

Department of Electrical Engineering, IIT Bombay

AUTUMN SEMESTER: JUL-NOV 2024

Mid-Semester Examination: EE 229 – Signal Processing I – (B. Tech. )

Maximum Marks: 50 (25 percent weight)

Date: Saturday 14 Sept. 2024

Time: 13:30 to 15:30 hours

**Instructions:**

1. Please begin the answer to each **main question** on a **fresh page** of the answer booklet.
2. This is a **closed book, closed notes** examination.
3. Show your reasoning and important steps clearly.
4. Unless otherwise stated, (a)treat  $x$  as the input to the system and  $y$ , the output (b)Stability is understood in the ‘Bounded input, bounded output’ (BIBO) sense. (c)  $j = \sqrt{-1}$

**Q1. (3 X 5 = 15 marks):**

In each of the parts (a) to (e) below, **either** provide a satisfactory **proof** of the statement A, if it is true, independent of example, ignoring the system S **or disprove** the statement A, **using the system S as a counter-example**, with proper explanation. State clearly, what you are doing.

(a) A: *A discrete linear system is stable in the ‘Bounded input, bounded output’ (BIBO) sense, if its impulse response is absolutely summable.*

$$S: y[n] = (n + 3)x[n] + (n + 2)x[n - 1]$$

(b) A: *A linear system always produces the output  $y[n] = 0 \quad \forall n$ , when the input is  $x[n] = 0 \quad \forall n$*

$$S: y[n] = 3x[n] + 2x[n - 1] + 9$$

(c) A: *A shift invariant system always produces the output  $y[n] = 0 \quad \forall n$ , when the input is  $x[n] = 0 \quad \forall n$*

$$S: y[n] = 3x[n] + 2x[n - 1] + 9$$

(d) A: *An additive system must also be homogeneous.*

$$S: y[n] = 3(\text{complex conjugate of } x[n]) + 4(\text{Real part of } x[n - 1]) + x[n - 2]$$

(e) A: *A homogeneous system must also be additive.*

$$S: y[n] = 2 \frac{x^2[n]}{x[n-1]}, \text{ if } x[n - 1] \neq 0; \quad y[n] = 0, \text{ if } x[n - 1] = 0$$

**Q2. (5 x 3 = 15 marks):**

S1 and S2 are two continuous time, linear shift invariant systems, whose impulse responses, respectively, are:  $e^{-2t}u(t)$  and  $e^{-5t}u(t)$ , where  $u(t)$  is the unit step function.

- (a) Let S1 and S2 be connected in series to form a composite linear shift invariant system. Obtain the impulse response of this composite linear shift invariant system. Why does the order, in which the two systems are connected in series, **not** matter?
- (b) Let an input  $x(t)$  be applied to the individual systems S1 and S2, as also to the composite series connected system described in Q2(a). Show that the output of the composite system is a linear combination of the outputs of the individual systems S1 and S2 and obtain the coefficients in this linear combination.
- (c) State and explain whether (I) the individual systems S1 and S2, as also (II) the composite series connected system of Q2(a), are (i) causal (ii) stable.

**Q3. (10 marks)** A linear, shift invariant, continuous time system has the frequency response:

$$H(j\Omega) = 7j, \forall (0 < \Omega < \pi); (-7j), \forall (-\pi < \Omega < 0); 0, \forall (|\Omega| > \pi)$$

- (a) Obtain its impulse response.
- (b) Obtain the output of this system, when the input is  $x(t) = 3 \cos(2t)$ .

**Q4. (10 marks)** A linear, shift invariant, causal discrete time system is described by:

$$y[n] = \alpha y[n-1] + x[n], \text{ where } \alpha \text{ is a real constant.}$$

- (a) Obtain its impulse response.
- (b) Obtain the condition on  $\alpha$ , for the system to be BIBO stable. Explain.
- (c) Obtain the response of the system to the unit step input.

**End of question paper – with best wishes.**

Q1 -

(a) We shall disprove the Statement 'A' using the System 'S' as counter-example

The impulse response of system 'S' is

$$h[n] = (n+3)\delta[n] + (n+2)\delta[n-1]$$
$$= 3\delta[n] + 2\delta[n-1]$$

This impulse response is absolutely summable.

Indeed  $\sum_{n \in \mathbb{Z}} |h[n]| = 3 + 2 = 5.$

However a bounded input, say  $x[n] = u[n]$  produces the unbounded output

$$(n+3)u[n] + (n+2)u[n-1]$$
$$= 3\delta[n] + (n+3+n+2)u[n-1]$$
$$= 3\delta[n] + (2n+5)u[n-1]$$

This discrete linear system is NOT BIBO Stable. This has happened as it is not shift-invariant.

