

Stochastic State Estimation (Kalman Filter)

Big picture
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Discrete-time Kalman Filter
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Properties
Example

Big picture

why are we learning this?

- ▶ state estimation in deterministic case:

$$\text{Plant: } x(k+1) = Ax(k) + Bu(k), \quad y(k) = Cx(k)$$

$$\text{Observer: } \hat{x}(k+1) = A\hat{x}(k) + Bu(k) + L(y(k) - C\hat{x}(k))$$

- ▶ L designed based on the error ($e(k) = x(k) - \hat{x}(k)$) dynamics:

$$e(k+1) = (A - LC)e(k) \tag{1}$$

to reach fast convergence of $\lim_{k \rightarrow \infty} e(k) = 0$

- ▶ L is not optimal when there is noise in the plant; actually $\lim_{k \rightarrow \infty} e(k) = 0$ isn't even a valid goal when there is noise
- ▶ Kalman Filter provides optimal state estimation under input and output noises

Problem statement

plant: $x(k+1) = A(k)x(k) + B(k)u(k) + B_w(k)w(k)$

$$y(k) = C(k)x(k) + v(k)$$

- ▶ $w(k)$ — s -dimensional input noise; $v(k)$ — r -dimensional measurement noise; $x(0)$ —unknown initial state
- ▶ assumptions: $x(0)$, $w(k)$, and $v(k)$ are independent and Gaussian distributed; $w(k)$ and $v(k)$ are white:

$$E[x(0)] = x_o, \quad E[(x(0) - x_o)(x(0) - x_o)^T] = X_0$$

$$E[w(k)] = 0, \quad E[v(k)] = 0, \quad E[w(k)v^T(j)] = 0 \quad \forall k, j$$

$$E[w(k)w^T(j)] = W(k)\delta_{kj}, \quad E[v(k)v^T(j)] = V(k)\delta_{kj}$$

Problem statement

- ▶ goal:

$$\text{minimize } E \left[\|x(k) - \hat{x}(k)\|^2 | Y_j \right], \quad Y_j = \{y(0), y(1), \dots, y(j)\}$$

- ▶ solution:

$$\hat{x}(k) = E[x(k) | Y_j]$$

- ▶ three classes of problems:

- ▶ $k > j$: prediction problem
- ▶ $k = j$: filtering problem
- ▶ $k < j$: smoothing problem

History

Rudolf Kalman:

- ▶ obtained B.S. in 1953 and M.S. in 1954 from MIT, and Ph.D. in 1957 from Columbia University, all in Electrical Engineering
- ▶ developed and implemented Kalman Filter in 1960, during the Apollo program, and furthermore in various famous programs including the NASA Space Shuttle, Navy submarines, etc.
- ▶ was awarded the National Medal of Science on Oct. 7, 2009 from U.S. president Barack Obama

Useful facts

assume x is Gaussian distributed

- ▶ if $y = Ax + B$ then

$$\begin{cases} X_{xy} = E \left[(x - E[x]) (y - E[y])^T \right] &= X_{xx} A^T \\ X_{yy} = E \left[(y - E[y]) (y - E[y])^T \right] &= A X_{xx} A^T \end{cases} \quad (2)$$

- ▶ if $y = Ax + B$ and $y' = A'x + B'$ then

$$X_{yy'} = A X_{xx} (A')^T, \quad X_{y'y} = A' X_{xx} A^T \quad (3)$$

- ▶ if $y = Ax + Bv$; v is Gaussian and independent of x , then

$$X_{yy} = A X_{xx} A^T + B X_{vv} B^T \quad (4)$$

- ▶ if $y = Ax + Bv$, $y' = A'x + B'v$; v is Gaussian and dependent of x , then

$$X_{yy'} = A X_{xx} (A')^T + A X_{xv} (B')^T + B X_{vx} (A')^T + B X_{vv} (B')^T \quad (5)$$

Derivation of Kalman Filter

- ▶ goal:

$$\text{minimize } E \left[\|x(k) - \hat{x}(k)\|^2 \middle| Y_k \right], \quad Y_k = \{y(0), y(1), \dots, y(k)\}$$

- ▶ the best estimate is the conditional expectation

$$\begin{aligned} E[x(k) | Y_k] &= E[x(k) | \{Y_{k-1}, y(k)\}] \\ &= E[x(k) | Y_{k-1}] + E[\tilde{x}(k) | Y_{k-1} | \tilde{y}(k) | Y_{k-1}] \end{aligned}$$

