LQG/Loop Transfer Recovery (LTR)

Big picture Loop transfer recovery Target feedback loop Fictitious KF

Big picture

Where are we now?

- ► LQ: optimal control, guaranteed robust stability under basic assumptions in stationary case
- KF: optimal state estimation, good properties from the duality between LQ and KF
- LQG: LQ+KF with separation theorem
- frequency-domain feedback design principles and implementations

Stability robustness of LQG was discussed in one of the homework problems: the nice robust stability in LQ (good gain and phase margins) is lost in LQG.

LQG/LTR is one combined scheme that uses many of the concepts learned so far.

Continuous-time stationary LQG solution

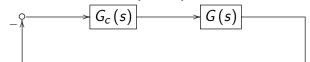
LQG gain
$$K$$

$$\begin{array}{c}
u(t) = -K\hat{x}(t|t) \\
\hline
u(t) = -K\hat{x}(t|t)
\end{array}$$
Kalman Filter
$$\begin{array}{c}
u(t) = -K\hat{x}(t|t) \\
\frac{d\hat{x}(t|t)}{dt} = A\hat{x}(t|t) + Bu(t) + F(y(t) - C\hat{x}(t|t)) \\
= (A - BK - FC)\hat{x}(t|t) + Fy(t)
\end{array}$$

$$\begin{array}{c}
-y(t) \\
-y(t) \\
\hline
G_{c}
\end{array}$$
Plant
$$\begin{array}{c}
y(t) \\
-y(t)
\end{array}$$
Plant
$$\begin{array}{c}
y(t) \\
-y(t)
\end{array}$$
Plant

(1)

Loop transfer recovery (LTR)



Theorem (Loop Transfer Recovery (LTR))

If a $m \times m$ dimensional G(s) has only minimum phase transmission zeros, then the open-loop transfer function

$$G(s) G_c(s) = \left[C(sI - A)^{-1} B \right] \left[K(sI - A + BK + FC)^{-1} F \right]$$

$$\xrightarrow{\rho \to 0} C(sI - A)^{-1} F \quad (2)$$

K and ρ are from the LQ [(A, B) controllable, (A, C) observable]

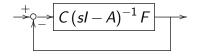
$$J = \int_0^\infty \left(x^T(t) C^T C x(t) + \rho u^T(t) N u(t) \right) dt \tag{3}$$

$$\dot{x}(t) = Ax(t) + Bu(t) \tag{4}$$

Loop transfer recovery (LTR)

$$\xrightarrow{+} G_c(s) \longrightarrow G(s) = C(sI - A)^{-1}B$$

converges, as ho
ightarrow 0, to the *target feedback loop*



key concepts:

- regard LQG as an output feedback controller
- ▶ will design F such that $C(sI A)^{-1}F$ has a good loop shape
- not a conventional optimal control problem
- not even a stochastic control design method

Selection of F for the target feedback loop

standard KF procedure: given noise properties (W, V, etc), KF gain F comes from RE

fictitious KF for target feedback loop design: want to have good behavior in

$$\xrightarrow{+} C(sI - A)^{-1}F$$

select W and V to get a desired F (hence a fictitious KF problem):

$$\dot{x}(t) = Ax(t) + Lw(t),$$
 $E[w(t)w^{T}(t+\tau)] = I\delta(\tau)$
 $y(t) = Cx(t) + v(t),$ $E[v(t)v^{T}(t+\tau)] = \mu I\delta(\tau)$

which gives

$$F = \frac{1}{u}MC^{T}, \quad AM + M^{T}A + LL^{T} - \frac{1}{u}MC^{T}CM = 0, \ M > 0$$
 (5)

The target feedback loop from fictitious KF

$$\dot{x}(t) = Ax(t) + Lw(t),$$
 $E[w(t)w^{T}(t+\tau)] = I\delta(\tau)$
 $y(t) = Cx(t) + v(t),$ $E[v(t)v^{T}(t+\tau)] = \mu I\delta(\tau)$

Return difference equation for the fictitious KF is

$$[I_m + G_F(s)][I_m + G_F(-s)]^T = I_m + \frac{1}{\mu}[C\Phi(s)L][C\Phi(-s)L]^T$$

where
$$G_F(s) = C(sI - A)^{-1} F$$
 and $\Phi(s) = (sI - A)^{-1}$. Then

$$\sigma\left[I_{m}+G_{F}\left(j\omega\right)\right]=\sqrt{\lambda\left\{\left[I_{m}+G_{F}\left(j\omega\right)\right]\left[I_{m}+G_{F}\left(-j\omega\right)\right]^{T}\right\}}$$

$$=\sqrt{1+\frac{1}{\mu}\left\{\sigma\left[C\Phi\left(j\omega\right)L\right]\right\}^{2}}\geq1$$

The (nice) target feedback loop from fictitious KF

$$\sigma\left[I_{m}+G_{F}\left(j\omega\right)\right]=\sqrt{\lambda\left\{\left[I_{m}+G_{F}\left(j\omega\right)\right]\left[I_{m}+G_{F}\left(-j\omega\right)\right]^{T}\right\}}$$

$$=\sqrt{1+\frac{1}{\mu}\left\{\sigma\left[C\Phi\left(j\omega\right)L\right]\right\}^{2}}\geq1$$

gives:

- $\sigma_{\mathsf{max}} S(j\omega) = \sigma_{\mathsf{max}} [I + G_F(j\omega)]^{-1} \leq 1$, namely no disturbance amplification at any frequency
- $\sigma_{\max} T(j\omega) = \sigma_{\max} [I S(j\omega)] \le 2$, hence, guaranteed closed loop stable if $\sigma_{\max} \Delta(j\omega) < 1/2$