1	09/09/2021	Passband: -we to we
		2 region: Stop band
	UNIT-3 - FIR-Filler Design	Remaining region: Stop band
DESCRIPTION OF THE PERSON OF T		
	Remove unwanted signals Filter	HAPF:
Appropriate Control of	Response Filter	Magnitude response of ideal high pass litter
ACCEPTAGE OF THE PARTY OF THE P	FIR - Finite Impulse Response Filter	high pass liter
CONTRACTOR CONTRACTOR	→ IIR - Infinite Impulse Response Filter	
PART WAS ASSESSED.	Methods	T-Wc O Wc T
TO CHANGE OF THE PARTY OF THE P	FIR Filter Design - Methods	me - cont off / sed.
Three recentains	Fourier series	Pass band: we to x;
AND DESCRIPTION OF THE PERSON	Windowing 7 1 2 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	-just x Z w Z - w L w C < W
The second second	Freq. sampling	Ha (w) = S Ce
	, ,	Desired 0, -we & w & we
	Design of FIR Filters using Windowing Method:	Ass bouse Destages
	Maria Variante de la companya della companya della companya de la companya della	
	# LPF: HLP (ejw)	Step & the design response
	The state of the s	Step 1 Identify the design response [Hd (w)] And Identify the design response [Hd (w)] * BPF: Range of freq. for which the filter passes the signal passes
	Liphase delay	I trey for which the litter passe
	Total And	* BPF Range MBPle 100)
	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Magnitude
	Magnitude response of ideal low pass litter	ideal band pass fitter
	H _a (w)= } Ce -jval , -w, ∠w∠w,	
	o , -t < w < -w & w < < w < t	· To -we 2 -we we we so we
		De land on de 180 °
		Pass band: we, to we =

When
$$(n) = \begin{cases} 0.5 \times 0.5 \cos \frac{2n\pi}{N-1} \end{cases}$$
, $(\frac{N-1}{2}) \leq n \leq (\frac{N-1}{2})$

Step 4 Transfer function $H(z)$ is obtained by faking z have been $y \in \mathbb{R}$ to draw the sum of $y \in \mathbb{R}$ to draw the $y \in \mathbb{R}$ to the fifth $y \in \mathbb{R}$ to the first $y \in \mathbb{R}$ to the first

h(0)= h(4)

Step 1:

To find hy (n)

hy (n) =
$$\frac{1}{2\pi} \int_{-\pi}^{\pi} H_{1} |w\rangle_{e}^{-j\omega_{e}} d\omega$$

TFT

$$\frac{\sum_{k=0}^{\infty} f_{k}(0)}{f_{k}(0)} = \frac{1}{K[0-2]} \left[\frac{\sin(0-2) \cdot \sqrt{4}}{4} \right]$$

$$= \frac{-1}{2K} \sin\left(\frac{-2K}{4}\right)$$

$$= \frac{-1}{2K} \cdot \sin\left(\frac{K}{2}\right)$$

$$= \frac{1}{2K} \cdot 1 \quad \left[K = 3.14 \right]$$

$$h_{k}(0) = 0.159$$

$$\frac{1}{2K} \left[\frac{1}{2K} \right] = 0.159$$

$$\frac{1}{2K} \left[\frac{1}{2K} \right] = \frac{1}{2K} \left[\frac{\sin((1-2) \cdot \sqrt{4})}{x} \right]$$

$$\frac{1}{2K} \left[\frac{1}{2K} \right] = \frac{1}{2K} \left[\frac{\sin((1-2) \cdot \sqrt{4})}{x} \right]$$

$$\frac{1}{2K} \left[\frac{-4K}{4} \right] \Rightarrow \frac{\sin((1-2) \cdot \sqrt{4})}{x}$$

$$\frac{1}{2K} A = 343$$

$$\frac{1}{2K} A = 343$$

$$\frac{1}{3} = 0.707$$

$$\frac{1}{3} = 0.225$$

$$|H(\omega)| = e^{\int \omega \left(\frac{N-1}{2}\right)} \int h(n) \cos \left[\left(\frac{N-1}{2}\right)^{2}\right] d\sin \theta dx \text{ and } h(n) = h(n) d \cos \theta dx$$

$$|H(\omega)| = e^{\int \omega \left(\frac{N-1}{2}\right)} \int h(n) \cos \theta dx + h(n) \cos \omega + h(n) \cos \omega + h(n) \cos \omega dx$$

