last update
december 11, 2014

reference
V 1.4
page
1/7

DATASHEET
SMT172

Preliminary

Features and Highlights

- World's most energy efficient temperature sensor

- Wide temperature range: -45 °C to 130 °C

- Extreme low noise: less than 0.002°C (3.6 ms measurement time)

- Low inaccuracy: 0.25°C (-10 °C to 100 °C)

- Ultra low current (60 µA active or 220 nA average)[1]

- Wide supply voltage range: 2.7 V to 5.5 V

- Excellent long term stability

- Direct interface with Microcontroller (MCU) without ADC

- Wide range of package options
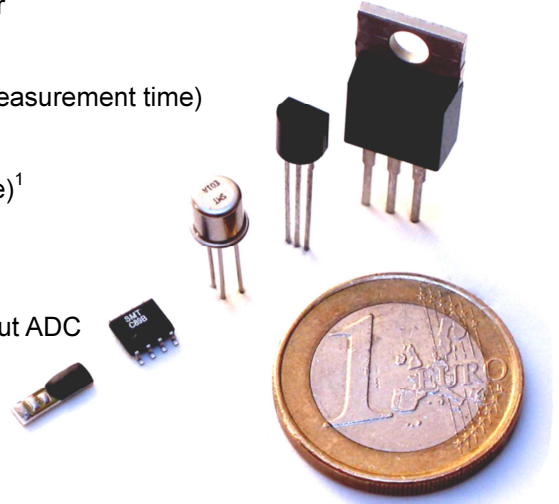
Application

- Ultra low power applications: wearable electronics, wireless sensor networks

- Medical applications: body temperature monitoring

- Instrumentation: (Bio)chemical analysis, Precision equipment

- Environmental monitoring (indoor / outdoor)

- Industrial applications: process monitoring / controlling

Introduction

The SMT172 is an ultra-low power, high accuracy temperature sensor that combines the ease of use with the world's leading performance over a wide temperature range. Using the most recent advances in the silicon temperature sensing technology, the SMT172 has applied some really sophisticated IC design techniques as well as high-precision calibration methods, to achieve an absolute inaccuracy of less than 0.25°C in the range of -10 °C to 100 °C

The SMT172 operates with a supply voltage from 2.7 V to 5.5 V. The typical active current of only 60 µA, the high speed conversion over 4000 outputs per second (at room temperature) and an extremely low noise makes this sensor the most energy efficient temperature sensor in the world.

The SMT172 has a pulse width modulated (PWM) output signal, where the duty cycle is proportional to the measured temperature value. This makes it possible that the sensor can interface directly to a MCU without using an Analog-to-Digital Converter (ADC). Today, the hardware Timer in a MCU to read our PWM signal has become available almost universally, fast in speed and low in cost. Therefore it is extremely easy for any user to get started with this sensor and achieve a very quick time to market.

[1]See Specification Section for detailed measurement conditions

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last update
december 11, 2014

reference
V1.4

page
2/7

SMT172

Preliminary

Absolute Maximum Rating

T_A= 25°C. All voltages are referenced to GND, unless o therwise noted.

Power supply voltage	-0.5 V to 7 V
Output Pin load	50 mA
ESD protection (HBM)	+2000 V
Junction temperature	+200°C
Soldering temperature (SOIC, SOT)	+260°C (10 s)

Specification

T_A= -45°C to 130°C, Vcc=2.7 V to 5.5 V, unless otherwise noted.

Parameter	Min	Typ	Max	Unit	Conditions
Supply Voltage	2.7		5.5	V	
Active current[1]		50		µA	T_A = -45 °C, Vcc = 2.7 V, no load at the output pin
		60		µA	T_A = 25 °C, Vcc = 3.3 V, no load at the output pin
		70		µA	T_A = 25 °C, Vcc = 5.5 V, no load at the output pin
Average current		220		nA	T_A = 25 °C, Vcc = 3.3 V, one sample per second, each sample is based on average of 16 output periods.
Power down current			1	µA	When controlling with dedicated PD pin, Vcc = 3.3 V
			0	µA	When controlling with Vcc pin
Inaccuracy[2]			0.25	°C	-10 °C to 100 °C
			0.8	°C	-45 °C to 130 °C
Noise[3]			0.002	°C	T_A = 25 °C, Vcc = 5 V, 3.6 ms measurement time (8 periods)
Output frequency	0.5		7	kHz	
PSRR at DC			0.1	°C/V	
Repeatability[4]			0.01	°C	T_A = 25°C
Long term drift			0.05	°C	Measured under 200 °C stress test condition for 48 h
Output impedance			100	Ω	
Operating Temperature	-45		130	°C	
Storage Temperature	-50		150	°C	

[1] Continuous conversion

[2] TO-18 package, all errors included. For other types of package, see section "understanding the specifications"-"package induced error". For an inaccuracy of 0.1 °C another convers ion formula is needed; please contact the factory.

