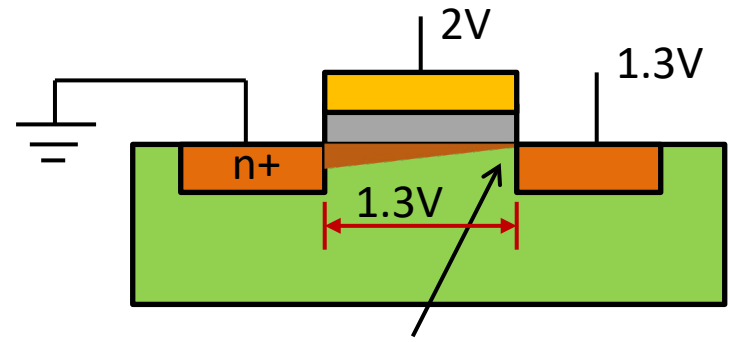
Start to pinch off



Start to pinch off



- NO pinch off
- Velocity saturation starts



- Starts to pinch off
- Velocity saturation

Outline

Nonideal Effects:

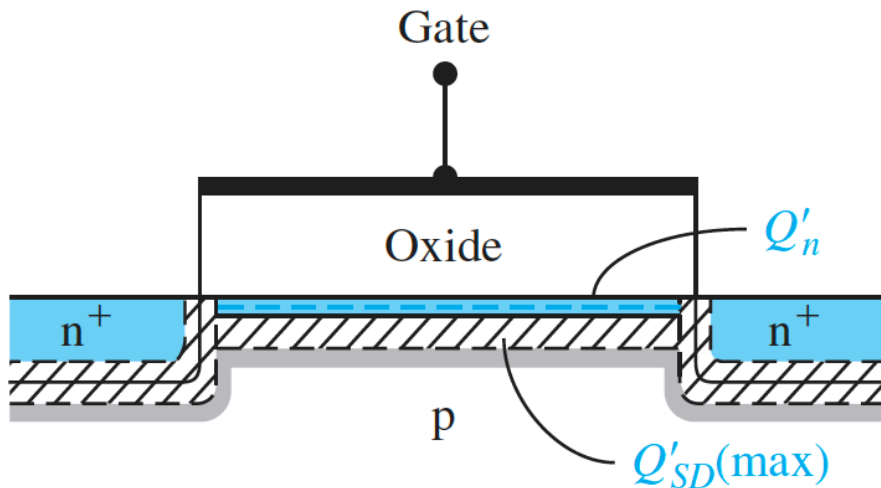
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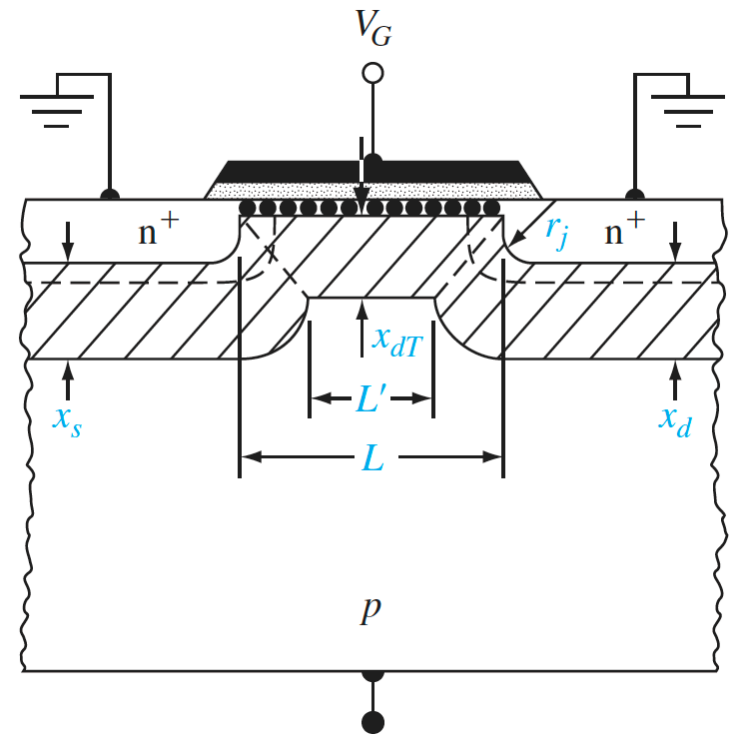
11.3 Velocity Saturation