(b) We shall prove the Statement 'A' independent of example.

Indeed with input  $x[n] = 0 \forall n$ , output  $y[n]$ ,

$$2x[n] = x[n] = 0 \forall n$$

whereupon, linearity of the system

$$\Rightarrow 2x[n] \text{ produces } 2y[n]$$

$$\text{and hence } 2y[n] = y[n] \forall n$$

$$\Rightarrow y[n] = 0 \forall n$$

Q1-

(c) We shall disprove Statement A', using system 'S' as counter-example.

Indeed, S is a shift-invariant system

$$\text{Since } x[n-D] \rightarrow 3x[n-D] + 2x[n-D-1] + 9 \\ = y[n-D]$$

However the input  $x[n] = 0 \quad \forall n$  produces  $y[n] = 9 \quad \forall n$ , not  $y[n] = 0 \quad \forall n$ .

In fact, 'S' is not linear.

(d) We shall disprove Statement A', using system 'S' as counter-example.

$$\text{Indeed let } x[n] \xrightarrow{S} y[n]$$

$$jx[n] \xrightarrow{S} 3(jx[n]) + 4\operatorname{Re}(jx[n-1]) + jx[n-2]$$

$$\neq j \{ 3x[n] + 4\operatorname{Re}(x[n-1]) + x[n-2] \}.$$

The system is not homogeneous, as exemplified by scaling by  $j$

(e) We shall disprove Statement A' using system 'S' as counter-example

$$\text{Indeed let } x[n] \rightarrow y[n]$$

$$\text{For any nonzero constant } c, \quad cx[n] \rightarrow cy[n]$$

$$\text{On the other hand } x[n] = 0 \quad \forall n \Rightarrow y[n] = 0 \quad \forall n$$

The system 'S' is homogeneous. It is clearly not additive as

$$x_{1,2}[n] \rightarrow y_{1,2}[n] \\ \Rightarrow x_1[n] + x_2[n] \rightarrow y_1[n] + y_2[n]$$

Q2 -

(a) The impulse response of this composite linear shift-invariant system is the convolution of  $e^{-2t}u(t)$  and  $e^{-5t}u(t)$ . It does not matter, in which order they are connected, since convolution is commutative.

The composite system has impulse response

$$h(t) = \int_{-\infty}^{\infty} e^{-2\lambda} u(\lambda) e^{-5(t-\lambda)} u(t-\lambda) d\lambda$$

$$= \left\{ \int_0^t e^{-2\lambda} \cdot e^{-5t+5\lambda} d\lambda \right\} u(t)$$

$$= e^{-5t} u(t) \cdot \int_0^t e^{3\lambda} d\lambda$$

$$= e^{-5t} u(t) \left. \frac{e^{3\lambda}}{3} \right|_0^t$$

$$= \frac{1}{3} e^{-5t} u(t) (e^{3t} - 1) = \frac{1}{3} e^{-2t} u(t) - \frac{1}{3} e^{-5t} u(t),$$

$$= \frac{1}{3} (\text{Impulse response of } s_1) - \frac{1}{3} (\text{Impulse response of } s_2)$$



Q2-

(b) clearly  $x(t) \rightarrow \boxed{s_1} \rightarrow x(t) * e^{-2t} u(t)$

$$x(t) \rightarrow \boxed{s_2} \rightarrow x(t) * e^{-5t} u(t)$$

$$x(t) \rightarrow \boxed{s_1} \rightarrow \boxed{s_2} \rightarrow x(t) * h(t)$$

However

$$\begin{aligned} & x(t) * h(t) \\ &= \frac{1}{3} [x(t) * e^{-2t} u(t)] \\ &\quad - \frac{1}{3} [x(t) * e^{-5t} u(t)] \\ &= \frac{1}{3} (\text{output of } s_1) \\ &\quad - \frac{1}{3} (\text{output of } s_2) \end{aligned}$$

which proves the statement  
with coefficients

$\frac{1}{3}$  associated with output of  $s_1$   
 $-\frac{1}{3}$  associated with output of  $s_2$ .

Q2-  
(c)

Each of the systems

$S_1, S_2$ , Composite Series  
Connected System

is causal

as the impulse response is

zero  $\forall t < 0$ .

