Digital Signal Processing

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Contents

***Abstract*—This manual provides a simple introduction to digital signal processing.**

1 Software Installation

Run the following commands

```
sudo apt−get update
sudo apt−get install libffi−dev libsndfile1 python3
    −scipy python3−numpy python3−matplotlib
sudo pip install cffi pysoundfile
```

2 Digital Filter

2.1 Download the sound file from

```
wget https://raw.githubusercontent.com/
    gadepall/
EE1310/master/filter/codes/Sound_Noise.wav
```

2.2 You will find a spectrogram at https://academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find?

Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.

2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal

#read .wav file
input_signal,fs = sf.read('Sound_Noise.wav'
    )

#sampling frequency of Input signal
sampl_freq=fs

#order of the filter
order=4

#cutoff frquency 4kHz
cutoff_freq=4000.0

#digital frequency
Wn=2*cutoff_freq/sampl_freq

# b and a are numerator and denominator
    polynomials respectively
b, a = signal.butter(order,Wn, 'low')

#filter the input signal with butterworth filter
output_signal = signal.filtfilt(b, a,
    input_signal)
#output_signal = signal.lfilter(b, a,
    input_signal)

#write the output signal into .wav file
sf.write('Sound_With_ReducedNoise.wav',
    output_signal, fs)
```

2.4 The output of the python script in Problem 2.3 is the audio file

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Sound_With_ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?
Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch $x(n)$.

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2), \\ y(n) = 0, n < 0 \tag{3.2}$$

Sketch $y(n)$.
Solution: The following code yields Fig. 3.2.

```
wget https://github.com/gadepall/EE1310/raw/
    master/filter/codes/xnyn.py
```

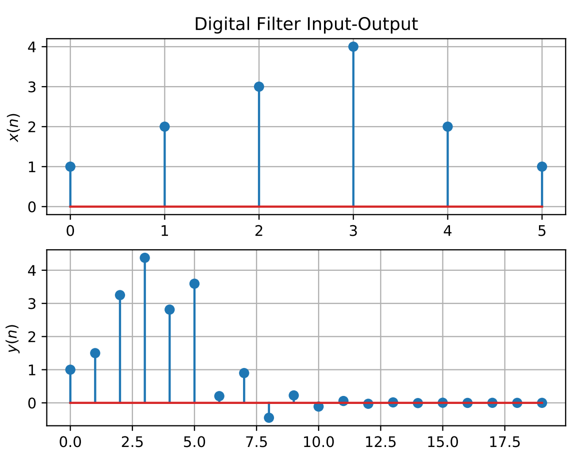

Fig. 3.2

3.3 Repeat the above exercise using a C code. **solution:** The following code is the implementation in C.