- ▶ introduce some notations:

a priori estimation $\hat{x}(k|k-1) = E[x(k) | Y_{k-1}] = \hat{x}(k) |_{y(0), \dots, y(k-1)}$

a posteriori estimation $\hat{x}(k|k) = E[x(k) | Y_k] = \hat{x}(k) |_{y(0), \dots, y(k)}$

a priori covariance $M(k) = E[\tilde{x}(k) | Y_{k-1} \tilde{x}^T(k) | Y_{k-1}]$

a posteriori covariance $Z(k) = E[\tilde{x}(k) | Y_k \tilde{x}^T(k) | Y_k]$

Derivation of Kalman Filter

KF gain update

to get $E [\tilde{x}(k) | Y_{k-1} | \tilde{y}(k) | Y_{k-1}]$ in

$$E [x(k) | Y_k] = E [x(k) | Y_{k-1}] + E [\tilde{x}(k) | Y_{k-1} | \tilde{y}(k) | Y_{k-1}]$$

we need $X_{\tilde{x}(k) | Y_{k-1} \tilde{y}(k) | Y_{k-1}}$ and $X_{\tilde{y}(k) | Y_{k-1} \tilde{y}(k) | Y_{k-1}}^{-1}$

$y(k) = C(k)x(k) + v(k)$ gives

$$\begin{aligned}\hat{y}(k) | Y_{k-1} &= C(k) \hat{x}(k|k-1) + \hat{v}(k) | Y_{k-1} = C(k) \hat{x}(k|k-1) \\ \Rightarrow \tilde{y}(k) | Y_{k-1} &= C(k) \tilde{x}(k|k-1) + v(k)\end{aligned}$$

hence

$$X_{\tilde{x}(k) | Y_{k-1} \tilde{y}(k) | Y_{k-1}} = M(k) C^T(k) \quad (6)$$

$$X_{\tilde{y}(k) | Y_{k-1} \tilde{y}(k) | Y_{k-1}} = C(k) M(k) C^T(k) + V(k) \quad (7)$$

Derivation of Kalman Filter

KF gain update

$$\tilde{y}(k) | \gamma_{k-1} = C(k) \tilde{x}(k|k-1) + v(k)$$

unbiased estimation: $E[\hat{x}(k|k-1)] = E[x] \Rightarrow$

$$E[\tilde{y}(k) | \gamma_{k-1}] = E[\tilde{x}(k) | \gamma_{k-1}] + E[v(k) | \gamma_{k-1}] = 0$$

thus

$$\begin{aligned} & E[\tilde{x}(k) | \gamma_{k-1} | \tilde{y}(k) | \gamma_{k-1}] \\ &= E[\tilde{x}(k) | \gamma_{k-1}] + \overset{0}{\cancel{X_{\tilde{x}(k) | \gamma_{k-1}}}} X_{\tilde{x}(k) | \gamma_{k-1} \tilde{y}(k) | \gamma_{k-1}}^{-1} X_{\tilde{y}(k) | \gamma_{k-1} \tilde{y}(k) | \gamma_{k-1}} (\tilde{y}(k) | \gamma_{k-1} - 0) \\ &= M(k) C^T(k) [C(k) M(k) C^T(k) + V(k)]^{-1} (y(k) - \hat{y}(k) | \gamma_{k-1}) \end{aligned}$$

Derivation of Kalman Filter

KF gain update

$$E[x(k) | Y_k] = E[x(k) | Y_{k-1}] + E[\tilde{x}(k) | Y_{k-1} | \tilde{y}(k) | Y_{k-1}]$$

now becomes

$$\begin{aligned} \hat{x}(k|k) &= \hat{x}(k|k-1) \\ &+ \underbrace{M(k) C^T (C M(k) C^T + V(k))^{-1}}_{F(k)} (y(k) - C \hat{x}(k|k-1)) \end{aligned}$$

namely

$$\boxed{\begin{cases} \hat{x}(k|k) &= \hat{x}(k|k-1) + F(k) (y(k) - C(k) \hat{x}(k|k-1)) \\ F(k) &= M(k) C^T(k) (C(k) M(k) C^T(k) + V(k))^{-1} \end{cases}} \quad (8)$$

Derivation of Kalman Filter

KF covariance update

now for the variance update:

$$\begin{aligned} E \left[\tilde{x}(k) |_{Y_k} \tilde{x}(k)^T |_{Y_k} \right] &= E \left[\tilde{x}(k) |_{\{Y_{k-1}, y(k)\}} \tilde{x}(k)^T |_{\{Y_{k-1}, y(k)\}} \right] \\ &= E \left[\tilde{x}(k) |_{Y_{k-1}} \tilde{x}(k)^T |_{Y_{k-1}} \right] \\ &\quad - X_{\tilde{x}(k)|_{Y_{k-1}} \tilde{y}(k)|_{Y_{k-1}}} X_{\tilde{y}(k)|_{Y_{k-1}} \tilde{y}(k)|_{Y_{k-1}}}^{-1} X_{\tilde{y}(k)|_{Y_{k-1}} \tilde{x}(k)|_{Y_{k-1}}} \end{aligned}$$

or, using the introduced notations,

$$Z(k) = M(k) - M(k) C^T(k) (C(k) M(k) C^T(k) + V(k))^{-1} C(k) M(k)$$

Derivation of Kalman Filter

KF covariance update

the connection between $Z(k)$ and $M(k)$:

$$x(k) = A(k-1)x(k-1) + B(k-1)u(k-1) + B_w(k-1)w(k-1)$$

$$\Rightarrow \hat{x}(k|k-1) = A(k-1)\hat{x}(k-1|k-1) + B(k-1)u(k-1)$$

$$\Rightarrow \tilde{x}(k|k-1) = A(k-1)\tilde{x}(k-1|k-1) + B_w(k-1)w(k-1)$$

thus $M(k) = \text{Cov}[\tilde{x}(k|k-1)]$ is [using useful fact (4)]

$$M(k) = A(k-1)Z(k-1)A^T(k-1) + B_w(k-1)W(k-1)B_w^T(k-1)$$

$$\text{with } M(0) = E[\tilde{x}(0|-1)\tilde{x}(0|-1)^T] = X_0$$

The full set of KF equations

$$\begin{aligned}\hat{x}(k|k) &= \hat{x}(k|k-1) + F(k) \overbrace{[y(k) - C(k) \hat{x}(k|k-1)]}^{e_y(k)} \\ \hat{x}(k|k-1) &= A(k-1) \hat{x}(k-1|k-1) + B(k-1) u(k-1) \\ F(k) &= M(k) C^T(k) [C(k) M(k) C^T(k) + V(k)]^{-1} \\ M(k) &= A(k-1) Z(k-1) A^T(k-1) + B_w(k-1) W(k-1) B_w^T(k-1) \\ Z(k) &= M(k) - M(k) C^T(k) \dots \\ &\quad \cdot (C(k) M(k) C^T(k) + V(k))^{-1} C(k) M(k) \\ &= (I - F(k) C(k)) M(k)\end{aligned}$$

with initial conditions $\hat{x}(0|-1) = x_0$ and $M(0) = X_0$.

The full set of KF equations

in a shifted index:

$$\hat{x}(k+1|k+1) = \hat{x}(k+1|k) + F(k+1)[y(k+1) - C(k+1)\hat{x}(k+1|k)]$$

$$\hat{x}(k+1|k) = A(k)\hat{x}(k|k) + B(k)u(k)$$

$$F(k+1) = M(k+1)C^T(k+1) \left[C(k+1)M(k+1)C^T(k+1) + V(k+1) \right]^{-1}$$

$$M(k+1) = A(k)Z(k)A^T(k) + B_w(k)W(k)B_w^T(k) \quad (9)$$

$$Z(k+1) = M(k+1) - M(k+1)C^T(k+1) \dots \quad (10)$$

$$\begin{aligned} & \cdot \left(C(k+1)M(k+1)C^T(k+1) + V(k+1) \right)^{-1} C(k+1)M(k+1) \\ & = (I - F(k+1)C(k+1))M(k+1) \end{aligned} \quad (11)$$

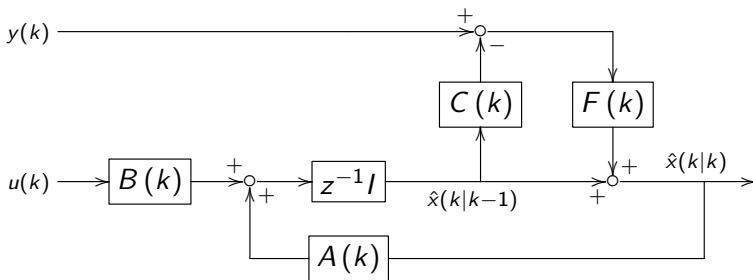
combining (9) and (10) gives the Riccati equation:

$$\begin{aligned} M(k+1) &= A(k)M(k)A^T(k) + B_w(k)W(k)B_w^T(k) \\ &- A(k)M(k)C^T(k) \left[C(k)M(k)C^T(k) + V(k) \right]^{-1} C(k)M(k)A^T(k) \end{aligned} \quad (12)$$

The full set of KF equations

Several remarks

- ▶ $F(k)$, $M(k)$, and $Z(k)$ can be obtained offline first
- ▶ Kalman Filter (KF) is linear, and optimal for Gaussian. More advanced nonlinear estimation won't improve the results here.
- ▶ KF works for time-varying systems
- ▶ the block diagram of KF is:



Steady-state KF

assumptions:

- ▶ system is time-invariant: $A(k)$, $B(k)$, $B_w(k)$, and $C(k)$ are constant;
- ▶ noise is stationary: $V \succ 0$ and $W \succ 0$ do not depend on time.

