# Introduction to Modern Controls Modeling of Dynamic Systems

## Why modeling?

#### Modeling of physical systems:

- a vital component of modern engineering
- often consists of complex coupled differential equations
- only when we have good understanding of a system can we optimally control it:
  - can simulate and predict actual system response, and
  - design model-based controllers

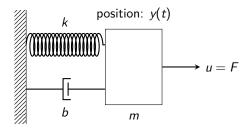
## Two general approaches of modeling

- based on physics:
  - using fundamental engineering principles such as Newton's laws, energy conservation, etc
- based on measurement data:
  - using input-output response of the system
  - a field itself known as system identification

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#### often the tools are combined in practice

## Example: Mass spring damper

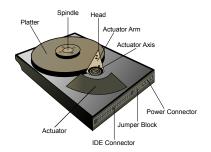


Newton's second law gives

$$m\ddot{y}(t) + b\dot{y}(t) + ky(t) = u(t), \ y(0) = y_0, \ \dot{y}(0) = \dot{y}_0$$

ullet modeled as a second-order ODE with input u(t) and output y(t)

## Example: HDD



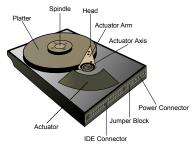
Newton's second law for rotation

$$\sum_{i} \tau_{i} = \underbrace{J}_{\text{moment of inertia angular acceleration}} \alpha$$
 net torque

ullet letting heta :=output and au :=input yields

$$\ddot{\theta} = \alpha = \frac{1}{J}\tau$$

## Example: HDD



$$\ddot{\theta} = \alpha = \frac{1}{J}\tau \Leftrightarrow \Theta(s) = \frac{1}{Js^2}T(s)$$

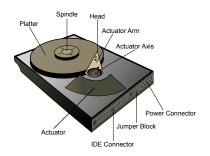
with damping:

$$\ddot{\theta} + 2\zeta\omega_n\dot{\theta} + \omega_n^2\theta = \kappa\tau \Leftrightarrow \Theta(s) = \frac{\kappa}{s^2 + 2\zeta\omega_n s + \omega_n^2} T(s)$$

• with multiple modes:

$$\ddot{\theta}_i + 2\zeta_i \omega_i \dot{\theta}_i + \omega_i^2 \theta_i = \kappa_i \tau \Leftrightarrow \Theta_i(s) = \frac{\kappa_i}{s^2 + 2\zeta_i \omega_i s + \omega_i^2} T(s)$$

## Example: HDD

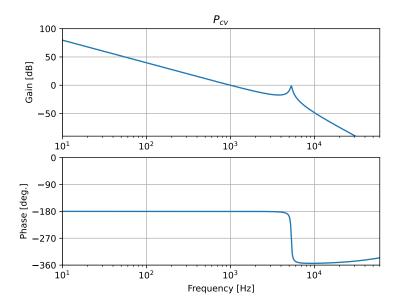


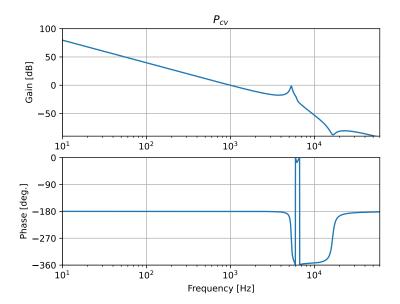
$$\ddot{\theta}_i + 2\zeta_i\omega_i\dot{\theta}_i + \omega_i^2\theta_i = \kappa_i\tau \Leftrightarrow \Theta_i(s) = \frac{\kappa_i}{s^2 + 2\zeta_i\omega_is + \omega_i^2}T(s)$$

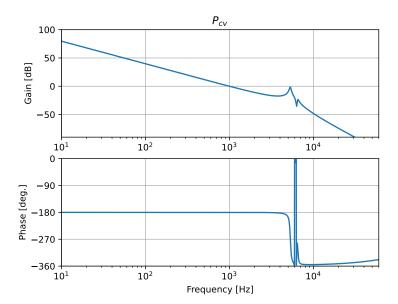
• final model:

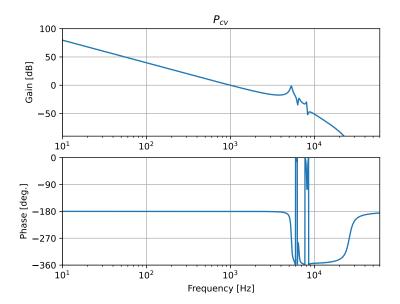
$$\Theta(s) = \sum_{i=1}^{n} \frac{\kappa_i}{s^2 + 2\zeta_i \omega_i s + \omega_i^2} T(s)$$