```
https://github.com/MA20BTECH11020/
    EE−3900/blob/main/EE3900−main/
    filter/codes/3_3.py
```

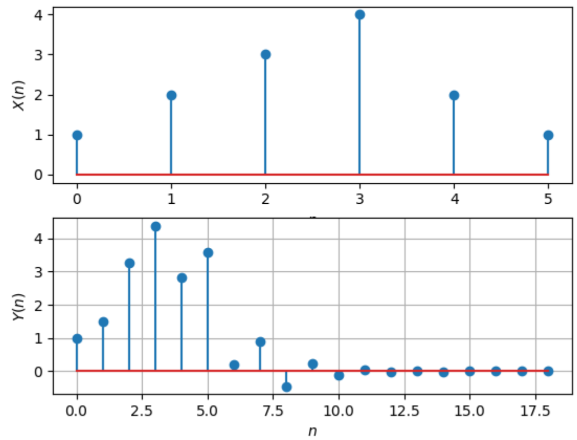

Fig. 3.3

4 $\mathcal{Z}$-transform

4.1 The $\mathcal{Z}$-transform of $x(n)$ is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \tag{4.1}$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \tag{4.2}$$

and find

$$\mathcal{Z}\{x(n-k)\} \tag{4.3}$$

Solution: From (4.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n} \tag{4.4}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n} \tag{4.5}$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \tag{4.6}$$

4.2 Obtain $X(z)$ for $x(n)$ defined in problem 3.1.
Solution:

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \tag{4.7}$$

But

$$x(n) = \{1, 2, 3, 4, 2, 1\} \tag{4.8}$$

so,

$$X(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5} \quad (4.9)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \quad (4.10)$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z) \quad (4.11)$$

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (4.12)$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases} \quad (4.13)$$

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & \text{otherwise} \end{cases} \quad (4.14)$$

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (4.15)$$

Solution: It is easy to show that

$$\delta(n) \overset{\mathcal{Z}}{\rightleftharpoons} 1 \quad (4.16)$$

and from (4.14),

$$U(z) = \sum_{n=0}^{\infty} z^{-n} \quad (4.17)$$

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (4.18)$$

using the formula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \overset{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \quad (4.19)$$

Solution: let

$$f(n) = a^n u(n) \quad (4.20)$$

$$f(n) = \left\{ \begin{array}{ll} a^n, & \text{if } n > 0 \\ 0, & \text{otherwise} \end{array} \right\} \quad (4.21)$$

Now the Z- Transform of f(n) is

$$F(z) = \mathcal{Z}\{f(n)\} = \sum_{n=-\infty}^{\infty} f(n)z^{-n} \quad (4.22)$$

$$F(z) = \sum_{n=0}^{\infty} a^n z^{-n} \quad (4.23)$$

This forms a infinite Geometric Progression.

$$F(z) = \frac{1}{1 - az^{-1}} \text{ for } z < a. \quad (4.24)$$

4.6 Let

$$H\left(e^{j\omega}\right) = H\left(z = e^{j\omega}\right). \quad (4.25)$$

Plot $\left|H\left(e^{j\omega}\right)\right|$.Is it periodic? If so, find the period. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of $x(n)$.

Solution:

$$|H(e^{j\omega})| = \frac{|(1 + cos(2\omega) - isin(2\omega))|}{|(1 + \frac{1}{2}cos(\omega)| - i\frac{1}{2}sin(\omega))|} \quad (4.26)$$

$$|H(e^{j\omega})| = \frac{|2cos(\omega)|}{|\sqrt{\frac{5}{4} + cos(\omega)}|} \quad (4.27)$$

$$|H(e^{j\omega})| = \frac{|4cos(\omega)|}{|\sqrt{5 + 4cos(\omega)}|} \quad (4.28)$$

we can see that the period of $H(e^{j\omega})$ is same as of $cos(\omega)$ which is 2π.
Hence the period is 2π.
The following code plots Fig.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/dtft.
    py
```

4.7 Express $x(n)$ in terms of $H\left(e^{j\omega}\right)$.

Solution: We have,

$$H(e^{\jmath\omega}) = \sum_{k=-\infty}^{\infty} h(k)e^{-\jmath\omega k} \quad (4.29)$$

However,

$$\int_{-\pi}^{\pi} e^{\jmath\omega(n-k)}d\omega = \begin{cases} 2\pi & n = k \\ 0 & \text{otherwise} \end{cases} \quad (4.30)$$

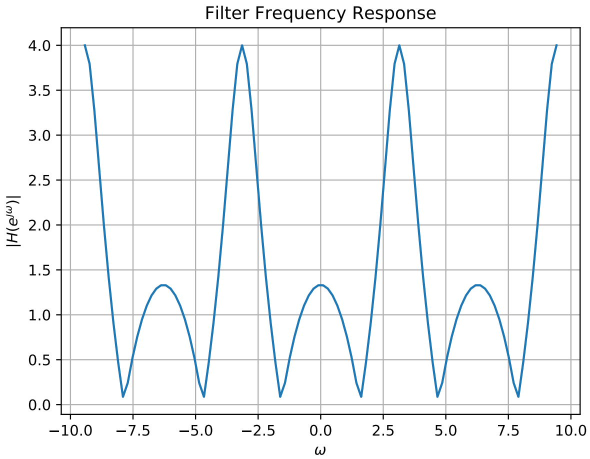

Fig. 4.6: $\left|H\left(e^{j\omega}\right)\right|$

and so,

$$\frac{1}{2\pi}\int_{-\pi}^{\pi} H(e^{\jmath\omega})e^{j\omega n}d\omega \tag{4.31}$$

$$= \frac{1}{2\pi}\sum_{k=-\infty}^{\infty}\int_{-\pi}^{\pi} h(k)e^{\jmath\omega(n-k)}d\omega \tag{4.32}$$

$$= \frac{1}{2\pi}2\pi h(n) = h(n) \tag{4.33}$$

which is known as the Inverse Discrete Fourier Transform. Thus,

$$h(n) = \frac{1}{2\pi}\int_{-\pi}^{\pi} H(e^{\jmath\omega})e^{\jmath\omega n}d\omega \tag{4.34}$$

$$= \frac{1}{2\pi}\int_{-\pi}^{\pi} \frac{1+e^{-2\jmath\omega}}{1+\frac{1}{2}e^{-\jmath\omega}}e^{\jmath\omega n}d\omega \tag{4.35}$$