$$|H(\omega)| = e^{\int \omega \left(\frac{N-1}{2}\right)} \int h(n) \cos \theta dx + h(n) \cos \omega dx$$

$$|H(\omega)| = e^{\int \omega \left(\frac{N-1}{2}\right)} \int h(n) d \cos \theta dx + h(n) \cos \theta dx + h(n) \cos \theta dx + h(n) \cos \theta dx$$

$$|H(\omega)| = e^{\int \omega \left(\frac{N-1}{2}\right)} \int h(n) d \cos \theta dx + h(n) \cos$$

$$h_{d}(n) = |FT| [H_{d}(w)]$$

$$h_{(n)} = h_{(n-1-n)}$$

$$h_{(n)} = h_{(n-2)} ... w(n)$$

$$h_{d}(n) = \frac{1}{\pi(n-2)} ... sin(n-2) \pi/4$$

$$h_{d}(n) = 0.159 = h_{d}(4)$$

$$h_{d}(1) : h_{d}(3) = 0.225$$

$$h_{d}(3) = 0.25$$

$$u(n) = \begin{cases} 0.54 - 0.46 & (0.52\pi) \\ 0.54 - 0.46 & (0.52\pi) \end{cases}$$

$$v(1) = 0.54 - 0.46 - 0.46 & (0.52\pi)$$

$$v(1) = 0.54 - 0.46 & (0.52\pi)$$

$$v(2) = 0.54 - 0.46 & (0.52\pi)$$

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$$v(3) = 0.54 - 0.46 & (0.52\pi)$$

$$v(4) = 0.54 - 0.46 & (0.52\pi)$$

$$v(5) = 0.54 - 0.46 & (0.52\pi)$$

(2) Same sum but use Hamming window:

h(2) = hd(2), w(2) = 0.25 * 1 h(2) = 0.25(Q:)

(Design a linear phase FIR little using Hamming windom for the foll specifications N = 7.

Ha(w) e^{-j3w} , $-\frac{\pi}{8} \le |w| \le \frac{\pi}{8}$ Plot the magnitude response of the fitter.

-> Desired response available from - 1/8 to 1/8 => Low pass

h(n): h11n). w(~)

h(0) = hd(0). & w(0)

h(0) = 0.012 = h(4)

(1) w. (1) & = (1)N

h(1) = 0.1215 = h(3)

=0.159 × 0.08

= 0.225 x Mas 0.59

$$h(0) = h(1) = 0.0980 \times 0.08 \Rightarrow 0.0018$$

$$h(1) = h(1) \Rightarrow 0.12 \times 0.31 \Rightarrow 0.034$$

$$h(2) = h(4) \Rightarrow 0.1218 \times 0.77 \Rightarrow 0.0937$$

$$h(3) \Rightarrow 0.125 \times 1 = 0.125$$

$$h(n) = \begin{cases} 0.0018 & 0.034 & 0.0937 \\ 0.0937 & 0.025 \end{cases} = 0.0185 & 0.0937 \\ 0.0937 & 0.0937 & 0.09$$

alo) = h(3) ora = 18 Voju = 0.125

 $a(1)=2h(3-1)=2h(2)=2\times0.0937=0.1874$

 $A(2) = 2h(3-2) = 2h(1) = 2 \times 0.034 = 0.068$ $\alpha(3) = 2h(3-3) = 2h(0) = 2 \times 0.008 = 0.0036$

H(ejw) = 0.125+0.1874 cos w + 0.068 cos 2w + 0.0036 cos 3w 20 30 40 60 80 90 100 H[ejw) 0.384 6.3508 H(ejw) 1-4.1 -4.5 dB w=0: H(e) = 0.125+ 0.1874 + 0.068 + 0.0036 = \(0.384 \)

= H[e; w)= a(0)+a(1) cos w/h+ a(2) cos 2w+a(3) cos 3w

To convert into dB. 10 log (0.384) = [-4.1 &B

w=20: H(ejw)=0.125 + 0.1874 (cos (0.939) + 0.068 (0.7660) + 0.034 (0.5) = 0.125 + 0.1759 + 0.0520 + 0.018 = 0.3548 (0.809 (0.3548) = -4.5 dB -10 -15 10 20 30 40 50

3 Design a linear phase FIR filter using Hamming & Hanning window for the foll specifications Find the values of h(n) go No 11. Find HIZ) Plot the magnitude response

$$f(\lambda m) = \begin{cases} 0 & \text{otherwise} \end{cases}$$

=> High pass liter







