

Linear Systems: Stability



1. Definitions in Lyapunov stability analysis

2. Stability of LTI systems: method of eigenvalue/pole locations

Finite dimensional vector norms

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default in this set of notes: $\|\cdot\| = \|\cdot\|_2$

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For an n -th order unforced system

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- ▶ the condition must be satisfied by all $t \geq 0$
- ▶ if a system starts at equilibrium state, it stays there

Equilibrium state of a linear system

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$$\dot{x}(t) = A(t)x(t), \quad x(t_0) = x_0$$

- ▶ origin $x_e = 0$ is always an equilibrium state
- ▶ when $A(t)$ is singular, multiple equilibrium states exist

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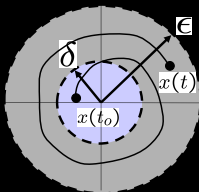


Figure: Stable s.i.L: $\|x(t_0)\| < \delta \Rightarrow \|x(t)\| < \epsilon \forall t \geq t_0$.

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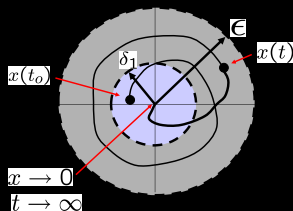


Figure: Asymptotically stable i.s.L: $\|x(t_0)\| < \delta \Rightarrow \|x(t)\| \rightarrow 0$.

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- ▶ $e^{\sigma t} \rightarrow 0$ if $\sigma < 0$; $e^{\lambda t} \rightarrow 0$ if $\lambda < 0$
- ▶ $\lambda^k \rightarrow 0$ if $|\lambda| < 1$; $r^k \rightarrow 0$ if $|r| = |\sqrt{\sigma^2 + \omega^2}| = |\lambda| < 1$

Stability of the origin for $\dot{x} = Ax$

stability at 0	$\lambda_i(A)$
unstable	$\text{Re}\{\lambda_i\} > 0$ for some λ_i or $\text{Re}\{\lambda_i\} \leq 0$ for all λ_i 's but for a repeated λ_m on the imaginary axis with multiplicity m , nullity $(A - \lambda_m I) < m$ (Jordan form)
stable i.s.L	$\text{Re}\{\lambda_i\} \leq 0$ for all λ_i 's and \forall repeated λ_m on the imaginary axis with multiplicity m , nullity $(A - \lambda_m I) = m$ (diagonal form)
asymptotically stable	$\text{Re}\{\lambda_i\} < 0 \forall \lambda_i$ (A is then called Hurwitz stable)

Example (Unstable moving mass)

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- ▶ i.e., two repeated eigenvalues but needs a generalized eigenvector \Rightarrow Jordan form after similarity transform
- ▶ verify by checking $e^{At} = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix}$: t grows unbounded

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Routh-Hurwitz criterion

- ▶ the Routh Test (by E.J. Routh, in 1877): a simple algebraic procedure to determine how many roots a given polynomial

$$A(s) = a_n s^n + a_{n-1} s^{n-1} + \cdots + a_1 s + a_0$$

has in the closed right-half complex plane, without the need to explicitly solve for the roots

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- ▶ popular if stability is the only concern and no details on eigenvalues (e.g., speed of response) are needed

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- ▶ simply apply the Routh Test to $A(s) = \det(sI - A)$
- ▶ recap: the poles of transfer function $G(s) = C(sI - A)^{-1}B + D$ come from $\det(sI - A)$ in computing the inverse $(sI - A)^{-1}$

The Routh Array

for $A(s) = a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0$, construct

$$\begin{array}{c|cccccc} s^n & a_n & a_{n-2} & a_{n-4} & a_{n-6} & \cdots \\ s^{n-1} & a_{n-1} & a_{n-3} & a_{n-5} & a_{n-7} & \cdots \\ s^{n-2} & q_{n-2} & q_{n-4} & q_{n-6} & \cdots & \\ s^{n-3} & q_{n-3} & q_{n-5} & q_{n-7} & \cdots & \\ \vdots & \vdots & \vdots & \vdots & & \\ s^1 & x_2 & x_0 & & & \\ s^0 & x_0 & & & & \end{array}$$

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 s^1 & x_2 & x_0 & & & \\
 s^0 & x_0 & & & &
 \end{array}$$

- ▶ first two rows contain the coefficients of $A(s)$
- ▶ third row constructed from the previous two rows via

$$\begin{array}{c|ccc}
 \cdot & a & b & x & \cdot \\
 \cdot & c & d & y & \cdot \\
 \cdot & \frac{bc - ad}{c} & \frac{xc - ay}{c} & \cdot & \\
 \cdot & \cdot & \cdot & \cdot & \cdot
 \end{array}$$

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- All roots of $A(s)$ are on the left half s-plane if and only if all elements of the first column of the Routh array are positive.