[3] Noise level will be reduced by averaging multiple consecutive samples, for instance noise can be reduced to 0.0004 °C by taking average in 0.1s, so the measurement time should always be provided when mentioning noise figures. The lower limit of the noise is determined by the flicker noise of the sensor, where further averaging will no longer reduce the noise.

[4] Repeatability is defined as difference between multiple measurements on the same temperature point during multiple temperature cycles.

last update
december 11, 2014

reference
V1.4

page
3/7

SMT172

Preliminary

Output Signal

According to tradition, the Smartec temperature sensors have a duty cycle (PWM) output that can be directly interfaced with a microcontroller without the use of extra components. The output is a square wave with a well-defined temperature-dependent duty cycle. In general, the duty cycle of the output signal is defined by a linear equation:

$$DC = 0.32 + 0.0047 \times T$$

where DC = Valid Duty Cycle
T = Temperature in ℃

A simple calculation shows that, i.e. at 0℃, DC= 0.32 (32%); at 130℃, DC= 0.931 (93.1%).

Temperature is then derived from the measured duty cycle by:

$$T = \frac{DC - 0.32}{0.0047} = 212.77 \times DC - 68.085 \tag{1}$$

The frequency of the output of the sensor is fixed and contains no temperature information. Only the duty cycle contains temperature information in accordance to the formula given above. The output signal may show a low frequency jitter or drift. Therefore most oscilloscopes and counters are not suited for verifying the accuracy of these sensors. However, the duty-cycle value is guaranteed to be accurate within the values specified for each type (housing).
A higher accuracy can be achieved when a second order conversion formula is used, an inaccuracy of 0.1 ℃ can be reached in the range of -20 ℃ to 80 ℃. Please contact the factory.

Valid Duty Cycle

A valid duty cycle in equation (1) is defined as the average of individual duty cycles from 8 consequent output periods. This is due to the internal working principle of the SMT172 sensor. The difference of duty cycle between individual periods within the 8 period output can be relatively large and also different from sensor to sensor, but the averaged value (valid duty cycle) is very stable and precise.
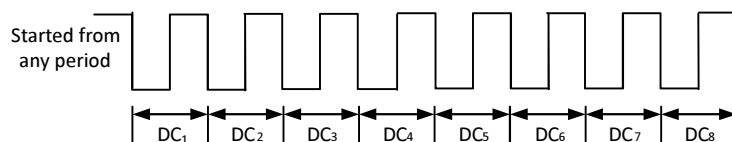
Therefore a valid duty cycle is: $DC = \frac{\sum_{i=1}^{8} DC_i}{8}$

$DC_i = \frac{t_{Hi}}{t_{Li} + t_{Hi}}$ Where t_{Hi}: time interval of high cycle

t_{Li}: time interval of low cycle

DC_i duty cycle of individual period i

DC the final duty cycle

For improved noise performance, a measurement of multiples (*N* times) of 8 periods is recommended. Measurement always starts on the negative edge of the output signal.

last update
december 11, 2014

reference
V1.4

page
4/7

SMT172

Preliminary

Understanding the specifications

Noise

From the theory of signal processing it can be derived that there is a fixed ratio between the sensor's signal frequency, the sampling rate and the sampling noise. The sampling rate limits the measurement accuracy to:

$$T_{err} = 200 \frac{t_s}{\sqrt{6 t_m t_p}}$$

Where T_{err} = measurement uncertainty (= standard deviation of the sampling noise)
t_s = microcontrollers sampling rate
t_m = total measurement time
t_p = period of the sensor output

Note:
The above mentioned error T_{err} is NOT related to the intrinsic accuracy of the sensor. It just indicates how the uncertainty (standard deviation) is influenced when a microcontroller samples a time signal.

Package induced error

When applying high stress package materials, extra errors will occur and therefore system designers should be aware of this effect. The TO-18 package has the minimum package induced errors. All other packages can have a slightly bigger error on top of the error in the specifications. **STILL UNDER INVESTIGATION** but based on the recent measurements on the plastic versions TO92 and TO220 the error will be less than 0.45 ℃ (-10℃ to 100 ℃).