5 Impulse Response

5.1 Using long divsion, find

$$h(n), n < 5 \tag{5.1}$$

for $H(z)$ in (4.12)

Solution: from (4.12)

$$H(z) = \frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{5.2}$$

$$\begin{array}{r} 1+\frac{1}{2}z^{-1}\,\overline{)1+z^{-2}}\,(2z^{-1} \\ \underline{2z^{-1}+z^{-2}} \\ 1+\frac{1}{2}z^{-1}\,)1-2z^{-1}(-4 \\ \underline{-4-2z^{-1}} \\ 5+0z^{-1} \end{array}$$

Hence by long division will be

$$H(z) = 2z^{-1} - 4 + \frac{5}{1+\frac{1}{2}z^{-1}} \tag{5.3}$$

5.2 Find an expression for $h(n)$ using $H(z)$, given that

$$h(n) \overset{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.4}$$

and there is a one to one relationship between $h(n)$ and $H(z)$. $h(n)$ is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.12),

$$H(z) = \frac{1}{1+\frac{1}{2}z^{-1}} + \frac{z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{5.5}$$

$$\implies h(n) = \left(-\frac{1}{2}\right)^{n}u(n) + \left(-\frac{1}{2}\right)^{n-2}u(n-2) \tag{5.6}$$

using (4.19) and (4.6).

5.3 Sketch $h(n)$. Is it bounded? Convergent?

Solution: The following code plots Fig. 5.3.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/hn.py
```

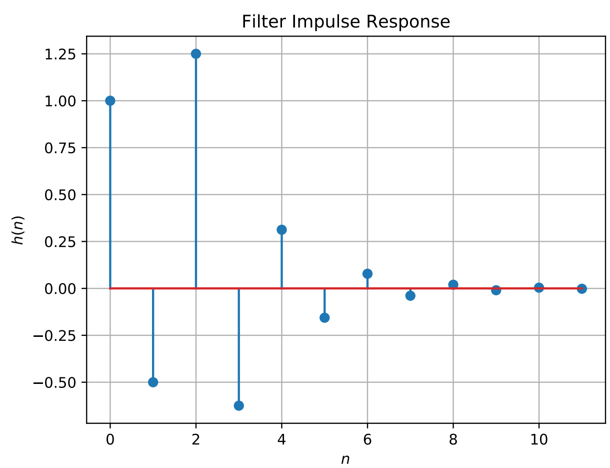

Fig. 5.3: $h(n)$ as the inverse of $H(z)$

5.4 The system with $h(n)$ is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.7}$$

Is the system defined by (3.2) stable for the impulse response in (5.4)?

Solution: from 5.6

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.8)$$

then

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=0}^{\infty}\left(-\frac{1}{2}\right)^n + \sum_{n=2}^{\infty}\left(-\frac{1}{2}\right)^{n-2} \quad (5.9)$$

$$\sum_{n=-\infty}^{\infty} h(n) = \frac{4}{3} \quad (5.10)$$

since

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \quad (5.11)$$

$h(n)$ is stable.

5.5 Compute and sketch $h(n)$ using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.12)$$

This is the definition of $h(n)$.
Solution: The following code plots Fig. 5.5. Note that this is the same as Fig. 5.3.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/hndef
    .py
```

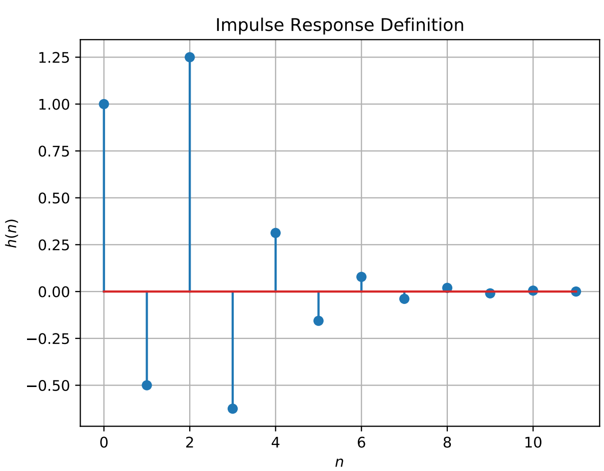

Fig. 5.5: $h(n)$ from the definition

5.6 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.13)$$

Comment. The operation in (5.13) is known as *convolution*.