KF equations become:

$$\begin{aligned}\hat{x}(k+1|k+1) &= \hat{x}(k+1|k) + F(k+1)[y(k+1) - C\hat{x}(k+1|k)] \\ &= A\hat{x}(k|k) + Bu(k) + F(k+1)[y(k+1) - C\hat{x}(k+1|k)] \\ F(k+1) &= M(k+1)C^T [CM(k+1)C^T + V]^{-1} \\ M(k+1) &= AZ(k)A^T + B_wWB_w^T; \quad M(0) = X_0 \\ Z(k+1) &= M(k+1) - M(k+1)C^T [CM(k+1)C^T + V]^{-1} CM(k+1)\end{aligned}$$

with Riccati equation (RE):

$$M(k+1) = AM(k)A^T + B_wWB_w^T - AM(k)C^T [CM(k)C^T + V]^{-1} CM(k)A^T$$

Steady-state KF

- if
- ▶ (A, C) is observable or detectable
 - ▶ (A, B_w) is controllable (disturbable) or stabilizable

then $M(k)$ in the RE converges to some steady-state value M_s
and KF can be implemented by

$$\hat{x}(k+1|k+1) = \hat{x}(k+1|k) + F_s [y(k+1) - C\hat{x}(k+1|k)]$$

$$\hat{x}(k+1|k) = A\hat{x}(k|k) + Bu(k)$$

$$F_s = M_s C^T [CM_s C^T + V]^{-1}$$

M_s is the positive definite solution of the algebraic Riccati equation:

$$M_s = AM_s A^T + B_w W B_w^T - AM_s C^T [CM_s C^T + V]^{-1} CM_s A^T$$

Duality with LQ

The steady-state condition is obtained by comparing the RE in LQ and KF discrete-time LQ:

$$P(k) = A^T P(k+1)A - A^T P(k+1)B[R + B^T P(k+1)B]^{-1} B^T P(k+1)A + Q$$

discrete-time KF (12):

$$M(k+1) = AM(k)A^T - AM(k)C^T [CM(k)C^T + V]^{-1} CM(k)A^T + B_w WB_w^T$$

discrete-time LQ	discrete-time KF
A	A^T
B	C^T
C	B_w
R	V
$Q = C^T C$	$B_w WB_w^T$
P	M
backward recursion	forward recursion

Duality with LQ

discrete-time LQ	discrete-time KF
A	A^T
B	C^T
C	B_w
$Q = C^T C$	$B_w W B_w^T$

steady-state conditions for discrete-time LQ:

- ▶ (A, B) controllable or stabilizable
- ▶ (A, C) observable or detectable

steady-state conditions for discrete-time KF:

- ▶ (A^T, C^T) controllable or stabilizable $\Leftrightarrow (A, C)$ observable or detectable
- ▶ (A^T, B_w^T) observable or detectable $\Leftrightarrow (A, B_w)$ controllable or stabilizable

Duality with LQ

discrete-time LQ	discrete-time KF
A	A^T
B	C^T
C	B_w
R	V
$Q = C^T C$	$B_w W B_w^T$
P	M
backward recursion	forward recursion

- ▶ LQ: stable closed-loop “A” matrix is

$$A - BK_s = \underline{A - B[R + B^T P_s B]^{-1} B^T P_s A}$$

- ▶ KF: stable KF “A” matrix is

$$\begin{aligned}
 \hat{x}(k+1|k) &= A\hat{x}(k|k) + Bu(k) \\
 &= A\hat{x}(k|k-1) + AF_s[y(k) - C\hat{x}(k|k-1)] + Bu(k) \\
 &= \underline{\left[A - AM_s C^T (CM_s C^T + V)^{-1} C \right]} \hat{x}(k|k-1) + \dots
 \end{aligned}$$

Purpose of each condition

- ▶ (A, C) observable or detectable: assures the existence of the steady-state Riccati solution
- ▶ (A, B_w) controllable or stabilizable: assures that the steady-state solution is positive definite and that the KF dynamics is stable

Remark

- KF: stable KF “A” matrix is

$$\begin{aligned}\hat{x}(k+1|k) &= \left[A - AM_s C^T (CM_s C^T + V)^{-1} C \right] \hat{x}(k|k-1) + \dots \\ &= \underline{(A - AF_s C)} \hat{x}(k|k-1) + \dots\end{aligned}$$

in the form of $\hat{x}(k|k)$ dynamics:

$$\begin{aligned}\hat{x}(k+1|k+1) &= \hat{x}(k+1|k) + F_s [y(k+1) - C\hat{x}(k+1|k)] \\ &= \underline{(A - F_s CA)} \hat{x}(k|k) + (I - F_s C) Bu(k) + F_s y(k+1) \\ &= \left[A - M_s C^T (CM_s C^T + V)^{-1} CA \right] \hat{x}(k|k) + \dots\end{aligned}$$

- can show that $\text{eig}(A - AF_s C) = \text{eig}(A - F_s CA)$

$$\text{hint: } \det(I + MN) = \det(I + NM) \Rightarrow \det[I - z^{-1}A(I - F_s C)] = \det[I - (I - F_s C)z^{-1}A]$$