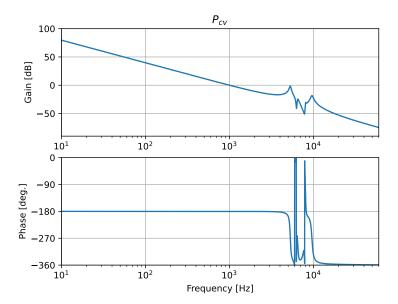
```
import numpy as np
import matplotlib.pyplot as plt
from scipy import signal
import control as ct
num_sector = 420 # Number of sector
num_rpm = 7200 # Number of RPM
Kp_vcm = 3.7976e+07 # VCM gain
omega_vcm = np.array([0, 5300, 6100, 6500, 8050, 9600, 14800, 17400,
                     21000, 26000, 26600, 29000, 32200, 38300, 43300,
                     → 44800]) * 2 * np.pi
kappa_vcm = np.array([1, -1.0, +0.1, -0.1, 0.04, -0.7, -0.1])
                     0.2, -1.0, +3.0, -3.2, 2.1, -1.5, +2.0, -0.2,
                     \rightarrow +0.3, -0.51)
zeta_vcm = np.array([0, 0.02, 0.04, 0.02, 0.01, 0.03, 0.01,
                    0.02, 0.02, 0.012, 0.007, 0.01, 0.03, 0.01, 0.01,
                    \rightarrow 0.011)
Sys_Pc_vcm_c1 = ct.TransferFunction([], [1]) # Create an empty
for i in range(len(omega_vcm)):
    Sys_Pc_vcm_c1 = Sys_Pc_vcm_c1 + ct.TransferFunction(np.array(
        [0, 0, kappa_vcm[i]]) * Kp_vcm, np.array([1, 2 * zeta_vcm[i] *
        → omega_vcm[i], (omega_vcm[i]) ** 2]))
```

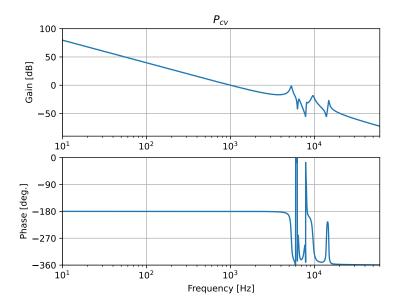


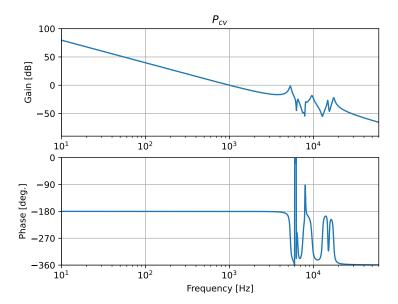


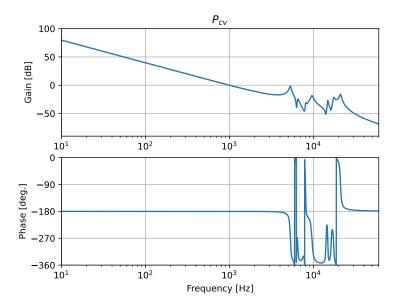


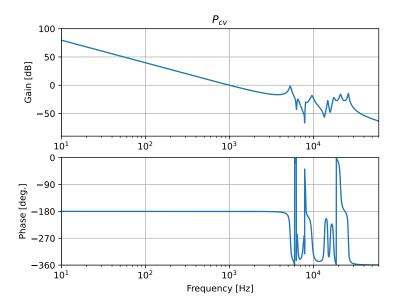


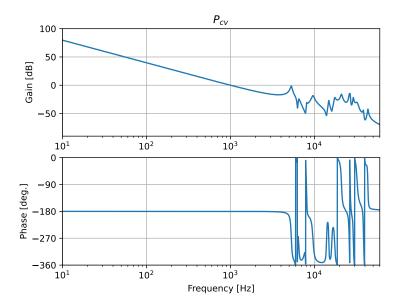






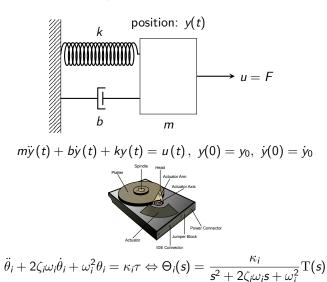






## Models of continuous-time systems

modeled as differential equations:



## Models of continuous-time systems

General continuous-time systems:

$$\frac{d^{n}y(t)}{dt^{n}} + a_{n-1}\frac{d^{n-1}y(t)}{dt^{n-1}} + \cdots + a_{0}y(t) = b_{m}\frac{d^{m}u(t)}{dt^{m}} + b_{m-1}\frac{d^{m-1}u(t)}{dt^{m-1}} + \cdots + b_{0}u(t)$$

with the initial conditions  $y(0) = y_0, \dots, y^{(n)}(0) = y_0^{(n)}$ .