Solution: The following code plots Fig. 5.6. Note that this is the same as $y(n)$ in Fig. 3.2.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/
    ynconv.py
```

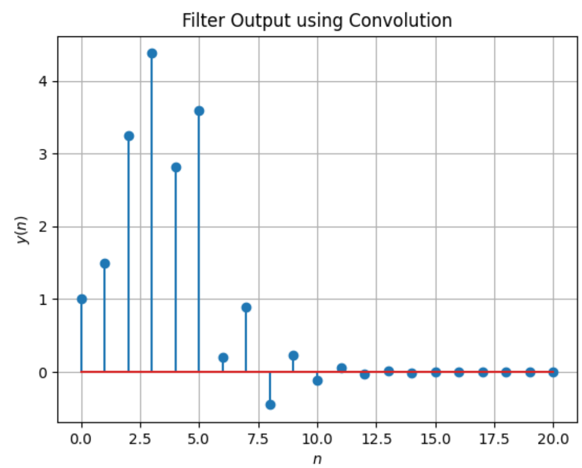

Fig. 5.6: $y(n)$ from the definition of convolution

5.7 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (5.14)$$

Solution: from 5.13 ww know that

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.15)$$

now consider

$$t = n - k \quad (5.16)$$

5.15 will transform into

$$y(n) = \sum_{n-t=-\infty}^{\infty} x(n-t)h(t) \quad (5.17)$$

since n is finite and $-\infty < \infty$, 5.17 is equivalent to

$$y(n) = \sum_{t=-\infty}^{\infty} x(n-t)h(t) \quad (5.18)$$

hence proved.

6 DFT and FFT

6.1 Compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n) e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \tag{6.1}$$

and $H(k)$ using $h(n)$.
Solution: The following code plots $X(k)$ and $H(k)$.

```
https://github.com/MA20BTECH11020/
    EE−3900/blob/main/EE3900−main/
    filter/codes/XkHk_dft.py
```

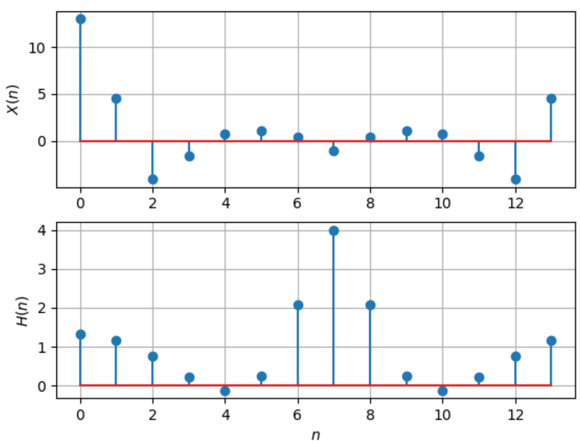

Fig. 6.1

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

Solution: The following code plots $Y(k)$.

```
https://github.com/MA20BTECH11020/
    EE−3900/blob/main/EE3900−main/
    filter/codes/Yk.py
```

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1 \tag{6.3}$$

Solution: The following code plots Fig. 5.6. Note that this is the same as $y(n)$ in Fig. 3.2.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/yndft.
    py
```