Remark

intuition of guaranteed KF stability: ARE \Rightarrow Lyapunov equation

$$\begin{aligned}M_s &= AM_s A^T + B_w W B_w^T - AM_s C^T \left[CM_s C^T + V \right]^{-1} CM_s A^T \\&= AM_s A^T + B_w W B_w^T - \underbrace{A M_s C^T \left[CM_s C^T + V \right]^{-1} \left[CM_s C^T + V \right]}_{F_s} \underbrace{\left[CM_s C^T + V \right]^{-1} CM_s A^T}_{F_s^T} \\&= (A - AF_s C) M_s (A - AF_s C)^T + 2AF_s CM_s A^T - AF_s CM_s C^T F_s^T A^T \\&\quad + B_w W B_w^T - AF_s \left[CM_s C^T + V \right] F_s^T A^T \\&= (A - AF_s C) M_s (A - AF_s C)^T + AF_s V F_s^T A^T + B_w W B_w^T \\&\iff (A - AF_s C) M_s (A - AF_s C)^T - M_s = -AF_s V F_s^T A^T - B_w W B_w^T\end{aligned}$$

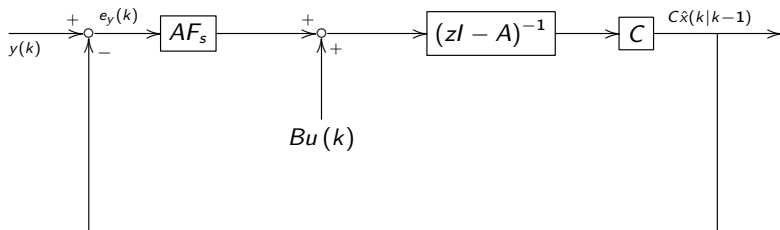
which is a Lyapunov equation with the right hand side being negative semidefinite and $M_s \succ 0$.

Return difference equation

KF dynamics

$$\begin{aligned}\hat{x}(k+1|k) &= A\hat{x}(k|k) + Bu(k) \\ &= A\hat{x}(k|k-1) + AF_s[y(k) - C\hat{x}(k|k-1)] + Bu(k)\end{aligned}$$

$$[zI - A]\mathcal{Z}\{\hat{x}(k|k-1)\} = AF_s\mathcal{Z}\{y(k) - C\hat{x}(k|k-1)\} + B\mathcal{Z}\{u(k)\}$$



let $G(z) = C(zI - A)^{-1}B_w$

ARE \Rightarrow return difference equation (RDE) (see course reader)

$$[I + C(zI - A)^{-1}AF_s] (V + CM_sC^T) [I + C(z^{-1}I - A)^{-1}AF_s]^T = V + G(z)WG^T(z^{-1})$$

Symmetric root locus for KF

- KF eigenvalues:

$$\begin{aligned}\det \left[I + C(zI - A)^{-1}AF_s \right] &= \det \left[I + \frac{(zI - A)^{-1}AF_s C}{1} \right] \\ &= \frac{\det (zI - A + AF_s C)}{\det (zI - A)} \triangleq \frac{\beta(z)}{\phi(z)}\end{aligned}$$

- taking determinants in RDE gives

$$\beta(z)\beta(z^{-1}) = \phi(z)\phi(z^{-1}) \frac{\det (V + G(z)WG^T(z^{-1}))}{\det (V + CMCT^T)}$$

- single-output case: KF poles come from $\beta(z)\beta(z^{-1}) = 0$, i.e.

$$\det \left(V + G(z)WG^T(z^{-1}) \right) = V \left(1 + G(z) \frac{W}{V} G^T(z^{-1}) \right) = 0$$

- $W/V \rightarrow 0$: KF poles \rightarrow stable poles of $G(z)G^T(z^{-1})$
- $W/V \rightarrow \infty$: KF poles \rightarrow stable zeros of $G(z)G^T(z^{-1})$

Continuous-time KF

summary of solutions

system:

$$\begin{aligned}\dot{x}(t) &= Ax(t) + Bu(t) + B_w w(t) \\ y(t) &= Cx(t) + v(t)\end{aligned}$$

assumptions: same as discrete-time KF

aim: minimize $J = \|x(t) - \hat{x}(t)\|_2^2 \big|_{\{y(\tau): 0 \leq \tau \leq t\}}$

continuous-time KF:

$$\frac{d\hat{x}(t|t)}{dt} = A\hat{x}(t|t) + Bu(t) + F(t)[y(t) - C\hat{x}(t|t)], \quad \hat{x}(0|0) = x_0$$

$$F(t) = M(t)C^T V^{-1}$$

$$\frac{dM(t)}{dt} = AM(t) + M(t)A^T + B_w W B_w^T - M(t)C^T V^{-1} C M(t), \quad M(0) = X_0$$