The impulse responses of  $S_1, S_2$   
are absolutely integrable

Indeed  $+$

$$\int_{-\infty}^{+\infty} |e^{-2t} u(t)| dt = \int_0^{\infty} e^{-2\lambda} d\lambda = \frac{e^{-2\lambda}}{-2} \Big|_0^{\infty} = \frac{1}{2}.$$

$$\int_{-\infty}^{+\infty} |e^{-5t} u(t)| dt = \int_0^{\infty} e^{-5\lambda} d\lambda = \frac{e^{-5\lambda}}{-5} \Big|_0^{\infty} = \frac{1}{5}.$$

As regards the composite system

$$\begin{aligned} \int_{-\infty}^{+\infty} |h(t)| dt &= \int_{-\infty}^{+\infty} \left| \frac{1}{3} e^{-2t} u(t) - \frac{1}{3} e^{-5t} u(t) \right| dt \\ &\leq \int_0^{\infty} \frac{1}{3} e^{-2\lambda} d\lambda + \int_0^{\infty} \frac{1}{3} e^{-5\lambda} d\lambda = \frac{1}{3} \left( \frac{1}{2} + \frac{1}{5} \right) < \infty \end{aligned}$$

whereupon  $h(t)$  is absolutely integrable  
and the composite system,  
BIBO Stable

$$\begin{aligned}
 Q3 - H(j\Omega) &= 7j \quad \forall 0 < \Omega < \pi \\
 &= -7j \quad \forall -\pi < \Omega < 0 \\
 &= 0 \quad \forall |\Omega| > \pi.
 \end{aligned}$$

(a) Impulse response =  
Inverse Fourier Transform of  $H(j\Omega)$

$$\begin{aligned}
 &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} H(j\Omega) e^{j\Omega t} d\Omega \\
 &= \frac{1}{2\pi} 7j \cdot \left\{ \int_0^{\pi} e^{j\Omega t} d\Omega - \int_0^0 e^{j\Omega t} d\Omega \right\} \\
 &= \frac{1}{2\pi} 7j \left\{ \int_0^{\pi} e^{j\Omega t} d\Omega + \int_0^{-\pi} e^{-j\Omega t} d\Omega \right\} \\
 &= \frac{1}{2\pi} 7j \left\{ \int_0^{\pi} e^{j\Omega t} d\Omega - \int_0^{\pi} e^{-j\Omega t} d\Omega \right\} \\
 &= \frac{7j}{2\pi} \int_0^{\pi} (e^{j\Omega t} - e^{-j\Omega t}) d\Omega \\
 &= \frac{7j}{2\pi} \int_0^{\pi} 2j \sin \Omega t d\Omega = \frac{-7j \cdot 2j \cos \Omega t}{2\pi} \Big|_0^{\pi} \\
 &= 7 \left\{ \frac{\cos \pi t - 1}{t} \right\} = -\frac{7}{t} 2 \sin^2 \frac{\pi}{2} t \\
 &= -\frac{14}{t} \sin^2 \frac{\pi}{2} t.
 \end{aligned}$$



$$Q3-(b) \quad x(t) = 3 \cos 2t$$

$$= \frac{3}{2} \left\{ e^{j2t} + e^{-j2t} \right\}$$

$$0 < 2 < \pi, \quad -\pi < -2 < 0$$

Accordingly, from the frequency response,

$$\text{the output} = \frac{7j}{2} \cdot 3 e^{j2t} - \frac{7j}{2} \cdot 3 e^{-j2t}$$

$$= -(7 \times 3) \left( \frac{e^{j2t} - e^{-j2t}}{2j} \right)$$

$$= -21 \sin 2t.$$

Q4- (a) Let the impulse response be  $h[n]$

Then from the causality of the system,  $h[n] = 0 \forall n < 0$

$$\text{Further } h[n] = \alpha h[n-1] + x[n] \quad \left| \begin{array}{l} x[n] = \delta[n] \\ \delta[n] \end{array} \right.$$

$$\text{For } n=0, \quad h[0] = \delta[0] = 1$$

$$n=1, \quad h[1] = \alpha h[0] + 0 = \alpha$$

$$n > 1 \quad h[n] = \alpha^n$$

$$\text{The impulse response} = \alpha^n u[n]$$

(b) For the system to be stable,  $h[n]$  must be absolutely summable.

$$\Rightarrow \sum_{n=-\infty}^{+\infty} |h[n]| = \sum_{n=0}^{\infty} |\alpha| < \infty$$

$$\Rightarrow |\alpha| < 1$$

(c) with a unit step input  $u[n]$ , the output would be  $\sum_{k=-\infty}^{+\infty} \alpha^k u[k] u[n-k]$

$$= \left( \sum_{k=0}^n \alpha^k \right) u[n] = \frac{1 - \alpha^{n+1}}{1 - \alpha} \cdot u[n]$$