Introduction to Modern Controls Inverse Laplace transform



Common Laplace transform pairs

f(t)	F(s)	f(t)	F(s)
-(')	7 (3)	7(0)	7 (3)
$\sin \omega t$	$rac{\omega}{s^2 + \omega^2}$	e^{-at}	$\frac{1}{s+a}$
$\cos \omega t$	$\frac{s}{s^2 + \omega^2}$	t	$\frac{1}{s^2}$
tx(t)	$-\frac{dX(s)}{\int_{-\infty}^{\infty}ds}$	t^2	$\frac{2}{s^3}$
$\frac{x(t)}{t}$	$\int_{s}^{\infty} X(s) ds$	te ^{-at}	$\frac{1}{(s+a)^2}$
δ (t)	1	$e^{-at}\sin(\omega t)$	$\frac{\omega}{(s+a)^2+\omega^2}$
1 (t)	1 s	$e^{-at}\cos(\omega t)$	$\frac{s+a}{(s+a)^2+\omega^2}$

Overview of inverse Laplace transform: modularity and decomposition

• goal: to break a large Laplace transform into small blocks, so that we can use elemental examples of Laplace transfer functions:

$$G(s) = \frac{B(s)}{A(s)} = \frac{B_1(s)}{A_1(s)} + \frac{B_2(s)}{A_2(s)} + \dots$$

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 we will use examples to demonstrate strategies for common fractional expansions

Real and distinct roots in A(s)

example 1

$$G(s) = \frac{B(s)}{A(s)} = \frac{32}{s(s+4)(s+8)} = \frac{K_1}{s} + \frac{K_2}{s+4} + \frac{K_3}{s+8}$$

residues:

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$$K_1 = \lim_{s \to 0} sG(s) = 1$$

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•
$$K_3 = \lim_{s \to -8} (s+8) G(s) = 1$$

Coding partial fraction expansions

```
syms s
G = 32/s/(s+4)/(s+8)
partfrac(G)

# Python
import sympy
s = sympy.symbols('s')
G = 32/s/(s+4)/(s+8)
```

print(sympy.apart(G))

1/(s + 8) - 2/(s + 4) + 1/s

% MATLAB

Real and repeated roots in A(s)

example 2

$$G(s) = \frac{2}{(s+1)(s+2)^2} = \frac{K_1}{s+1} + \frac{K_2}{s+2} + \frac{K_3}{(s+2)^2}$$

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$$K_3 = \lim_{s \to -2} (s+2)^2 G(s) = -2$$

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- $K_1 = \lim_{s \to -1} (s+1) G(s) = 2$

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- $K_3 = \lim_{s \to -2} (s+2)^2 G(s) = -2$
- $K_1 = \lim_{s \to -1} (s+1) G(s) = 2$
- for K_2 , we hit both sides with $(s+2)^2$ then differentiate once w.r.t. s, to get

$$K_2 = \lim_{s \to -2} \frac{d}{ds} (s+2)^2 G(s) = -2$$

Coding partial fraction expansions

```
import sympy
s = sympy.symbols('s')
G = 2/(s+1)/(s+2)**2
print(sympy.apart(G))

-2/(s + 2) - 2/(s + 2)**2 + 2/(s + 1)
```

Python

example 1: Let $a > 0, b > 0, y(0) = y_0 \in \mathbb{R}$, obtain the solution to the ODE:

$$\dot{y}(t) = -ay(t) + b1(t)$$

where
$$1(t) = \left\{ egin{array}{ll} 1, & t \geq 0 \\ 0, & t < 0 \end{array} \right.$$

• Laplace transform: $\mathcal{L}\{\dot{y}(t)\} = sY(s) - y(0)$

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$$Y(s) = \frac{1}{s+a}y(0) + \frac{b}{s(s+a)} = \frac{1}{s+a}y(0) + \frac{b}{a}\left(\frac{1}{s} - \frac{1}{s+a}\right)$$

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• apply inverse Laplace transform: $y(t) = \mathcal{L}^{-1}\{Y(s)\} = \dots$

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- solution:

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- using final value theorem,

$$\lim_{t\to\infty} y(t) = \lim_{s\to 0} sY(s) = \frac{b}{a}$$

example 2: Let $a > 0, b > 0, y(0) = y_0 \in \mathbb{R}$, obtain the solution to the ODE:

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- Q: what's the initial value from initial value theorem? what does the impulse do to the initial condition?