## Models of discrete-time systems

#### General discrete-time systems

- inputs and outputs defined at discrete time instances k = 1, 2, ...
- described by ordinary difference equations in the form of

$$y(k)+a_{n-1}y(k-1)+\cdots+a_0y(k-n)=b_mu(k+m-n)+\cdots+b_0u(k-n)$$

#### Example: bank statements

- $x(k+1) = (1+\rho)x(k) + u(k), x(0) = x_0$
- k month counter;  $\rho$  interest rate; x(k) wealth at the beginning of month k; u(k) money saved at the end of month k;  $x_0$  initial wealth in account

# Model properties: static v.s. dynamic, causal v.s. acausal

$$u \longrightarrow \overline{\mathcal{M}} \longrightarrow y$$

Model  $\mathcal{M}$  is said to be

- memoryless or static if y(t) depends only on u(t)
- dynamic (has memory) if y at time t depends on input values at other times
- e.g.:  $y(t) = \mathcal{M}(u(t)) = \gamma u(t), \ y(t) = \int_0^t u(\tau) d\tau, \ y(k) = \sum_{i=0}^k u(i)$
- causal if y(t) depends on  $u(\tau)$  for  $\tau \leq t$
- strictly causal if y(t) depends on  $u(\tau)$  for  $\tau < t$ , e.g.: y(t) = u(t-10)

## Linearity and time-invariance

#### The system ${\mathcal M}$ is called

• linear if satisfying the superposition property:

$$\mathcal{M}(\alpha_1 u_1(t) + \alpha_2 u_2(t)) = \alpha_1 \mathcal{M}(u_1(t)) + \alpha_2 \mathcal{M}(u_2(t))$$

for any input signals  $u_1(t)$  and  $u_2(t)$ , and any real numbers  $\alpha_1$  and  $\alpha_2$ 

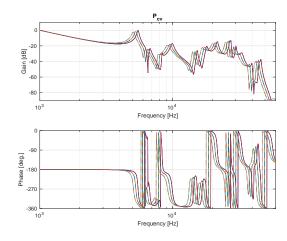
- time-invariant if its properties do not change with respect to time
- e.g.,  $\dot{y}(t) = Ay(t) + Bu(t)$  is linear and time-invariant
- $\dot{y}(t) = 2y(t) \sin(y(t))u(t)$  is nonlinear, yet time-invariant
- $\dot{y}(t) = 2y(t) t\sin(y(t))u(t)$  is time-varying
- assuming the same initial conditions, if we shift u(t) by a constant time interval, i.e., consider  $\mathcal{M}(u(t+\tau_0))$ , then  $\mathcal{M}$  is time-invariant if the output  $\mathcal{M}(u(t+\tau_0)) = y(t+\tau_0)$

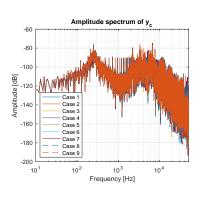
### George Box

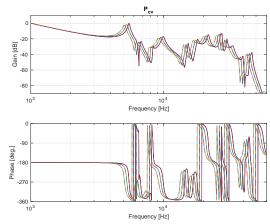
- "All Models are Wrong, but Some are Useful"
  - statistical models always fall short of the complexities of reality but can still be useful nonetheless
  - a dynamic system may simply be too complex (consider the neural system of human brains)
  - or there are inevitable hardware uncertainties such as the fatigue of gears or bearings in a car

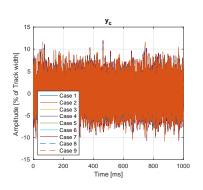


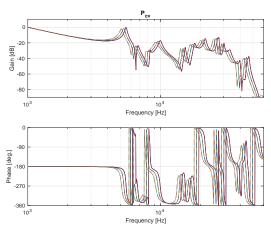
- temperature influence
- manufacturing variations
- but, control works!













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#### Benchmark Problem for Magnetic-Head Positioning Control System in HDDs

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Abstract: In the "Cloud Ers", the data capacity of the hard disk drive (HDD) must grow to develop the cloud. As a result, we must improve the positioning accuracy of the magnetic-lead in the HDD. To encourage research about magnetic-lead meditioning control, we release a the magnetic-head positioning control system for the latest HDDs with our designed controller In this paper, a control design method with the decoupling filter is also presented for this

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Keywords: Precision control, Data storage, Positioning systems, Actuators, Servo

According to a major data-storage device manufacturer. pendent on the hard disk drive (HDD) capacity growth are rapidly increasing. To solve this issue, we are going

decreases (Yamaguchi and Atsumi (2008); Abrahamson In order to encourage research about the magnetic-head representatives of major universities and an HDD man-ufacturer with HDD serve research in Japan has developed an open-source HDD benchmark uroblem and re-

This paper presents the details of the benchmark problems

2. HARD DISK DRIVE

Figure 1 shows a picture of the HDD with the cover opened. The HDD consists of a voice coil motor (VCM).



Fig. 2. Magnetic-head positioning system

PZT actuators, a head-stack assembly (BSA), magnetic

bruds, disks, and a spinsile motor. Most of the latest HDDs for cloud storage employ beliam-scaled technology (Account et al. (2022)). This means that flow-induced

centers. To compensate for the external vibrations, the latest HDDs combow the dual-stage actuator system that

consists of the VCM and the PZT actuators. Figure 2

illustrates the magnetic-head positioning control system

In the marnetir-head positioning system, the controlled

Fig. 3. Thus, the magnetic-head position signal is only

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