The Routh Array

Example ($A(s) = 2s^4 + s^3 + 3s^2 + 5s + 10$)

$$\begin{array}{c|ccc} s^4 & 2 & 3 & 10 \\ s^3 & 1 & 5 & 0 \\ s^2 & 3 - \frac{2 \times 5}{1} = -7 & 10 & 0 \\ s^1 & 5 - \frac{1 \times 10}{-7} & 0 & 0 \\ s^0 & 10 & 0 & 0 \end{array}$$

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- ▶ two sign changes in the first column
- ▶ unstable and two roots in the right half side of s-plane

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special cases:

- ▶ If the 1st element in any one row of Routh's array is zero, one can replace the zero with a small number ϵ and proceed further.

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- ▶ If the 1st element in any one row of Routh's array is zero, one can replace the zero with a small number ϵ and proceed further.
- ▶ There are other possible complications, which we will not pursue further. See, e.g. "Automatic Control Systems", by Kuo, 7th ed., pp. 339-340.

Stability of the origin for $x(k+1) = f(x(k), k)$

- stability analysis follows analogously for nonlinear time-varying discrete-time systems of the form

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- ▶ equilibrium point x_e :

$$f(x_e, k) = x_e, \quad \forall k$$

- ▶ without loss of generality, 0 is assumed an equilibrium point

Stability of the origin for $x(k+1) = Ax(k)$

stability at 0	$\lambda_i(A)$
unstable	$ \lambda_i > 1$ for some λ_i or $ \lambda_i \leq 1$ for all λ_i 's but for a repeated λ_m on the unit circle with multiplicity m , $\text{nullity}(A - \lambda_m I) < m$ (Jordan form)
stable i.s.L	$ \lambda_i \leq 1$ for all λ_i 's but for any repeated λ_m on the unit circle with multiplicity m , $\text{nullity}(A - \lambda_m I) = m$ (diagonal form)
asymptotically stable	$ \lambda_i < 1 \ \forall \lambda_i$ (such a matrix is called Schur stable)

Routh-Hurwitz criterion for DT LTI systems

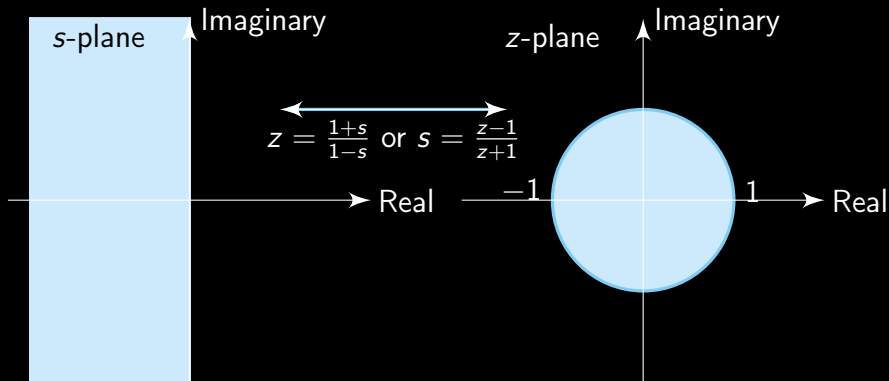
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- ▶ the stability domain $|\lambda_i| < 1$ is a unit disk
- ▶ Routh array validates stability in the left-half plane
- ▶ bilinear transformation maps the closed left half s -plane to the closed unit disk in z -plane



Routh-Hurwitz criterion for DT LTI systems

- ▶ Given $A(z) = z^n + a_1 z^{n-1} + \dots + a_n$, procedures of Routh-Hurwitz test:
 - ▶ apply bilinear transform
$$A(z)|_{z=\frac{1+s}{1-s}} = \left(\frac{1+s}{1-s}\right)^n + a_1 \left(\frac{1+s}{1-s}\right)^{n-1} + \dots + a_n =: \frac{A^*(s)}{(1-s)^n}$$
 - ▶ apply Routh test to
$$A^*(s) = a_n^* s^n + a_{n-1}^* s^{n-1} + \dots + a_0^* = A(z)|_{z=\frac{1+s}{1-s}} (1-s)^n$$

Routh-Hurwitz criterion for DT LTI systems

Example ($A(z) = z^3 + 0.8z^2 + 0.6z + 0.5$)

$$\blacktriangleright A^*(s) = A(z)|_{z=\frac{1+s}{1-s}} (1-s)^3 = (1+s)^3 + 0.8(1+s)^2(1-s) + 0.6(1+s)(1-s)^2 + 0.5(1-s)^3 = 0.3s^3 + 3.1s^2 + 1.7s + 2.9$$

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► Routh array

$$\begin{array}{c|cc} s^3 & 0.3 & 1.7 \\ s^2 & 3.1 & 2.9 \\ s & 1.7 - \frac{0.3 \times 2.9}{3.1} = 1.42 & 0 \\ s^0 & 2.9 & 0 \end{array}$$

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► all elements in first column are positive \Rightarrow roots of $A(z)$ are all in the unit circle