Long-term drift

This drift strongly depends on the operating condition. The measured hysteresis in a thermal cycle (TO-18 packaged samples) is less than 0.01 ℃ over the whole temperature range. Even at extreme condition (TO-18 samples heated to 200 ℃ for 48 hours), the drift is still less than 0.05 ℃ over the whole temperature range.

last update
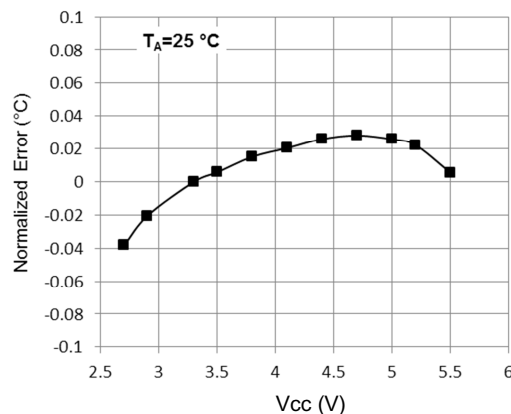december 11, 2014

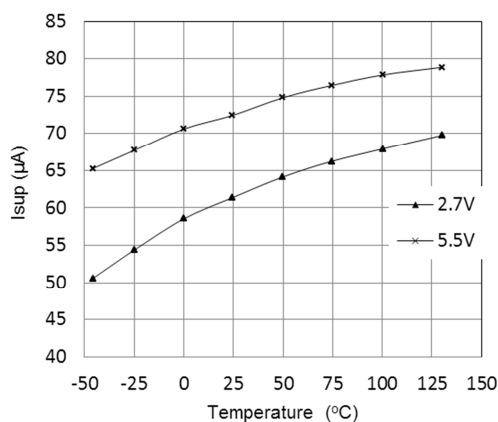reference
V1.4

page
5/7

SMT172

Preliminary

Typical Performance Characteristics
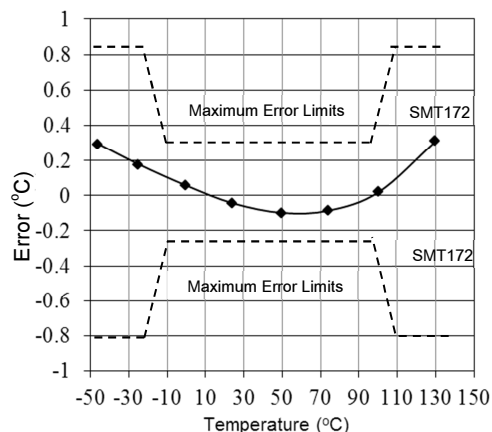
Inaccuracy vs. Temperature

Normalized Error vs. Supply Voltage

Supply current vs. Temperature

Application Information

Temperature measurement

The SMT172 measures the temperature of its bipolar transistors with high precision. Due to the great thermal conducting property of single crystalline silicon, we can assume the temperature difference within the sensor die to be negligible. However the thermal property of the package material, the shape and the size of soldering pads, the neighbouring components on the PCB as well as the presence of dedicated thermal sinks are all affecting the die temperature that the sensor is measuring. Therefore a good thermal path between the die and the objects under measurement should be carefully designed and considered. When measuring temperature of solid or liquid targets, it helps to have a good thermal contact between the sensor and the target. This can be achieved with metals and thermal paste. When measuring air

last update
december 11, 2014

reference
V1.4

page
6/7

SMT172

Preliminary

temperatures, it is important to isolate the sensor from the rest of the measurement system, so that the heating from the surrounding circuit components has only a small influence on the sensor temperature.

Self-Heating

All electronic circuits consume power, and all power becomes heat. Depending on the thermal resistance to the environment and the related thermal mass on the heat path, this heat will cause an extra temperature rise of the sensor die and will influence the final reading. Although the ultra-low power consumption of SMT172 sensor minimizes this effect greatly, it is always important to take this into account when designing a temperature measurement system. Design considerations like optimal thermal contact with the environment and powering down the sensors whenever possible (see SMTAS08) are all useful techniques to minimize this effect.
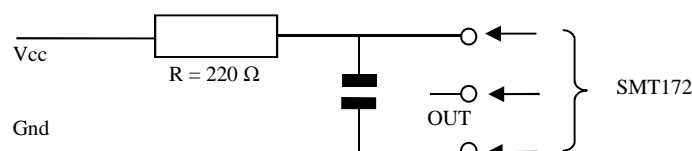
Thermal response time

The thermal response time of the temperature sensor is determined by both the thermal conductance and the thermal mass between the heat source and the sensor die. Depending on the packaging material and the immerging substances, this can vary in a wide range from sub-second to hundreds of seconds. The following table illustrates the time constant (the time required to reach 63 % of an instantaneous temperature change) of TO-18 packaged sensors.