11.4 Short Channel Effect

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A long channel device



A short channel device

11.4 Short Channel Effect

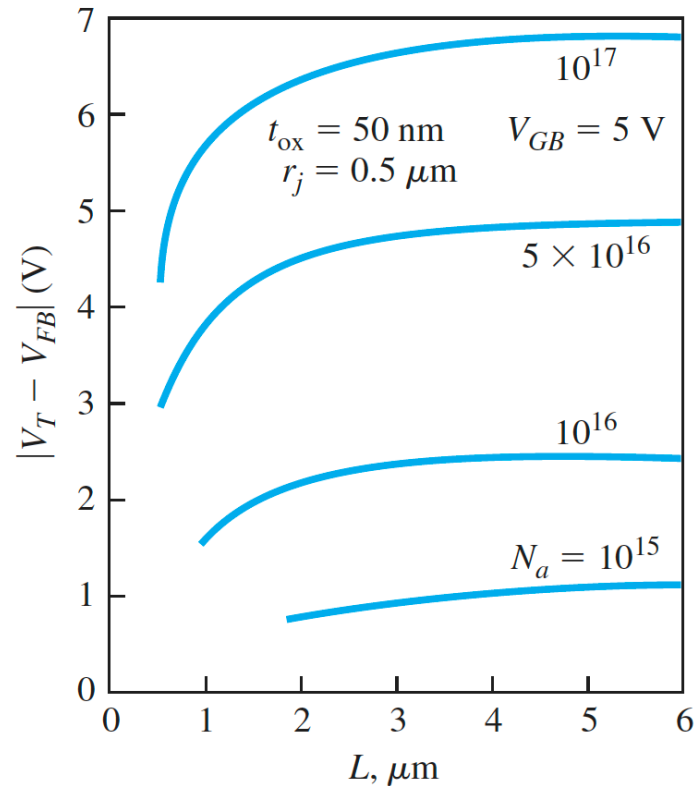


Figure 11.16 | Threshold voltage versus channel length for various substrate dopings. (From Yau [26].)

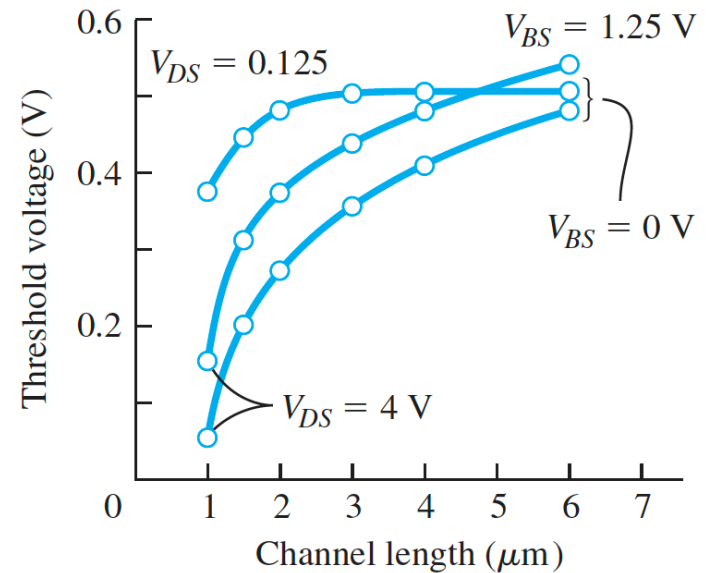


Figure 11.17 | Threshold voltage versus channel length for two values of drain-to-source and body-to-source voltage. (From Yang [25].)