Continuous-time KF: steady state

assumptions: (A, C) observable or detectable;
 (A, B_w) controllable or stabilizable

asymptotically stable steady-state KF:

$$\frac{d\hat{x}(t|t)}{dt} = A\hat{x}(t|t) + Bu(t) + F_s[y(t) - C\hat{x}(t|t)]$$

$$F_s = M_s C^T V^{-1}$$

$$AM_s + M_s A^T + B_w W B_w^T - M_s C^T V^{-1} C M_s = 0$$

duality with LQ:

$$\begin{array}{c} \text{Continuous-Time LQ} \\ \hline A^T P_s + P_s A + Q - P_s B R^{-1} B^T P_s = 0 \\ K = R^{-1} B^T P_s \end{array}$$

Continuous-time KF: return difference equality

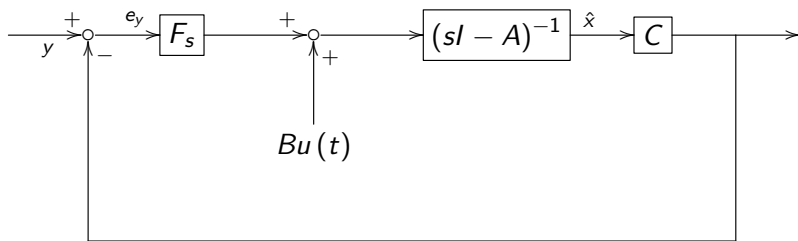
analogy to LQ gives the return difference equality:

$$\left[I + C (sI - A)^{-1} F_s \right] V \left[I + F_s^T (-sI - A)^{-T} C^T \right] = V + G(s) W G^T(-s)$$

where $G(s) = C (sI - A)^{-1} B_w$, hence:

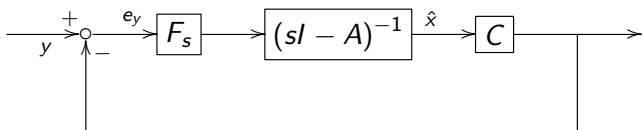
$$\left[I + C (j\omega I - A)^{-1} F_s \right] V \left[I + C (-j\omega I - A)^{-1} F_s \right]^T = V + G(j\omega) W G^T(-j\omega)$$

observation 1: $\frac{d\hat{x}(t|t)}{dt} = A\hat{x}(t|t) + Bu(t) + \underbrace{F_s [y(t) - C\hat{x}(t|t)]}_{e_y(t)}$



Continuous-time KF: properties

observation 1:



► transfer function from y to e_y : $\left[I + C (j\omega I - A)^{-1} F_s \right]^{-1}$

► spectral density relation:

$$\Phi_{e_y e_y}(\omega) = \underbrace{\left[I + C (j\omega I - A)^{-1} F_s \right]^{-1}}_{G_{y \rightarrow e_y}(j\omega)} \Phi_{yy}(\omega) \underbrace{\left\{ \left[I + C (-j\omega I - A)^{-1} F_s \right]^{-1} \right\}^T}_{G_{y \rightarrow e_y}(-j\omega)^T}$$

Continuous-time KF: properties

observation 2:

$$\begin{cases} \dot{x}(t) = Ax(t) + Bu(t) + B_w w(t) \\ y(t) = Cx(t) + v(t) \end{cases} \Rightarrow \Phi_{yy}(\omega) = G(j\omega) W G^T(-j\omega) + V$$

from observations 1 and 2:

$$\left[I + C(j\omega I - A)^{-1} F_s \right] V \left[I + C(-j\omega I - A)^{-1} F_s \right]^T = V + G(j\omega) W G^T(-j\omega)$$

thus says

$$\begin{aligned} \Phi_{e_y e_y}(\omega) &= \left[I + C(j\omega I - A)^{-1} F_s \right]^{-1} \Phi_{yy}(\omega) \left\{ \left[I + C(-j\omega I - A)^{-1} F_s \right]^{-1} \right\}^T \\ &= V \end{aligned}$$

namely, the estimation error is white!

Continuous-time KF: symmetric root locus

taking determinants of RDE gives:

$$\begin{aligned}\det \left[I + C (sI - A)^{-1} F_s \right] \det V \det \left[I + C (-sI - A)^{-1} F_s \right]^T \\ = \det \left[V + G(s) W G^T(-s) \right]\end{aligned}$$

for single-output systems:

$$\det \left[I + C (sI - A)^{-1} F_s \right] \det \left[I + C (-sI - A)^{-1} F_s \right]^T = 1 + G(s) \frac{W}{V} G^T(-s)$$

Continuous-time KF: symmetric root locus

the left hand side of

$$\det \left[I + C (sI - A)^{-1} F_s \right] \det \left[I + C (-sI - A)^{-1} F_s \right]^T = 1 + G(s) \frac{W}{V} G^T(-s)$$

determines the KF eigenvalues:

$$\begin{aligned} \det \left[I + C (sI - A)^{-1} F_s \right] &= \det \left[I + (sI - A)^{-1} F_s C \right] \\ &= \det \left[(sI - A)^{-1} \right] \det [sI - A + F_s C] \\ &= \frac{\det [sI - (A - F_s C)]}{\det (sI - A)} \end{aligned}$$

hence looking at $1 + G(s) \frac{W}{V} G^T(-s)$, we have:

- ▶ $W/V \rightarrow 0$: KF poles \rightarrow stable poles of $G(s) G^T(-s)$
- ▶ $W/V \rightarrow \infty$: KF poles \rightarrow stable zeros of $G(s) G^T(-s)$

Summary

1. Big picture

2. Problem statement

3. Discrete-time KF

- Gain update

- Covariance update

- Steady-state KF

- Duality with LQ

4. Continuous-time KF

- Solution

- Steady-state solution and conditions

- Properties: return difference equality, symmetric root locus...