Conditions of installation	Time constant (s) (TO-18)
In an aluminium block	0.6
In a bath filled with oil that is stirred constantly	1.4
In air that moves at 3 m/s: - Without heat sink - With heat sink	13.5 5
In non-moving air: - Without heat sink - With heat sink	60 100

Supply decoupling

It is common practice for precision analogue ICs to use a decoupling capacitor between Vcc and GND pins. This capacitor ensures a better overall EMI/EMC performance. When applied, this capacitor should be a ceramic type and have a value of approximately 100 nF. The location should be as close to the sensor as possible.
It is advised to also add a series resistor of 220 Ohm, so a low pass filter is created. This will enhance the EMC performance and it will limit the maximum current in case of faults or wrong connections.
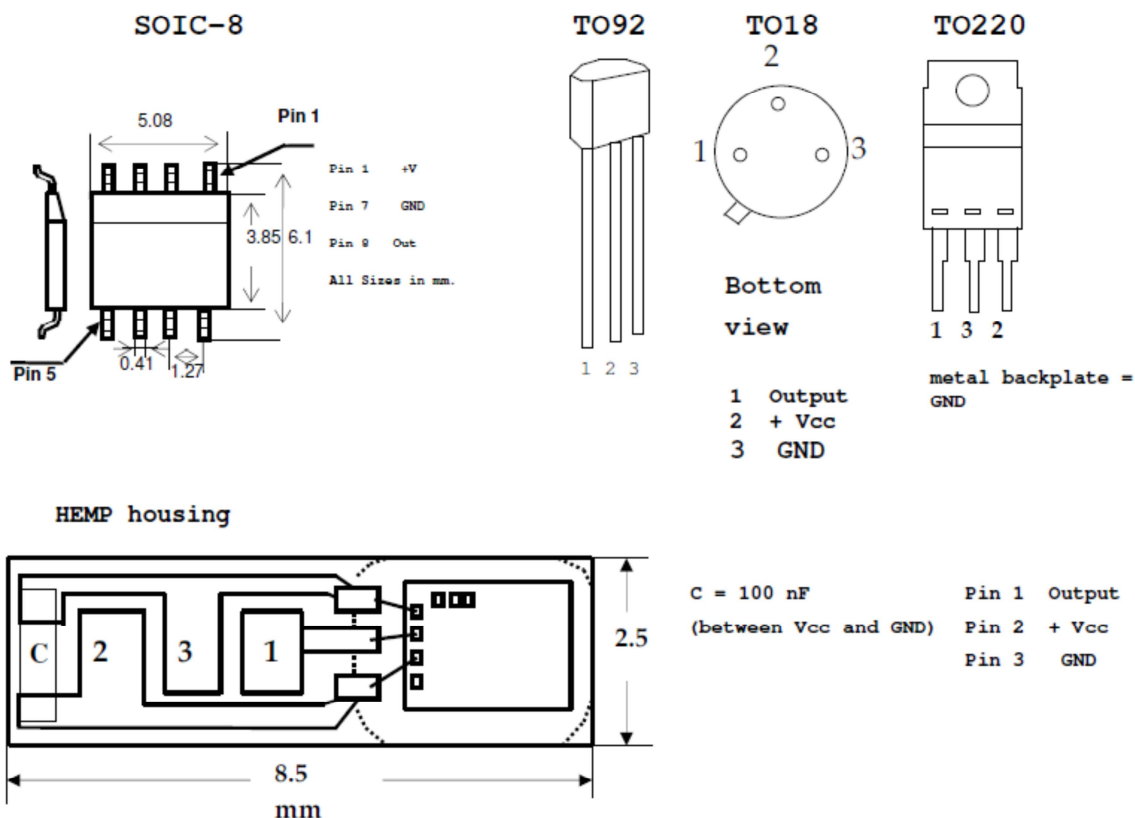
Vcc   R = 220 Ω   OUT   SMT172

Gnd

last update
december 11, 2014

reference
V1.4

page
7/7

SMT172

Preliminary

Packaging

Pin-out and housing.

SOIC-8

Pin 1 +V
Pin 7 GND
Pin 9 Out
All Sizes in mm.

TO92 TO18 TO220

Bottom view

1 Output
2 + Vcc
3 GND

metal backplate = GND

HEMP housing

C = 100 nF
(between Vcc and GND)

Pin 1 Output
Pin 2 + Vcc
Pin 3 GND

Ordering code:

SMT172-SOT89	**SMT172 in SOT89 encapsulation (Not yet available)**
SMT172-TO18	**SMT172 in TO-18 encapsulation**
SMT172-TO92	**SMT172 in TO-92 encapsulation**
SMT172-TO220	**SMT172 in TO-220 encapsulation**
SMT172-HEMP	**SMT172 in HEMP encapsulation**
SMT172-DIE	**SMT172 DIE (die size 1.1 x 1.5 mm)**

Related products

SMTAS04	**evaluation board for 4 sensors input (RS232)**
SMTAS04USB	**evaluation board for 4 sensors input (USB connection)**
SMTAS08	**evaluation board for 8 sensors input (RS232)**
SMTAS08USB	**evaluation board for 8 sensors input (USB connection)**