Connecting two domains

n-th order differential equation:

$$\frac{d^{n}y}{dt^{n}} + a_{n-1}\frac{d^{n-1}y}{dt^{n-1}} + \dots + a_{1}\dot{y} + a_{0}y = b_{m}\frac{d^{m}u}{dt^{m}} + b_{m-1}\frac{d^{m-1}u}{dt^{m-1}} + \dots + b_{1}\dot{u} + b_{o}u$$
where $y(0) = 0, \frac{dy}{dt}|_{t=0} = 0, \dots, \frac{d^{n-1}y}{dt^{n-1}}|_{t=0} = 0$

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where
$$y(0) = 0$$
, $\frac{dy}{dt}|_{t=0} = 0$, ..., $\frac{d^{n-1}y}{dt^{n-1}}|_{t=0} = 0$

applying Laplace transform yields

$$(s^{n} + a_{n-1}s^{n-1} + \dots + a_{0})Y(s) = (b_{m}s^{m} + b_{m-1}s^{m-1} + \dots + b_{0})U(s)$$

$$\Rightarrow Y(s) = \frac{b_{m}s^{m} + b_{m-1}s^{m-1} + \dots + b_{0}}{s^{n} + a_{n-1}s^{n-1} + \dots + a_{0}}U(s)$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_0}$$

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 - ▶ proper if $n \ge m$
 - ▶ strictly proper if n > m
- examples: $G_1(s) = K$, $G_2(s) = \frac{k}{s+a}$

Coding transfer functions in Python

```
import control as co
import matplotlib.pyplot as plt
import numpy as np
num = [1,2] # Numerator co-efficients
den = [1,2,3] # Denominator co-efficients
sys_tf = co.tf(num,den)
print(sys_tf)
poles = co.pole(sys_tf)
zeros = co.zero(sys_tf)
print('\nSystem Poles = ', poles, '\nSystem Zeros = ', zeros)
T,yout = co.step_response(sys_tf)
plt.figure(1, figsize = (6,4))
plt.plot(T,yout)
plt.grid(True)
plt.ylabel("y")
plt.xlabel("Time (sec)")
plt.show()
```

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sys_tf = co.tf(num,den)
T,yout = co.step_response(sys_tf)
u1 = np.full((1, len(T)), 2) # Create an array of 2's
u2 = np.sin(T)
T, yout_u1 = co.forced_response(sys_tf,T,u1) # Response to input 1
T,yout_u2 = co.forced_response(sys_tf,T,u2) # Response to input 2
```

The DC gain

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_0}$$

 DC gain: the ratio of a stable system's output to its input after all transients have decayed

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- DC gain: the ratio of a stable system's output to its input after all transients have decayed
- can use the Final Value Theorem to find the DC gain:

$$\underline{\mathsf{DC}\ \mathsf{gain}\ \mathsf{of}\ \mathit{G}(\mathit{s})} = \lim_{s \to 0} \mathit{sY}(\mathit{s}) = \lim_{s \to 0} \mathit{sG}(\mathit{s}) \frac{1}{\mathit{s}} = \lim_{s \to 0} \mathit{G}(\mathit{s})$$

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$$\underline{\mathsf{DC} \ \mathsf{gain} \ \mathsf{of} \ \mathsf{G}(s)} = \lim_{s \to 0} s \mathsf{Y}(s) = \lim_{s \to 0} s \mathsf{G}(s) \frac{1}{s} = \lim_{s \to 0} \mathsf{G}(s)$$

• example: find the DC gain of $G_1(s) = K$ and $G_2(s) = \frac{k}{s+a}$. Try (i) solve the ODE and (ii) the Final Value Theorem

The DC gain in Matlab and Python

```
% MATLAB

s = tf('s');

G = (2*s+3)/(4*s^2+3*s+1);

dcgain(G)
```

```
# Python
import control as co
s = co.tf('s')
G = (2*s+3)/(4*s**2+3*s+1);
print(co.dcgain(G))
```

The DC gain in Matlab and Python

find the DC gain of the system corresponding to $Y_2(s) = \frac{3}{s-2}$

```
# Python
import control as co
H = co.tf([0, 3],[1, -2])
print(co.dcgain(H))
T, yout = co.step_response(H)
print(yout)
```

exercise: verify the result in Matlab

The DC gain in Matlab and Python

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```

- exercise: verify the result in Matlab
- is the result correct?