*KF equations using other notation systems

- ▶ the field of information fusion often uses a different set of notations
- ▶ system model:

$$\begin{aligned}x(k+1) &= A(k)x(k) + B(k)u(k) + w(k) \\ z(k) &= H(k)x(k) + v(k)\end{aligned}$$

where $E[w(k)w(i)^T] = Q(k)\delta(k-i)$,
 $E[v(k)v(i)^T] = R(k)\delta(k-i)$, and $E[w(k)v(i)^T] = 0 \forall k, i$.

- ▶ state prediction covariance:

$$P(k+1|k) \triangleq E\left\{(x(k+1) - \hat{x}(k+1|k))(x(k+1) - \hat{x}(k+1|k))^T\right\}$$

- ▶ Kalman Filter gain: denoted as $K(k)$ (sometimes also as $W(k)$)

*KF equations using other notation systems

► prediction

$$\hat{x}(k|k-1) = A(k-1) \hat{x}(k-1|k-1) + B(k-1) u(k-1)$$

$$P(k|k-1) = A(k-1) P(k-1|k-1) A^T(k-1) + Q(k-1)$$

► correction/update

$$\hat{x}(k|k) = \hat{x}(k|k-1) + K(k) [z(k) - H(k) \hat{x}(k|k-1)]$$

$$K(k) = P(k|k-1) H^T(k) [H(k) P(k|k-1) H^T(k) + R(k)]^{-1}$$

$$P(k|k) = P(k|k-1) - P(k|k-1) H^T(k) \dots$$

$$\begin{aligned} & \cdot (H(k) P(k|k-1) H^T(k) + R(k))^{-1} H(k) P(k|k-1) \\ & = (I - K(k) H(k)) P(k|k-1) \end{aligned}$$

with initial conditions $\hat{x}(0|-1) = x_o$ and $P(0|0) = X_0$.

*KF equations using other notation systems

in a shifted index:

$$\hat{x}(k+1|k+1) = \hat{x}(k+1|k) + K(k+1) [z(k+1) - H(k+1) \hat{x}(k+1|k)]$$

$$\hat{x}(k+1|k) = A(k) \hat{x}(k|k) + B(k) u(k)$$

$$K(k+1) = P(k+1|k) H^T(k+1) \left[H(k+1) P(k+1|k) H^T(k+1) + R(k+1) \right]^{-1}$$

$$P(k+1|k) = A(k) P(k|k) A^T(k) + Q(k)$$

$$P(k+1|k+1) = P(k+1|k) - P(k+1|k) H^T(k+1) \dots$$

$$\cdot \left(H(k+1) P(k+1|k) H^T(k+1) + R(k+1) \right)^{-1} \dots$$

$$\cdot H(k+1) P(k+1|k)$$

$$= (I - K(k+1) H(k+1)) P(k+1|k)$$

*The information form of the KF, aka, information filter

- ▶ system model (leveraging the benefits of the two notation systems):

$$\begin{aligned}x(k+1) &= A(k)x(k) + B(k)u(k) + w(k) \\ y(k) &= C(k)x(k) + v(k)\end{aligned}$$

where $E[w(k)w(i)^T] = Q(k)\delta(k-i)$,
 $E[v(k)v(i)^T] = R(k)\delta(k-i)$, and $E[w(k)v(i)^T] = 0 \forall k, i$.

- ▶ state prediction covariance:

$$P(k+1|k) \triangleq E\left\{(x(k+1) - \hat{x}(k+1|k))(x(k+1) - \hat{x}(k+1|k))^T\right\}$$

*The information form of the KF, aka, information filter

► KF:

$$\hat{x}(k|k-1) = A(k-1) \hat{x}(k-1|k-1) + B(k-1) u(k-1)$$

$$P(k|k-1) = A(k-1) P(k-1|k-1) A^T(k-1) + Q(k-1) \quad \uparrow \text{predict}$$

$$\hat{x}(k|k) = \hat{x}(k|k-1) + K(k) [y(k) - C(k) \hat{x}(k|k-1)] \quad \downarrow \text{correct}$$

$$K(k) = P(k|k-1) C^T(k) [C(k) P(k|k-1) C^T(k) + R(k)]^{-1}$$

$$P(k|k) = P(k|k-1) - P(k|k-1) C^T(k) \dots$$

$$\cdot (C(k) P(k|k-1) C^T(k) + R(k))^{-1} \dots$$

$$\cdot C(k) P(k|k-1)$$

$$= (I - K(k) C(k)) P(k|k-1)$$

with initial conditions $\hat{x}(0|-1) = x_0$ and $P(0|0) = X_0$.