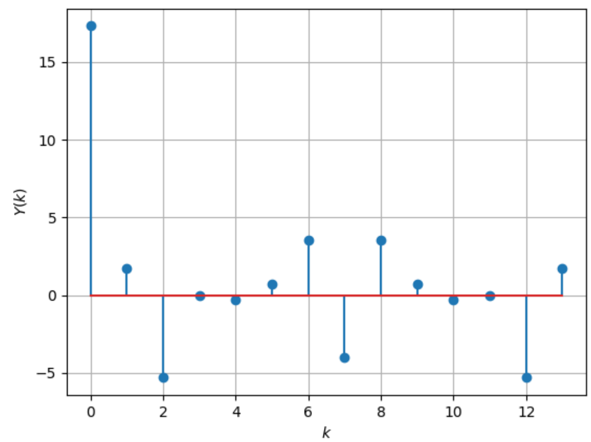

Fig. 6.2

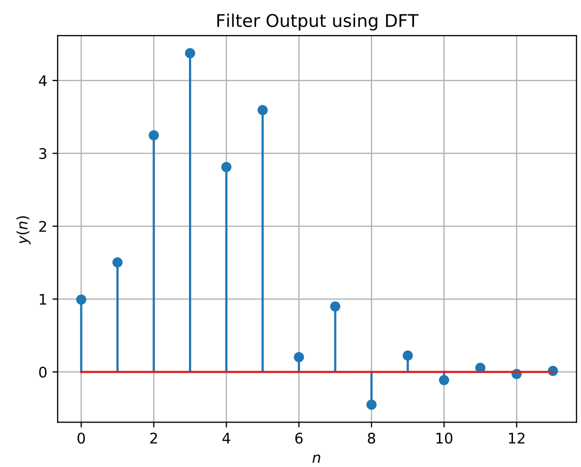

Fig. 6.3: $y(n)$ from the DFT

6.4 compare $y(n)$ obtained in 6.3 and 3.3.
Solution:

6.5 Repeat the previous exercise by computing $X(k), H(k)$ and $y(n)$ through FFT and IFFT.
Solution: The following code plots $X(n)$, $H(n)$ and $y(n)$ by fft.

```
https://github.com/MA20BTECH11020/
    EE−3900/blob/main/EE3900−main/
    filter/codes/Xk_Hk_fft.py
```

6.6 Wherever possible, express all the above equations as matrix equations.
Solution:

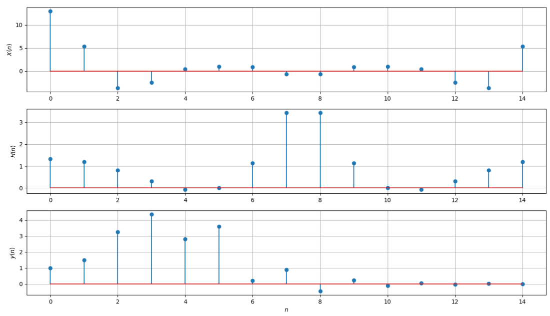

Fig. 6.5: $X(k)$, $H(k)$ and $y(n)$ from fft and IFFT

$$x = \begin{pmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \\ 0 \\ . \\ . \\ . \\ 0 \end{pmatrix}$$

$$J = \begin{pmatrix} 0 \\ e^{\frac{-2\pi j(1)k}{N}} \\ e^{\frac{-2\pi j(2)k}{N}} \\ . \\ . \\ . \\ e^{\frac{-2\pi j(N-1)k}{N}} \end{pmatrix}$$

$$X(k) = x^T J \tag{6.4}$$

$$h = \begin{pmatrix} h[0] \\ h[1] \\ . \\ . \\ . \\ h[N-1] \end{pmatrix}$$

$$H(k) = h^T J \tag{6.5}$$

$$y = \begin{pmatrix} h[0]x[0] \\ h[1]x[1] \\ . \\ . \\ . \\ h[N-1]x[N-1] \end{pmatrix}$$

$$Y(k) = y^T J \tag{6.6}$$

7 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

```
output_signal = signal.lfilter(b, a,
    input_signal)
```

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m)\, y(n-m) = \sum_{k=0}^{N} b(k)\, x(n-k) \tag{7.1}$$

where the input signal is $x(n)$ and the output signal is $y(n)$ with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.
Solution:

```
https://github.com/MA20BTECH11020/
    EE-3900/blob/main/EE3900-main/
    filter/codes/7_1.py
```

7.2 Repeat all the exercises in the previous sections for the above a and b.

7.3 What is the sampling frequency of the input signal?
Solution: run the following code to Sampling frequency(fs)=44.1kHZ.

7.4 What is type, order and cutoff-frequency of the above butterworth filter
Solution: The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

7.5 Modifying the code with different input parameters and to get the best possible output.