*The information form of the KF, aka, information filter

- ▶ standard KF form computes the inverse of the innovation covariance matrix, i.e.,

$$\left[C(k) P(k|k-1) C^T(k) + R(k) \right]^{-1}$$

which can be expensive, e.g.:

- ▶ for high-dimension systems and when nondiagonal elements appear
- ▶ when many sensors are available to create a fusion of information
- ▶ from the Matrix Inversion Lemma,
 $\left[C(k) P(k|k-1) C^T(k) + R(k) \right]^{-1}$ will involve the inverse of the covariance $P(k|k-1)^{-1}$, the information matrix, aka the Fisher information

*The information form of the KF, aka, information filter

- ▶ let's focus on the Fisher information. By using the Matrix Inversion Lemma, we can rewrite the innovation covariance equation as

$$P(k|k)^{-1} = P(k|k-1)^{-1} + C(k)^T R(k)^{-1} C(k)$$

b/c $(A + BCB^T)^{-1} = A^{-1} - A^{-1}B(B^T A^{-1}B + C^{-1})^{-1}B^T A^{-1}$ and hence

$$\left(P_{k|k-1}^{-1} + C_k^T R_k^{-1} C_k\right)^{-1} = P_{k|k-1} - P_{k|k-1} C_k^T \left(C_k P_{k|k-1} C_k^T + R_k\right)^{-1} C_k P_{k|k-1}$$

Here, for brevity, we introduced the shorthand $P_{i|j} \triangleq P(i|j)$

*The information form of the KF, aka, information filter

► define

$$\mathcal{I}(i|j) = P(i|j)^{-1}$$

we have

$$\begin{aligned}\mathcal{I}(k|k) &= \mathcal{I}(k|k-1) + C(k)^T R(k)^{-1} C(k) \\ \mathcal{I}(k+1|k) &= P(k+1|k)^{-1} = \left(A_k P_{k|k} A_k^T + Q_k \right)^{-1} \\ &= Q_k^{-1} - Q_k^{-1} A_k \left(A_k^T Q_k^{-1} A_k + \mathcal{I}_{k|k}^{-1} \right)^{-1} A_k^T Q_k^{-1}\end{aligned}$$

*The information form of the KF, aka, information filter

► and the KF gain is

$$\begin{aligned}
 K(k) &= P_{k|k-1} C_k^T \left[C_k P_{k|k-1} C_k^T + R_k \right]^{-1} \\
 &= \mathcal{I}_{k|k-1}^{-1} C_k^T \left[R_k^{-1} - R_k^{-1} C_k \underbrace{\left(C_k^T R_k^{-1} C_k + \underbrace{P_{k|k-1}^{-1}}_{\mathcal{I}_{k|k-1}} \right)^{-1}}_{\mathcal{I}_{k|k}^{-1}} C_k^T R_k^{-1} \right] \\
 &= \left[\mathcal{I}_{k|k-1}^{-1} - \mathcal{I}_{k|k-1}^{-1} C_k^T R_k^{-1} C_k \mathcal{I}_{k|k}^{-1} \right] C_k^T R_k^{-1} \\
 &= \left[\mathcal{I}_{k|k-1}^{-1} - \mathcal{I}_{k|k-1}^{-1} (\mathcal{I}_{k|k} - \mathcal{I}_{k|k-1}) \mathcal{I}_{k|k}^{-1} \right] C_k^T R_k^{-1} \\
 &= \mathcal{I}_{k|k}^{-1} C_k^T R_k^{-1}
 \end{aligned}$$

*The full set of the information filter

$$\hat{x}_{k|k-1} = \hat{x}_{k-1|k-1} + B_{k-1}u_{k-1}$$

$$\mathcal{I}_{k|k-1} = Q_{k-1}^{-1} - Q_{k-1}^{-1}A_{k-1} \left(A_{k-1}^T Q_{k-1}^{-1} A_{k-1} + \mathcal{I}_{k-1|k-1}^{-1} \right)^{-1} A_{k-1}^T Q_{k-1}^{-1}$$

$$\mathcal{I}_{k|k} = \mathcal{I}_{k|k-1} + C_k^T R_k^{-1} C_k$$

$$K_k = \mathcal{I}_{k|k}^{-1} C_k^T R_k^{-1}$$

$$\hat{x}_{k|k} = \hat{x}_{k|k-1} + K_k [y_k - C_k \hat{x}_{k|k-1}]$$

with initial conditions $\hat{x}(0|-1) = x_o$ and $\mathcal{I}(0|0) = X_0^{-1}$.